

$$1) \quad x(h+1) = \begin{bmatrix} 1 & 0 \\ 1 & 1 \end{bmatrix} x(h) + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u(h)$$

$$u(h) = -Lx(h) + Kr(h) = -[l_1 \ l_2] x(h) + Kr(h)$$

$$\Rightarrow x(h+1) = \underbrace{\begin{bmatrix} 1-l_1 & -l_2 \\ 1 & 1 \end{bmatrix}}_{A_c} x(h) + \underbrace{\begin{bmatrix} K \\ 0 \end{bmatrix}}_{B_r} r(h)$$

a) Poles = eigenvalues of A_c

$$\begin{aligned} \det(\lambda I - A_c) &= \det \begin{bmatrix} \lambda - 1 + l_1 & l_2 \\ -1 & \lambda - 1 \end{bmatrix} \\ &= (\lambda - 1)(\lambda - 1 - l_1) + l_2 \\ &= \lambda^2 + \lambda(l_1 - 2) + l_2 - l_1 + 1 \\ &= (\lambda - 0.9)^2 \\ &= \lambda^2 - 1.8\lambda + 0.81 \end{aligned}$$

$$\Rightarrow \underline{l_1 = 0.2} \quad \text{and} \quad \underline{l_2 = 0.81 + l_1 - 1 = 0.01}$$

$$b) \quad x(h+1) = A_c x(h) + B_r r(h)$$

$$y(h) = c x(h)$$

$$\text{stable \& SS} \Rightarrow x(h+1) = x(h) = \bar{x}$$

$$(I - A_c) \bar{x} = B_r \bar{r}$$

$$\Rightarrow \bar{y} = c(I - A_c)^{-1} B_r \bar{r}$$

$$= [0 \ 1] \begin{bmatrix} l_1 & l_2 \\ -1 & 0 \end{bmatrix}^{-1} \begin{bmatrix} K \\ 0 \end{bmatrix} \bar{r}$$

$$= [0 \ 1] \underbrace{\frac{1}{l_2} \begin{bmatrix} 0 & -l_2 \\ 1 & 1 \end{bmatrix}}_{= 1} \begin{bmatrix} K \\ 0 \end{bmatrix} \bar{r}$$

$$= 1 \Rightarrow 100K = 1 \Rightarrow \underline{\underline{K = 0.01}}$$

2) If reachable then poles can be placed arbitrarily

Controllability matrix

$$\mathcal{C} = [B \ AB \ A^2B] = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}$$

full rank \Rightarrow reachable

$$3) G(p) = \begin{bmatrix} 0 & \frac{2}{p+1} \\ 0 & \frac{1}{p} \\ \frac{1}{p+1} & \frac{2}{p+1} \end{bmatrix}$$

a) Largest minors (order 2)

$$0, -\frac{1}{p(p+1)}, \frac{-2}{(p+1)^2}$$

Order 1 = elements of G

The pole polynomial is the least common divisor of all minors, i.e.

$$P(p) = p(p+1)^2$$

\therefore 3 poles: 0, -1, -1

The minors are consequently

$$\frac{0}{p(p)} \quad , \quad \frac{p+1}{p(p)} \quad , \quad \frac{2p}{p(p)}$$

Zero polynomial is the largest common divisor, which is 1 in this case

\Rightarrow No zeros

$$b) \quad y_1 = \frac{2}{p+1} u_2$$

$$y_2 = \frac{1}{p} u_2$$

$$y_3 = \frac{1}{p+1} u_1 + \frac{2}{p+1} u_2$$

Let, for example

$$x_1 = \frac{1}{p+1} u_2, \quad x_2 = \frac{1}{p} u_2, \quad x_3 = \frac{1}{p+1} u_1$$

$$\Rightarrow \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} = \begin{bmatrix} -1 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & -1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 0 & 1 \\ 0 & 1 \\ 1 & 0 \end{bmatrix} \begin{bmatrix} u_1 \\ u_2 \end{bmatrix}$$

$$\begin{bmatrix} y_1 \\ y_2 \\ y_3 \end{bmatrix} = \begin{bmatrix} 2 & 0 & 0 \\ 0 & 1 & 0 \\ 2 & 0 & 1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix}$$

Minimal since 3 states and 3 poles

(observable and controllable equivalently)

$$4) \text{ a) } y(k) = \frac{1}{q-0.5} u(k) \Rightarrow y(k+1) - 0.5y(k) = u(k)$$

$$\therefore x(k+1) = 0.5x(k) + u(k)$$

$$y(k) = x(k)$$

$$b) x_I(k) = \frac{1}{q-1} (r(k) - y(k))$$

$$\Rightarrow x_I(k+1) - x_I(k) = r(k) - x(k)$$

$$\therefore \underbrace{\begin{bmatrix} x(k+1) \\ x_I(k+1) \end{bmatrix}}_{x_e(k+1)} = \underbrace{\begin{bmatrix} 0.5 & 0 \\ -1 & 1 \end{bmatrix}}_{A_e} \underbrace{\begin{bmatrix} x(k) \\ x_I(k) \end{bmatrix}}_{x_e(k)} + \underbrace{\begin{bmatrix} 1 \\ 0 \end{bmatrix}}_{B_e} u(k) + \underbrace{\begin{bmatrix} 0 \\ 1 \end{bmatrix}}_{K_r} r(k)$$

$$\therefore y(k) = \underbrace{\begin{bmatrix} 1 & 0 \end{bmatrix}}_{C_e} x_e(k)$$

c) Optimal control law (Assume $r=0$ and $\lambda_e = [l_1, l_2]$)

$$\begin{aligned} u(k) &= -\lambda_e x_e(k) \\ &= -l_1 x(k) - l_2 x_I(k) \\ &= -l_1 y(k) + l_2 \frac{1}{q-1} y(k) \end{aligned}$$

Block scheme gives

$$u(k) = -\left(K_p + K_I \frac{1}{q-1}\right) y(k)$$

$$\therefore K_p = l_1 \text{ and } K_I = -l_2$$

d) The obtained LQ controller guarantees a stable system

$$5) G(s) = \frac{1}{s} e^{-0.6s}$$

Time delay $T_d = 0.6$, Sampling time $h=1$

We recognise this process as a pure integration ($'/(s)$) with a time delay, i.e.

$$a) \dot{x}(t) = ax(t) + bu^*(t), u^*(t) = u(t-0.6)$$

$$y(t) = cx(t)$$

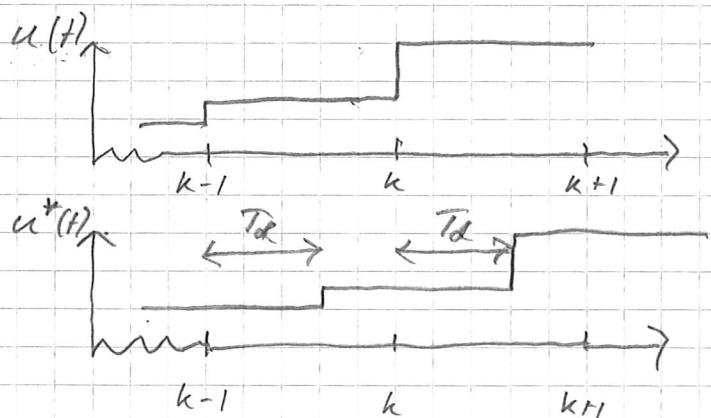
with $a=0, b=1, c=1$

b) Given the state x_0 at time t_0 the analytical solution is

$$x(t) = e^{a(t-t_0)} x(t_0) + \int_{t_0}^t e^{a(t-\tau)} Bu^*(\tau) d\tau$$

Let $t_0 = kh = k$ and $t = (k+1)h = k+1$.

Study input signal on the interval



Consequently, splitting the interval into two

$$x(k+1) = x(k) + \int_k^{k+T_d} Bu(k-1) d\tau + \int_{k+T_d}^{k+1} Bu(k) d\tau$$

$$x(k+1) = x(k) + T_d u(k-1) + (1-T_d)u(k)$$

Inroduce delayed input state

$$x_d(k) = u(k-1)$$

Then with $T_d = 0.6$

$$\begin{bmatrix} x(k+1) \\ x_d(k+1) \end{bmatrix} = \begin{bmatrix} 1 & 0.6 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} x(k) \\ x_d(k) \end{bmatrix} + \begin{bmatrix} 0.4 \\ 1 \end{bmatrix} u(k)$$

$$6) \quad x(u+1) = \underbrace{\begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix}}_A x(u) + \underbrace{\begin{bmatrix} 0 \\ 1 \end{bmatrix}}_B u(u) + v_1(u)$$

$$y(u) = \underbrace{\begin{bmatrix} 1 & 0 \end{bmatrix}}_C x(u) + v_2(u)$$

$$v_1 \sim \text{WGN} \left(\begin{bmatrix} 0 \\ 0 \end{bmatrix}, \begin{bmatrix} 1 & 0 \\ 0 & 2 \end{bmatrix} \right) \Rightarrow R_1 = \begin{bmatrix} 1 & 0 \\ 0 & 2 \end{bmatrix}$$

$$v_2 \sim \text{WGN}(0, 1) \Rightarrow R_2 = 1$$

$$E\{v_1 v_2^T\} = \begin{bmatrix} 0 \\ 0 \end{bmatrix} \Rightarrow R_{12} = \begin{bmatrix} 0 \\ 0 \end{bmatrix}$$

$$a) \quad O = \begin{bmatrix} C \\ CA \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}$$

$\text{rank } O = 2 \Rightarrow \text{observable}$

b) Standard Kalman (predictor) in ss:

$$\hat{x}_p(u+1) = A\hat{x}_p(u) + B u(u) + K(y(u) - C\hat{x}_p(u))$$

$$\text{where } \hat{x}_p(u) = \hat{x}(u|u-1)$$

$\text{Var}\{x_p(u)\} = P$ where P is given by

$$P = A P A^T + R_1 - K(C P A^T + R_{12}^T)$$

$$K = (A P C^T + R_{12}) / (R_2 + C P C^T)^{-1}$$

$$\begin{aligned} &= \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} P_{11} & P_{12} \\ P_{12} & P_{22} \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} \left(\begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} + \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} P_{11} & P_{12} \\ P_{12} & P_{22} \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} \right)^{-1} \\ &= \begin{bmatrix} P_{12} \\ 0 \end{bmatrix} \frac{1}{1 + P_{11}} \end{aligned}$$

$$\begin{bmatrix} p_{11} & p_{12} \\ p_{21} & p_{22} \end{bmatrix} = \underbrace{\begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} p_{11} & p_{12} \\ p_{12} & p_{22} \end{bmatrix} \begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix}}_{\begin{bmatrix} p_{22} & 0 \\ 0 & 0 \end{bmatrix}} + \begin{bmatrix} 1 & 0 \\ 0 & 2 \end{bmatrix}$$

$$= -\frac{1}{1+p_{11}} \underbrace{\begin{bmatrix} p_{12} \\ 0 \end{bmatrix} \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} p_{11} & p_{12} \\ p_{12} & p_{22} \end{bmatrix} \begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix}}_{\begin{bmatrix} p_{12}^2 & 0 \\ 0 & 0 \end{bmatrix}}$$

$$\begin{aligned} \Rightarrow p_{11} &= p_{22} + 1 - \frac{p_{12}^2}{1+p_{11}} \\ p_{12} &= 0 \\ p_{22} &= 2 \end{aligned} \quad \left. \begin{array}{l} \\ \\ \end{array} \right\} P = \begin{bmatrix} 3 & 0 \\ 0 & 2 \end{bmatrix} \Rightarrow K = \begin{bmatrix} 0 \\ 0 \end{bmatrix}$$

c) Filter case

$$(1) \hat{x}_f(k+1) = \hat{x}_p(k+1) + K_f(y(k)) - C\hat{x}_p(k)$$

where $\hat{x}_f(k) = \hat{x}(k|k)$ and

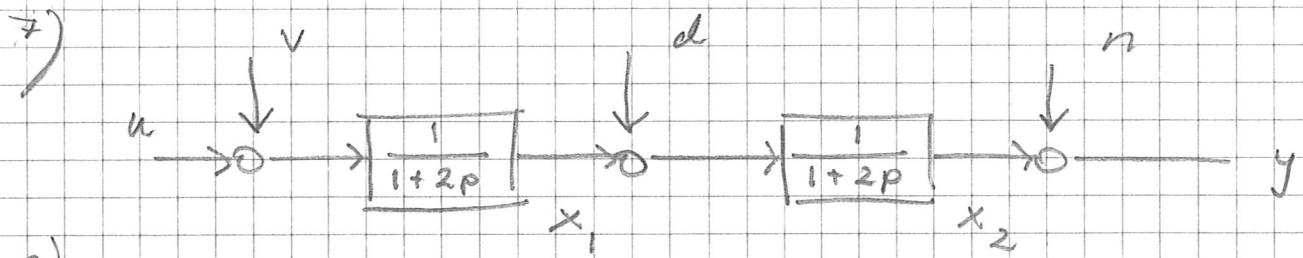
$$(2) \hat{x}_p(k+1) = A\hat{x}_f(k) + B u(k)$$

$$\begin{aligned} \Rightarrow K_f &= P C^T (C P C^T + R_2)^{-1} \\ &= \begin{bmatrix} 3 & 0 \\ 0 & 2 \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} \frac{1}{1+p_{11}} = \begin{bmatrix} 3/4 \\ 0 \end{bmatrix} \end{aligned}$$

$$\text{Var}\{\hat{x}_f\} = P_f = P - P C^T (C P C^T + R_2)^{-1} C P$$

$$= \begin{bmatrix} 3 & 0 \\ 0 & 2 \end{bmatrix} - \begin{bmatrix} 3/4 \\ 0 \end{bmatrix} \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} 3 & 0 \\ 0 & 2 \end{bmatrix} = \begin{bmatrix} 0.75 & 0 \\ 0 & 2 \end{bmatrix}$$

$\text{Var}\{\hat{x}_f\}$ unchanged but $\text{Var}\{\hat{x}\}$ reduced by 75%



a)

With x_1 and x_2 as above

$$x_1 = \frac{1}{1+2p} (u+v) \Leftrightarrow \dot{x}_1 = -\frac{1}{2} x_1 + \frac{1}{2} u + \frac{1}{2} v$$

$$x_2 = \frac{1}{1+2p} (x_1+d) \Leftrightarrow \dot{x}_2 = -\frac{1}{2} x_2 + \frac{1}{2} x_1 + \frac{1}{2} d$$

Spectral factorization

$$\frac{\Phi_d(\omega)}{d} = \frac{1}{\omega^2 + 1} \Leftrightarrow d(t) = \frac{1}{t^2 + 1} \nabla(t)$$

if $\nabla(t)$ white noise with unit intensity

$$\text{Let } d = x_3 \Rightarrow \dot{x}_3 = -x_3 + \nabla$$

$$\therefore \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} -0.5 & 0 & 0 \\ 0.5 & -0.5 & 0.5 \\ 0 & 0 & -1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 0.5 \\ 0 \\ 0 \end{bmatrix} u + \begin{bmatrix} 0.5 & 0 \\ 0 & 0 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} v \\ \nabla \end{bmatrix}$$

$$y = [0 \ 1 \ 0] \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + n(t)$$

where $e = \begin{bmatrix} v \\ \nabla \end{bmatrix}$ have intensity $\begin{bmatrix} \Phi_v & 0 \\ 0 & \Phi_\nabla \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}$

$$b) \quad \Phi_n = \frac{w^2 + 4}{w^2 + 9}$$

Spectral factorization again gives

$$n = \frac{P+2}{P+3} v_2, \quad v_2 \text{ white noise with unit intensity}$$

$$n = \frac{P+3}{P+3} v_2 - \underbrace{\frac{1}{P+3} v_2}_0 = v_2 - x_4$$

$$\Rightarrow x_4 = -3x_4 + v_2$$

$$\therefore \dot{x} = \begin{bmatrix} -0.5 & 0 & 0 & 0 \\ 0.5 & -0.5 & 0.5 & 0 \\ 0 & 0 & -1 & 0 \\ 0 & 0 & 0 & -3 \end{bmatrix} x + \begin{bmatrix} 0.5 \\ 0 \\ 0 \\ 0 \end{bmatrix} u + \begin{bmatrix} 0.5 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix} \begin{bmatrix} v \\ v \\ v \\ v_2 \end{bmatrix}$$

$$y = [0 \ 1 \ 0 \ -1] x + v_2$$

c) v, d, n independent \Rightarrow

v, D, v_2 — " — "

$$\Rightarrow R_1 = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}$$

$$R_{12} = \begin{bmatrix} \Phi_{vv_2} \\ \Phi_{vV_2} \\ \Phi_{V_2} \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}, \quad R_2 = \Phi_{V_2} = 1$$

$$\therefore R = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 1 \\ 0 & 0 & 1 & 1 \end{bmatrix}$$