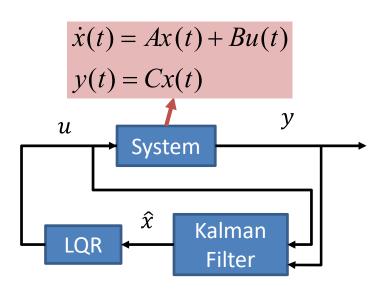
### **Full State Estimator**





#### Dynamics of estimated equation can be written as

$$\begin{split} \dot{\hat{x}} &= A\hat{x} + Bu + K_e(y - \hat{y}); \quad \hat{y} = C\hat{x} \\ \dot{\hat{x}} &= A\hat{x} + Bu + K_ey - K_eC\hat{x} \\ &= (A - K_eC)\hat{x} + \begin{bmatrix} B & K_e \end{bmatrix} \begin{bmatrix} u \\ y \end{bmatrix} \end{split}$$

is stable, then  $\hat{x}$  steadily converge to x.

Lets define an error,  $e = x - \hat{x}$ 

$$\dot{e} = Ax + Bu - A\hat{x} - Bu + K_e C\hat{x} - K_e y$$

$$= A(x - \hat{x}) + K_e C(\hat{x} - x)$$

$$= (A - K_e C)e$$

# Kalman Filter for LTI System

Consider a LTI System

$$\dot{x} = Ax + Bu + v(t)$$
$$y = Cx + w(t)$$

 $v(t) \rightarrow Gaussian input noise,$  $w(t) \rightarrow Gaussian Measurement Noise$ 

Assume

- 1. (A,C) is observable
- 2. w(t) and v(t) are independent white noises with the following properties

$$E[v(t)] = 0; E[w(t)] = 0$$
  

$$E[v(t)v^{T}(t)] = Q\partial(t - \tau), Q = Q^{T} \ge 0$$
  

$$E[w(t)w^{T}(t)] = R\partial(t - \tau), R = R^{T} \ge 0$$

The problem of Kalman Filter is to design a state feedback estimator to estimate the state x(t) by  $\hat{x}(t)$  such that the estimator error covariance is minimized, i.e., the following index is minimized

$$J_e = E[\{x(t) - \hat{x}(t)\}^T \{x(t) - \hat{x}(t)\}]$$

# **Construction of Steady State Kalman Filter**

Kanmal filter is a state observer with a specially selected observer gain (or Kalman filter gain). It has the dynamic equation:

$$\dot{\hat{x}} = A\hat{x} + Bu + k_e(y - \hat{y})$$

$$\dot{y} = c\hat{x}; \quad u = -k_e\hat{x};$$

With the Kalman filter gain  $K_e$  being given as

$$k_e = P_e C^T R^{-1}$$

where  $P_e$  is the positive definite of the following function Riccati Equation

$$P_{e}A^{T} + AP_{e} - P_{e}C^{T}R^{-1}CP_{e} + Q = 0$$

Let  $e = x - \hat{x}$ . The above Kalman Filter has the following properties:

$$\lim_{t\to\infty} E[e(t)] = \lim_{t\to\infty} E[x - \hat{x}] = 0$$

$$\lim_{t\to\infty} J_e = \lim_{t\to\infty} E[e^T e] = trace(P_e)$$

### Kalman Filter and LQR- they are dual

**LQR** 

$$K = R^{-1}B^{T}P$$

$$PA + A^{T}P - PBR^{-1}B^{T}P + Q = 0$$

$$J_{optimal} = trace(P)$$

#### Kalman Filter

$$K_e = P_e C^T R^{-1}$$
 
$$P_e A^T + A P_e - P_e C^T R^{-1} C P_e + Q = 0$$
 
$$J_{optimal} = trace(P_e)$$

$$A^{T} < -> A$$

$$B^{T} < -> C$$

$$K^{T} < -> K_{e}$$

$$P < -> P_{e}$$

### **Kalman Filter - Example**

Design a Kalman filter for the following aircraft system

$$A = \begin{bmatrix} -1.7 & 50 & 260 \\ 0.22 & -1.4 & -32 \\ 0 & 0 & -12 \end{bmatrix}; B = \begin{bmatrix} -272 \\ 0 \\ 14 \end{bmatrix}; F = \begin{bmatrix} 0.02 & 0.1 \\ -0.0035 & 0.04 \\ 0 & 0 \end{bmatrix};$$

$$C = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \end{bmatrix}; D = 0$$

Solution:

See the example 7.6 in "Modern Control Engineering" by Ashish Tewari.

### Linear Quadratic Gaussian (LQG) Control

It is often in control system design for a real life problem that one cannot measure all the state variables of the given plant. Thus, LQR, although it has a very impressive gain and phase margins, is impractical as it utilizes all state variables in the feedback, i.e., u = -Kx. In most practical situations, only partial information of the state of the given plant is accessible or can be measured for feedback. The natural questions one would ask:

- 1. Can we recover or estimate the state variables of the plant through the partially measurable information? The answer is yes. The solution is Kalman Filter.
- 2. Can we replace x the control law in LQR, i.e., u = -Kx, by the estimated state to carry out a meaningful control system design? The answer is yes. The solution is called LQG.
- 3. Do we still have impressive propeties associated with LQG? The answer is no. Any solution? Yes. Its called loop transfer recovery (LTR) technique.

## Solution of LQG Problem – Separation Principle

#### Step-1: Design an LQR control law u = -Kx which solves the following problem

$$\dot{x} = Ax + Bu \quad J(x, u, Q, R) = \int_{0}^{\infty} (x^{T}Qx + u^{T}Ru)dt \quad i.e., compute$$

$$PA + A^{T}P - PBR^{-1}B^{T}P + Q = 0, \ P > 0, \ K = R^{-1}B^{T}P$$

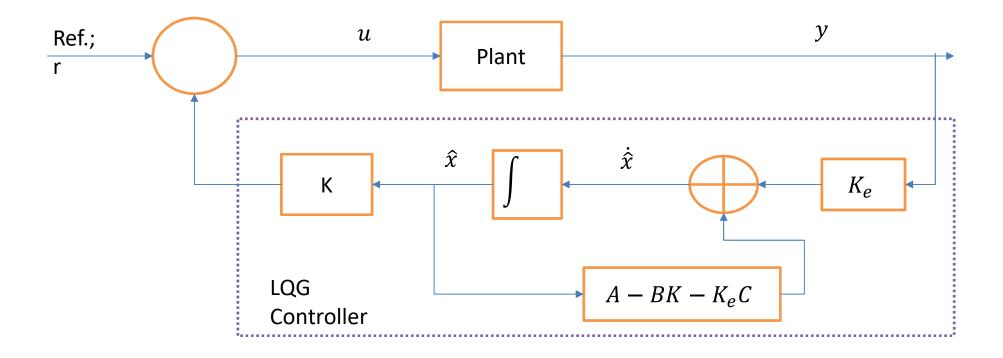
#### Step-2: Design a Kalman Filter for the given plant:

$$\begin{split} \dot{\hat{x}} &= A\hat{x} + Bu + K_e(y - \hat{y}) \\ \hat{y} &= c\hat{x} \quad where \quad K_e = P_eC^TR^{-1}, \quad P_eA^T + AP_e - P_eC^TR_e^{-1}CP_e + Q_e = 0, \quad P_e > 0 \end{split}$$

### Step-3: The LQG control law is given by $u=-K\widehat{x}$

$$\begin{vmatrix} \dot{\hat{x}} = A\hat{x} + Bu + K_e(y - c\hat{x}) \\ u = -K\hat{x} \end{vmatrix} \Rightarrow \begin{cases} \dot{\hat{x}} = (A - BK - K_eC)\hat{x} + K_ey \\ u = -K\hat{x} \end{cases}$$

# **Solution of LQG Problem**



Closed-loop dynamics of the given plant together with LQG controller:

$$\dot{\hat{x}} = (A - BK - K_eC)\hat{x} + K_ey$$

$$u = -K\hat{x} + r$$

### **Solution of LQG Problem**

Let Define an error variable  $e = x - \hat{x}$ 

$$\begin{split} \dot{e} &= \dot{x} - \dot{\hat{x}} \\ &= Ax + Bu + v(t) - (A - BK - K_e C)\hat{x} - K_e y \\ &= Ax + B(-K\hat{x} + r) + v(t) - (A - BK - K_e C)\hat{x} - K_e(Cx + w(t)) \\ &= Ax - BK\hat{x} + Br + v(t) - A\hat{x} + BK\hat{x} + K_eC\hat{x} - K_eCx - K_ew(t) \\ &= A(x - \hat{x}) - K_eC(x - \hat{x}) + Br + v(t) - K_ew(t) \\ &= (A - K_eC)e + Br + v(t) - K_ew(t) \end{split}$$

And

$$\dot{x} = Ax + Bu + v(t) = Ax - BK\hat{x} + Br + v(t)$$

$$= Ax - BK(x - e) + Br + v(t)$$

$$= (A - BK)x + BKe + Br + v(t)$$

Clearly, the above closed loop system is characterised by the following state-space equation

$$\begin{bmatrix} \dot{x} \\ \dot{e} \end{bmatrix} = \begin{bmatrix} A - BK & BK \\ 0 & A - K_e C \end{bmatrix} \begin{bmatrix} x \\ e \end{bmatrix} + \begin{bmatrix} B \\ B \end{bmatrix} r + \begin{bmatrix} v \\ v - K_e w \end{bmatrix}$$

$$y = \begin{bmatrix} c & 0 \end{bmatrix} \begin{bmatrix} x \\ e \end{bmatrix} + w$$
Kalman Filter Eigenvalues

LQR Eigenvalues

# **Complete Diagram of LQG Control**

