

# Lecture 18: Linear quadratic regulator

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# Outline

- continuous-time LQR problem
- dynamic programming solution
- infinite horizon LQR
- discrete-time LQR and nonlinear LQR.

The slide is partially adapted from Lecture notes by Prof. Stephen Boyd <sup>1</sup>.

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<sup>1</sup><http://stanford.edu/class/ee363/lectures/clqr.pdf>

# Introduction

Consider the linear time invariant (LTI) system

$$\dot{x} = Ax + Bu$$

Control objective: bring the system from an arbitrary initial state to the origin.

- A LTI system, if **stabilizable**, can always be stabilized by a linear feedback law  $u = Kx$ .
- If it is controllable, the close-loop poles can be placed ***anywhere*** on the S-plane. Thus the convergence can be made arbitrarily fast, which means ***large input amplitudes***.

E.g.  $\dot{x} = x + u$ , choosing  $u = -kx$  and  $k = 1.1, 1000, 10000$ .

# Formulating an optimization problem

Considerations:

- The bounded input amplitudes.
- The speed of convergence.

Naturally an optimization problem, let's see how it is formulated.

**Criterion 1: Fast convergence.**

$$\int_0^T x^T(t) Q x(t) dt$$

where  $Q$  is a ***nonnegative-definite symmetric*** matrix and  $T$  is the ***final time***.

- $x^T Q x$  : a measure of how much  $x$  deviates from the origin.
- $Q$ : weighting matrix.
- the integral is the cumulative deviation of  $x$  from the zero during the interval  $[0, T]$ .

# Formulating an optimization problem

Criterion 2: Bound the input amplitudes.

$$\int_0^T u^T(t) R u(t) dt$$

where  $R$  is a nonnegative definite symmetric matrix.

- $u^T R u$ : a measure of how much  $u$  deviates from the zero.
- To reduce the amplitudes we have the total value as small as possible.

Criterion 3: Converge to the origin. the terminal cost, if  $x(T) \neq 0$ .

$$x(T)^T Q_f x(T)$$

Remark: Weighting matrices  $Q, R, Q_f$  may need to be selected via trials and errors.

# Formulating an optimization problem

problem: Choosing  $u : [0, T] \rightarrow \mathbf{R}^m$  to minimize the total cost

$$J = \int_0^T \left( x(\tau)^T Q x(\tau) + u(\tau)^T R u(\tau) \right) d\tau + x(T)^T Q_f x(T)$$

- $T$  is time horizon.
- $Q = Q^T \geq 0$ ,  $Q_f = Q_f^T \geq 0$ , and  $R = R^T > 0$  are **state cost**, **final state cost**, and **input cost** matrices.

... an **infinite-dimensional problem**: (the input  $u : [0, T] \rightarrow \mathbf{R}^m$  is variable)

LQR Problem: linear quadratic regulator (regulating the controller).

## LQR solution via DP

we'll solve LQR problem using dynamic programming  
for  $0 \leq t \leq T$  we define the **value function**  $V_t : \mathbf{R}^n \rightarrow \mathbf{R}$  by

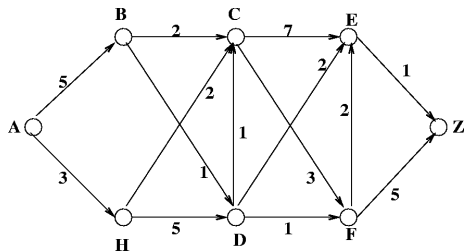
$$V_t(z) = \min_u \int_t^T x(\tau)^T Q x(\tau) + u(\tau)^T R u(\tau) d\tau + x(T)^T Q_f x(T)$$

subject to  $x(t) = z$ ,  $\dot{x} = Ax + Bu$ .

$V_t(z)$  = "minimal total cost for completing the task starting from state  $z$  at time  $t$ ".

- minimum is taken over all possible input signals  $u : [t, T] \rightarrow \mathbf{R}^m$
- $V_t(z)$  is also called **minimum LQR cost-to-go** from  $z$  at time  $t$ .
- $V_T(z) = z^T Q_f z$  where  $z = x(T)$  is the final state.

# A simple dynamic programming example





## LQR solution via DP

Let  $V_t$  be the optimal cost to go from time  $t$ .

fact

$V_t$  is quadratic, i.e.,  $V_t(z) = z^T P_t z$ , where  $P_t = P_t^T \geq 0$  (***symmetric, positive definite***).

What do we know about  $V_t$ ?

## LQR solution via DP

We start with  $x(t) = z$ , let's take  $u(t) = w \in \mathbf{R}^m$ , a constant, over the time interval  $[t, t+h]$ , where  $h > 0$  is small

The cost incurred over  $[t, t+h]$  is and the state is

## LQR solution via DP

Recall  $V_t = x(t)^T P_t x(t)$ , and  $P_{t+h} \approx P_t + h\dot{P}_t$ .

min-cost-to-go from where we land is approximately

$$V_{t+h}(x(t+h)) \approx V_{t+h}(z + h(Az + Bw))$$

cost plus incurred cost at  $t$  is approximately

# LQR solution via DP

To optimize

$$V_t(z) \approx \min_w \left[ z^T P_t z + h \left( z^T Q z + w^T R w + (Az + Bw)^T P_t z + z^T P_t (Az + Bw) + z^T \dot{P}_t z \right) \right]$$

minimize over  $w$  to get (approximately) optimal  $w$ :

## HJB equation

now let's substitute  $w^*$  into equation.

$$z^T P_t z = V_t(z) \approx z^T P_t z + \\ + h \left( z^T Q z + w^{*T} R w^* + (Az + Bw^*)^T P_t z + z^T P_t (Az + Bw^*) + z^T \dot{P}_t z \right)$$

after simplification,

$$-\dot{P}_t = A^T P_t + P_t A - P_t B R^{-1} B^T P_t + Q$$

which is the ***Riccati differential equation*** for the LQR problem  
we can solve it (numerically) using the ***final condition***  $P_T = Q_f$ .

# Summary of LQR for LTI system via DP

- 1 solve Riccati differential equation

$$-\dot{P}_t = A^T P_t + P_t A - P_t B R^{-1} B^T P_t + Q, \quad P_T = Q_f$$

(backward in time)

- 2 optimal  $u$  is  $u_{\text{lqr}}(t) = K_t x(t)$ ,  $K_t := -R^{-1} B^T P_t$

DP method readily extends to time-varying  $A$ ,  $B$ ,  $Q$ ,  $R$ , and tracking problem (define  $e(t) = x(t) - x_d(t)$  and drive the tracking error to 0.)

## Solve $P(t)$ backward in time

1. select a time step  $\delta t$  and total number of steps  $N = \frac{T}{\delta t}$ . By final condition  $P_N = P(T) = Q_f$ .
2.  $-\dot{P}_t = A^T P_t + P_t A - P_t B R^{-1} B^T P_t + Q$  in numerical computation via Euler approximation of  $\dot{P}_t$ .

## Steady-state regulator

$$-\dot{P}_t = A^T P_t + P_t A - P_t B R^{-1} B^T P_t + Q, \quad P_T = Q_f$$

Usually  $P_t$  rapidly converges as  $t$  decreases below  $T$ .

by convergence,  $\dot{P}_t = 0, P_t = P_{t+} = P$  as the limit of  $P_t$  which satisfies a

quadratic matrix equation — algebraic Riccati equation

- $P$  can be found by direct method or numerically integrating the Riccati differential equation. In matlab,  $[X, L, G] = \text{care}(A, B, Q, R, S, E)$  computes the unique solution  $X$  of the continuous-time algebraic Riccati equation.
- for  $t$  not close to horizon  $T$ , LQR optimal input is approximately a linear, constant state feedback

$$u(t) = K_{ss} x(t), \quad K_{ss} = -R^{-1} B^T P_{ss}$$



## Variation: Infinite horizon

The infinite horizon cost function

$$J = \int_0^{\infty} x(\tau)^T Q x(\tau) + u(\tau)^T R u(\tau) d\tau$$

The value function as

$$V(z) = \min_u \int_0^{\infty} x(\tau)^T Q x(\tau) + u(\tau)^T R u(\tau) d\tau$$

subject to  $x(0) = z$ ,  $\dot{x} = Ax + Bu$

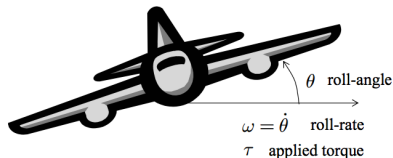
- we assume that  $(A, B)$  is controllable, so  $V$  is finite for all  $z$
- $V$  is quadratic and **time-invariant**:  $V(z) = z^T P z$ , where  $P = P^T \geq 0$
- the optimal  $u$  is  $u(t) = Kx(t)$ , where  $K = -R^{-1}B^T P$  (i.e., a constant linear state feedback)

For infinite horizon, in matlab,  $K = \text{lqr}(A, B, Q, R)$ .

# Remarks

- For finite-horizon  $[0, T]$  optimal control, need to solve  $P(t)$  backward in time and apply  $u(t) = -R^{-1}B^T P(t)x$  which is a ***time varying*** feedback control.
- For infinite-horizon  $[0, \infty)$  optimal control,  $P$  is a constant and  $u$  is a ***time-invariant*** control input.
- Require  $R > 0$ ,  $Q \geq 0$ , and there exists  $Q = H^T H$ .

## Example: Aircraft roll-dynamics <sup>2</sup>



$$\dot{\theta} = \omega$$

$$\dot{\omega} = -.875\omega - 20\tau$$

$$\dot{\tau} = -50\tau + 50u$$

$$x = [\theta \quad \omega \quad \tau]'$$

with

$$A = \begin{bmatrix} 0 & 1 & 0 \\ 0 & -.875 & -20 \\ 0 & 0 & -50 \end{bmatrix}, \quad B = \begin{bmatrix} 0 \\ 0 \\ 50 \end{bmatrix}.$$

Consider

$$J := \int_0^\infty \|x\|^2 + \rho \|u\|^2 dt,$$

where  $\rho$  is a positive constant, corresponding to  $Q = \text{diag}(1, 1)$ ;  $R = \rho$ .

<sup>2</sup>Hespanha, J. "Lecture notes on lqr/lqr controller design." notes, online note (2005).

## Example

The control is  $u := -Kx$ ,  $K := R^{-1}B^T P$  where  $P$  solves the algebraic Riccati equation

$$A^T P + PA - PBR^{-1}B^T P + Q = 0$$

# Discrete-time LQR

$$x_{t+1} = Ax_t + Bu_t$$

and objective function

$$J = \sum_{\tau=0}^{N-1} (x_{\tau}^T Q x_{\tau} + u_{\tau}^T R u_{\tau}) + x_N^T Q_f x_N$$

For  $t = 0, \dots, N$  define the value function

$$V_t(z) = \min_{u_t, \dots, u_{N-1}} \sum_{\tau=t}^{N-1} (x_{\tau}^T Q x_{\tau} + u_{\tau}^T R u_{\tau}) + x_N^T Q_f x_N$$

subject to  $x(t) = z$  and the dynamics.

- $V_t(z)$  is the minimum LQR cost-to-go, starting from  $z$  at time  $t$ ;
- $V_0(x_0)$  is the min cost (from state  $x_0$  at time 0.)

# Discrete-time LQR

- $V_t$  is quadratic, i.e.,  $V_t(z) = z^T P_t z$  where  $P_t = P_t^T \geq 0$ .
- $P_T = Q_f$  and can be computed backward in time and

$$P_t = Q + A^T P_{t+1} A - A^T P_{t+1} B (R + B^T P_{t+1} B)^{-1} B^T P_{t+1} A$$

# Discrete-time LQR

$$V_t = z^T Q z + \min_u (u^T R u + V_{t+1}(z_{t+1}))$$

since  $z_{t+1} = Az + Bu$ ,

$$V_t = z^T Q z + \min_u (u^T R u + V_{t+1}(Az + Bu))$$

is the dynamic programming, Bellman or Hamilton-Jacobi equation.

## Discrete-time LQR

$$V_t = z^T Qz + \min_u (u^T Ru + V_{t+1}(z_{t+1}))$$

since  $z_{t+1} = Az + Bu$ ,

$$V_t = z^T Qz + \min_u (u^T Ru + V_{t+1}(Az + Bu))$$

is the dynamic programming, Bellman or Hamilton-Jacobi equation.  
minimize  $u$  gives optimal input at time  $t$

$$u_t^{\text{lqr}} = \arg \min_u (u^T Ru + V_{t+1}(Az + Bu))$$

take derivative and since  $V_{t+1}(z) = z^T P_{t+1}z$

$$\frac{\partial (u^T Ru + (Az + Bu)^T P_{t+1}(Az + Bu))}{\partial u} = 0$$

$$u^* = -(R + B^T P_{t+1} B)^{-1} B^T P_{t+1} A z$$

put in  $V_t$  obtain the relation between  $P_t$  and  $P_{t+1}$



# Conclusion

- continuous and discrete LQR for time invariant systems.
- extension to for time varying systems is straightforward.

Further readings ...

- the dual problem: linear quadratic estimator and Kalman filter.
- linear quadratic gaussian (LQG) control.