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<i>Assignment/Lab Number:</i>	4
<i>Assignment/Lab Title:</i>	Discrete-Time Fourier Transform

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A) Discrete-Time Fourier Transform (DTFT)

Part 1 - DTFT

```
clc
clear
N0 = 128;
n = 0:N0-1;
ohm = (2*pi/128).*(-64:63);

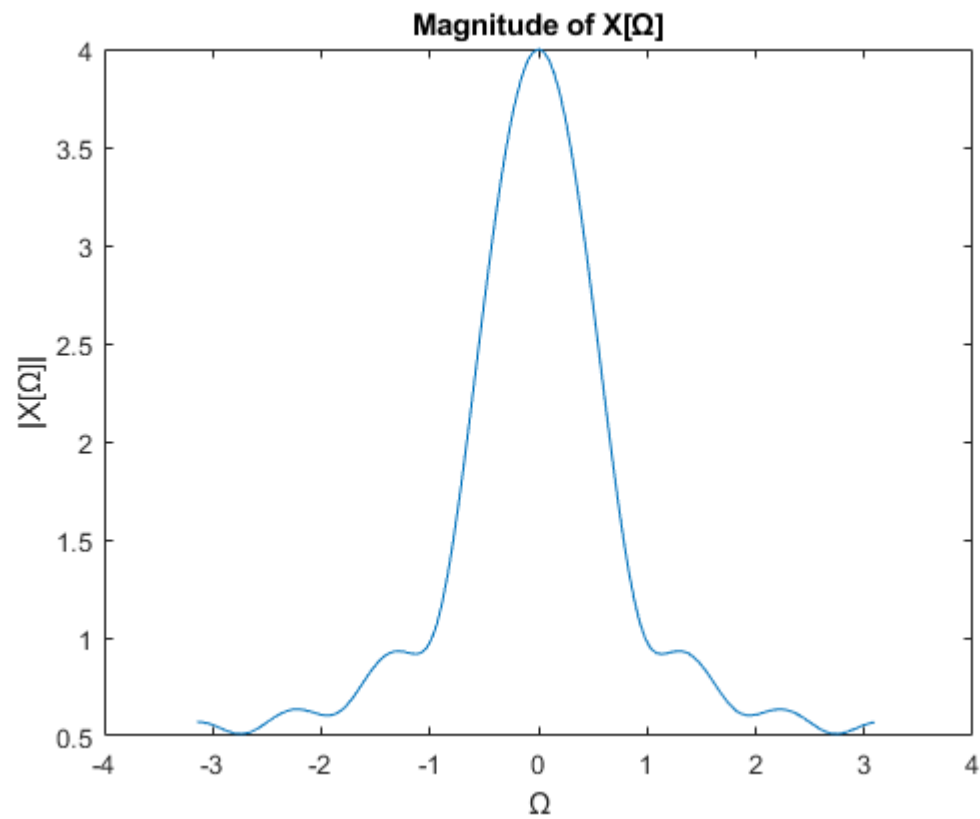
x = [1 (6/7) (5/7) (4/7) (3/7) (2/7) (1/7) zeros(1, 121)];
X = fft(x);

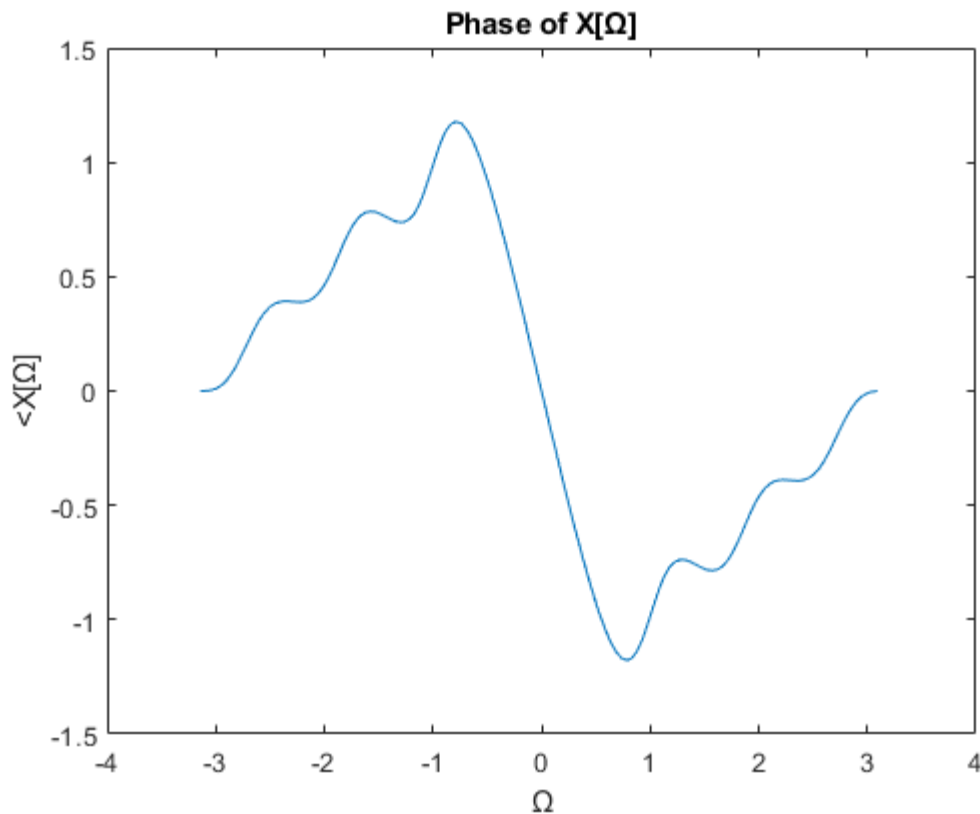
figure;
plot(ohm, fftshift(abs(X)));
title('Magnitude of X[Ω]')
xlabel('Ω')
ylabel('|X[Ω]|')

figure;
plot(ohm, fftshift(angle(X)));
title('Phase of X[Ω]')
xlabel('Ω')
ylabel('∠X[Ω]')

% figure;
% plot(ohm, abs(X));
% title('Magnitude of X[Ω]')
% xlabel('Ω')
% ylabel('|X[Ω]|')
%
% figure;
```

```
% plot(ohm, angle(X));  
% title('Phase of X[Ω]')  
% xlabel('Ω')  
% ylabel('<X[Ω]')
```





Part 2 - FTD by Hand

$$X(\Omega) = \sum_{n=-\infty}^{\infty} x(n) e^{j\Omega n}$$

$$x(n) = \left(\frac{-1}{7}n + 1\right)(u(n) - u(n-6))$$

$$\begin{aligned} X(\Omega) &= \sum_{n=-\infty}^{\infty} \left(\frac{-1}{7}n + 1\right)(u(n) - u(n-6)) e^{-j\Omega n} \\ &= \sum_{n=0}^6 \left(\frac{-1}{7}n + 1\right) e^{-j\Omega(n+6)} \end{aligned}$$

$$\begin{aligned} X(\Omega) &= 1 + \frac{6}{7} e^{-j\Omega} + \frac{5}{7} e^{-2j\Omega} + \frac{4}{7} e^{-3j\Omega} \\ &\quad + \frac{3}{7} e^{-4j\Omega} + \frac{2}{7} e^{-5j\Omega} + \frac{1}{7} e^{-6j\Omega} \end{aligned}$$

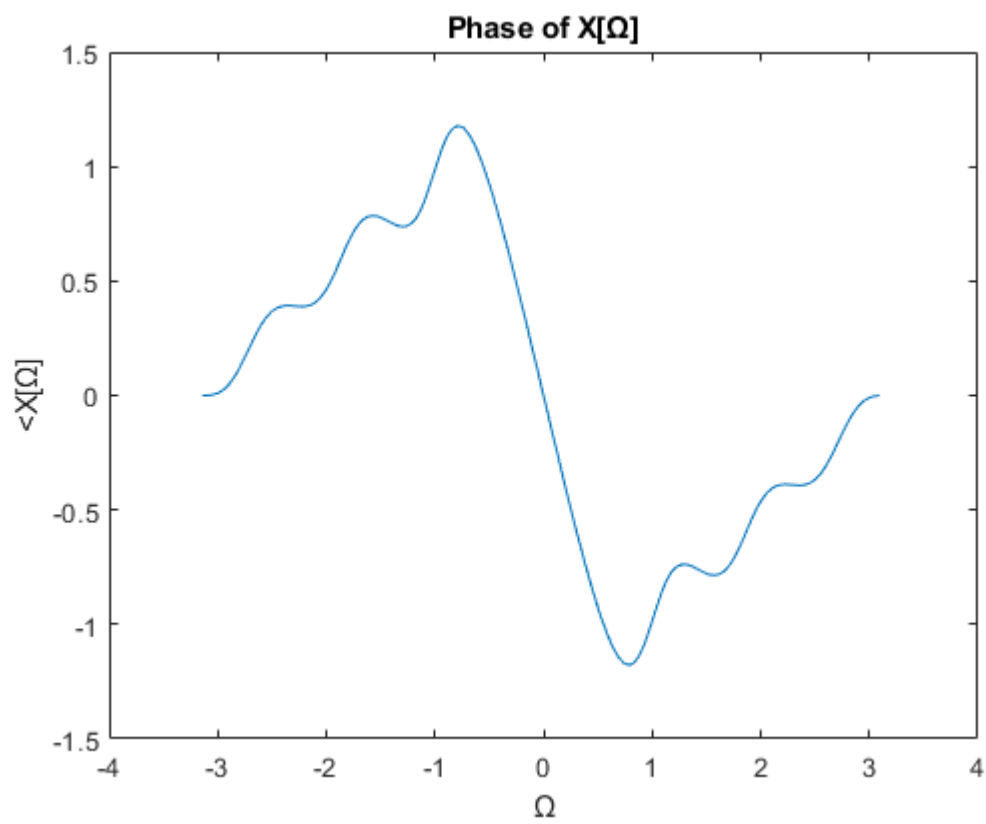
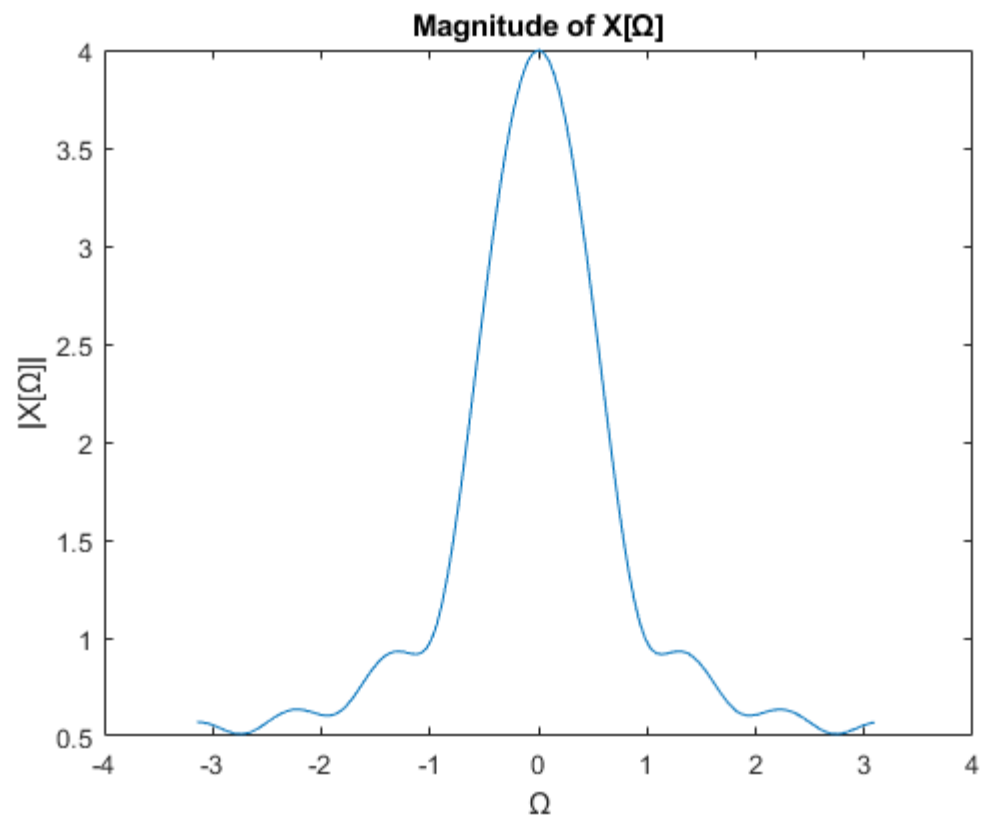
```
clc
clear
```

```
ohm = (2*pi/128).*(-64:63);
```

```
X = 1 + (6/7).*exp(-1i.*ohm) ...  
    + (5/7).*exp(-1i.*2.*ohm) ...  
    + (4/7).*exp(-1i.*3.*ohm) ...  
    + (3/7).*exp(-1i.*4.*ohm) ...  
    + (2/7).*exp(-1i.*5.*ohm) ...  
    + (1/7).*exp(-1i.*6.*ohm);
```

```
figure;  
plot(ohm, abs(X));  
title('Magnitude of X[Ω]')  
xlabel('Ω')  
ylabel('|X[Ω]|')
```

```
figure;  
plot(ohm, angle(X));  
title('Phase of X[Ω]')  
xlabel('Ω')  
ylabel('<X[Ω]')
```



Response: Yes, the results of part 2 are consistent with part 1.

Part 3 - IFFT of Hand Calculated DTFT

```
clc
clear
N0 = 128;
n = 0:N0-1;
ohm = (2*pi/128).*(-64:63);

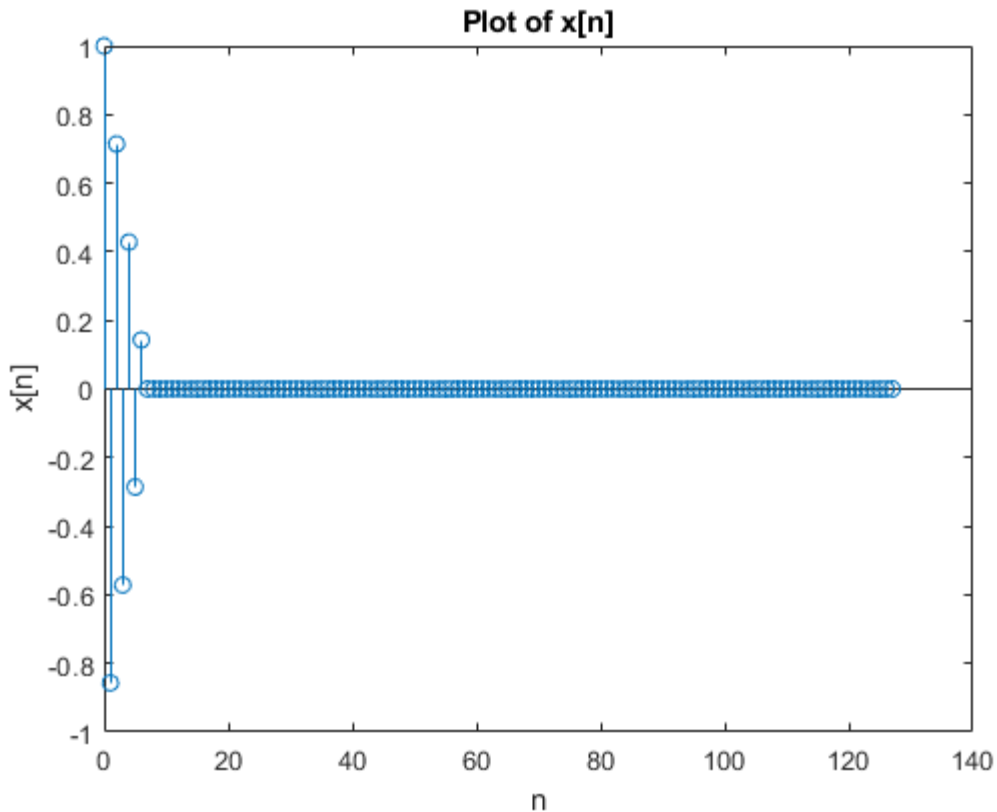
X = 1 + (6/7).*exp(-1i.*ohm) ...
    + (5/7).*exp(-1i.*2.*ohm) ...
    + (4/7).*exp(-1i.*3.*ohm) ...
    + (3/7).*exp(-1i.*4.*ohm) ...
    + (2/7).*exp(-1i.*5.*ohm) ...
    + (1/7).*exp(-1i.*6.*ohm);

% X = fftshift(X);

x = ifft(X);

figure;
stem(n, x);
title('Plot of x[n]')
xlabel('n')
ylabel('x[n]')
```

Warning: Using only the real component of complex data.



Response: No, the result isn't the same as $x[n]$ since the definition of the Fourier Transform involves the multiplication by a complex exponential thus resulting in a shift when switching between domains.

B) Time Convolution

Part 1 - DTFT plot of $x[n]$

```

clc
clear
n = (0:15);
u_c = @(t) 1.0.*(t>=0);
u = @(n) u_c(n).*(mod(n,1) == 0);
x = sin(2*pi*n/10).*(u(n) - u(n-10));

omega= linspace(-pi,pi,1001);
W_omega = exp(-1i).^((0:length(x)-1)*omega);
X = (x*W_omega);

```

```
figure;
stem(n, x);
title('Plot of x[n]')
xlabel('n')
ylabel('x[n]')
```

```
figure;
plot(omega, abs(X));
title('Magnitude of X[Ω]')
xlabel('Ω')
ylabel('|X[Ω]|')
```

```
figure;
plot(omega, angle(X));
title('Phase of X[Ω]')
xlabel('Ω')
ylabel('∠X[Ω]')
```

```
% Part 2 - DTFT plot of h[n]
n = (0:9);
x = u(n)-u(n-9);
```

```
omega= linspace(-pi,pi,1001);
W_omega = exp(-1i).^((0:length(x)-1)*omega);
H = (x*W_omega);
```

```
figure;
```

```

stem(n, x);
title('Plot of h[n]')
xlabel('n')
ylabel('h[n]')

figure;
plot(omega, abs(H));
title('Magnitude of H[Ω]')
xlabel('Ω')
ylabel('|H[Ω]|')

```

```

figure;
plot(omega, angle(H));
title('Phase of H[Ω]')
xlabel('Ω')
ylabel('∠H[Ω]')

```

% Part 3 - Convolution Plot of X[Ω] and H[Ω]

```
Y = X.*H;
```

```

figure;
plot(omega, abs(Y));
title('Magnitude of Y[Ω]')
xlabel('Ω')
ylabel('|Y[Ω]|')

```

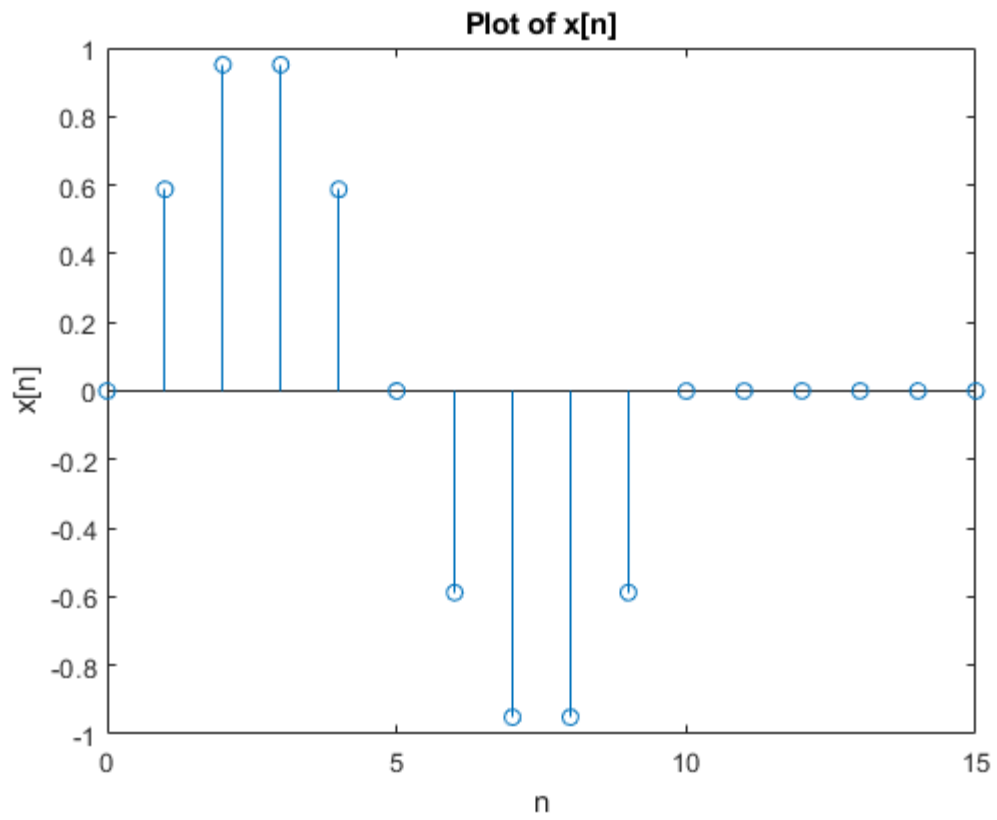
```

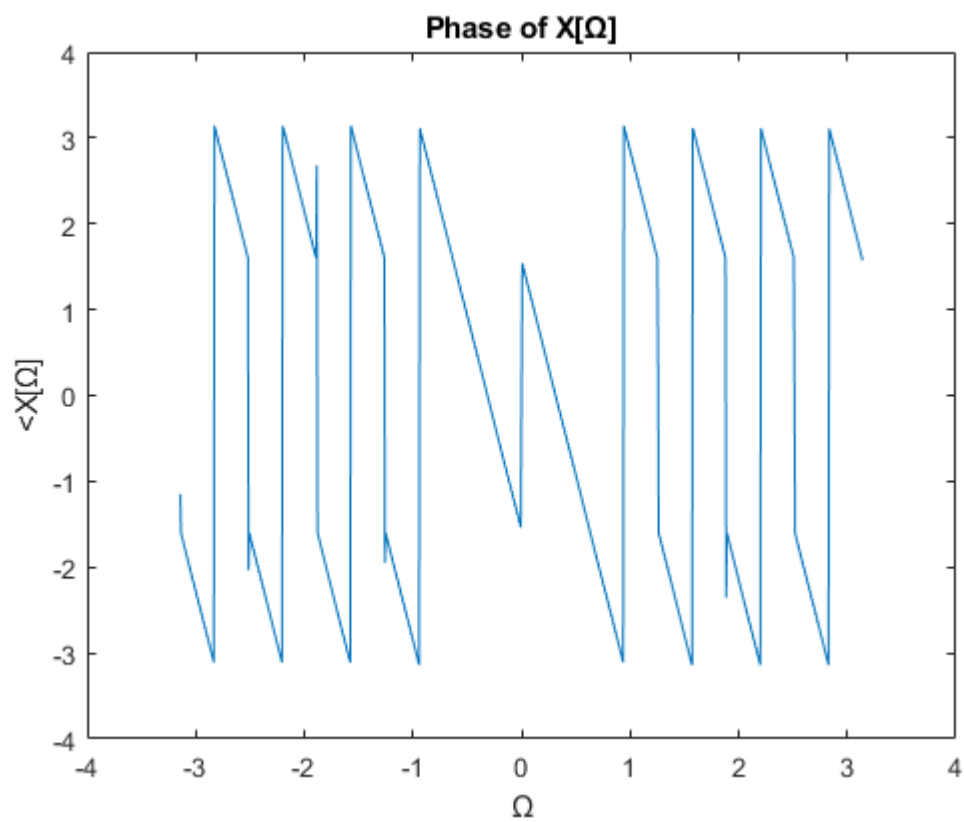
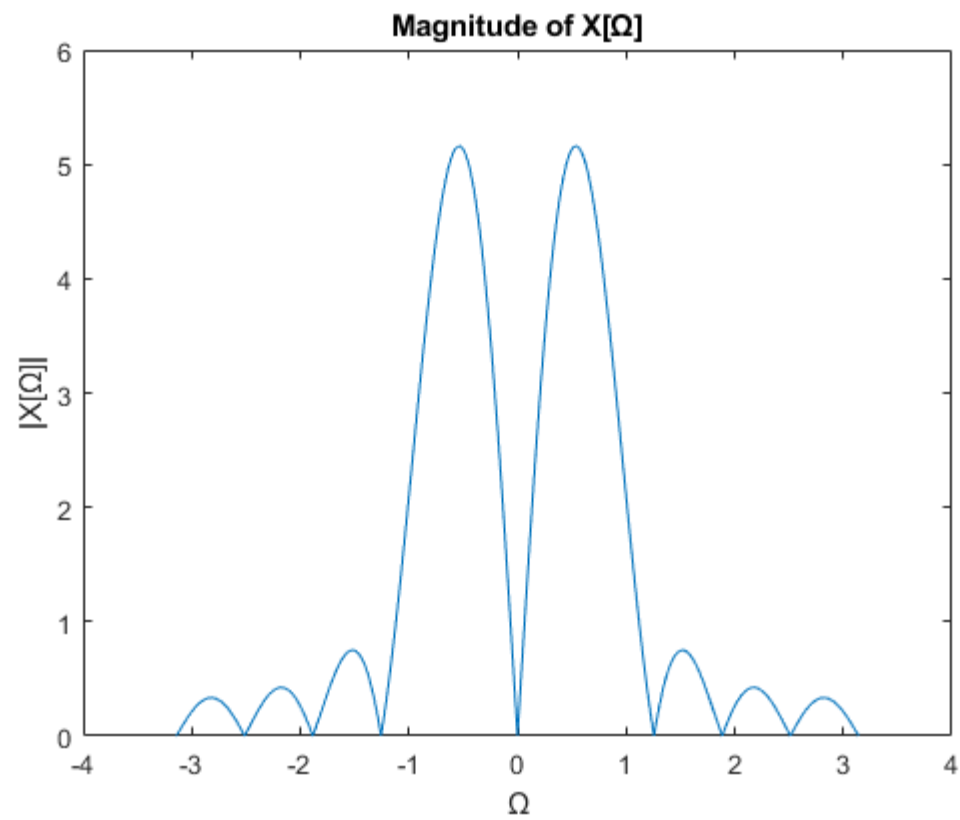
figure;
plot(omega, angle(Y));
title('Phase of Y[Ω]')

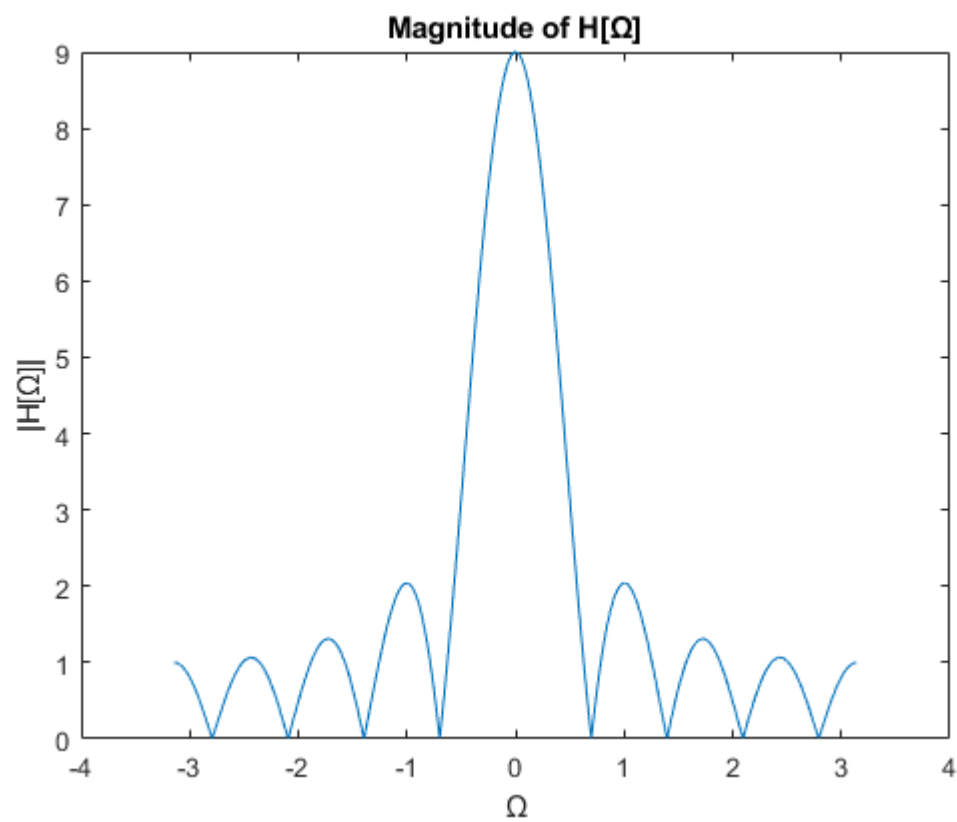
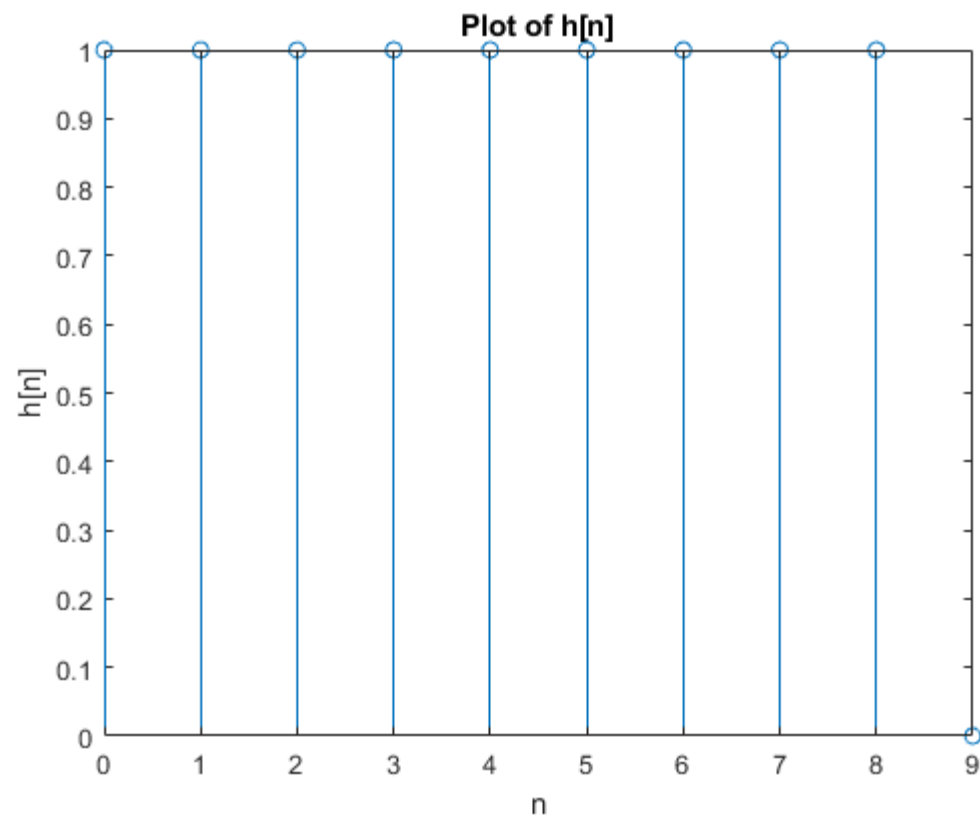
```

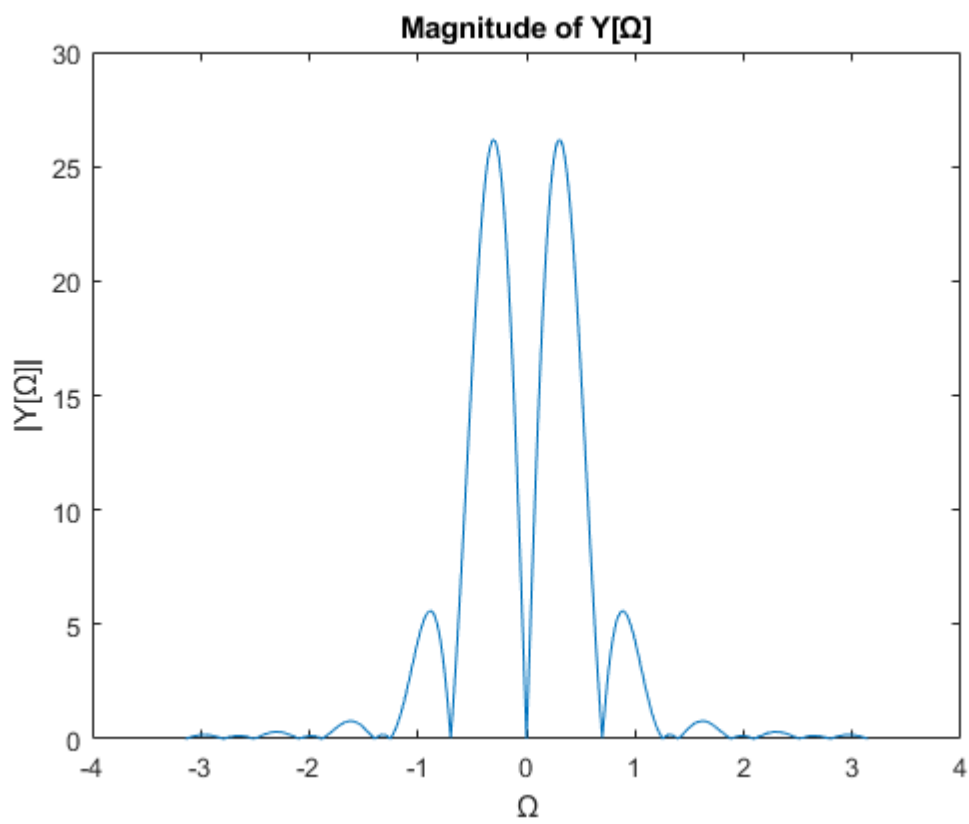
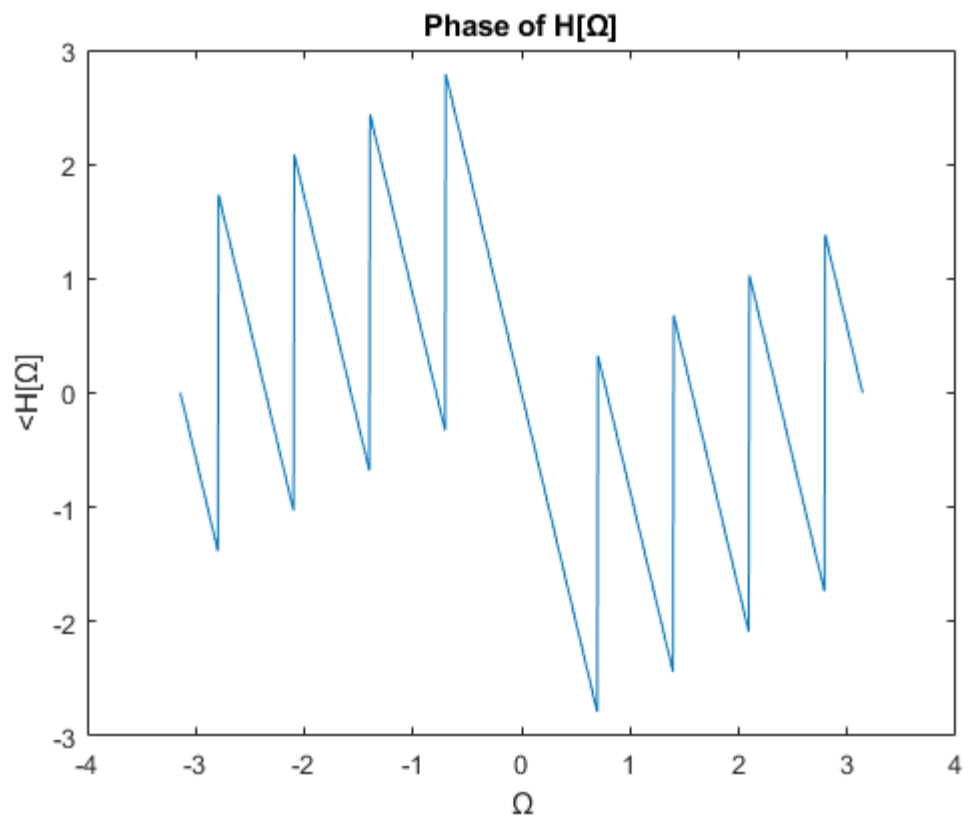
```
xlabel('Ω')
```

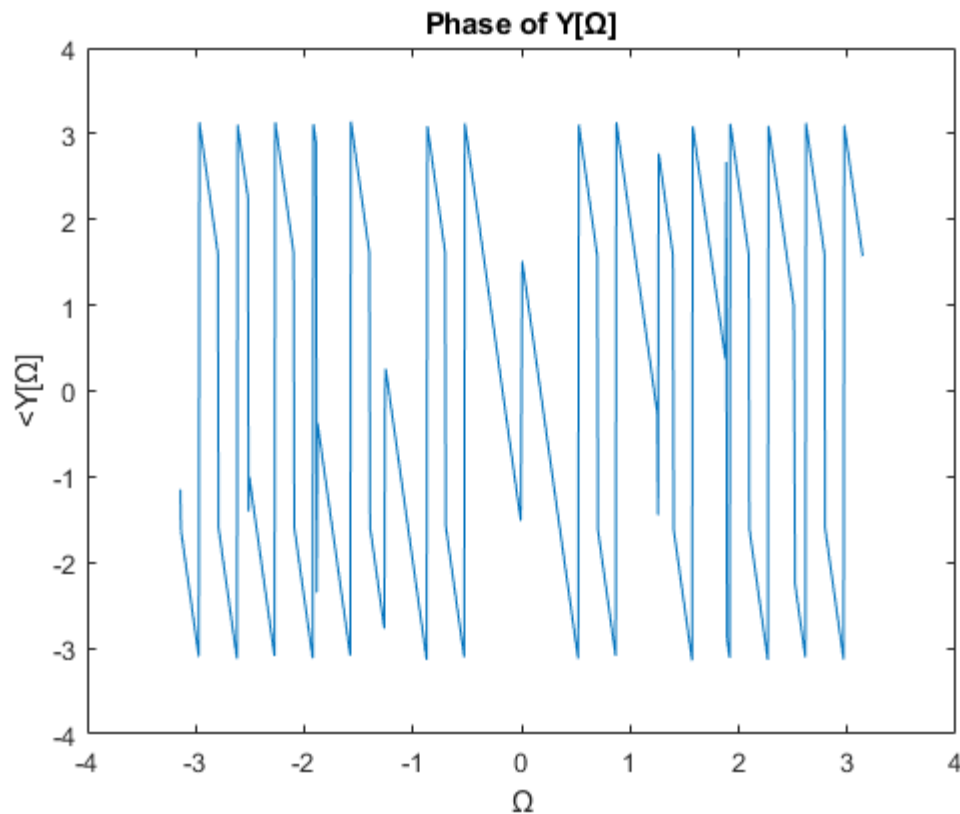
```
ylabel('<Y[Ω]')
```











Part 4 - Convolution Plot of $x[n]$ and $h[n]$ by conv command

```
clc
clear
n = (0:15);
u_c = @(t) 1.0.*(t>=0);
u = @(n) u_c(n).*(mod(n,1) == 0);

h = u(0:9);
x = sin(2*pi*n/10).*(u(n) - u(n-10));
n=0:24;

y = conv(x, h);

figure;
stem(n, y);
```



```

title('plot of y[n] using conv command')
xlabel('n')
ylabel('y[n]')

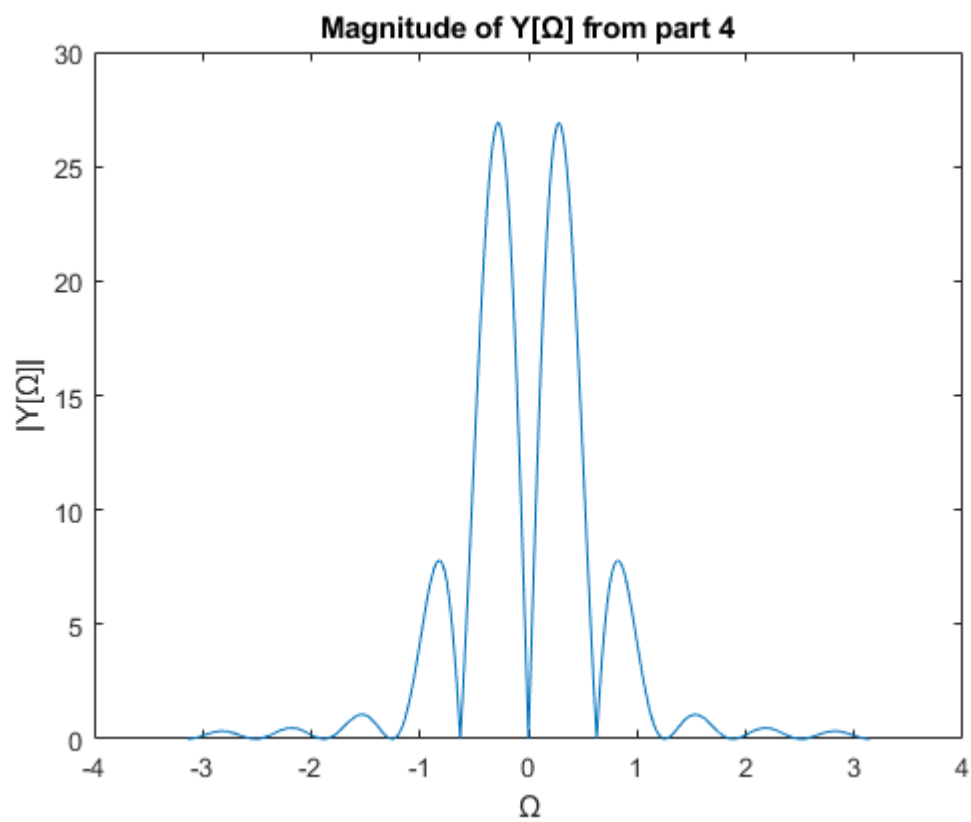
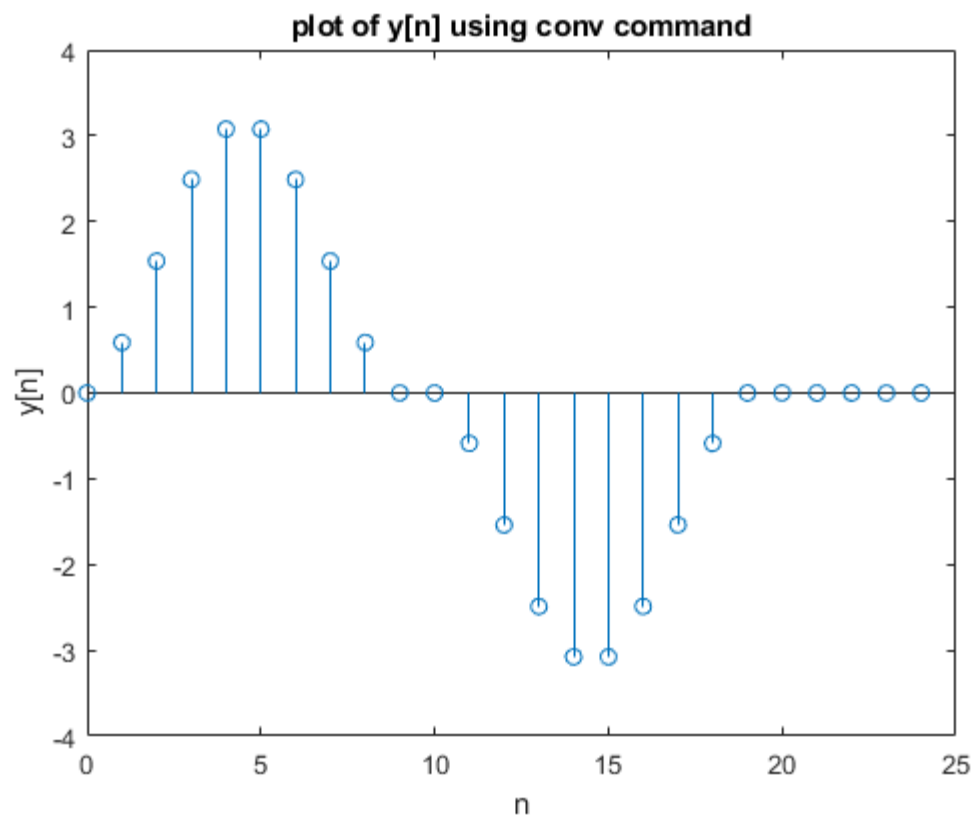
% Part 5 - DTFT Plot of y[n] from Part 4

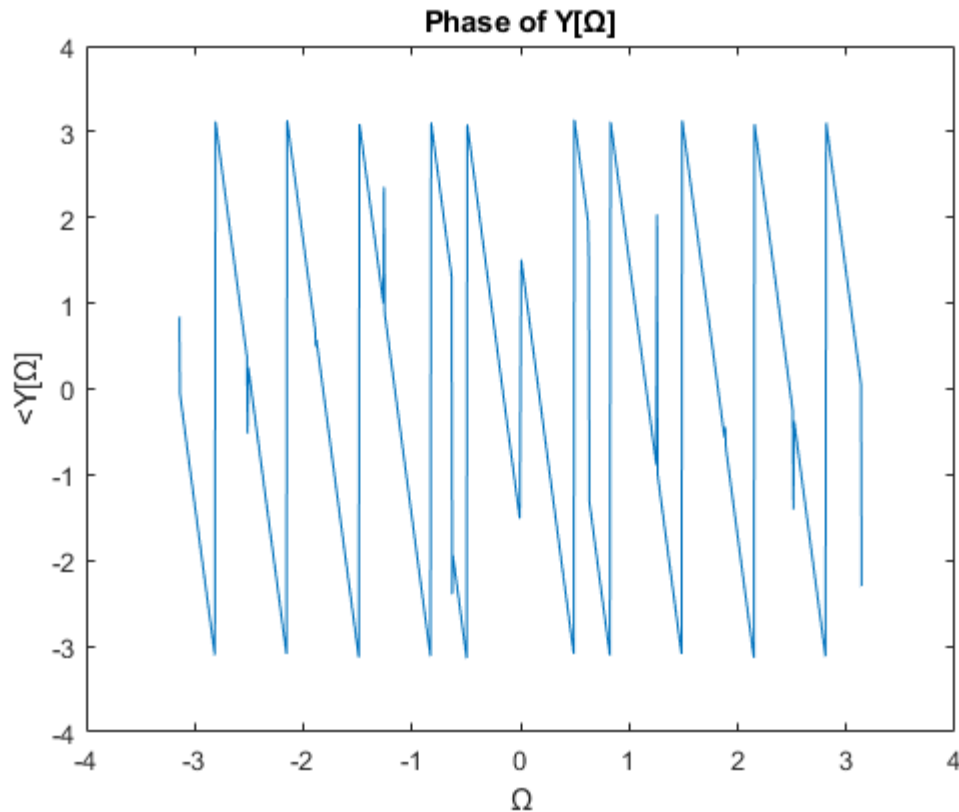
omega= linspace(-pi,pi,1001);
W_omega = exp(-1i).^((0:length(y)-1)*omega);
Y = (y*W_omega);

figure;
plot(omega, abs(Y));
title('Magnitude of Y[Ω] from part 4')
xlabel('Ω')
ylabel('|Y[Ω]|')

figure;
plot(omega, angle(Y));
title('Phase of Y[Ω]')
xlabel('Ω')
ylabel('<Y[Ω]')

```





Part 6 – Question Response

Yes, the same results were achieved in part 3 and 5, this is due to a property of the Fourier transform; The convolution of two functions in the time domain is equivalent to the product in the frequency domain.

C) FIR Filter Design by Frequency Sampling

Part 1 - High Pass FIR Filter

```
ohm0 = 2*pi/3;
N = 35;
n = 0:N-1;
Omega = linspace(0,2*pi*(1-1/N),N);
H_d = @(Omega) (mod(Omega,2*pi)>ohm0).*(mod(Omega,2*pi)<2*pi-ohm0);

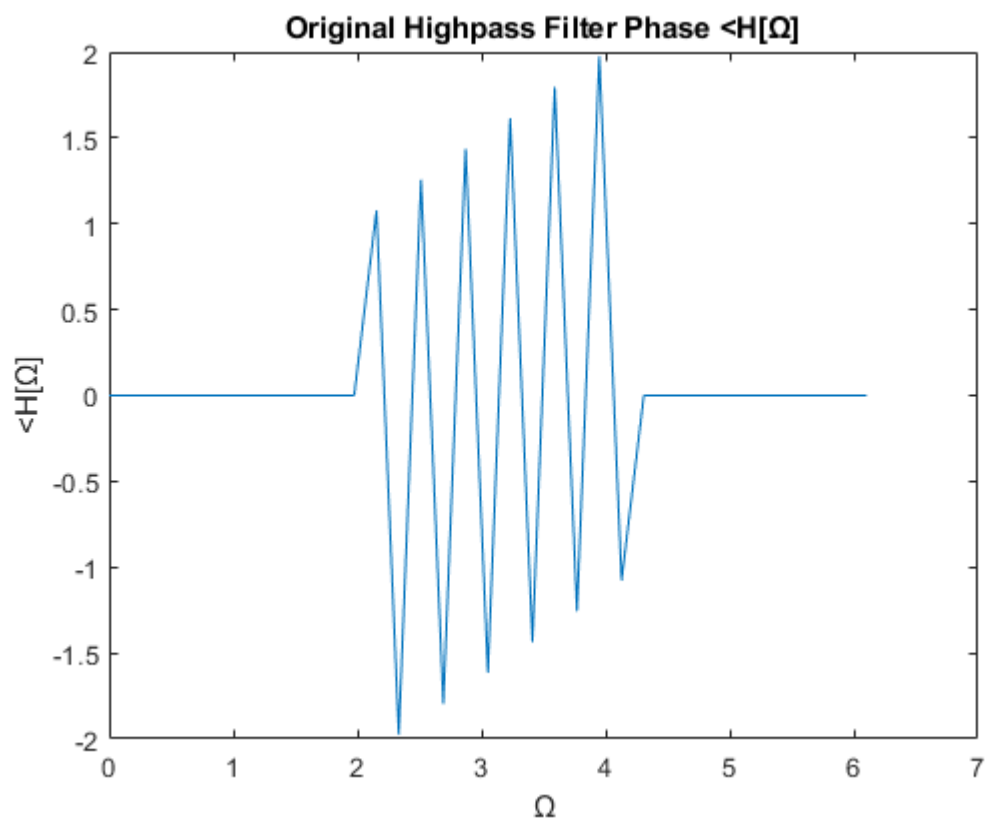
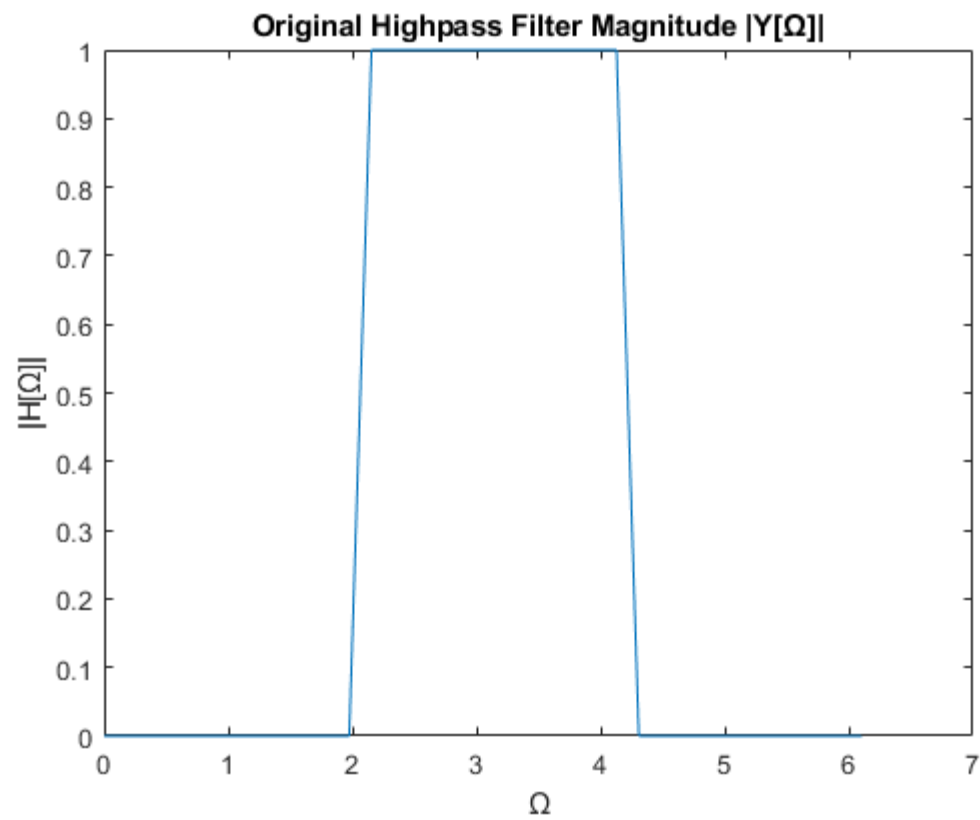
H = H_d(Omega).*exp(-1i.*Omega.*((N-1)/2));
h = ifft(H);
```

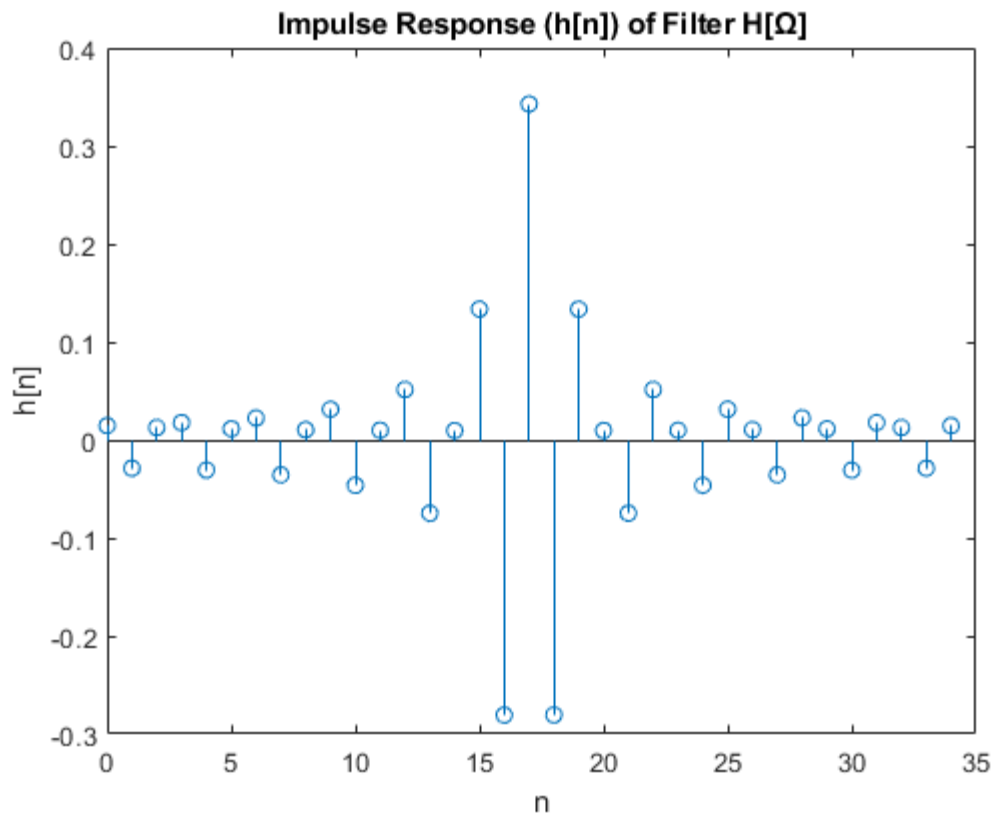
```
figure;  
plot(Omega, abs(H));  
title('Original Highpass Filter Magnitude  $|Y[\Omega]|$ ')  
xlabel('Ω')  
ylabel('H[Ω]')
```

```
figure;  
plot(Omega, angle(H));  
title('Original Highpass Filter Phase  $\angle H[\Omega]$ ')  
xlabel('Ω')  
ylabel('∠H[Ω]')
```

```
figure;  
stem(n, h);  
title('Impulse Response (h[n]) of Filter H[Ω]')  
xlabel('n')  
ylabel('h[n]')
```

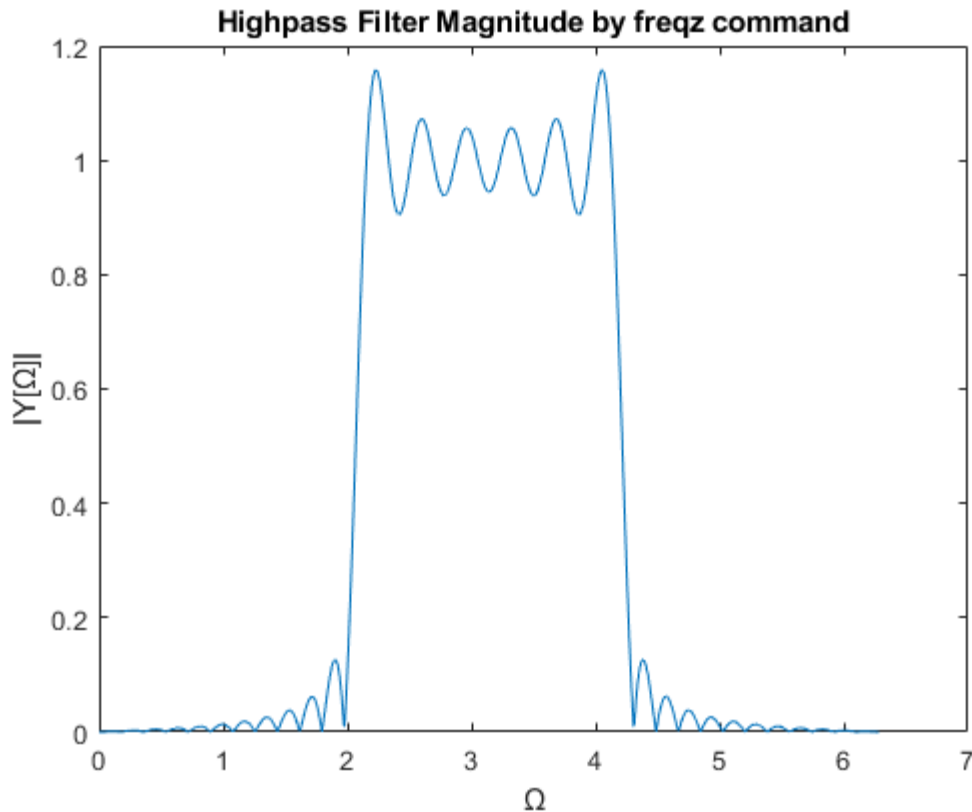
Warning: Using only the real component of complex data.





Part 2 - Frequency Response from $h[n]$ by freqz Command

```
H = freqz(h,1, 0:2*pi/1001:2*pi);
figure;
plot(0:2*pi/1001:2*pi, abs(H));
title('Highpass Filter Magnitude by freqz command')
xlabel('Ω')
ylabel('|Y[Ω]|')
```



Part 3 – Question Response

The result in part 2 is different from the ideal filter we started with in terms of the frequencies it permits. The ideal filter has a strict cut-off frequency, whereas the result in part 2 has ripples beyond the desired cut-off frequency. This is due to the impulse response of an ideal high pass filter being noncausal and therefore physically unrealizable.

Part 4 – Increase to 71 points

```
clc
clear
ohm0 = 2*pi/3;
N = 71;
n = 0:N-1;
Omega = linspace(0,2*pi*(1-1/N),N);
H_d = @(Omega) (mod(Omega,2*pi)>ohm0).*(mod(Omega,2*pi)<2*pi-ohm0);

H = H_d(Omega).*exp(-1i.*Omega.*((N-1)/2));
```

```

h = ifft(H);

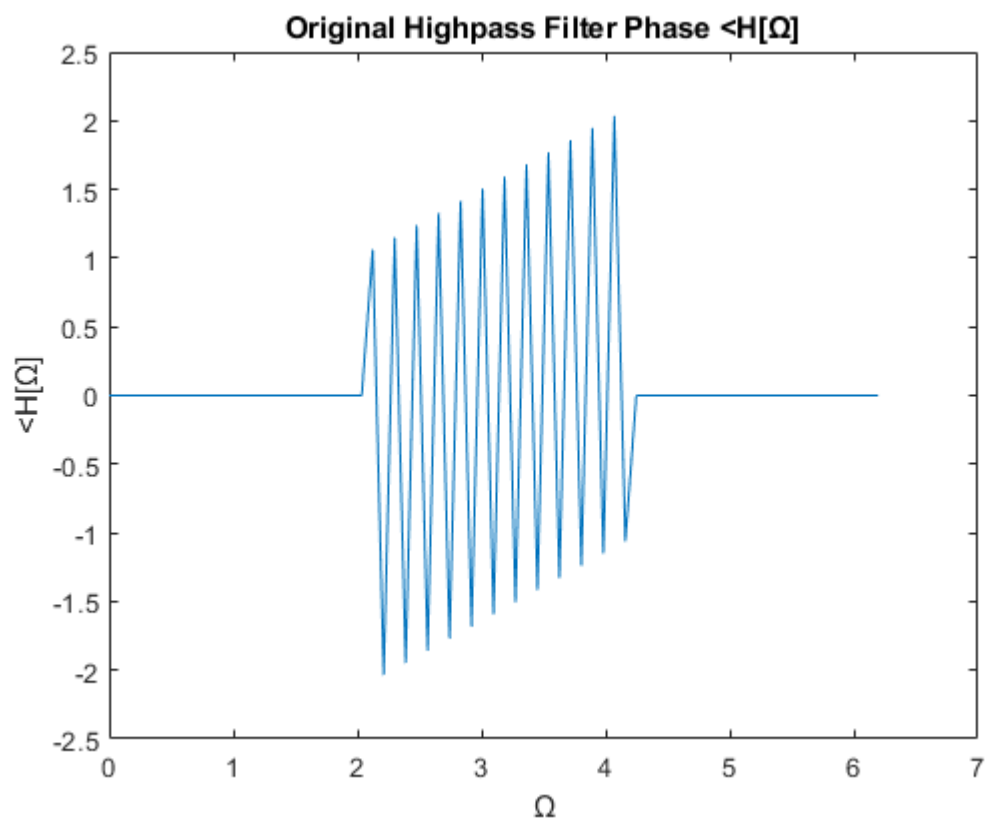
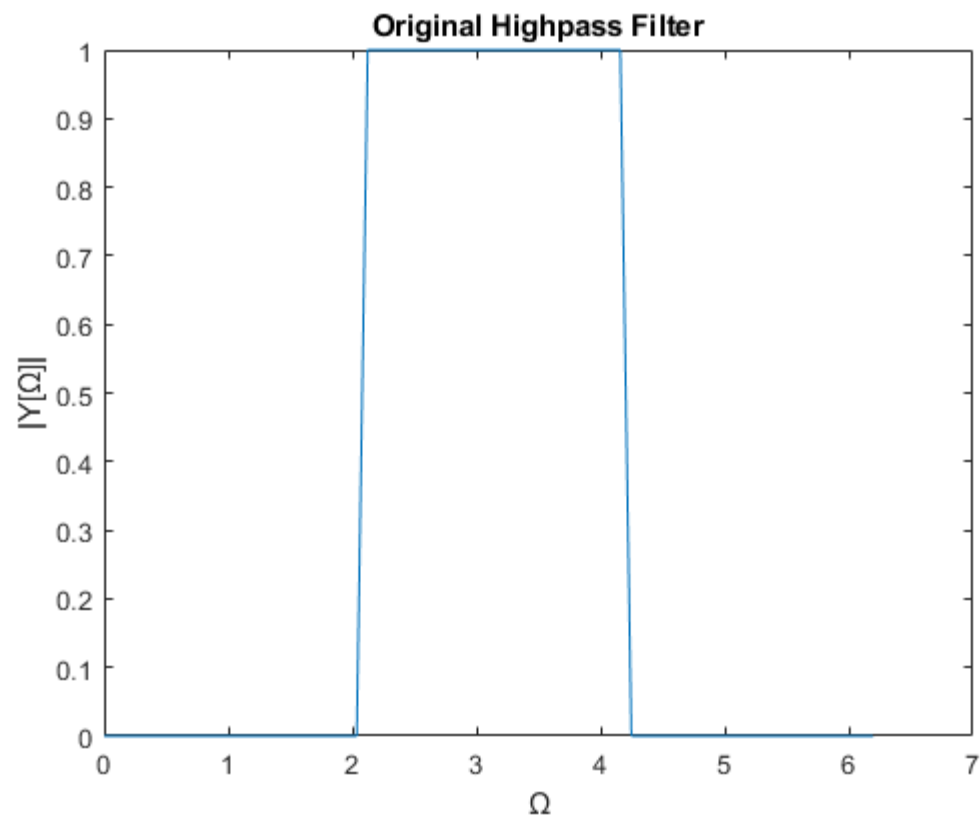
figure;
plot(Omega, abs(H));
title('Original Highpass Filter')
xlabel('Ω')
ylabel('|Y[Ω]|')

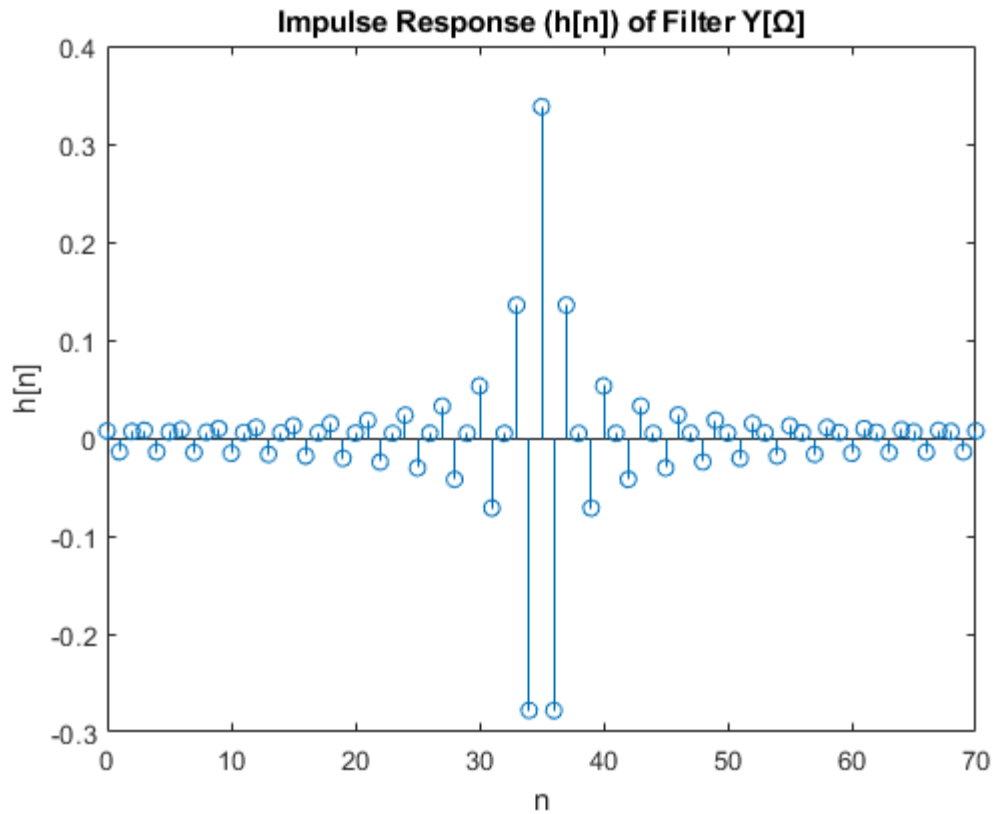
figure;
plot(Omega, angle(H));
title('Original Highpass Filter Phase <H[Ω]')
xlabel('Ω')
ylabel('<H[Ω]')

figure;
stem(n, h);
title('Impulse Response (h[n]) of Filter Y[Ω]')
xlabel('n')
ylabel('h[n]')

```

Warning: Using only the real component of complex data.





Part 5 – Question Response

Increasing N increases the 'resolution' since we get a more accurate representation of the impulse response due to the existence of more points and a more unique signal representation.