

# Lecture notes: Linear control theory and state space design

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This lecture note is based on

- Chapter 6 of John Lygeros and Federico A. Ramponi, (2015). Lecture Notes on Linear System Theory.
- Karl Johan Aström Richard M. Murray, *Feedback Systems, An introduction to Scientists and Engineers*. Chapter 3. Cruise control example, Chap5, The Matrix Exponential.

$$\dot{x} = Ax$$

If there are control input

$$\dot{x} = Ax + Bu$$

- $u$ : input vector (can be multi-input).

# Reachability

let us include the input.

$$\dot{x} = Ax + Bu$$

**Reachable set:** the set of all points  $x_f$  such that given  $x(0)$  there exists  $u(t), t \in [0, T]$  that steers the system from  $x(0)$  to  $x(T) = x_f$ .

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**Reachability** A linear system is reachable if for **any**  $x_f \in R^n$  there exists a  $T > 0$  and  $u : [0, T] \rightarrow R^m$  such that the corresponding solution satisfies

$$x(0) = x_0, x(T) = x_f$$

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**Reachability** A linear system is reachable if for **any**  $x_f \in \mathbb{R}^n$  there exists a  $T > 0$  and  $u : [0, T] \rightarrow \mathbb{R}$  such that the corresponding solution satisfies

$$x(0) = x_0, x(T) = x_f$$

equivalent to: for  $x(0) = 0$ , the reachable set is  $\mathbb{R}^n$ .

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- How to decide if a LTI system is **reachable**?
- Given  $x_0$  and  $x_f$ , how to design the input  $u$  to **stabilize to**  $x_f$  under disturbances?

# Reachability matrix

Q: How to decide if a LTI system is **reachable**?

Under zero initial condition, the solution of  $\dot{x} = Ax + Bu$  is

$$x(t) = \int_0^t e^{A(t-\tau)} Bu(\tau) d\tau.$$

$$sX(s) - x(0) = \underbrace{AX(s)}_{\sim} + BU(s)$$

$$X(s) =$$

The solution under an impulse  $u(t) = \delta(t)$ :

$$x_{\delta}(t) = \int_0^t e^{A(t-\tau)} B \delta(\tau) d\tau = e^{At} B$$

$$u = \dot{\delta}(t) \quad x_{\dot{\delta}}(t) = A e^{At} B$$

$$u = \ddot{\delta}(t) \quad x_{\ddot{\delta}}(t) = A^2 e^{At} B$$

$$\vdots$$
$$u = \delta^{(n)}(t) \quad x_{\delta^{(n)}}(t) = A^{(n-1)} e^{At} B$$



particular input:  $u(t) = a_0 \delta(t) + a_1 \dot{\delta}(t) + \dots + a_{n-1} \delta^{(n-1)}(t)$

$$a_0 e^{At} B + a_1 A e^{At} B + a_2 A^2 e^{At} B + \dots + a_{n-1} A^{n-1} e^{At} B$$

$$t := T \quad : \quad \equiv X_f \quad \quad \underline{e^{AT}} = I$$

$$\lim_{t \rightarrow 0^+} ( \quad ) = a_0 B + a_1 A B + a_2 A^2 B + \dots + a_{n-1} A^{n-1} B$$

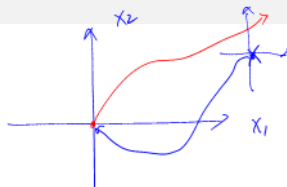
$$= X_f$$

$$\underbrace{[B \quad AB \quad \dots \quad A^{(n-1)} B]}_{\text{Full rank.}} \begin{bmatrix} a_0 \\ \vdots \\ a_{n-1} \end{bmatrix} = \underbrace{X_f}_{\in \mathbb{R}^n}$$

# Reachability rank condition

**Theorem** A linear system is **reachable** if and only if the reachability matrix  $[B, AB, \dots, A^{n-1}B]$  is **invertible**.

# Controllability



**Controllability** A linear system is **controllable** if for any  $x_0 \in R^n$  there exists a  $T > 0$  and  $u : [0, T] \rightarrow R$  such that the corresponding solution satisfies

$$x(T) = \underline{x_e}.$$

where  $x_e$  is the equilibrium (usually the origin.)

$$\begin{aligned} \dot{x} &= \underline{A}x + \underline{B}u \\ \underline{A}x &= 0 \\ x_e &= 0 \end{aligned}$$

# Reachability $\implies$ Controllability

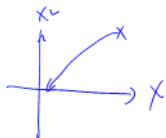
- The reachability implies the controllability.
- The controllability does not imply the reachability.

$$\begin{aligned}\dot{x}_1 &= x_1 + x_2 + u \\ u &= -x_2 - 2x_1 \\ \hline \dot{x}_1 &= -x_1\end{aligned}$$

ex.

$$\dot{x} = \begin{bmatrix} 1 & 1 \\ 0 & -1 \end{bmatrix} x + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u = \begin{matrix} x_1 + x_2 + u \\ -x_2 \end{matrix}$$

- Is the system reachable?  $\text{rank}[B \ AB] = \text{rank} \begin{bmatrix} 1 & 1 \\ 0 & 0 \end{bmatrix} = 1 < 2$
- Is the system controllable at the equilibrium of unforced system  $\dot{x} = Ax$ ?



$$\begin{aligned}\dot{x}_2 &= -x_2 \\ e^{-t} x_{2(0)} \\ \boxed{\dot{x}_1 = x_1 + u} & \quad \text{when } x_2 = 0.\end{aligned}$$

# Feedback control

Consider a system

$$\dot{x} = Ax + Bu$$

Objective: design the control input  $u$  such that the system eventually converge to the origin?

- Is the system controllable?  $\text{rank}(B) =$
- Consider a feedback control:  $u = Kx$ , selecting the value for  $K$  so that the system is stabilized to the origin.  $\rightarrow$  Feedback

$$\dot{x} = (A + BK)x$$

$\Delta$   
 $A'$  eigenvalues of  $A'$  have neg. real parts.

# Feedback control

exercise:

$$\dot{x} = \begin{bmatrix} 1 & 1 \\ 0 & -1 \end{bmatrix} x + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u$$

How to design the control  $u$  to stabilize the system to the origin?

# Error dynamics

Consider a system

$$\dot{x} = Ax + Bu$$

Objective: design the control input  $u$  such that the system eventually converge to **the goal**  $x_f$ ?

- Is the system reachable? *rank.*
- Will the same feedback control:  $u = Kx$  do that job?

① change of state variable:  $e = x - x_f$  error signal  
state

② design control  $u$  such that  $\lim_{t \rightarrow \infty} e(t) = 0$

$$\begin{aligned}\dot{e} &= \dot{x} - \dot{x}_f = \dot{x} = Ax + Bu \\ &= A(e + x_f) + Bu\end{aligned}$$

$$\dot{e} = Ae + Bv$$

goal:  $\lim_{t \rightarrow \infty} e(t) = 0$

$$v = \underset{\Delta}{K} e$$

$$\dot{e} = Ae + \underbrace{Bu + Ax_f}_{\text{stable}} \quad Bv \uparrow$$

input for the error dynamics

$$\underset{\checkmark}{B} \underset{\checkmark}{K} e = B \underset{\Delta}{u} + A \underset{\checkmark}{x_f} \quad \text{solve for } u.$$

$$\boxed{u = \underset{\uparrow}{K} x + \underset{\uparrow}{K_r} \cdot \underset{\Delta}{x_f}} \quad \begin{matrix} \uparrow \\ \text{gain} \end{matrix} \quad \begin{matrix} \uparrow \\ \text{reference signal} \end{matrix}$$

$$BKe = BKx + BK_r \cdot r + Ax_f$$

$$\cancel{BKx} - BKx_f = \cancel{BKx} + \underline{BK_r \cdot r} + Ax_f$$

$$\text{let } r = x_f$$

$$BK_r x_f + Ax_f + BKx_f = 0$$



$$\underline{[BK_r + A + BK]} x_f = 0$$

$$\underline{BK_r} + A + BK = 0 \rightarrow \text{zero matrix}$$

$$K_r = \cancel{-B^T(A+BK)}$$

$$= - \frac{1}{(A+BK)^T B}$$

# Error dynamics

Consider a system

$$\dot{x} = Ax + Bu$$

Objective: design the control input  $u$  such that the system eventually converge to **the goal**  $x_f$ ?

- Is the system reachable?
- Will the same feedback control:  $u = Kx$  do that job?

Introducing Error dynamics: Let  $e = x - x_f$  called the **error state**.

Q: How does the error evolves over time?

Goal: Design  $u$  so that the error converges to the zero.

# Error dynamics

$$\dot{x} = \begin{bmatrix} 1 & 1 \\ 0 & -1 \end{bmatrix} x + \begin{bmatrix} 1 \\ 1 \end{bmatrix} u$$

$$\begin{bmatrix} 1 \\ 1 \end{bmatrix} [k_1 \ k_2]$$

design a controller to reach the goal  $x_f = (10, 10)^T$ .

$$K = [k_1 \ k_2]$$

$$u = Kx + K_r x_f$$

$$A = \begin{bmatrix} 1 & 1 \\ 0 & -1 \end{bmatrix}$$

$$A + BK = \begin{bmatrix} 1 & 1 \\ 0 & -1 \end{bmatrix} + \begin{bmatrix} k_1 & k_2 \\ k_1 & k_2 \end{bmatrix}$$

$$= \begin{bmatrix} 1+k_1 & 1+k_2 \\ k_1 & -1+k_2 \end{bmatrix}$$

stable matrix

$$K_r = \frac{1}{(A+BK)^{-1}B}$$

# Trajectory tracking

Consider a system

$$\dot{x} = Ax + Bu$$



Objective: design the control input  $u$  such that the system eventually converge to a **desired trajectory**  $x^d(t)$ ?

① Feasible ② ... optimal ③ obstacle free

Similar to previous case, define  $e(t) = x(t) - x^d(t)$  and aims to drive the error to the origin.

$$\dot{x}^d = Ax^d + Bu^d$$

↑  
Feasible  
control input

Q: How does the error evolves over time?

$$\begin{aligned}\dot{e}(t) &= \dot{x}(t) - \dot{x}^d(t) \\ &= \underline{Ax} + Bu - \underline{Ax^d} - Bu^d\end{aligned}$$

$$\dot{e} = Ae + \underline{\underline{B(u - u^d)}}$$

$$\downarrow$$

$$Bv$$

$$v = ke$$

$$\dot{e} = \underbrace{(A+BK)}_{\text{stable}} e$$

$$u - u^d = v = ke$$

$$u(t) = ke(t) + u^d(t)$$

FB

FF (Feed-forward)  
Reference input.

# Trajectory tracking

$$\dot{x} = \begin{bmatrix} 1 & 1 \\ 0 & -1 \end{bmatrix} x + \begin{bmatrix} 1 \\ 1 \end{bmatrix} u$$

Design a controller so that the system track the trajectory of  $\underline{x}_1 = \sin t$  and  $\underline{x}_2 = \sin 2t$ .

$$x_1^d = \sin t : \quad \dot{x}_1^d = \cos t = x_1^d + x_2^d + u^d$$

$$\cos t = \sin t + \sin 2t + u^d$$

$$x_2^d = \sin 2t : \quad \dot{x}_2^d = 2 \cos 2t = -\sin 2t + u^d$$

Non-feasible traj.

Design feasible traj.

note: goal-reaching is a spectral case of trajectory tracking.

## Example: Feedback control, pole placement

$$\begin{bmatrix} \dot{x}_1(t) \\ \dot{x}_2(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -1 & 0 \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u(t)$$

Choose a feedback gain  $K$  such that the closed loop system is stable and has good control performance (pole placement.)



# Summary

- stability criteria of linear time invariant systems.
- reachability, controllability.
- Stabilization, goal-reaching, and trajectory tracking.