Impedance 
$$Z_n(2) = \frac{\widetilde{V}_n(2)}{\widetilde{T}_n(2)} = Z_n \left[ \frac{1 + \widetilde{\Gamma}_n(2)}{1 - \widetilde{\Gamma}_n(2)} \right]$$

At the boundary:

$$\widetilde{V}_{o}(Z_{o_{i}}) = \widetilde{V}_{i}(Z_{o_{i}})$$
 and  $\widetilde{T}_{o}(Z_{o_{i}}) = \widetilde{T}_{i}(Z_{o_{i}})$ 

$$\tilde{V}_{i}(\tilde{z}_{0i}) = \tilde{V}_{i}^{\dagger} e^{-j\delta_{i}} \tilde{z}_{0i} (1 + \tilde{r}_{i}(\tilde{z}_{0i}))$$

$$\frac{\hat{T}_{o}(z_{o_{i}}) = \hat{v}_{o}^{+}}{\frac{2}{2}_{o}} e^{-3\delta_{o}z_{o_{i}}} (1 - \hat{v}_{o}(z_{o_{i}}))$$

$$\widetilde{T}_{i}(\overline{Z}_{0i}) = \frac{\widetilde{C}_{i}^{+}}{\overline{Z}_{i}^{+}} e^{-i\theta_{i} \overline{Z}_{0i}} (1 - \widetilde{\Gamma}_{i}(\overline{Z}_{0i}))$$

$$\widetilde{Z}_{o}(Z_{o_{i}}) = \widetilde{V}_{o}(Z_{o_{i}}) = \frac{\widetilde{V}_{o}^{\dagger} e^{-j\beta_{o} Z_{o_{i}}} (1 + \widetilde{\Gamma}_{o}(Z_{o_{i}}))}{\widetilde{V}_{o}^{\dagger} (Z_{o_{i}})} = Z_{o} \left[ \frac{1 + \widetilde{\Gamma}_{o}(Z_{o_{i}})}{1 - \widetilde{\Gamma}(Z_{o_{i}})} \right]$$

Since 
$$\widetilde{V}_0(z_{0i}) = \widetilde{V}_i(z_{0i})$$
 and  $\widetilde{T}_0(z_{0i}) = \widetilde{V}_i(z_{0i})$   
 $\widetilde{Z}_0(z_{0i}) = \widetilde{Z}_i(z_{0i})$ 

$$\mathcal{Z}_{0}\left[\frac{1+\widetilde{\Gamma_{0}}\left(2_{0}\right)}{1-\widetilde{F_{0}}\left(2_{0}\right)}\right]=\mathcal{Z}_{1}$$

$$\int_{0}^{\infty} (\bar{z}_{01}) = \frac{Z_{1}(z_{01}) - Z_{0}}{Z_{1}(z_{01}) + Z_{0}}$$

$$\tilde{V}_{0}(z) = \frac{\tilde{V}_{0}^{-}}{\tilde{V}_{0}^{+}} e^{2i\theta_{0}z} \rightarrow \tilde{V}_{0}(z) = \tilde{V}_{0}^{+} e^{2i\theta_{0}z_{01}} \left[ 1 + \frac{Z_{1}(z_{01}) - Z_{0}}{Z_{1}(Z_{01}) + Z_{0}} \right] \text{ Since } \tilde{V}_{0}^{+} \text{ is Known}$$

$$\widetilde{\Gamma}_{1}(z) = \widetilde{V}_{1}^{-} e^{2j\theta_{1}z}$$

$$\widetilde{V}_{1}^{+} e^{2j\theta_{1}z}$$

$$\widetilde{V}_{1}^{+}(z) = \widetilde{V}_{1}^{-} e^{2j\theta_{1}z_{12}} \left[ 1 + \frac{\mathcal{I}_{2}(z_{12}) + \mathcal{I}_{1}}{\mathcal{I}_{2}(z_{12}) - \mathcal{I}_{1}} \right]$$

$$V_{2}^{+}(z) = V_{2}^{-} e^{-\frac{2}{3}\delta_{2}t_{23}} + \frac{2}{2}(z_{23}) + 2$$

$$\frac{\overset{\sim}{V_0}(z)}{\overset{\sim}{V_0}(z)} e^{-j\beta_0 z_{01}} = \overset{\sim}{\Gamma_0}(z_{01}) \Rightarrow \frac{\overset{\sim}{V_0}(z)}{\overset{\sim}{V_0}(z)} = e^{j\beta_0 z_{01}} \left[ \frac{z_1(z_{01}) - \overline{z}_0}{\overline{z}_1(z_{01}) + \overline{z}_0} \right]$$

$$\frac{\tilde{V}_{1}(z)}{\tilde{V}_{1}^{+}(z)}e^{-j\beta_{1}z_{12}} = \tilde{\Gamma}_{1}(z_{12}) \Rightarrow \frac{\tilde{V}_{1}(z)}{\tilde{V}_{1}^{+}(z)} = e^{-j\beta_{1}z_{12}} \left[ \frac{z_{2}(z_{12}) - z_{1}}{z_{2}(z_{12}) + z_{1}} \right]$$

9.2.2.  
1. 
$$V(z,t)=V^{\dagger}[\cos(\omega t-Bz)+p\cos(\omega t+Bz)]$$
  
 $\cos(d\pm\beta)=\cos d\cos\beta$   $\mp \sinh d\sin\beta$ 

$$V(2,t) = V^{+}[\cos(\omega t)\cos(\beta z) + \sin(\omega t)\sin(\beta z) + p\cos(\omega t)\cos(\beta z)$$
  
 $V(2,t) = V^{+}[(1+p)\cos(\omega t)\cos(\beta z) + (1-p)\sin(\omega t)\sin(\beta z)]$   
 $V(2,t) = V^{+}[(1+p)\cos(\omega t)\cos(\beta z) + (1-p)\sin(\omega t)\sin(\beta z)]$ 

 $V(2t) = A\cos(\omega t)\cos(\beta z) + B\sin(\omega t)\sin(\beta z)$  where  $A = V^{+}(1+\rho)$  and  $B = V^{+}(1-\rho)$ 

3. This plot shows 2 appositely travelled waves which will produce a Standing wave. The VSWR can be calculated from the wave amplitudes to determine the impedance minimatch and describe how efficiently radio-frequency power is transmitted.