

ECC, Probability and LDPC

CoCoNut, 2016
Emmanuela Orsini

Previously

[8, 3, 5] Linear Code over $GF(3)$

$$G = \begin{bmatrix} 1 & 0 & 0 & 1 & 1 & 0 & 2 & 2 \\ 0 & 1 & 0 & 2 & 1 & 2 & 2 & 0 \\ 0 & 0 & 1 & 0 & 2 & 1 & 2 & 2 \end{bmatrix}$$

$$H = \begin{bmatrix} 1 & 0 & 0 & 0 & 0 & 1 & 2 & 2 \\ 0 & 1 & 0 & 0 & 0 & 0 & 1 & 2 \\ 0 & 0 & 1 & 0 & 0 & 2 & 1 & 2 \\ 0 & 0 & 0 & 1 & 0 & 1 & 1 & 0 \\ 0 & 0 & 0 & 0 & 1 & 2 & 2 & 2 \end{bmatrix}$$

(0 0 0 0 0 2 0 0)	(0 0 0 0 2 0 1 0)	(0 0 1 0 0 2 0 0)	(0 2 0 0 1 0 0 0)	(0 0 0 1 2 0 0 0)
(2 0 1 2 1)	(2 1 1 1 1)	(2 0 2 2 1)	(0 2 0 0 1)	(0 0 0 1 2)
(0 0 0 0 0 0 2 0)	(0 0 0 0 0 1 0 1)	(0 0 0 0 0 1 1 0)	(1 0 0 0 1 0 0 0)	(0 0 0 1 0 0 1 0)
(1 2 2 2 1)	(0 2 1 1 1)	(0 1 0 2 1)	(1 0 0 0 1)	(2 1 1 2 2)
(0 0 0 0 0 0 0 2)	(0 0 0 2 0 2 0 0)	(0 0 1 0 0 0 2 0)	(2 0 0 0 1 0 0 0)	(0 0 0 0 1 2 0 0)
(1 1 1 0 1)	(2 0 1 1 1)	(1 2 0 2 1)	(2 0 0 0 1)	(2 0 1 2 2)
(0 0 0 0 1 0 0 0)	(0 0 0 2 0 0 2 0)	(0 0 2 0 0 2 0 0)	(2 0 0 0 0 0 1 0)	(0 0 0 0 0 2 0 2)
(0 0 0 0 1)	(1 2 0 1 1)	(2 0 0 2 1)	(1 1 1 1 2)	(0 1 2 2 2)
(0 0 0 0 0 0 1 0)	(0 0 0 0 2 1 0 0)	(0 0 0 2 1 0 0 0)	(1 0 0 0 0 0 1 0)	(0 0 0 0 1 0 2 0)
(2 1 1 1 2)	(1 0 2 1 1)	(0 0 0 2 1)	(0 1 1 1 2)	(1 2 2 2 2)
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(1 0 2 1 2)	(1 0 0 1 1)	(2 1 1 0 1)	(2 2 1 1 2)	(2 2 2 2 2)
(0 0 0 0 0 0 0 1)	(0 0 0 1 1 0 0 0)	(2 0 0 0 0 0 0 2)	(0 0 2 0 0 1 0 0)	(0 0 0 1 0 1 0 0)
(2 2 2 0 2)	(0 0 0 1 1)	(0 1 1 0 1)	(1 0 1 1 2)	(1 0 2 2 2)
(0 0 0 0 2 0 0 0)	(0 0 0 2 0 0 0 2)	(0 1 0 0 0 0 0 2)	(0 2 0 0 0 0 1 0)	(0 0 0 0 0 0 2 2)
(0 0 0 0 2)	(1 1 1 2 1)	(1 2 1 0 1)	(2 0 1 1 2)	(2 0 0 2 2)
(0 0 0 1 0 0 0 0)	(0 1 0 0 0 2 0 0)	(0 2 0 0 0 0 0 2)	(0 1 0 0 0 1 0 0)	(0 0 0 2 2 0 0 0)
(0 0 0 1 0)	(2 1 1 2 1)	(1 0 1 0 1)	(1 1 2 1 2)	(0 0 0 2 2)
(0 0 0 2 0 0 0 0)	(0 0 2 0 0 0 2 0)	(0 0 0 1 0 2 0 0)	(0 0 1 0 0 0 1 0)	(0 0 0 0 1 0 0 2)
(0 0 0 2 0)	(1 2 1 2 1)	(2 0 1 0 1)	(2 1 2 1 2)	(1 1 1 0 2)
(0 0 1 0 0 0 0 0)	(0 2 0 0 0 2 0 0)	(0 0 1 0 1 0 0 0)	(0 2 0 0 0 1 0 0)	(0 0 0 2 0 0 1 0)
(0 0 1 0 0)	(2 2 1 2 1)	(0 0 1 0 1)	(1 2 2 1 2)	(2 1 1 0 2)
(0 0 2 0 0 0 0 0)	(2 0 0 0 0 2 0 0)	(0 0 1 0 0 0 0 2)	(0 0 0 1 0 0 0 1)	(0 0 2 0 0 0 0 1)
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(1 0 0 0 0 0 0 0)	(1 0 0 0 0 0 2 0)	(0 0 2 0 1 0 0 0)	(0 0 2 0 0 0 1 0)	(2 0 0 0 0 0 0 1)
(1 0 0 0 0)	(2 2 2 2 1)	(0 0 2 0 1)	(2 1 0 1 2)	(1 2 2 0 2)
(2 0 0 0 0 0 0 0)	(2 0 0 0 0 0 2 0)	(0 0 2 0 0 0 0 2)	(0 0 0 0 0 2 2 0)	(1 0 0 0 0 0 0 1)
(2 0 0 0 0)	(0 2 2 2 1)	(1 1 0 0 1)	(0 2 0 1 2)	(0 2 2 0 2)
(0 0 0 1 0 0 0 2)	(0 1 0 0 0 0 2 0)	(0 1 0 0 1 0 0 0)	(0 0 1 0 0 1 0 0)	(0 0 0 2 0 1 0 0)
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```
r:=Vector(GF(3), [1,1,0,1,0,2,0,2]);
r*Transpose(H);
(1 2 2 0 2)
```

Consider the usual $[7, 4]_2$ Hamming code with parity-check matrix H , and suppose $\mathbf{r} = 1011000$ is the received word. Then the associated syndrome is

$$\mathbf{s} = \begin{pmatrix} 1 & 0 & 1 & 0 & 1 & 0 & 1 \\ 0 & 1 & 1 & 0 & 1 & 1 & 0 \\ 0 & 0 & 0 & 1 & 1 & 1 & 1 \end{pmatrix} \cdot \begin{pmatrix} 1 \\ 0 \\ 1 \\ 1 \\ 0 \\ 0 \\ 0 \end{pmatrix} = \begin{pmatrix} 0 \\ 1 \\ 1 \end{pmatrix}.$$

The decoding procedure for binary Hamming code is as follows:
Suppose a single error occurred in the i th component, then $\mathbf{s} = H\mathbf{e}_i = \mathbf{h}_i$, where \mathbf{h}_i is the i th column of H .

- Compute the syndrome $\mathbf{s} = H\mathbf{r}$;
- Find the column \mathbf{h}_i of H that matches the syndrome;
- Complement the i th bit of the received word.

The simplest way to add redundancy to a transmission is to repeat each symbol of the message a fixed number of times.

- $\mathcal{A} = \{a_1, \dots, a_q\}$
- A codeword is given by a symbol of \mathcal{A} repeated n times

$$C_q^{rep}(n) = \{\underbrace{a_1 \dots a_1}_n, \underbrace{a_2 \dots a_2}_n, \dots, \underbrace{a_q \dots a_q}_n\},$$

- Minimum distance? Length? Dimension? Rate?
- Decoding with a majority vote strategy (it can be proved equivalent to the MLD)

This lecture

A communication scheme

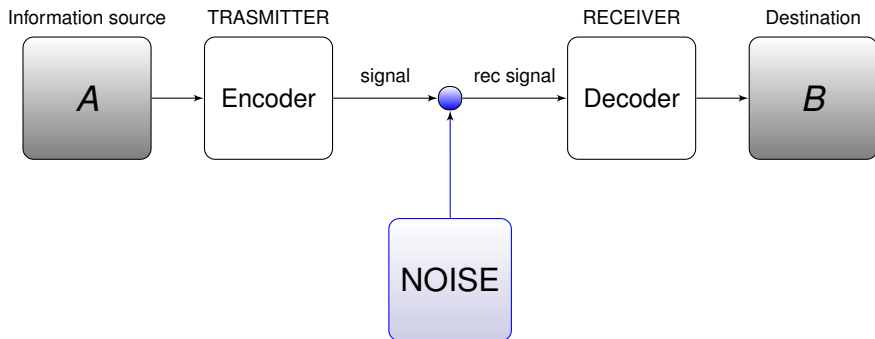


Figure: Communication channel model with noise

A communication scheme

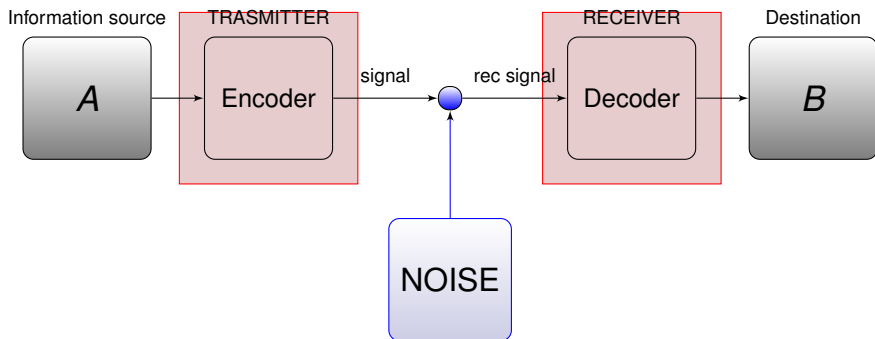
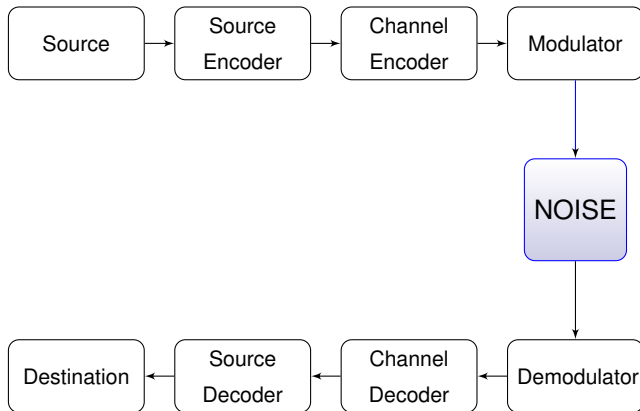
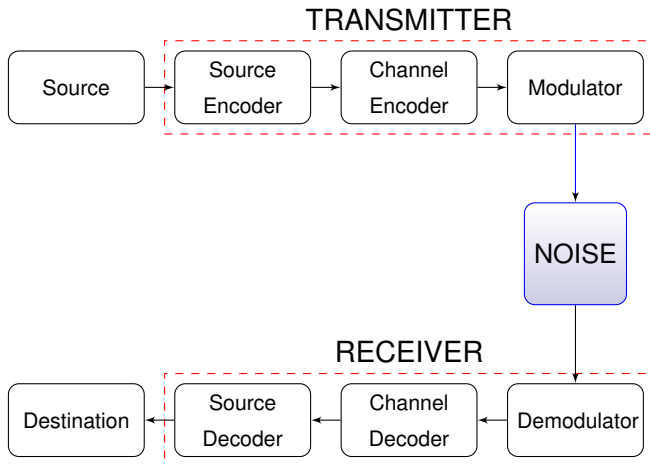


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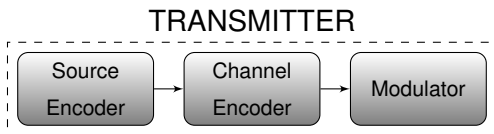
A more detailed communication scheme



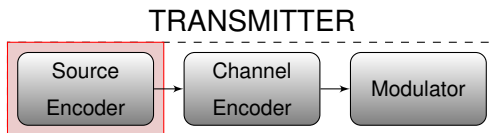
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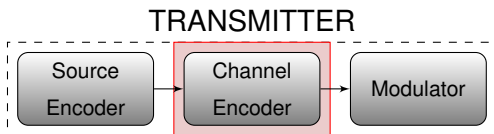


A more detailed communication scheme



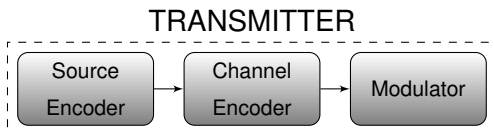
- *Source encoder*: it represents the source message in a compact way by removing unnecessary content (**data compression**)

A more detailed communication scheme



- *Source encoder*: it represents the source message in a compact way by removing unnecessary content (**data compression**)
- *Channel encoder*: it outputs a “channel codeword”, introducing a systematic redundancy to tolerate errors that might be introduced by the channel (*error correcting encoding*).

A more detailed communication scheme

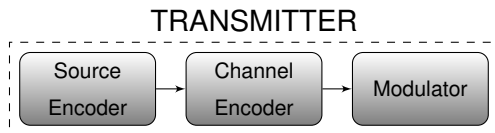


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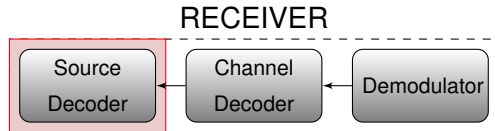


- *Channel decoder*: reverse operation of channel encoder

A more detailed communication scheme



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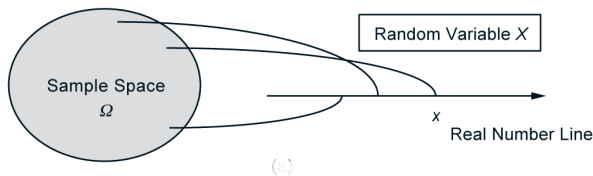
- *Channel decoder*: reverse operation of channel encoder
- *Source decoder*: reverse operation of source encoder

- Procedures for data compression
- How much can we compress? Is there a limit?
- How much redundancy is necessary to minimize the error probability in decoding?
- Is it true that to minimize the error probability we need to add a lot of redundancy, or we can achieve the same result more efficiently?

A **random variable** is a real-valued function of the experimental outcome:

$$X : \Omega \rightarrow E \subseteq \mathbb{R}.$$

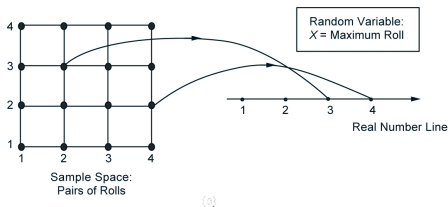
Visualization of a random variable: it is a function that assigns a numerical value to each possible outcome of the experiment



Discrete random variable: if the set of possible values is finite or at most countably finite.

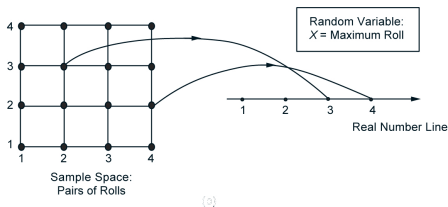
Random variables II

Example Consider two rolls of a 4-sided dice. Let X be the maximum of the two rolls. If the outcome is $(4, 2)$ then the value of X is 4, if the outcome is $(2, 3)$, the value of X is 3.



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- Let X be a discrete random variable, the *probability mass function* (probability for short) of X , denoted by p_X is the probability of the event $\{X = x\}$ consisting of all outcomes that give rise to a value of X equal to x :

$$p_X(x) = \Pr(X = x).$$

In many experiments, more than a single random variable is involved. For example, to calculate probabilities involving two random variables X and Y we need the **joint probability** of X and Y .

Definition

Let X and Y be two discrete random variables associated with the same experiment. The **joint probability** of X and Y is defined as

$$p_{X,Y}(x, y) = \Pr(X = x, Y = y),$$

for all pairs of numerical values (x, y) that X and Y can take.

- Let E be the set consisting of all (x, y) values, then

$$\Pr((X, Y) \in E) = \sum_{(x,y) \in E} p_{X,Y}(x, y).$$

- The **marginal** probability mass function of X and Y , denoted $p_X(x)$ and $p_Y(y)$ are given by

$$p_X(x) = \sum_y p_{X,Y}(x, y) \quad p_Y(y) = \sum_x p_{X,Y}(x, y)$$

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Example. The input source to a noisy channel is a r. v. X over a, b, c, d . The output for this channel is a r. v. Y over the same alphabet. The joint probability of these two random variables is:

$X \backslash Y$	a	b	c	d	$\Pr_X(x)$
a	$1/8$	$1/16$	$1/16$	$1/4$	$1/2$
b	$1/16$	$1/8$	$1/16$	0	$1/4$
c	$1/32$	$1/32$	$1/16$	0	$1/8$
d	$1/32$	$1/32$	$1/16$	0	$1/8$
$\Pr_Y(y)$	$1/4$	$1/4$	$1/4$	$1/4$	1

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$\Pr_Y(y)$	$1/4$	$1/4$	$1/4$	$1/4$	1

$X \backslash Y$	y_1	\dots	y_l	$\Pr_X(x)$
x_1	$p_{X,Y}(x_1, y_1)$	\dots	$p_{X,Y}(x_1, y_l)$	$p_X(x_1)$
x_2	$p_{X,Y}(x_2, y_1)$	\dots	$p_{X,Y}(x_2, y_l)$	$p_X(x_2)$
\vdots	\vdots	\vdots	\vdots	\vdots
x_h	$p_{X,Y}(x_h, y_1)$	\dots	$p_{X,Y}(x_h, y_l)$	$p_X(x_h)$
$\Pr_Y(y)$	$p_Y(y_1)$	\dots	$p_Y(y_n)$	1

- The **marginal** probability mass function of X and Y , denoted $p_X(x)$ and $p_Y(y)$ are given by

$$p_X(x) = \sum_y p_{X,Y}(x, y) \quad p_Y(y) = \sum_x p_{X,Y}(x, y)$$

Example. The input source to a noisy channel is a r. v. X over a, b, c, d . The output for this channel is a r. v. Y over the same alphabet. The joint probability of these two random variables is:

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$\Pr_Y(y)$	1/4	1/4	1/4	1/4	1

X \ Y	y_1	...	y_l	$\Pr_X(x)$
x_1	$p_{X,Y}(x_1, y_1)$...	$p_{X,Y}(x_1, y_l)$	$p_X(x_1)$
x_2	$p_{X,Y}(x_2, y_1)$...	$p_{X,Y}(x_2, y_l)$	$p_X(x_2)$
...
x_h	$p_{X,Y}(x_h, y_1)$...	$p_{X,Y}(x_h, y_l)$	$p_X(x_h)$
$\Pr_Y(y)$	$p_Y(y_1)$...	$p_Y(y_n)$	1

$$\Pr_Y(Y = a) = \Pr_{X,Y}(X = a, Y = a) +$$

$$\Pr_{X,Y}(X = b, Y = a) +$$

$$\Pr_{X,Y}(X = c, Y = a) +$$

$$\Pr_{X,Y}(X = d, Y = a) = 1/4$$

- Conditional probability:

$$p_{X|Y}(x, y) = \Pr(X = x | Y = y).$$

$$p_{X|Y}(x, y) = \frac{\Pr(X = x, Y = y)}{\Pr(Y = y)} = \frac{p_{X,Y}(x, y)}{p_Y(y)}.$$

- Multiplication rule:

$$p_{X,Y}(x, y) = p_{X|Y}(x, y) \cdot p_Y(y) \quad p_X(x) > 0$$

and

$$p_{X,Y}(x, y) = p_{Y|X}(x, y) \cdot p_X(x) \quad \text{if } p_Y > 0.$$

- Two random variables X and Y are **independent** if

$$p_{X|Y}(x, y) = p_X(x).$$

In this case we also have that

$$p_{Y|X}(x, y) = p_Y(y).$$

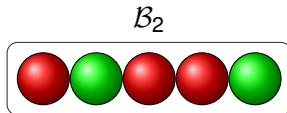
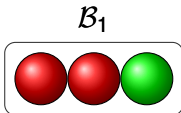
Problem 1 We choose one box at random and extract from it a ball. We want to compute the probability that the extracted ball is green.

Solution:

- X the random value that takes values in $\{1, 2\}$
- Y the random value that takes values $\{g, r\}$

We need to compute

$$\Pr(Y = g) = \Pr(X = 1) \cdot \Pr(Y = g|X = 1) + \Pr(X = 2) \cdot \Pr(Y = g|X = 2)$$



- **Sum rule:** We can rewrite the marginal probabilities: Given X and Y two random variables, and E_X , E_Y the set of all values x and y respectively

$$\begin{aligned}Pr_X(x) &= \sum_{y \in E_Y} \Pr(X = x, Y = y) = \\&\sum_{y \in E_Y} \Pr(Y = y) \cdot \Pr(X = x | Y = y)\end{aligned}$$

- **Bayes rule:** From the multiplication rule

$$p_{X,Y}(x, y) = p_{X|Y}(x, y) \cdot p_Y(y) \quad p_{X,Y}(x, y) = p_{Y|X}(x, y) \cdot p_X(x)$$

we obtain

$$p_{X|Y}(x, y) = \frac{p_{Y|X}(x, y) \cdot p_X(x)}{p_Y(y)}$$

Problem 2 We now want to compute the probability that \mathcal{B}_1 is the chosen box, knowing that the extracted ball is green.

Solution:

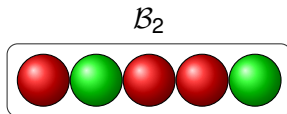
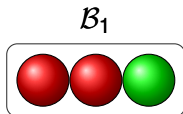
- X takes values $\{0, 1\}$
- Y takes values $\{g, r\}$

We need to compute $\Pr(X = 1 | Y = g)$. Using the multiplication rule,

$$\Pr(X = 1 | Y = g) \cdot \Pr(Y = g) = \Pr(Y = g | X = 1) \cdot \Pr(X = 1)$$

and from this

$$\Pr(X = 1 | Y = g) = \frac{\Pr(Y = g | X = 1) \cdot \Pr(X = 1)}{\Pr(Y = g)} = 5/11$$



Suppose a BSC with crossover probability p ($p < 1/2$).

$\mathbf{c} \in \mathbb{F}_2^n$ is sent and $\mathbf{r} \in \mathbb{F}_2^n$ is received

The probabilities

$$\Pr(\mathbf{c}|\mathbf{r}) \quad \Pr(\mathbf{c}) \quad \Pr(\mathbf{r}|\mathbf{c}) \quad \Pr(\mathbf{r})$$

are related by Bayes'rule

$$\Pr(\mathbf{c}|\mathbf{r}) = \frac{\Pr(\mathbf{r}|\mathbf{c}) \Pr(\mathbf{c})}{\Pr(\mathbf{r})}$$

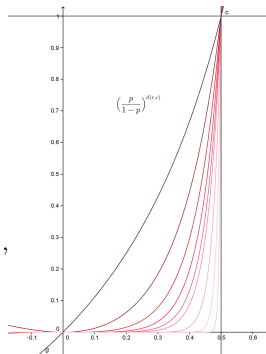
- $\hat{\mathbf{c}} = \max_{\mathbf{c} \in \mathcal{C}} \Pr(\mathbf{c}|\mathbf{r})$ Maximum a posteriori probability
- $\hat{\mathbf{c}} = \max_{\mathbf{c} \in \mathcal{C}} \Pr(\mathbf{r}|\mathbf{c})$ Maximum likelihood decoder

- $\mathbf{r} = r_1 \dots r_n$ and $\mathbf{c} = c_1 \dots c_n$
- $\Pr(\mathbf{r}|\mathbf{c}) = \prod_{i=1}^n \Pr(r_i|c_i)$ since we assumed that bit errors are independent.

$$\begin{cases} \Pr(r_i|c_i) = p & \text{if } r_i \neq c_i \\ \Pr(r_i|c_i) = 1 - p & \text{if } r_i = c_i \end{cases}$$

$$\Pr(\mathbf{r}|\mathbf{c}) = p^{d(\mathbf{r},\mathbf{c})} (1-p)^{n-d(\mathbf{r},\mathbf{c})} = (1-p)^n \left(\frac{p}{1-p} \right)^{d(\mathbf{r},\mathbf{c})},$$

Since $0 < p < 1/2$ and $0 < p/(1-p) < 1$, we can deduce that maximizing $\Pr(\mathbf{r}|\mathbf{c})$ is equivalent to minimizing $d(\mathbf{r}, \mathbf{c})$.



This means that on a BSC, maximum likelihood decoding and nearest decoding are the same.

- This result can be generalized to the q -ary symmetric channel.

- **Bit error probability p_b :** suppose that a codeword is represented by a binary vector \mathbf{c} of length N ; this is the *average* probability that a bit of \mathbf{r} is not equal to the corresponding bit of \mathbf{c} .
- **Block error probability p_B :** of a code and decoder, for a given channel, and for a given input probability $\Pr(c_i)$, is the probability that at least one symbol of a block of symbols is decoded incorrectly.
- **The optimal decoder:** is the one that minimizes the probability of block error. If a uniform input distribution on \mathbf{c} is assumed the optimal decoder is the MLD decoder.

PROBLEM 1

Compute the error probability of the code R_3 (the Repetition code of length 3) for a binary symmetric channel with noise level p .

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Solution: An error is made by R_3 if

- ❶ if 3 bits are flipped, this happens with probability p^3
- ❷ if exactly 2 bits are flipped, this happens with probability $3p^2(1 - p)$

$$p_b = p_B = 3p^2(1-p) + p^3 = 3p^2 - 2p^3 \approx 3p^2$$

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PROBLEM 2

Assuming $p = 0.1$, find how many repetitions are required to get the probability of error down to 10^{-15} .

PROBLEM 1

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Solution: An error is made by R_3 if

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- 2 if exactly 2 bits are flipped, this happens with probability $3p^2(1 - p)$

$$p_b = p_B = 3p^2(1-p) + p^3 = 3p^2 - 2p^3 \approx 3p^2$$

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Solution:

$$n \approx 61$$

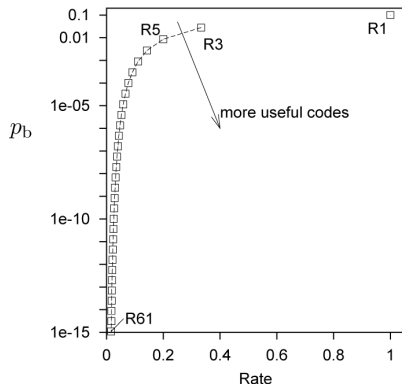
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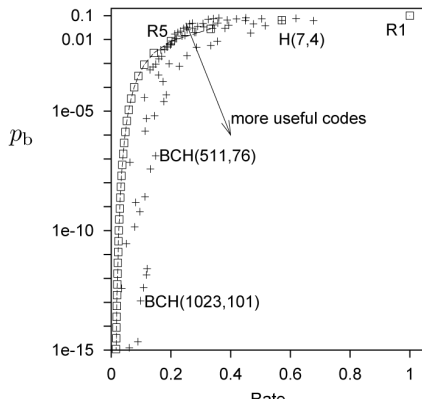
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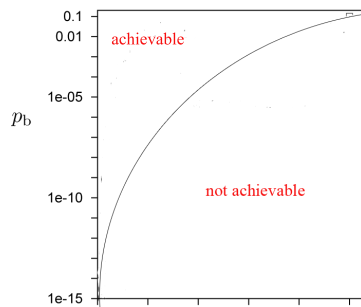
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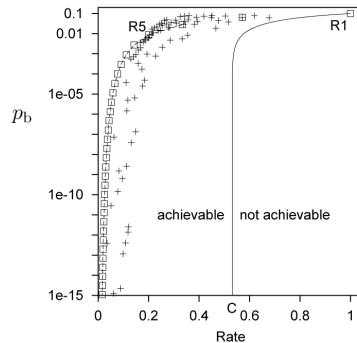
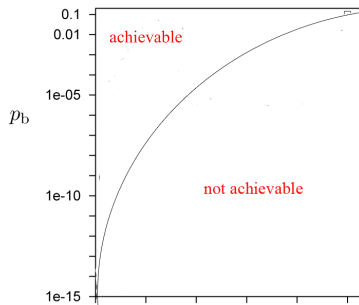
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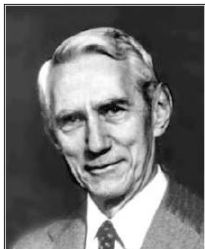


Good codes exist! but...



... but we do not know how to construct them

Shannon proved that for every DMC with a finite number of inputs and outputs points, one may define the notion of **channel capacity**.



Claude-Shannon 1916-2001

- If C is a code with rate $R > \mathcal{C}$, then the probability of error in decoding this code is bounded away from 0. (In other words, at any rate $R > \mathcal{C}$, reliable communication is not possible.)
- For any information rate $R < \mathcal{C}$ and any $\delta > 0$, there exists a code C of length n_δ and rate R , such that the probability of error in maximum likelihood decoding of this code is at most δ .

Proof: **Non-constructive!**

How can we find good codes?? Ingredients of Shannon's proof:

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- Large block length
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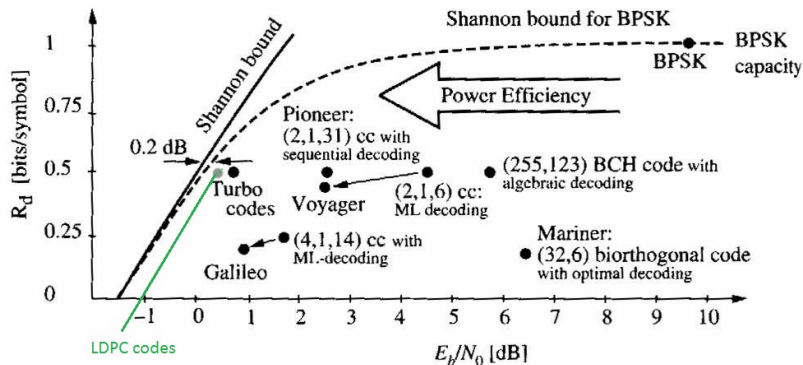
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State-of-art:

- Turbo codes and **LDPC** codes have brought Shannon limits to within reach on a wide range of channels



- LDPC codes are capacity-approaching codes
 - ◇ G.hn/G.9960 (ITU-T Standard for networking over power lines, phone lines and coaxial cable)
 - ◇ 802.3an (10 Giga-bit/s Ethernet over Twisted pair)
 - ◇ DVB-S2 / DVB-T2 / DVB-C2 (Digital video broadcasting, 2nd Generation) and DMB-T/H (Digital video broadcasting)
 - ◇ WiMAX (IEEE 802.16e standard for microwave communications)
 - ◇ IEEE 802.11n-2009 (Wi-Fi standard)