

# DSP & Digital Filters

Mike Brookes

## 1: Introduction

- Organization
- Signals
- Processing
- Syllabus
- Sequences
- Time Scaling
- z-Transform
- Region of Convergence
- z-Transform examples
- Rational z-Transforms
- Rational example
- Inverse z-Transform
- MATLAB routines
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- 18 lectures: feel free to ask questions

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- Exam + Formula Sheet (past exam papers + solutions on website)
- **Problems:** Mitra textbook contains many problems at the end of each chapter and also MATLAB exercises

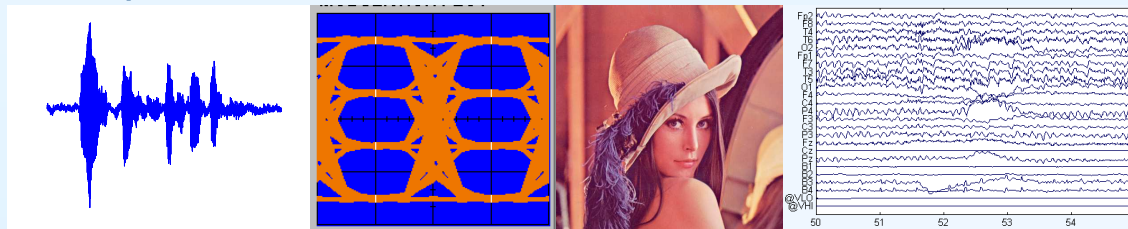
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- A signal is a numerical quantity that is a function of one or more independent variables such as time or position.

## Examples:



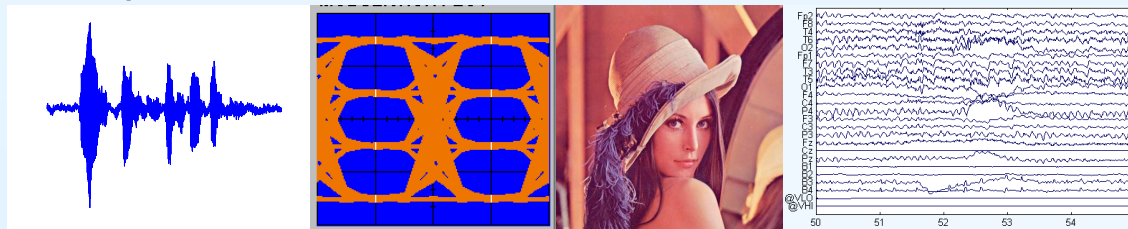
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- A signal is a numerical quantity that is a function of one or more independent variables such as time or position.
- Real-world signals are analog and vary continuously and take continuous values.

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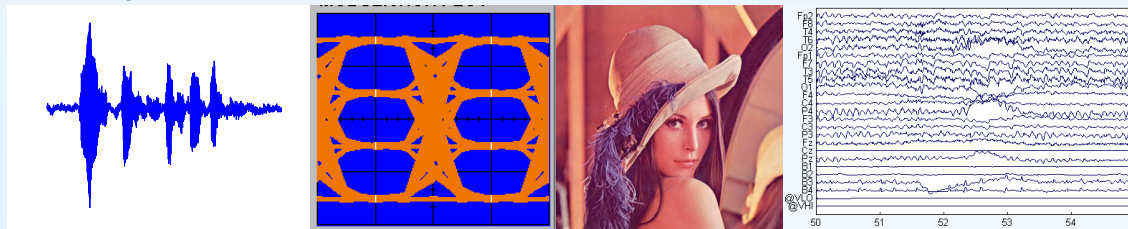
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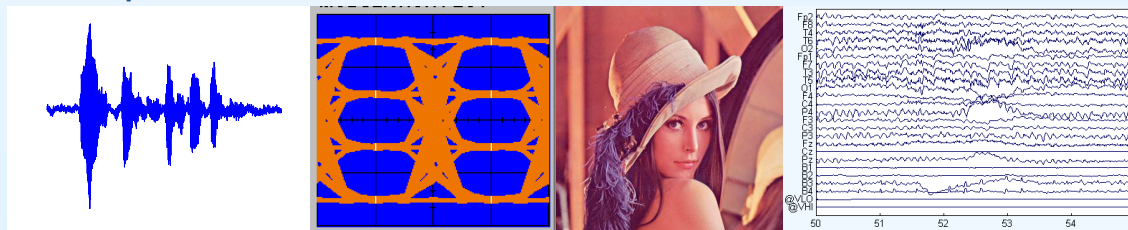
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## Examples:



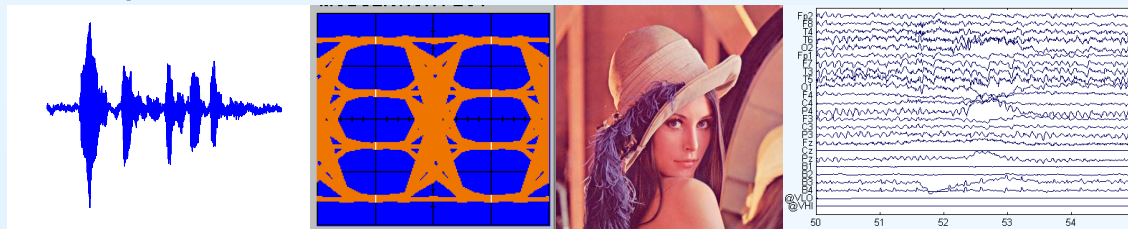
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- Digital signals are sampled at discrete times and are quantized to a finite number of discrete values
- We will mostly consider one-dimensionalal real-valued signals with regular sample instants; except in a few places, we will ignore the quantization.
  - Extension to multiple dimensions and complex-valued signals is mostly straightforward.

## Examples:



# Processing

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- Aims to “improve” a signal in some way or extract some information from it
- Examples:
  - Modulation/demodulation
  - Coding and decoding
  - Interference rejection and noise suppression
  - Signal detection, feature extraction
- We are concerned with linear, time-invariant processing

# Syllabus

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## Main topics:

- Introduction/Revision
- Transforms
- Discrete Time Systems
- Filter Design
  - FIR Filter Design
  - IIR Filter Design
- Multirate systems
  - Multirate Fundamentals
  - Multirate Filters
  - Subband processing



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- **Finite Energy:**  $\sum_{n=-\infty}^{\infty} |x[n]|^2 < \infty$
- **Absolutely Summable:**  $\sum_{n=-\infty}^{\infty} |x[n]| < \infty \Rightarrow \text{Finite energy}$

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Often easiest to scale time so that  $f_s = 1$  Hz. E.g. to design a 1 kHz low-pass filter for  $f_s = 44.1$  kHz we can design a 0.0227 Hz filter for  $f_s = 1$  Hz.

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Often easiest to scale time so that  $f_s = 1$  Hz. E.g. to design a 1 kHz low-pass filter for  $f_s = 44.1$  kHz we can design a 0.0227 Hz filter for  $f_s = 1$  Hz.

To scale back to real-world values: *Every quantity* of dimension (Time) <sup>$n$</sup>  is multiplied by  $T^n$  (or equivalently by  $f_s^{-n}$ ). Thus all *times* are multiplied by  $T$  and all *frequencies* and *angular frequencies* by  $T^{-1}$ .

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We use  $\Omega$  for “real” angular frequencies and  $\omega$  for normalized angular frequency. The units of  $\omega$  are “radians per sample”.



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**Warning:** Several MATLAB routines scale time so that  $f_s = 2$  Hz. Weird, non-standard and irritating.

# z-Transform

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The  $z$ -transform converts a sequence,  $\{x[n]\}$ , into a function,  $X(z)$ , of an arbitrary complex-valued variable  $z$ .

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- Definition:  $X(z) = \sum_{n=-\infty}^{+\infty} x[n]z^{-n}$

# Region of Convergence

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The set of  $z$  for which  $X(z)$  converges is its *Region of Convergence* (ROC).

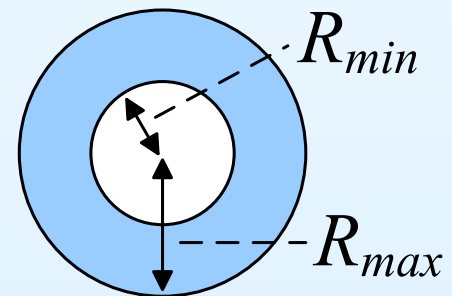
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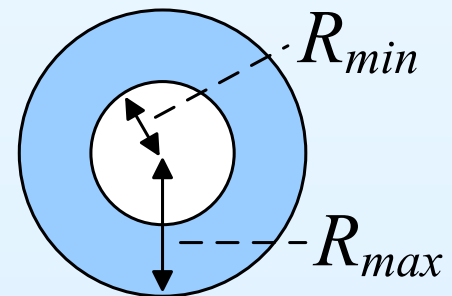
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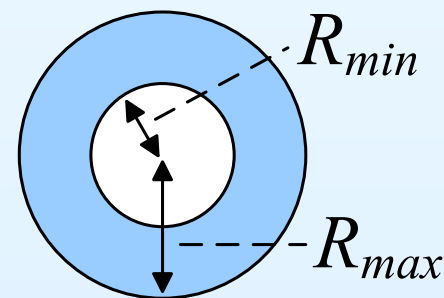
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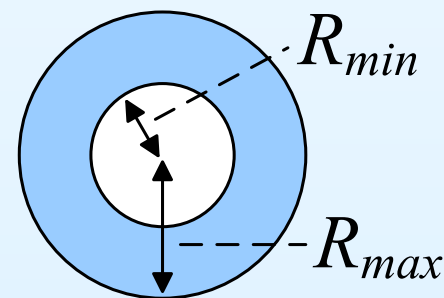
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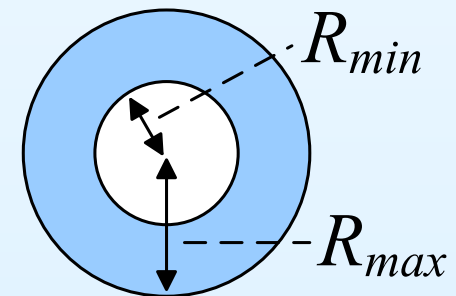
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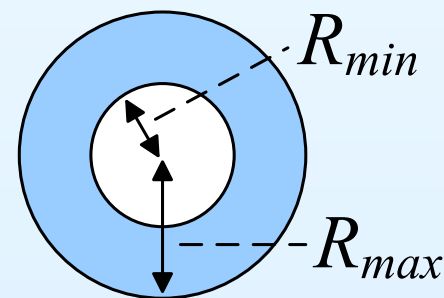
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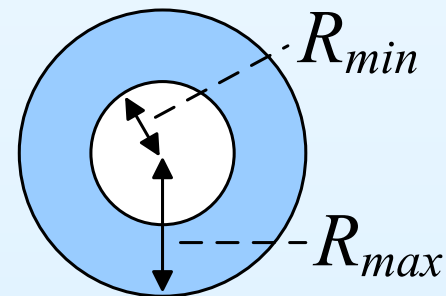
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# Region of Convergence

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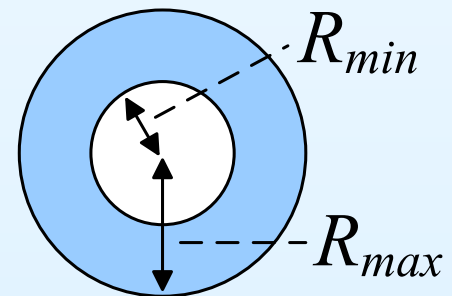
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  - **causal**  $\Leftrightarrow X(\infty)$  converges



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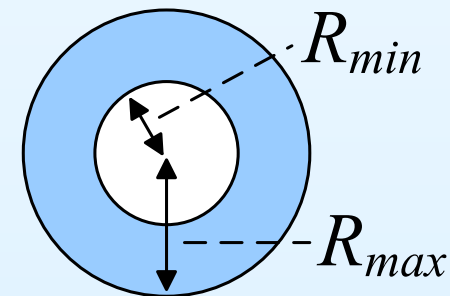
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- **left-sided**  $\Leftrightarrow R_{min} = 0$



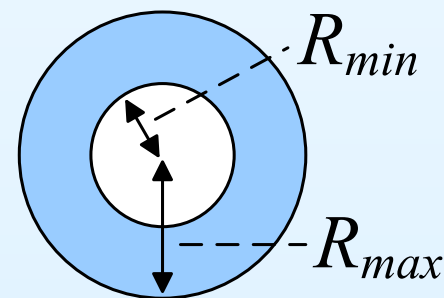
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## z-Transform examples

The sample at  $n = 0$  is indicated by an open circle.



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$$u[n] \quad \cdots \cdot \cdot \cdot \circ \uparrow \uparrow \uparrow \uparrow \cdots$$

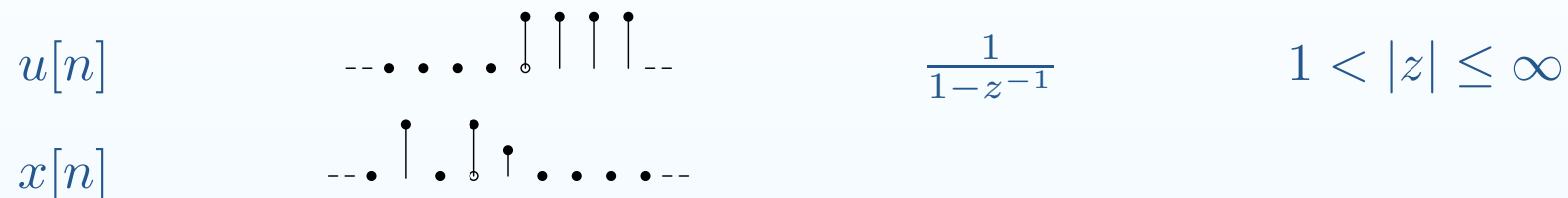
$$\frac{1}{1-z^{-1}}$$

$$1 < |z| \leq \infty$$

$$\text{Geometric Progression: } \sum_{n=q}^r \alpha^n z^{-n} = \frac{\alpha^q z^{-q} - \alpha^{r+1} z^{-r-1}}{1 - \alpha z^{-1}}$$

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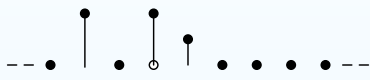
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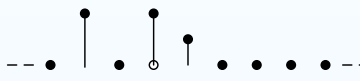
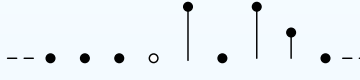
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$u[n]$		$\frac{1}{1-z^{-1}}$	$1 <  z  \leq \infty$
$x[n]$		$2z^2 + 2 + z^{-1}$	$0 <  z  < \infty$

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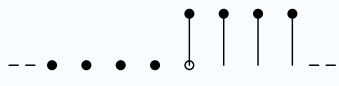

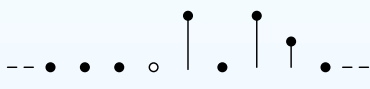

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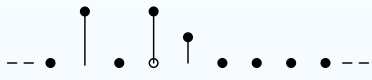
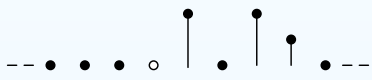


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$\alpha^n u[n]_{\alpha=0.8}$		$\frac{1}{1-\alpha z^{-1}}$	$\alpha <  z  \leq \infty$

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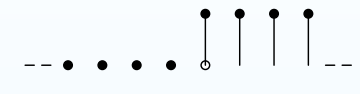
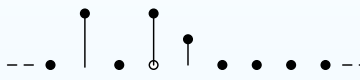


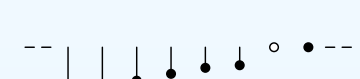
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$-\alpha^n u[-n-1]$			

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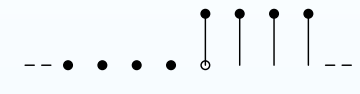
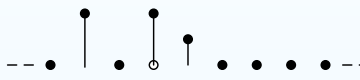


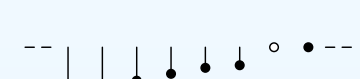
The sample at  $n = 0$  is indicated by an open circle.

$u[n]$		$\frac{1}{1-z^{-1}}$	$1 <  z  \leq \infty$
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$\alpha^n u[n]_{\alpha=0.8}$		$\frac{1}{1-\alpha z^{-1}}$	$\alpha <  z  \leq \infty$
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## z-Transform examples

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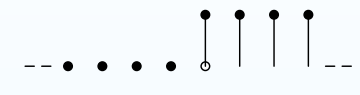
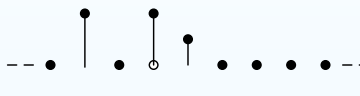
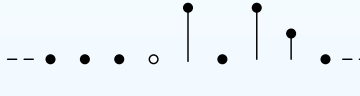



$u[n]$		$\frac{1}{1-z^{-1}}$	$1 <  z  \leq \infty$
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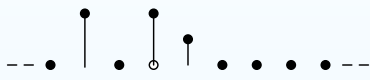
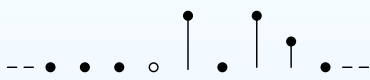



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## z-Transform examples

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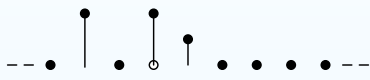
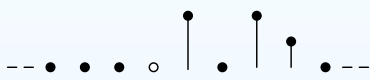




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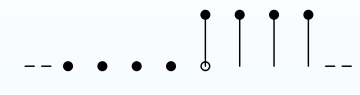
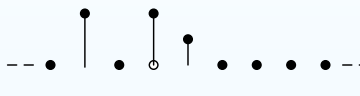
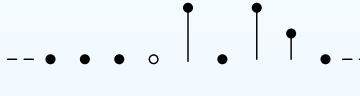



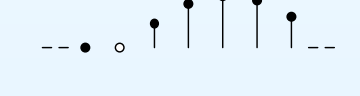
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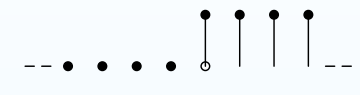
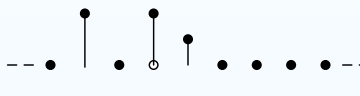
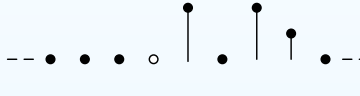



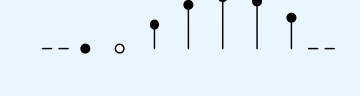
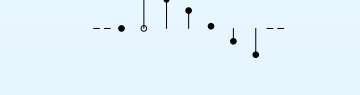
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Most  $z$ -transforms that we will meet are **rational polynomials** with real coefficients, usually one polynomial in  $z^{-1}$  divided by another.

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$$G(z) = g \frac{\prod_{m=1}^M (1 - z_m z^{-1})}{\prod_{k=1}^K (1 - p_k z^{-1})}$$

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**Note:** There are  $K - M$  zeros or  $M - K$  poles at  $z = 0$  (**easy to overlook**)

# Rational example

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$$G(z) = \frac{8-2z^{-1}}{4-4z^{-1}-3z^{-2}}$$

Poles/Zeros:  $G(z) = \frac{2z(z-0.25)}{(z+0.5)(z-1.5)}$

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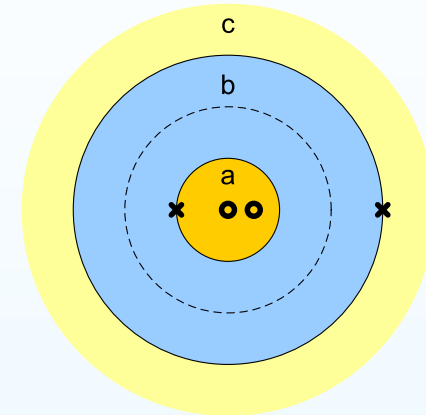
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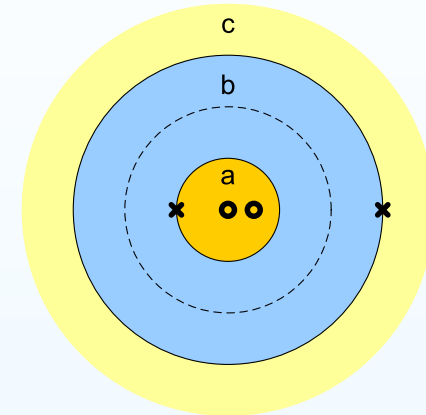
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Partial Fractions:  $G(z) = \frac{0.75}{1+0.5z^{-1}} + \frac{1.25}{1-1.5z^{-1}}$



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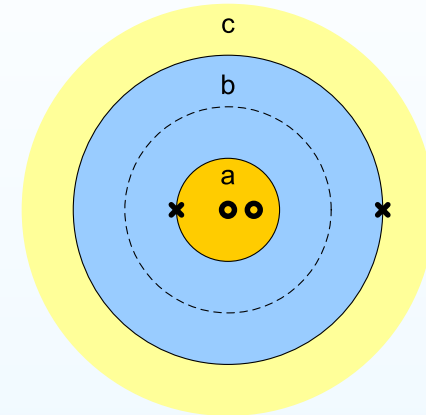
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Partial Fractions:  $G(z) = \frac{0.75}{1+0.5z^{-1}} + \frac{1.25}{1-1.5z^{-1}}$

ROC	ROC	$\frac{0.75}{1+0.5z^{-1}}$	$\frac{1.25}{1-1.5z^{-1}}$	$G(z)$
a	$0 \leq  z  < 0.5$			
b	$0.5 <  z  < 1.5$			
c	$1.5 <  z  \leq \infty$			

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# Inverse z-Transform

$g[n] = \frac{1}{2\pi j} \oint G(z) z^{n-1} dz$  where the integral is anti-clockwise around a circle within the ROC,  $z = Re^{j\theta}$ .

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**Proof:**

$$\frac{1}{2\pi j} \oint G(z) z^{n-1} dz = \frac{1}{2\pi j} \oint \left( \sum_{m=-\infty}^{\infty} g[m] z^{-m} \right) z^{n-1} dz$$



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$g[n] = \frac{1}{2\pi j} \oint G(z) z^{n-1} dz$  where the integral is anti-clockwise around a circle within the ROC,  $z = Re^{j\theta}$ .

**Proof:**

$$\begin{aligned} \frac{1}{2\pi j} \oint G(z) z^{n-1} dz &= \frac{1}{2\pi j} \oint \left( \sum_{m=-\infty}^{\infty} g[m] z^{-m} \right) z^{n-1} dz \\ &\stackrel{(i)}{=} \sum_{m=-\infty}^{\infty} g[m] \frac{1}{2\pi j} \oint z^{n-m-1} dz \end{aligned}$$

(i) depends on the circle with radius  $R$  lying within the ROC

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(ii) Cauchy's theorem:  $\frac{1}{2\pi j} \oint z^{k-1} dz = \delta[k]$  for  $z = Re^{j\theta}$  anti-clockwise.

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In practice use a combination of partial fractions and table of  $z$ -transforms.



# MATLAB routines

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tf2zp,zp2tf	$\frac{b(z^{-1})}{a(z^{-1})} \leftrightarrow \{z_m, p_k, g\}$
residuez	$\frac{b(z^{-1})}{a(z^{-1})} \rightarrow \sum_k \frac{r_k}{1-p_k z^{-1}}$
tf2sos,sos2tf	$\frac{b(z^{-1})}{a(z^{-1})} \leftrightarrow \prod_l \frac{b_{0,l}+b_{1,l}z^{-1}+b_{2,l}z^{-2}}{1+a_{1,l}z^{-1}+a_{2,l}z^{-2}}$
zp2sos,sos2zp	$\{z_m, p_k, g\} \leftrightarrow \prod_l \frac{b_{0,l}+b_{1,l}z^{-1}+b_{2,l}z^{-2}}{1+a_{1,l}z^{-1}+a_{2,l}z^{-2}}$
zp2ss,ss2zp	$\{z_m, p_k, g\} \leftrightarrow \begin{cases} x' = Ax + Bu \\ y = Cx + Du \end{cases}$
tf2ss,ss2tf	$\frac{b(z^{-1})}{a(z^{-1})} \leftrightarrow \begin{cases} x' = Ax + Bu \\ y = Cx + Du \end{cases}$

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- Time scaling: assume  $f_s = 1$  Hz so  $-\pi < \omega \leq \pi$

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- Time scaling: assume  $f_s = 1$  Hz so  $-\pi < \omega \leq \pi$
- z-transform:  $X(z) = \sum_{n=-\infty}^{+\infty} x[n]z^{-n}$

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- **ROC:**  $0 \leq R_{min} < |z| < R_{max} \leq \infty$

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  - **Causal:**  $\infty \in \text{ROC}$

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  - **Not unique** unless ROC is specified



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For further details see Mitra:1 & 6.