

CFA-4

Name: - Shaik Affan Azeeb

Roll No: 2022102054

Q1)

Given,

$$p(t) = \begin{cases} t/a & 0 \leq t \leq a \\ 1 & a \leq t \leq 1-a \\ (1-t)/a & 1-a \leq t \leq 1 \\ 0 & \text{else.} \end{cases}$$

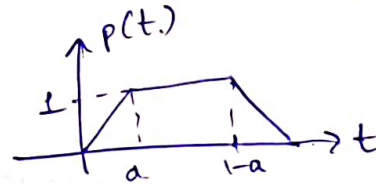
where $0 \leq a \leq \frac{1}{2}$

In 0 to $a \rightarrow$ straight line with slope $\frac{1}{a}$

In $a \leq t \leq 1-a \rightarrow$ constant line with amplitude $\underline{1}$.

In $1-a \leq t \leq 1 \rightarrow$ straight line with slope $= -\frac{1}{a}$ and y intercept as $\frac{1}{a}$

\Rightarrow plot =



\Rightarrow Now fourier transform :- $(p(t))$

$$p(f) = \int_{-\infty}^{\infty} p(t) e^{-j\omega t} dt$$

$$\Rightarrow p(f) = \int_0^a \frac{t}{a} e^{-j\omega t} dt + \int_a^{1-a} 1 e^{-j\omega t} dt + \int_{1-a}^1 \frac{1-t}{a} e^{-j\omega t} dt$$

$$= \frac{1}{a} \left[\frac{t e^{-j\omega t}}{-j\omega} - \frac{e^{-j\omega t}}{j^2 \omega^2} \right]_0^a + \frac{e^{-j\omega t}}{-j\omega} \Big|_a^{1-a} + \frac{1}{a} \left[\frac{(1-t) e^{-j\omega t}}{-j\omega} + \frac{e^{-j\omega t}}{(-j\omega)^2} \right]_{1-a}^1$$

$$= \frac{1}{a} \left[\frac{a e^{-j\omega a}}{-j\omega} + \frac{e^{-j\omega a}}{\omega^2} + \frac{1}{-\omega^2} \right] + \frac{e^{-j\omega(1-a)}}{-j\omega} + \frac{e^{-j\omega a}}{j\omega} + \frac{1}{a} \left[\frac{(-e^{-j\omega})}{\omega^2} - \frac{a e^{-j\omega(1-a)}}{-j\omega} + \frac{e^{-j\omega(1-a)}}{\omega^2} \right]$$

$$= \frac{1}{a} \left[\frac{a e^{-j\omega a}}{-j\omega} + \frac{e^{-j\omega a}}{\omega^2} - \frac{1}{\omega^2} \right] + \frac{e^{-j\omega(1-a)}}{-j\omega} + \frac{e^{-j\omega a}}{j\omega} - \frac{1}{a} \frac{e^{-j\omega}}{\omega^2} - \frac{1}{a} \left[\frac{a e^{-j\omega(1-a)}}{-j\omega} + \frac{e^{-j\omega(1-a)}}{-\omega^2} \right]$$

$$= \frac{\cancel{-e^{-j\omega a}}}{\cancel{j\omega}} + \frac{e^{-j\omega a}}{a\omega^2} - \frac{1}{a\omega^2} + \cancel{-\frac{e^{-j\omega(1-a)}}{j\omega}} + \frac{\cancel{e^{-j\omega a}}}{\cancel{j\omega}} - \frac{e^{-j\omega}}{a\omega^2} + \frac{\cancel{e^{-j\omega(1-a)}}}{\cancel{j\omega}} + \frac{e^{-j\omega(1-a)}}{a\omega^2}$$

$$= \frac{e^{-j\omega a}}{a\omega^2} - \frac{1}{a\omega^2} - \frac{e^{-j\omega}}{a\omega^2} + \frac{e^{-j\omega(1-a)}}{a\omega^2} \rightarrow \textcircled{1}$$

$$= \frac{e^{-j\omega a}}{a\omega^2} + \frac{e^{j\omega a}}{a\omega^2} + \frac{e^{-j\omega}}{a\omega^2} - \frac{e^{-j\omega}}{a\omega^2} - \frac{1}{a\omega^2}$$

we know that $\frac{e^{j\theta} + e^{-j\theta}}{2} = \cos\theta$, $\frac{e^{j\theta} - e^{-j\theta}}{2j} = \sin\theta$

on keeping $\omega = 2\pi f$

$$\Rightarrow \frac{e^{-j2\pi f a}}{a\omega^2} - \frac{1}{a\omega^2} - \frac{e^{-j2\pi f}}{a\omega^2} + \frac{e^{-j\omega(1-a)}}{a\omega^2}$$

Now taking $e^{-j\pi f}$ common from all terms

$$\rightarrow \frac{e^{-j\pi f}}{a\omega^2} \left[e^{-j2\pi f a} e^{j\pi f} - e^{j\pi f} - e^{-j\pi f} + e^{-j2\pi f(1-a)} e^{j\pi f} \right]$$

$$= e^{-j\pi f} \cdot (1-a) \operatorname{sinc}(1-a) \operatorname{sinc}(af)$$

$$\text{as } \frac{e^{j\pi f(1-a)} - e^{-j\pi f(1-a)}}{2j}$$

upon solving we get:-

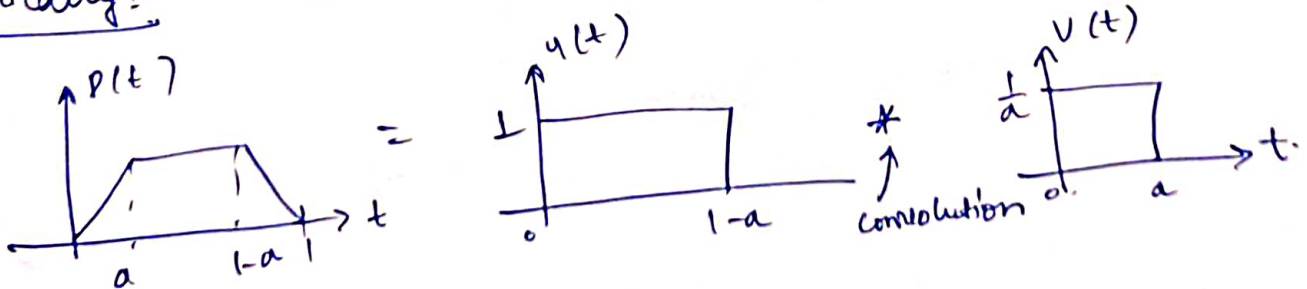
$$\frac{e^{-j\pi f} \left(\frac{e^{j\pi f(1-a)} - e^{-j\pi f(1-a)}}{2j} \right) \left(\frac{e^{j\pi fa} - e^{-j\pi fa}}{2j} \right)}{a 4\pi^2 f^2}$$

$$\text{as } \frac{e^{j\pi f(1-a)} - e^{-j\pi f(1-a)}}{2j} = \sin(\pi f(1-a)) \quad \left\langle \frac{e^{j\pi a} - e^{-j\pi a}}{2j} = \sin \pi a \right\rangle$$

$$\therefore = e^{-j\pi f} (1-a) \operatorname{sinc}(1-a) \operatorname{sinc}(af)$$

$$\Rightarrow \boxed{P(f) = (1-a) \operatorname{sinc}(1-a) \operatorname{sinc}(af) e^{-j\pi f}}$$

Graphically:-



$$P(t) = u(t) * v(t)$$

$$\Rightarrow \boxed{P(f) = U(f) \cdot V(f)}$$

$$\circ u(f) = (1-a) \operatorname{sinc}((1-a)f)$$

$$\text{and } v(f) = \operatorname{sinc}(af)$$

Now convolution will give centered version of $p(t)$

$$\text{with } F \cdot \tilde{f} = (1-a) \operatorname{sinc}((1-a)f) \operatorname{sinc}(af)$$

But $p(t)$ is delayed by $\frac{1}{2}$ from its centered version $\tilde{p}(t)$

\Rightarrow we get a phase of $e^{-j\pi f}$ in freq domain.

$$\circ P(f) = \tilde{p}(f) e^{-j\pi f} = (1-a) \operatorname{sinc}((1-a)f) \operatorname{sinc}(af) e^{-j\pi f}$$

b) Average symbol energy of equiprobable 4-PAM is

$$\frac{\sigma^2}{b} = \frac{(-1)^2 + (-3)^2 + 1^2 + 3^2}{4} = 5$$

$$\Rightarrow \frac{\sigma^2}{b} = 5, T=1$$

PSD [Power spectral density for modulated signal]

$$S_{xx}(f) = \frac{\sigma^2}{bT} |P(f)|^2$$

$$\Rightarrow S_u(f) = 5 \left| (1-a) \operatorname{sinc}((1-a)f) \operatorname{sinc}(af) \right|^2$$

$$S_u(f) = 5(1-a)^2 \operatorname{sinc}^2((1-a)f) \operatorname{sinc}^2(af)$$

③ 95% of B.W = B satisfies the eqⁿ: dropping constants from PSD :-

$$\int_{-\frac{B}{2}}^{\frac{B}{2}} |P(f)|^2 df = \frac{95}{100} \int_{-\infty}^{\infty} |P(f)|^2 df$$

Pulse Energy is given by :-

$$\int_{-\infty}^{\infty} |P(f)|^2 df = \int_{-\infty}^{\infty} |P(t)|^2 dt \quad \text{Parseval's thm}$$

$$\Rightarrow 2 \int_0^a \left(\frac{t}{a}\right)^2 dt + \int_a^{1-a} dt$$

$$= \frac{2}{a^2} \left(\frac{a^3}{3}\right) + 1 - 2a$$

$$= 1 - 2a + \frac{2a}{3} = \boxed{1 - \frac{4a}{3}}$$

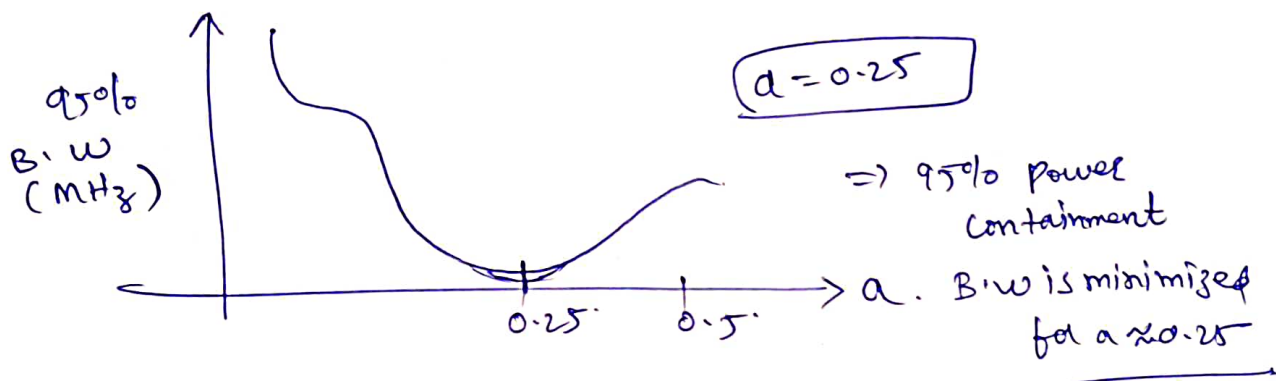
$$\Rightarrow \int_{-\infty}^{\infty} |P(f)|^2 df = 1 - \frac{4a}{3}$$

Now using symmetry of $|P(f)|^2$.

$$\begin{aligned} \int_0^{B/2} |P(f)|^2 df &= 2 \int_0^{B/2} (1-a)^2 \operatorname{sinc}^2((1-a)f) \operatorname{sinc}^2(af) df \\ &= 0.95 \left(1 - \frac{4a}{3}\right) \end{aligned}$$

$$\Rightarrow \int_0^{B/2} \text{sinc}^2((1-a)f) \text{sinc}^2 af \, df = \frac{0.95(1-\frac{4a}{3})}{2(1-a)^2}$$

Now on solving through matlab we get $a=0.25$



Unit time is 100ps

So unit B.W = 10 MHz.

\Rightarrow we plotted 10B with a . (10B MHz vs a)

Q2) Given,

Linear modulation with signalling pulse

$$p(t) = \text{sinc}(at) \text{sinc}(bt)$$

a) we have to find a & b so that $p(t)$ is Nyquist with 50% excess B.W for data rate of 40 Mbps using 16QAM.

we know: multiplication in time domain gives convolution in freqdomain

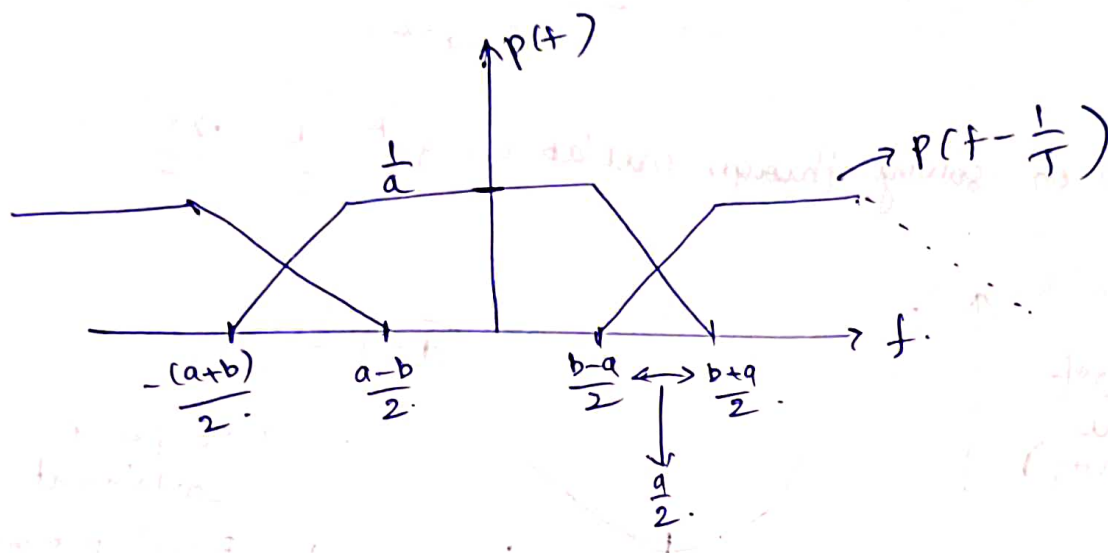
$$\Rightarrow p_1(t) = \text{sinc}(at), p_2(t) = \text{sinc}(bt)$$

$$P(f) = P_1(f) * P_2(f)$$

which results in trapezoid

$$\boxed{P \cdot f_0}$$

assuming $a \leq b$



$$R_b = \frac{\text{bit rate}}{\text{no of bits}} \quad (\text{symbol rate})$$

$$\text{No of bits of 16QAM} = \log_2 16 = 4$$

$$\text{Bit rate} = 40 \text{ Mbps} \quad R_b = \frac{40 \text{ Mbps}}{4} = 10 \text{ Mbps.}$$

$$\text{Hence } a = 10 \text{ MHz}$$

$p(t)$ has 50% extra B.W.

$$\Rightarrow B_T = \frac{(a+b)}{2} = \frac{R_b}{2} \left(1 + \frac{50}{100}\right) = \frac{R_b}{2} \left(\frac{3}{2}\right)$$

$$\Rightarrow 10 + b = 15 \Rightarrow \boxed{b = 5} \Rightarrow \boxed{b = 5 \text{ MHz}}$$

$\frac{a}{2} = \frac{1}{2T}$. < pulse has zeros at integer multiple of T , so we minimise RSS and get this

$$\Rightarrow \frac{1}{2T} (1 + \alpha) = \frac{a+b}{2}$$

$$\boxed{b = \frac{\alpha}{T}}$$

$$\frac{1}{T} = 10 \text{ M symbols/sec.}$$

$$\Rightarrow \frac{a}{2} = \frac{1}{2} \times 10 \Rightarrow \boxed{a = 10 \text{ MHz}}$$

$$b = \frac{\alpha}{T} = \frac{1}{2} \times 10 = \boxed{5 \text{ MHz}}$$

$$\alpha = 50\% = \frac{1}{2}$$

$$\Rightarrow \boxed{\text{occupied B.W.} = 15 \text{ MHz}}$$

⑥ Given,

Data rate for 16QAM = 40 Mbps.

" " " 8PSK = 18 Mbps

No of symbols = $\log_2^{16} = 4$

\Rightarrow symbol rate for 16QAM = $\frac{1}{T_1} = \frac{40}{\log_2^{16}} = 10^6 \text{ symbols/sec}$

\Rightarrow No of symbols = $\log_2^8 = 3$

\downarrow
symbol rate for 8PSK = $\frac{1}{T_2} = \frac{18}{\log_2^8} = 6 \text{ M symbols/sec.}$

So the conditions should be

$$a = 10, b = 10$$

$$a = 10, b = 6$$

$$a = 6, b = 10$$

$$a = 6, b = 6$$

we have to choose for $b > a$

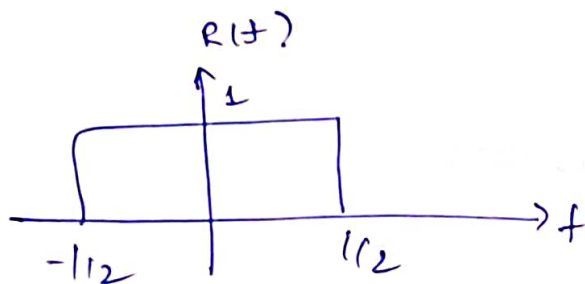
So $a = 6, b = 10$ is the best option.

Q3) Given,

$$R(f) = \mathcal{F}\left[\frac{1}{2} \left(\frac{1}{2}\right)\right](f)$$

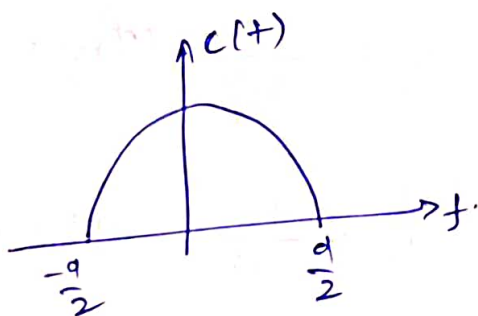
(a) $R(f) \mathcal{F}\{C(f)\}$ for $0 < a < 1$

$$R(f) = \mathcal{F}\left[\frac{1}{2} \left(\frac{1}{2}\right)\right](f)$$



$$C(f) = \frac{\pi}{2a} \cos\left(\frac{\pi f}{a}\right) \mathcal{F}\left[\frac{1}{2} \left(\frac{1}{2}\right)\right]$$

$C(f)$ is a cosine function with amp $\frac{\pi}{2a}$ from $-\frac{a}{2}$ to $\frac{a}{2}$



(b) we have to show Freq domain raised cosine pulse $S(f)$

$$S(f) = R(f) * C(f)$$

$S = R * C$ is symmetric as R, C are symmetric

$$S(f) = \int R(f-v) C(v) dv$$

$$= R(v+f) C(v) dv$$

using symmetry \perp

$$f \leq \left(\frac{1-q}{2}\right)$$

$$R(v-f)=1, dv > 0$$

$$\Rightarrow S(f) = \int_{-\frac{a}{2}}^{\frac{a}{2}} (dv) dv = 1$$

for $\frac{1-q}{2} < f < \frac{1+q}{2}$

$$S(f) = \int_{f-\frac{a}{2}}^{\frac{a}{2}} dv$$

$$= \frac{\frac{\pi}{2a}}{\frac{\pi}{a}} \left[\sin\left(\frac{\pi v}{a}\right) \right]_{f-\frac{1}{2}}^{\frac{a}{2}}$$

$$= \frac{1}{2} \left[1 - \sin\left(\frac{\pi}{a} \left(f - \frac{1}{2}\right)\right) \right]$$

lets check it is raised cosine or not, -

$$f = \frac{\pi}{a} \left(f - \left(\frac{1-q}{2}\right)\right)$$

$$\frac{\pi}{a} \left(f - \frac{1}{2}\right) = f - \frac{\pi}{2}$$

for $\frac{1-q}{2} < f < \frac{1+q}{2}$, $0 < f < \frac{\pi}{2}$

$$S(f) = \frac{1}{2} (1 - \sin(f - \frac{\pi}{2})) =$$

$$= \frac{1}{2} (1 + \cos f)$$

$$\text{for } F > (\frac{1+f}{2}) \quad S(f) = 0$$

$$\textcircled{c} \quad S(f) = \lambda(f) \lambda(f)$$

$$\lambda(f) = \int \lambda(t) e^{j2\pi ft} dt$$

$$= \frac{\pi}{2a} \int_{-\frac{a}{2}}^{\frac{a}{2}} \frac{1}{2} (e^{j\frac{\pi t}{a}} + e^{-j\frac{\pi t}{a}}) e^{j2\pi ft} dt$$

$$= \frac{\pi}{4a} \left[\frac{e^{j(\frac{\pi}{a} + 2\pi f)t}}{j(\frac{\pi}{a} + 2\pi f)} + \frac{e^{j(-\frac{\pi}{a} + 2\pi f)t}}{j(-\frac{\pi}{a} + 2\pi f)} \right]_{t=-\frac{a}{2}}^{\frac{a}{2}}$$

$$= \frac{\cos \pi a f}{1 - 4a^2 f^2}$$

$$\text{Time domain} = \lambda(f) = \sin(\pi f) \left(\frac{\cos \pi a f}{1 - 4a^2 f^2} \right)$$

Raised cosine L -

$S(f)$ decays as $1/f^3$

$$g(t) = s(t/T)$$

$$x(t) = \sum_n b(n) g(t - nT)$$

$$= \sum_n b(n) s\left(\frac{t}{T} - n\right)$$

So by finding t we realize that

$s\left(\frac{t}{T} - n\right)$ decays roughly as $\frac{1}{T^3}$.

The convergence of $\sum_n \frac{1}{n^3}$ implies that sum is the

precting eqⁿ.

∴ $s(t/T)$ is used for BPSK signalling at rate $1/T$
then magnitude of the transmitted waveform is always finite.

$$g(t) = s(t/T)$$

$$x(t) = \sum_n b(n) g(t - nT)$$

$$= \sum_n b(n) s\left(\frac{t}{T} - n\right)$$

so by finding t we realize that

$s\left(\frac{t}{T} - n\right)$ decays roughly as $\frac{1}{n^3}$.

The convergence of $\sum_n \frac{1}{n^3}$ implies that sum is finite

proving eqⁿ.

∴ $s(t/T)$ is used for BPSK signalling at rate $1/T$.
then magnitude of the transmitted waveform is always finite.