Contents

- Lab 9
- Exercise 9.1: (Effects of DFT size)
- 9.1a
- 9.1b
- 9.1c
- 9.1d
- Exercise 9.2: (Frequency-domain analysis using FFT)
- 9.2a
- 9.2b
- 9.2c

```
format compact
%Alexander Adams
%3769-0517
```

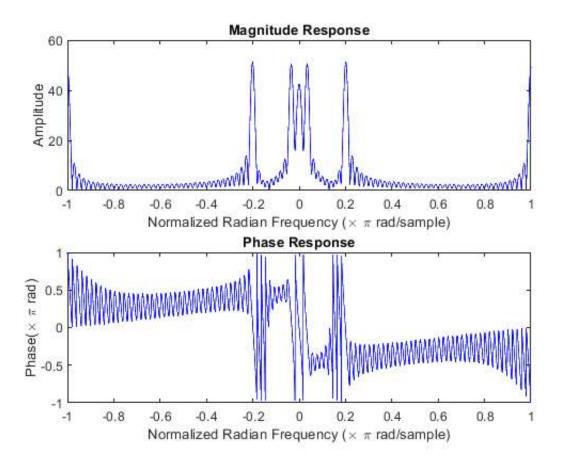
Lab 9

Exercise 9.1: (Effects of DFT size)

9.1a

```
x = [];
for n = 1:100
x(n) = 0.5 + \cos(pi*(n-1)/30) + \cos(pi*(n-1)/5) + \cos(pi*(n-1) + (2*pi/3));
end
type dtft
w = -pi:pi/2000:pi;
H = dtft(x, w);
figure(1);
subplot(2,1,1)
plot(w/pi,abs(H),'b-')
hold on;
title('Magnitude Response')
xlabel('Normalized Radian Frequency (\times \pi rad/sample)');
ylabel ('Amplitude');
subplot(2,1,2)
plot(w/pi,angle(H)/pi,'b-')
hold on;
title('Phase Response')
xlabel('Normalized Radian Frequency (\times \pi rad/sample)');
ylabel('Phase(\times \pi rad)');
```

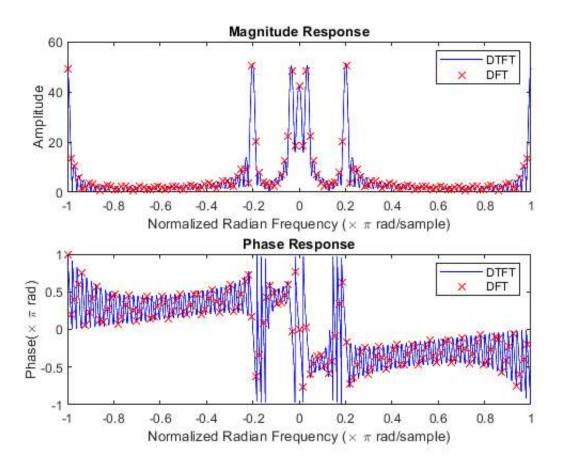
```
%dtft Calculates X(e^jw) using inputs impulse response vector and %frequency vector  X = zeros(1, length(w));  for k = 1: length(w) for r = 1: length(x)  X(k) = X(k) + (x(r) *exp(-1j*w(k) *(r-1)));  end end end end
```



9.1b

```
wfft = -pi:2*pi/128:pi-2*pi/128;
xfft = fft(x,128);
subplot(2,1,1)
plot(wfft/pi,fftshift(abs(xfft)),'rx');
hold off;
legend ('DTFT','DFT');
subplot(2,1,2)
plot(wfft/pi,fftshift(angle(xfft)/pi),'rx');
hold off;
legend ('DTFT','DFT');
% The DFT coeffecients are indeed frequency samples of the DTFT.
% The sample normalized radian frequency is 2*k/128 for the kth term.
```

```
% This can be seen as k0 is 0 while k1 is 0.01563 and k(-1) is -0.01562. 
% If you plug in k = 0, you get k0 = 0. 
% If you plug in k = 1, you get k1 = 2/128 = 0.015625 
% If you plug in k = -1, you get k(-1) = -2/128 = -0.015625
```

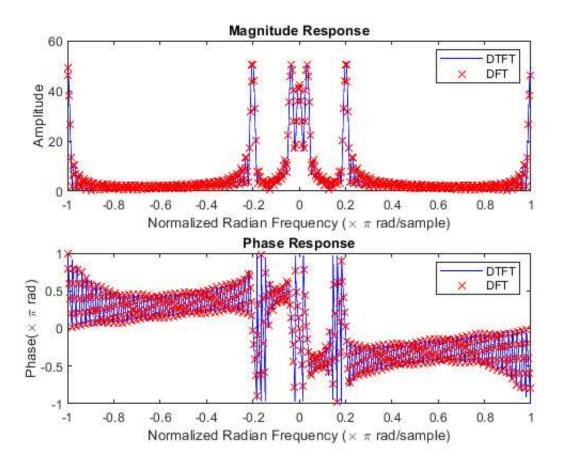


9.1c

```
wdft = -pi:2*pi/512:pi-2*pi/512;
xdft = fft(x, 512);
figure(2);
subplot(2,1,1)
plot(w/pi, abs(H), 'b-')
hold on;
plot(wdft/pi,fftshift(abs(xdft)),'rx');
hold off;
legend ('DTFT','DFT');
title('Magnitude Response')
xlabel('Normalized Radian Frequency (\times \pi rad/sample)');
ylabel ('Amplitude');
subplot(2,1,2)
plot(w/pi,angle(H)/pi,'b-')
hold on;
plot(wdft/pi,fftshift(angle(xdft)/pi),'rx');
hold off;
legend ('DTFT','DFT');
title('Phase Response')
```

```
xlabel('Normalized Radian Frequency (\times \pi rad/sample)');
ylabel('Phase(\times \pi rad)');

% The sample frequency from (b) is decreased by 4x for every point meaning
% that the frequency is now 2*k/512 for every kth term. The normalized radian frequency step
is 0.03906. As 512/128 = 4 this is how the sample
% frequency factor changes. If you use an even larger-size DFT, the
% sample frequency will become even smaller.
```



9.1d

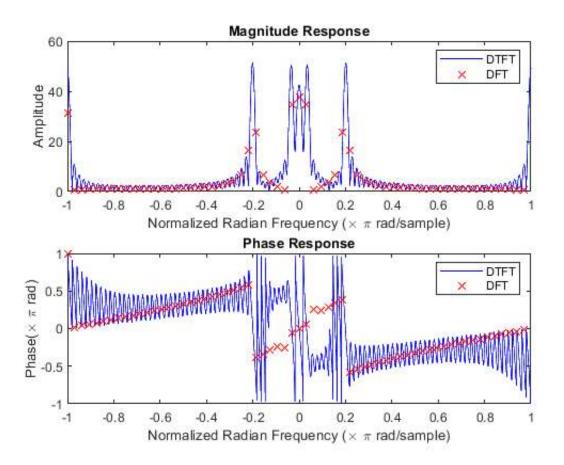
```
wdt = -pi:2*pi/64:pi-2*pi/64;
xdt = fft(x,64);

figure(3);
subplot(2,1,1)
plot(w/pi,abs(H),'b-')
hold on;
plot(wdt/pi,fftshift(abs(xdt)),'rx');
hold off;
legend ('DTFT','DFT');
title('Magnitude Response')
xlabel('Normalized Radian Frequency (\times \pi rad/sample)');
ylabel ('Amplitude');

subplot(2,1,2)
plot(w/pi,angle(H)/pi,'b-')
hold on;
```

```
plot(wdt/pi,fftshift(angle(xdt)/pi),'rx');
hold off;
legend ('DTFT','DFT');
title('Phase Response')
xlabel('Normalized Radian Frequency (\times \pi rad/sample)');
ylabel('Phase(\times \pi rad)');

% As shown in the figure, this DFT has a sample frequency that is twice as
% much as the 128 point DFT. The normalized radian frequency step for is
% 2/64 = 0.03125. However, many points for the DFT do not match the output
% of the DTFT. This is due to the size, which is 64 for the DFT,
% being smaller than the size of the signal, which is 100.
```



Exercise 9.2: (Frequency-domain analysis using FFT)

9.2a

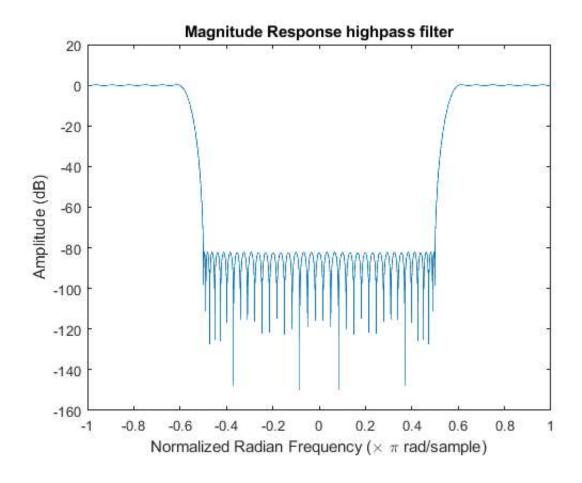
```
h
Hfilt = freqz(h,1,w);
figure(4);
plot(w/pi, 20*log10(abs(Hfilt)));
title('Magnitude Response highpass filter');
xlabel('Normalized Radian Frequency (\times \pi rad/sample)');
ylabel('Amplitude (dB)');

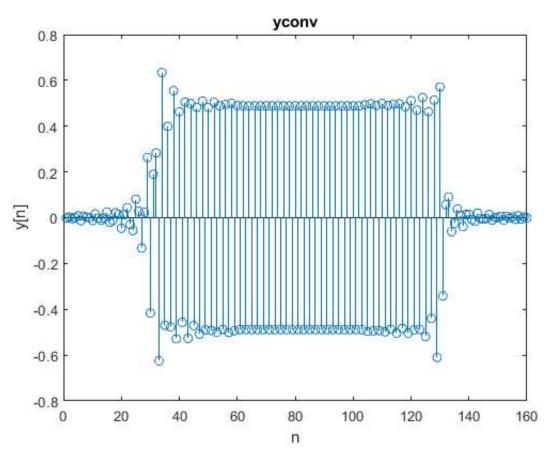
yconv = conv(x,h);
figure(5);
stem(1:length(yconv),yconv)
```

```
title('yconv');
xlabel('n');
ylabel('y[n]');

% yconv has length 160, which is 100+61-1 (L+N-1). The magnitude response
% of the filter shows that the filter oscillates at a low output for normalized
% frequency 0 to 0.5. At around n = 30 in yconv, the output rises and this
% corresponds to half the length of the filter which has length 61. For
% the next 100 samples (length of the signal), the magnitude of the output
% for yconv is higher than the previous lower frequencies. This
% demonstrates how the highpass filter affects the output. The remaining
% 30 points go back to oscillating at low output as it is half the length
% of the filter.
```

h =					
Columns 1 through 7					
-0.0003 0.0019	-0.0045	0.0055	-0.0024	-0.0029	0.0040
Columns 8 through 14					
0.0012 -0.0055	0.0013	0.0064	-0.0050	-0.0057	0.0094
Columns 15 through	21				
0.0027 -0.0135	0.0031	0.0159	-0.0118	-0.0150	0.0228
Columns 22 through	28				
0.0089 -0.0349	0.0049	0.0468	-0.0310	-0.0568	0.0848
Columns 29 through	35				
0.0634 -0.3109	0.4342	-0.3109	0.0634	0.0848	-0.0568
Columns 36 through	42				
-0.0310 0.0468	0.0049	-0.0349	0.0089	0.0228	-0.0150
Columns 43 through	49				
-0.0118 0.0159	0.0031	-0.0135	0.0027	0.0094	-0.0057
Columns 50 through	56				
-0.0050 0.0064	0.0013	-0.0055	0.0012	0.0040	-0.0029
Columns 57 through	61				
-0.0024 0.0055	-0.0045	0.0019	-0.0003		

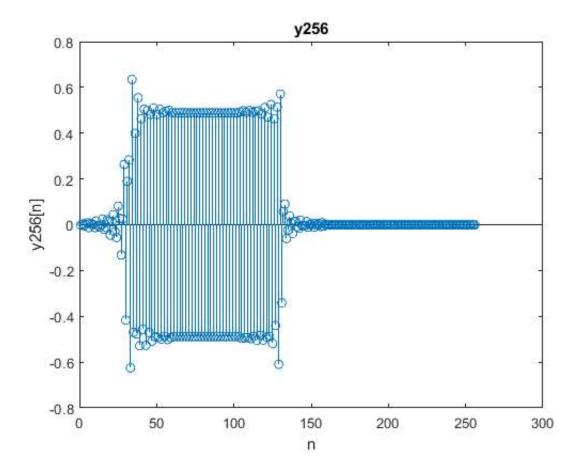




```
xd = fft(x,256);
hd = fft(h,256);
DFTy256 = xd.*hd;
y256 = ifft(DFTy256,256);

figure(6);
stem(1:length(y256),y256)
title('y256');
xlabel('n');
ylabel('y256[n]');

% The length of y256 is 256, hence its name while the length of yconv is
% 160. The output of each is identical for the
% first 160 points, which is the shorter of the two lengths. The points
% after this are virtually 0 for y256 while yconv does not have output,
% which shows that the output of each is the same.
```

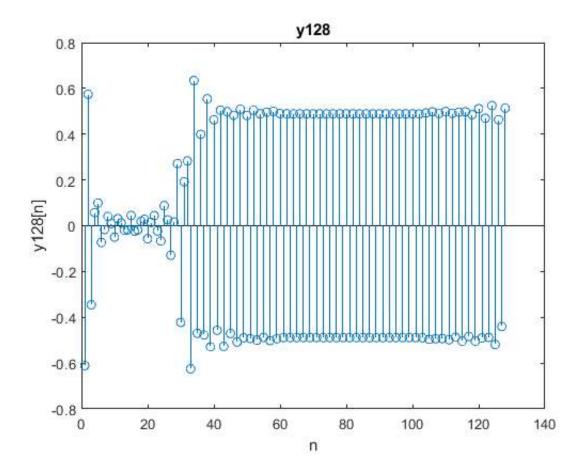


9.2c

```
xx = fft(x,128);
hx = fft(h,128);
DFTy128 = xx.*hx;
y128 = ifft(DFTy128,128);

figure(7);
stem(1:length(y128),y128)
title('y128');
xlabel('n');
```

```
ylabel('y128[n]');
% y128 is not the same as yconv. y128 only has length 128 while yconv has
% length 160. Additionally, the first 30 inputs of y128 are not as low
% output as a highpass filtered equation should function, such as yconv. The next ~100
% points are basically the same for the higher frequencies however.
```



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