
Optimal Flow control of a noise amplifier, using system identification

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Outline

- 1 Introduction
- 2 Configuration
 - Baseflow, equations
- 3 Model identification
 - Identification of the estimator
 - The ARMAX equation
 - Estimator performances
- 4 Control results
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 - Noise measurement influence
 - Controlling a non-linear flow
- 5 Conclusions

Objectives

Noise amplifier

- Globally stable
- Local amplification of any external perturbation

The challenge

- Control of the unsteadiness
- Model solely based on available data
- The upstream excitations are unknown
- The upstream excitations are always present

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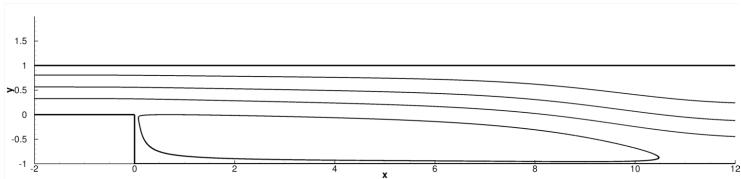
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Step configuration & equations

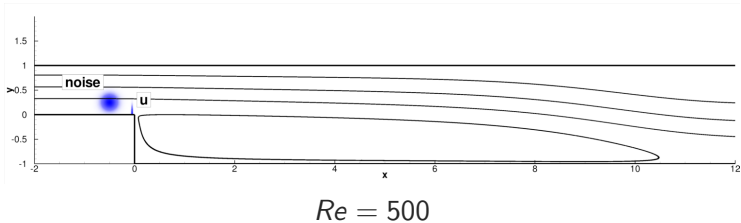


$Re = 500$

$$\begin{aligned}\partial_t \mathbf{v} + (\mathbf{v}_0 \cdot \nabla) \mathbf{v} + (\mathbf{v} \cdot \nabla) \mathbf{v}_0 &= -\nabla p + \frac{1}{Re} \Delta \mathbf{v} - \varepsilon (\mathbf{v} \cdot \nabla) \mathbf{v} \\ \nabla \cdot \mathbf{v} &= 0\end{aligned}$$

$$X(t+1) = AX(t)$$

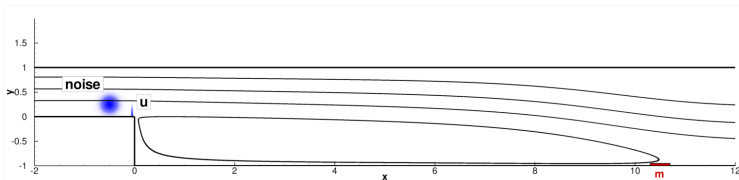
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$$X(t+1) = AX(t) + Bu(t) + B_w w(t)$$

Step configuration & equations

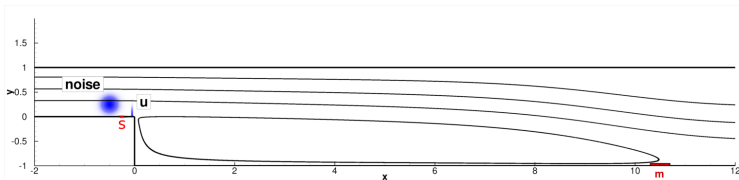


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$$\begin{aligned}X(t+1) &= AX(t) + Bu(t) + B_w w(t) \\ m(t) &= C_m X(t)\end{aligned}$$

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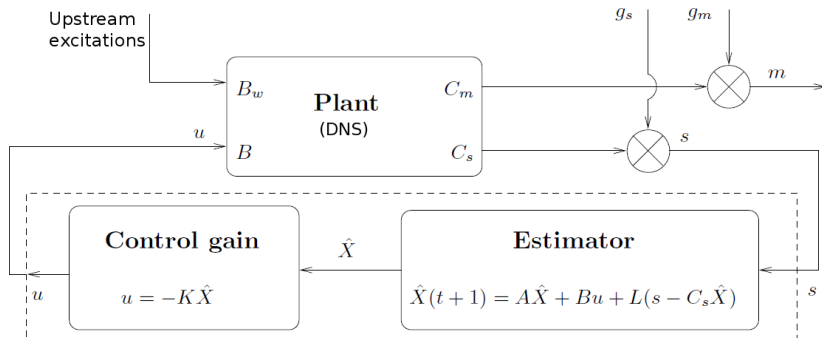
$$\begin{aligned}X(t+1) &= AX(t) + Bu(t) + B_w w(t) \\ m(t) &= C_m X(t) \\ s(t) &= C_s X(t)\end{aligned}$$

Linear Low Order Model

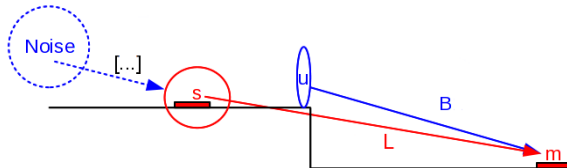
$$X(t+1) = AX(t) + Bu(t) + B_w w(t)$$

$$m(t) = C_m X(t) + g_m(t)$$

$$s(t) = C_s X(t) + g_s(t)$$



Identifying an estimator

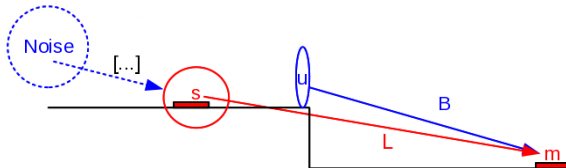


Estimator equation within the classical framework:

$$\begin{cases} \hat{X}(t+1) = \underbrace{(A - LC_s)}_{A_e} \hat{X}(t) + Bu(t) + Ls(t) \\ \hat{m}(t) = C_m \hat{X}(t) \end{cases}$$

$$\Rightarrow \hat{m}(t) = \sum_{k=0}^{\infty} \begin{bmatrix} C_m A_e^k B & C_m A_e^k L \end{bmatrix} \begin{bmatrix} u \\ s \end{bmatrix} (t-k)$$

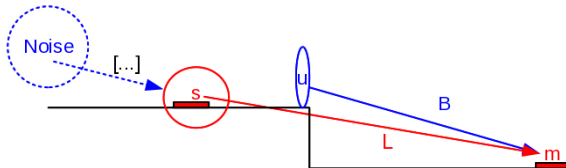
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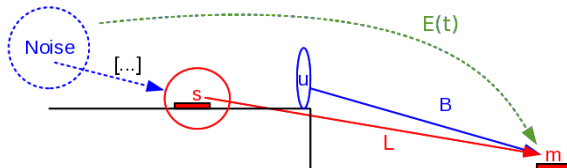
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Estimator equation within the classical framework:

$$\begin{cases} \hat{X}(t+1) = \underbrace{(A - LC_s)}_{A_e} \hat{X}(t) + Bu(t) + Ls(t) \\ \hat{m}(t) = C_m \hat{X}(t) \end{cases}$$
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Identification of the transfer function, using ARMAX model



ARMAX : Auto Regressive Moving Average eXogenous

$$m(t) + \underbrace{\sum_{k=1}^{n_a} a_k m(t-k)}_{\text{auto-regressive}} = \underbrace{\sum_{k=1}^{n_b} \mathbf{b}_k \begin{bmatrix} u \\ s \end{bmatrix} (t-k - \begin{bmatrix} n_{d_u} \\ n_{d_s} \end{bmatrix})}_{\text{eXogenous}} + \underbrace{E(t)}_{\text{MovingAverage}}$$

Learning dataset

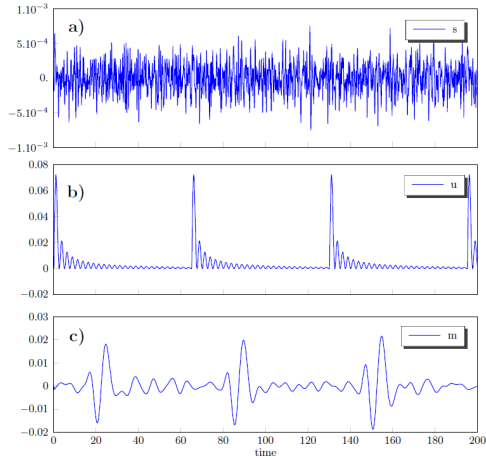


Figure 1: Learning dataset

Testing dataset

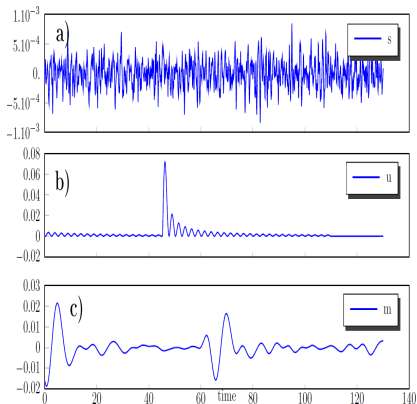


Figure 2: First testing dataset

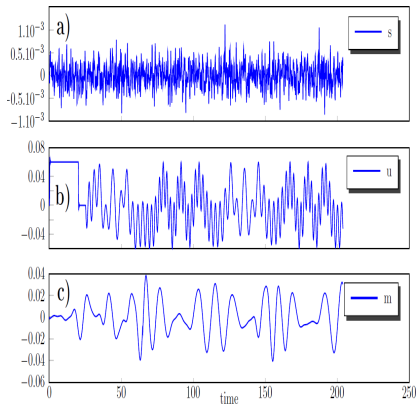


Figure 3: Second testing dataset

Identified model performances



Figure 4: Estimator performances evaluation

From the estimator to a full model

Estimator

$$\hat{X}(t+1) = \underbrace{(A - LC_s)}_{A_e} \hat{X}(t) + Bu(t) + Ls(t)$$

C_s is needed

- s is a measurement $\Rightarrow s \neq$ white noise
- s could be affected by the control u

Full model:

$$ROM : \begin{cases} X(t+1) &= AX(t) + Bu(t) + Lw(t) \\ s(t) &= C_s X(t) + w(t) \\ m(t) &= C_m X(t) + E(t) \end{cases}$$

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Linear regression to obtain C_s :

$$s(t) = C_s \hat{X}(t) + w(t)$$

$$\Leftrightarrow s(t) = f \left[\underbrace{s(t-1), s(t-2)}_{\text{sensor characterization}} \dots \underbrace{u(t-1), u(t-2)}_{\text{control effects}} \dots \right] + w(t)$$

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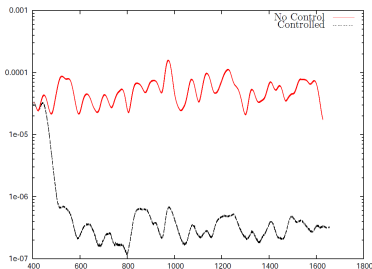
Linear regression to obtain C_s :

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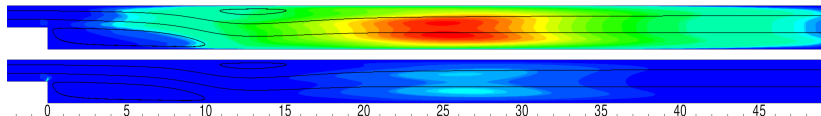
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Perturbations energy reduction



Total energy reduction (log scale)

- Total energy reduction: 99%
- Maximal kinetic energy reduction: 96% (at $x \approx 25$)
- The compensator consists of a 17×17 matrix



Mean kinetic energy reduction

Noise measurement influence onto the control performances

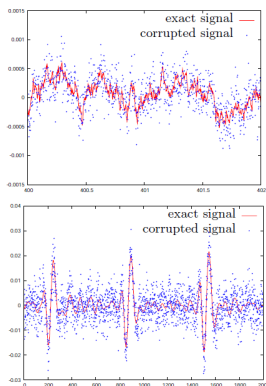


Figure 5: Noisy signals

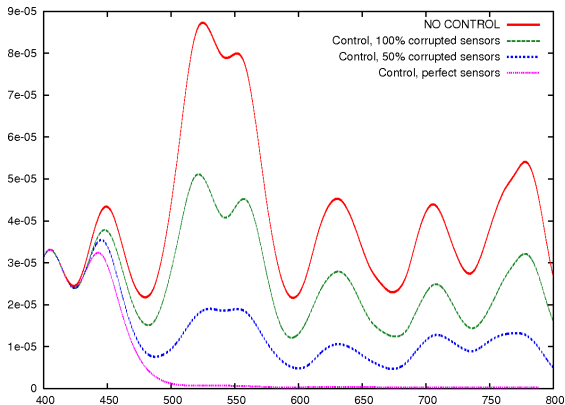
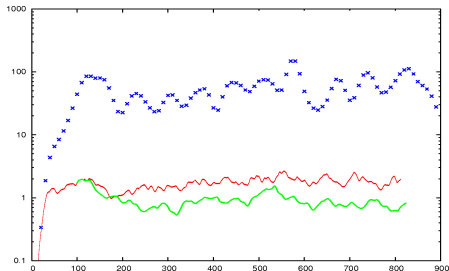


Figure 6: Influence of the noise measurement onto the final performances

Effects of the non-linearities onto the control performances



- Perturbations energy: $\times 10^6$
- 17 modes compensator

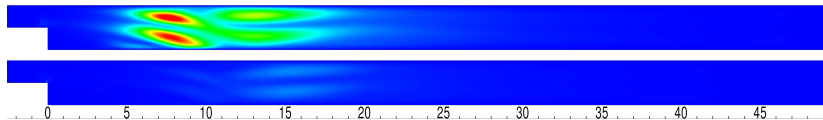
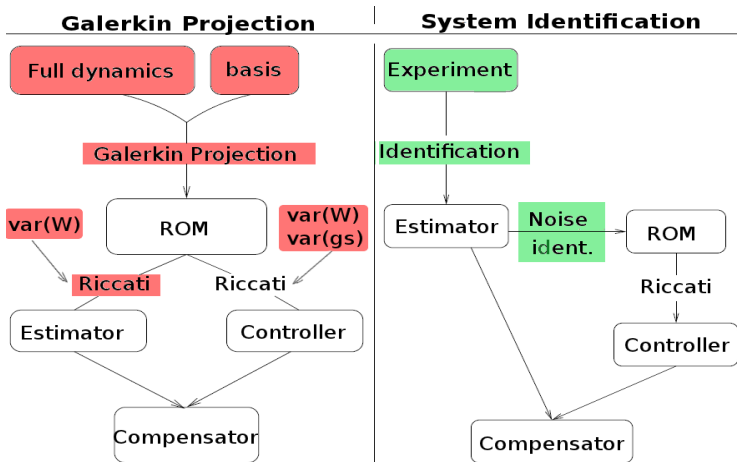


Figure 7: The mean turbulent kinetic energy is reduced by a factor of 98%, at $x \approx 8$

Comparison between classical Galerkin/POD method, & Identification method



Summary

- The model is identified rather than projected
- The process only requires data which can readily be extracted from a lab experiment
- Low dimensional controller
- The downstream sensor is only used for the identification process
- No knowledge about the excitations is required
- Good robustness against noise measurement, and against high amplitude excitations
- Low computational cost of the whole process

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The end

- Thank you for your attention

Appendix: Realization of the ARMAX equation

General Space state form

$$(ARMAX \text{ Estimator}) : \begin{cases} X(t+1) &= A_e X(t) + Bu(t) + Ls(t) \\ m(t) &= C_m X(t) + E(t). \end{cases}$$

State-Space Realization of ARMAX equation

$$\underbrace{\begin{bmatrix} m(t) \\ m(t-1) \\ \vdots \\ m(t-n_a+1) \\ \mathbf{u}_{t-n_{bu}+1}^t \\ \mathbf{s}_{t-n_{bs}+1}^t \end{bmatrix}}_{X(t+1)} = \underbrace{\begin{bmatrix} - & - & C_{ARMAX} & - & - \\ & 1 & 0 & & 0 \\ & & \ddots & & \\ \cdots & & & 1 & 0 \\ & & & Shift(n_{bu}) & \\ & & Shift(n_{bs}) & & \end{bmatrix}}_{A_e} \underbrace{\begin{bmatrix} m(t-1) \\ m(t-2) \\ \vdots \\ m(t-n_a) \\ \mathbf{u}_{t-n_{bu}}^{t-1} \\ \mathbf{s}_{t-n_{bs}}^{t-1} \end{bmatrix}}_{X(t)} + \underbrace{\begin{bmatrix} 0 \\ \vdots \\ \vdots \\ \vdots \\ 1 \\ 0 \\ \vdots \end{bmatrix}}_B u(t) + \underbrace{\begin{bmatrix} 0 \\ \vdots \\ \vdots \\ \vdots \\ 0 \\ 1 \\ \vdots \end{bmatrix}}_L s(t) + \underbrace{\begin{bmatrix} 1 \\ 0 \\ \vdots \end{bmatrix}}_{E(t)} E(t)$$

$$Shift(n) = \begin{pmatrix} 0 & 0 & 0 & 0 & \cdots \\ 1 & 0 & 0 & 0 & \cdots \\ 0 & 1 & 0 & 0 & \cdots \\ 0 & 0 & 1 & 0 & \cdots \\ & & \ddots & & \end{pmatrix} \in \mathbf{R}^{n \times n}, \quad \mathbf{M}_a^b = \begin{bmatrix} m(t=b) \\ \vdots \\ m(t=a+1) \\ m(t=a) \end{bmatrix}, \quad \mathbf{U}_a^b = \begin{bmatrix} u(t=b) \\ \vdots \\ u(t=a+1) \\ u(t=a) \end{bmatrix},$$