

Visual Recognition: Filtering and Transformations

Raquel Urtasun

TTI Chicago

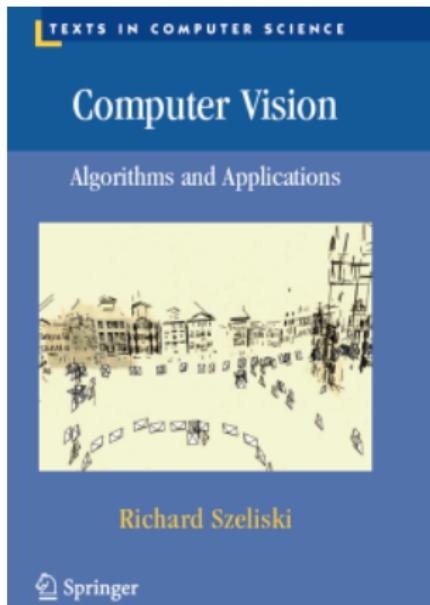
Jan 15, 2012

Today's lecture ...

- More on Image Filtering
- Additional transformations

Readings

- Chapter 2 and 3 of Rich Szeliski's book



- Available online [here](#)

Image Sub-Sampling

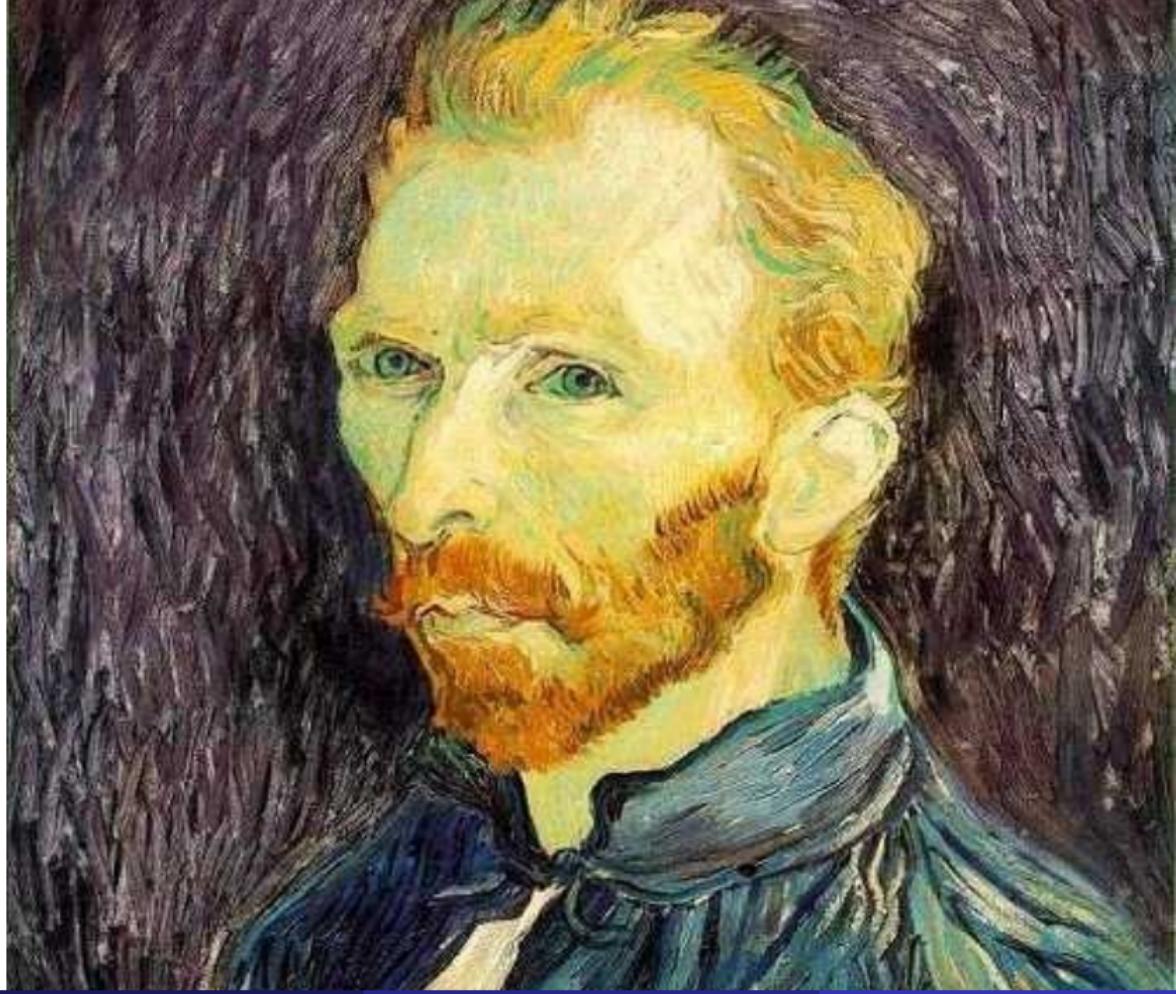
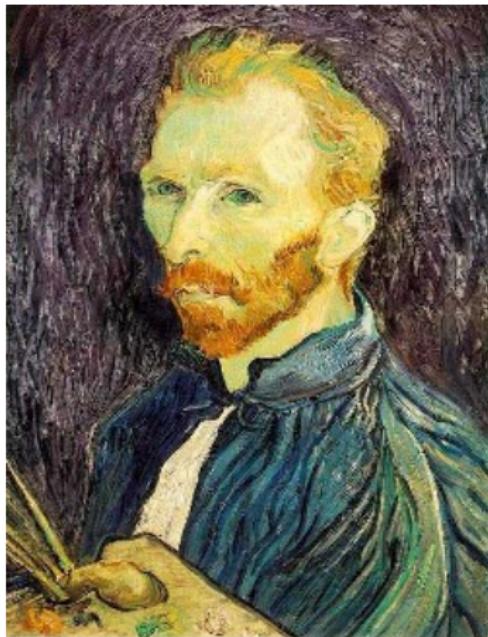


Image Sub-Sampling

- Throw away every other row and column to create a $1/2$ size image



1/4

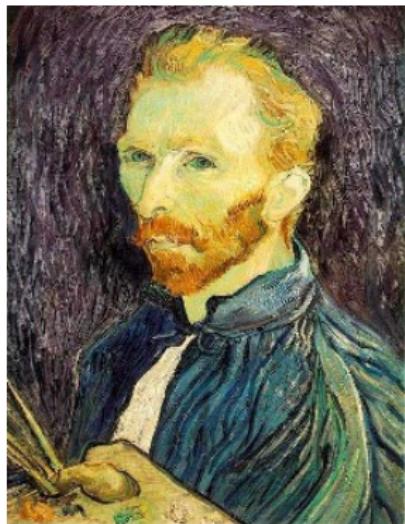


1/8

[Source: S. Seitz]

Image Sub-Sampling

- Why does this look so cruffy?



1/2



1/4 (2x zoom)



1/8 (4x zoom)

[Source: S. Seitz]

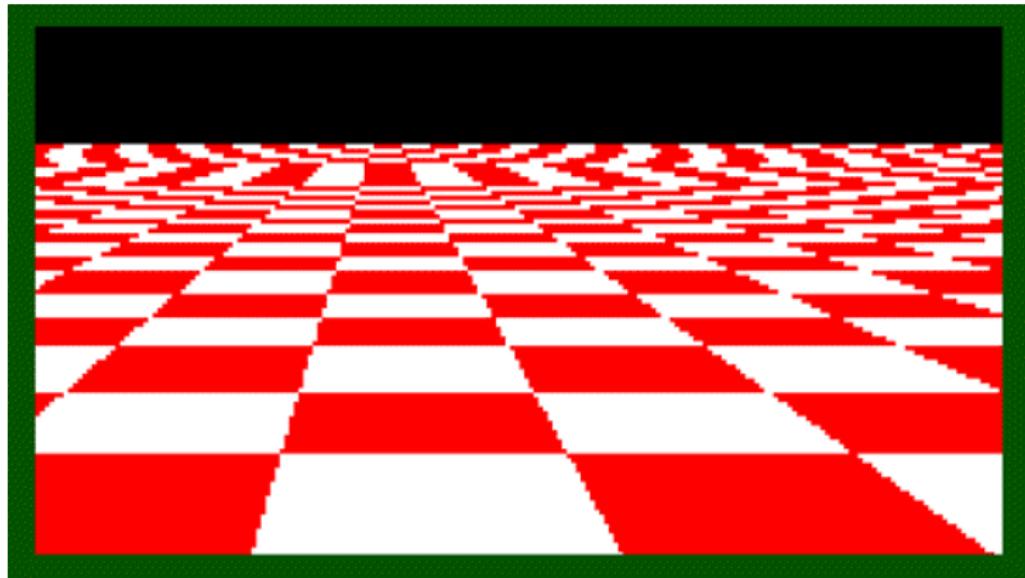
Image Sub-Sampling



[Source: F. Durand]

Even worse for synthetic images

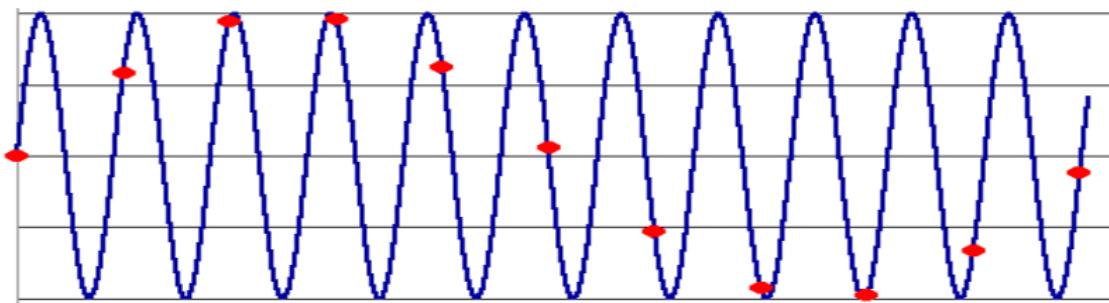
- What's happening?



[Source: L. Zhang]

Aliasing

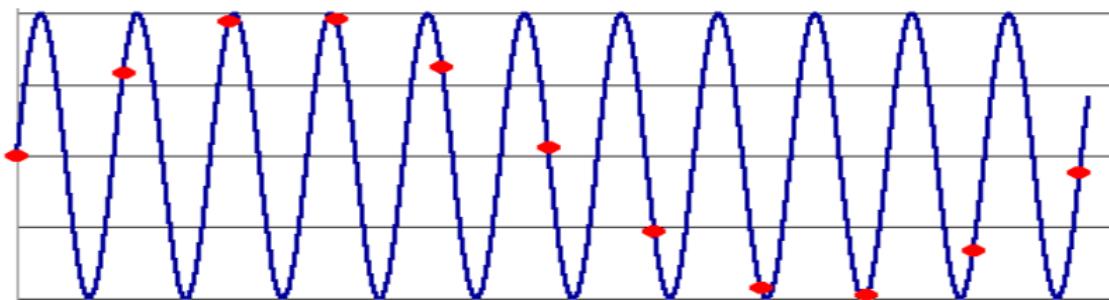
- Occurs when your sampling rate is not high enough to capture the amount of detail in your image



- To do sampling right, need to understand the structure of your signal/image

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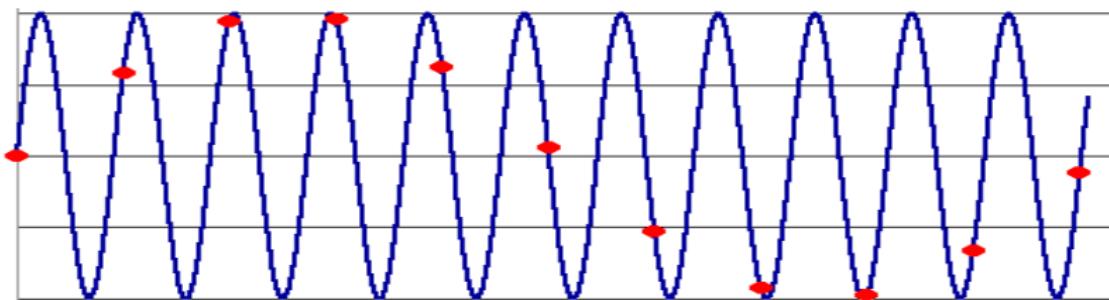
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- The minimum sampling rate is called the **Nyquist rate**

Aliasing problems

- **Shannons Sampling Theorem** shows that the minimum sampling

$$f_s \geq 2f_{max}$$

- If you haven't seen this... take a class on Fourier analysis... everyone should have at least one!

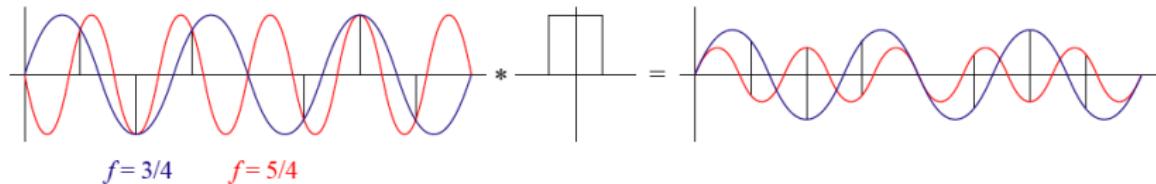
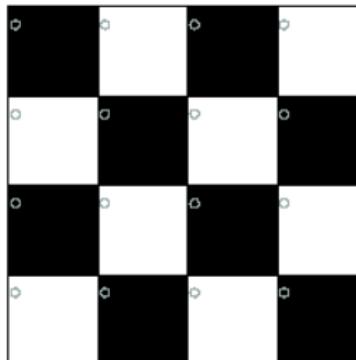
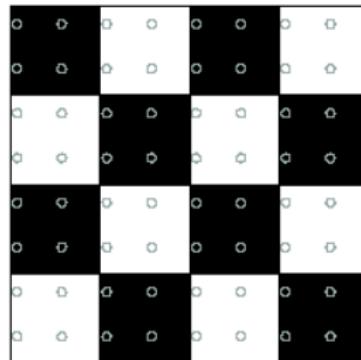
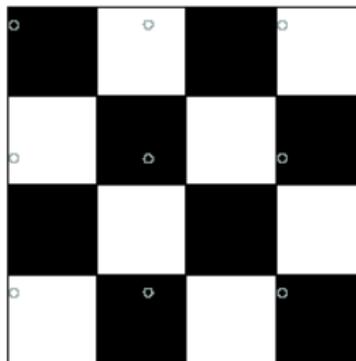
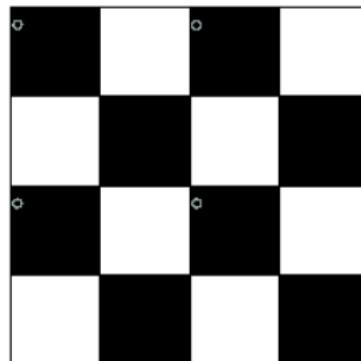


Figure: example of a 1D signal [R. Szeliski et al.]

Nyquist limit 2D example



Good sampling



Bad sampling

[Source: N. Snavely]

Going back to Downsampling ...

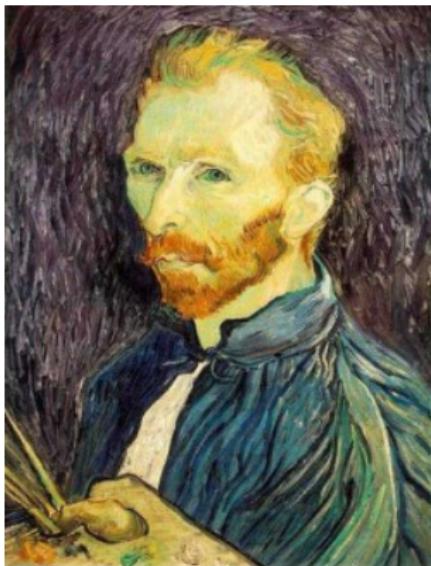
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- How can we fix this?

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Gaussian pre-filtering

- Solution: filter the image, then subsample



Gaussian 1/2



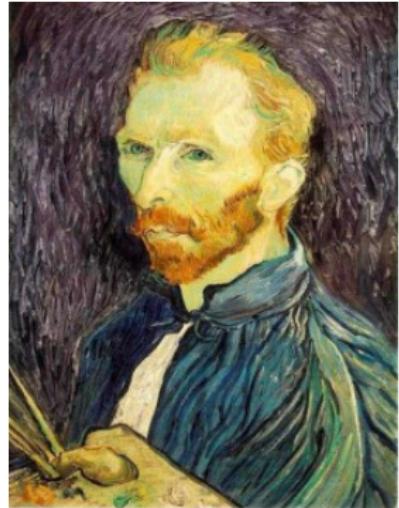
G 1/4



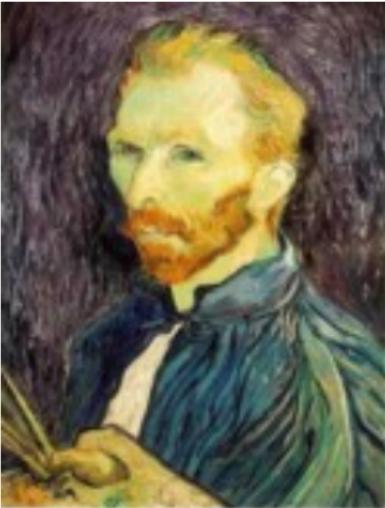
G 1/8

[Source: S. Seitz]

Subsampling with Gaussian pre-filtering



Gaussian 1/2



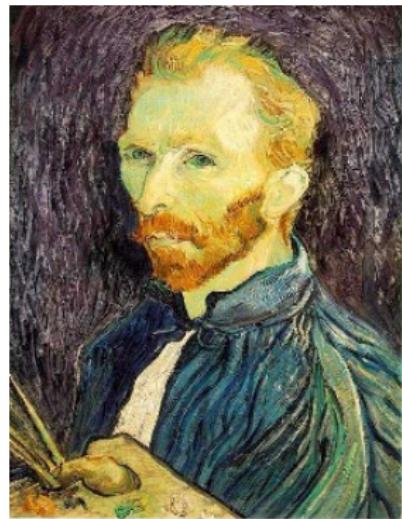
G 1/4



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[Source: S. Seitz]

Compare with ...



1/2



1/4 (2x zoom)



1/8 (4x zoom)

[Source: S. Seitz]

And in 2D...

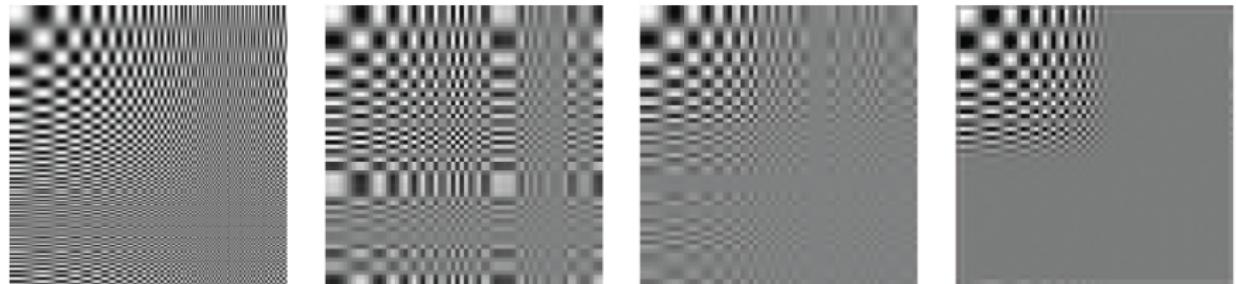
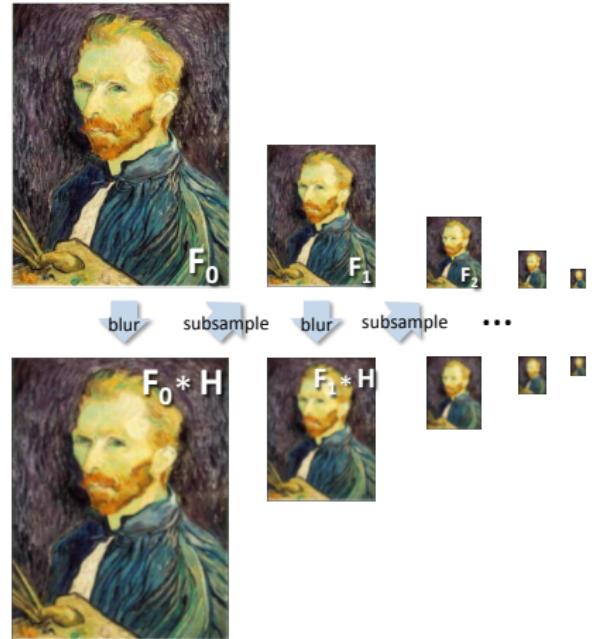


Figure: (a) Example of a 2D signal. (b-d) downsampled with different filters

[Source: R. Szeliski]

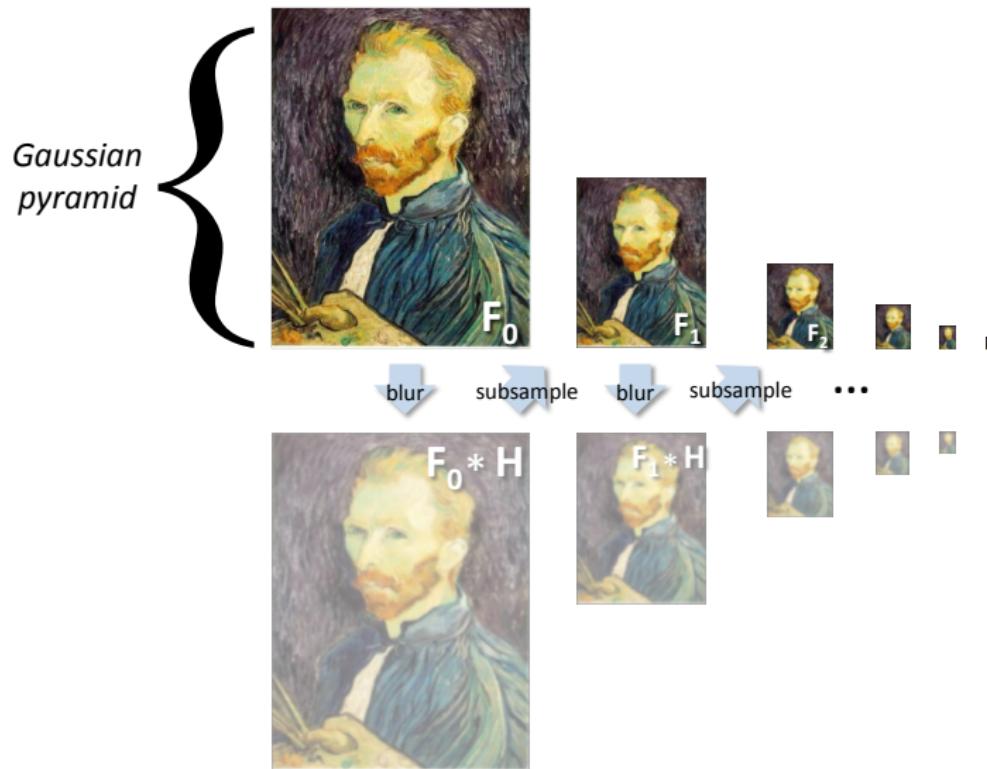
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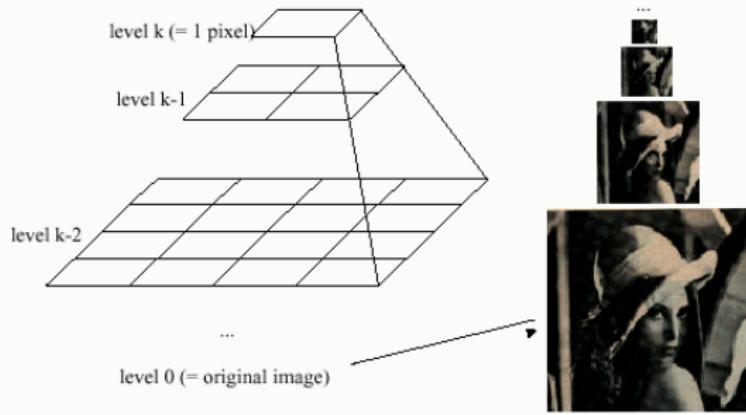
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Gaussian Pyramids [Burt and Adelson, 1983]

- In computer graphics, a *mip map* [Williams, 1983]
- A precursor to wavelet transform

Idea: Represent $N \times N$ image as a "pyramid" of $1 \times 1, 2 \times 2, 4 \times 4, \dots, 2^k \times 2^k$ images (assuming $N = 2^k$)



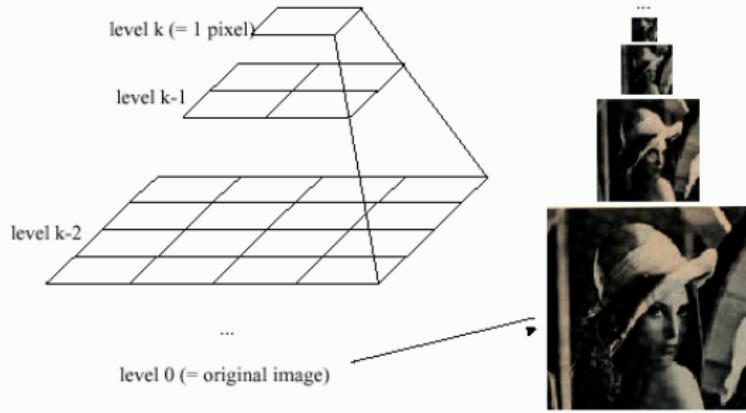
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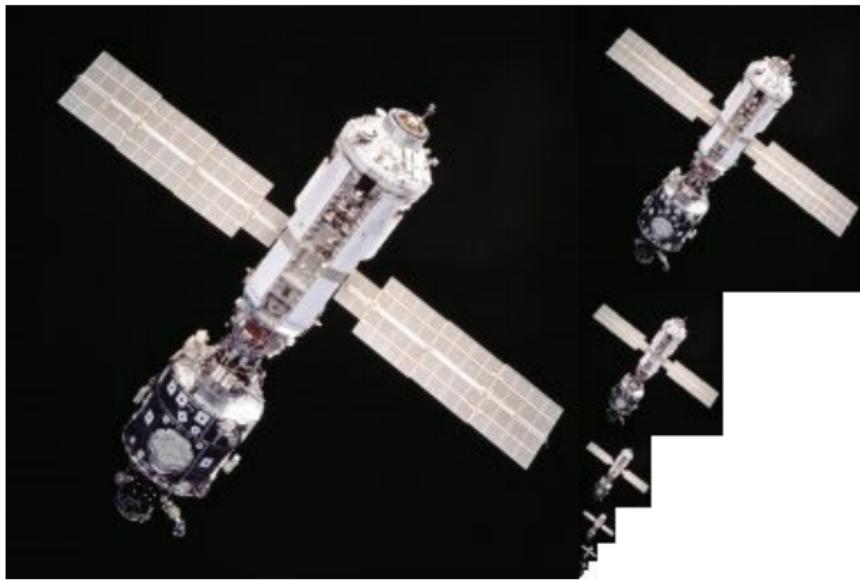
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[Source: S. Seitz]

Example of Gaussian Pyramid



[Source: N. Snavely]

Decimation or Sub-sampling

- **Decimation:** reduces resolution

$$g(i, j) = \sum_{k,l} f(k, l) h(i - k/r, j - l/r)$$

with r the down-sampling rate.

- Different filters exist to do this.

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Image Up-Sampling

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- This image is too small, how can we make it 10 times as big?

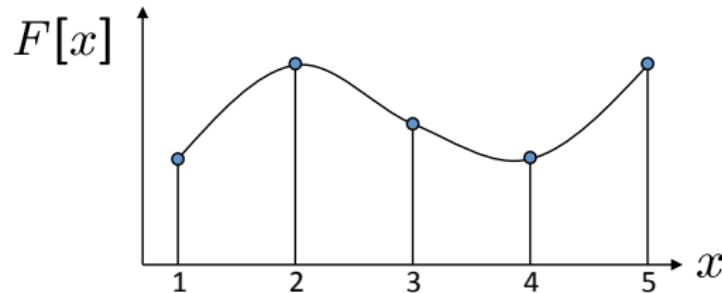


- Simplest approach: repeat each row and column 10 times (Nearest neighbor interpolation)



[Source: N. Snavely]

Image Interpolation



$d = 1$ in this example

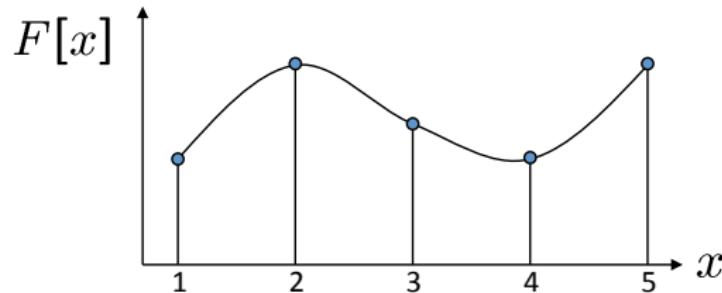
Recall how a digital image is formed

$$F[x, y] = \text{quantize}\{f(xd, yd)\}$$

- It is a discrete point-sampling of a continuous function
- If we could somehow reconstruct the original function, any new image could be generated, at any resolution and scale

[Source: N. Snavely, S. Seitz]

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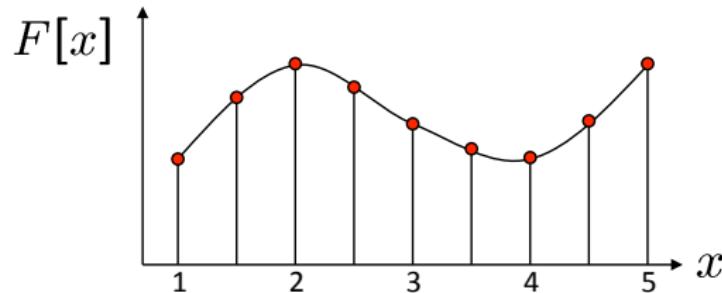
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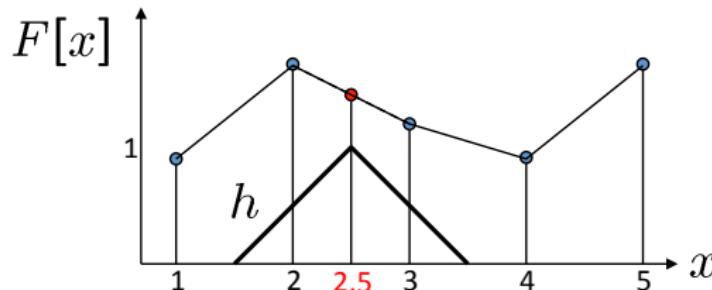
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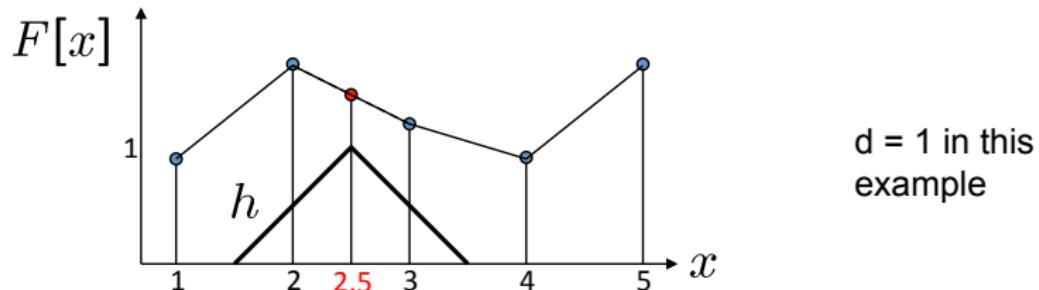


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- Guess an approximation: Can be done in a principled way via filtering

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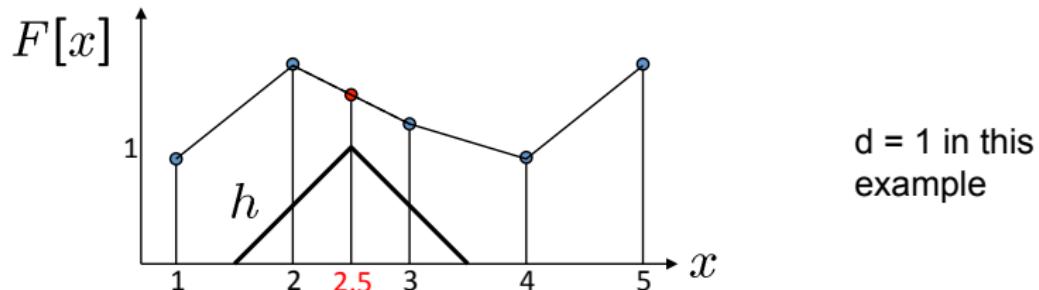


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$$f_F(x) = \begin{cases} F\left(\frac{x}{d}\right) & \text{if } \frac{x}{d} \text{ is an integer} \\ 0 & \text{otherwise} \end{cases}$$

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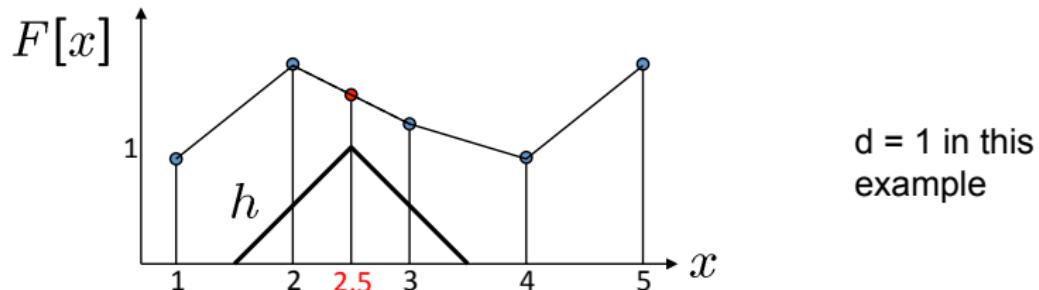
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$$\hat{f} = h * f_F$$

[Source: N. Snavely, S. Seitz]

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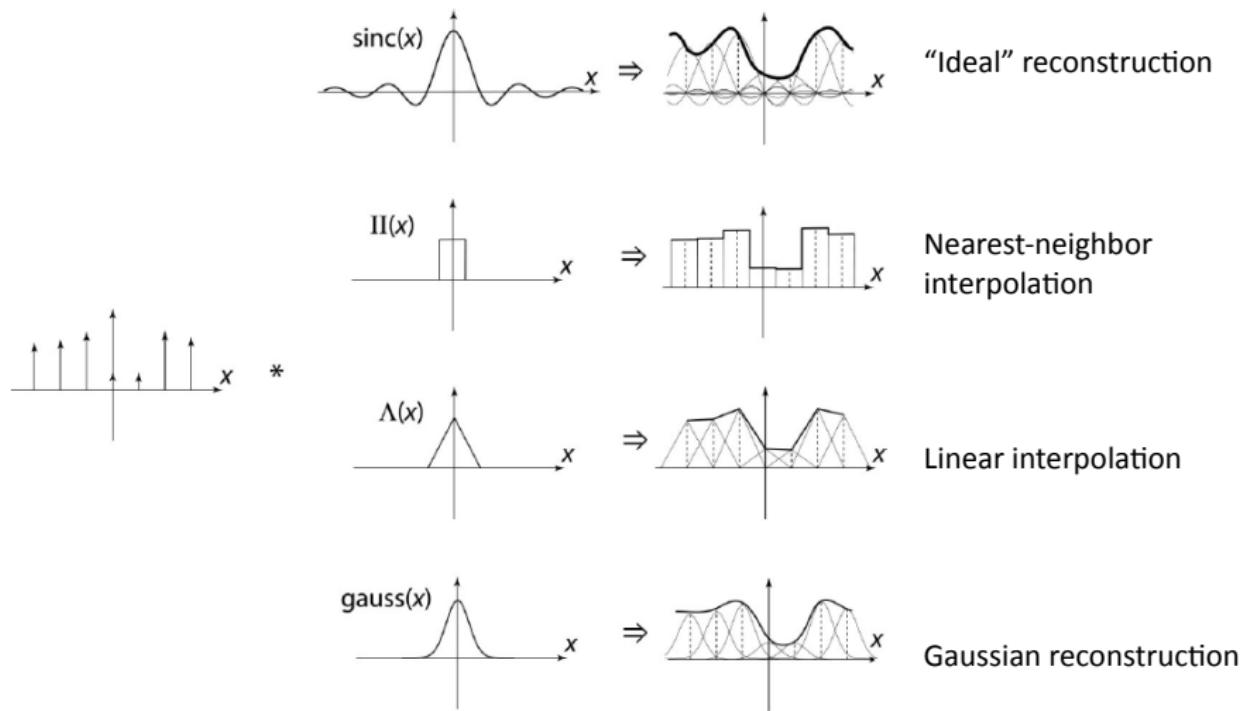
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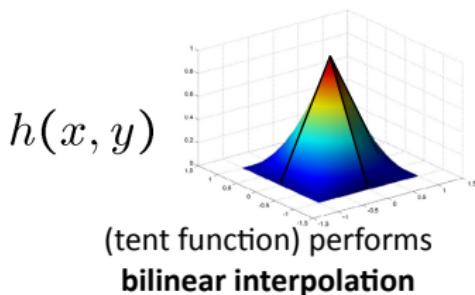
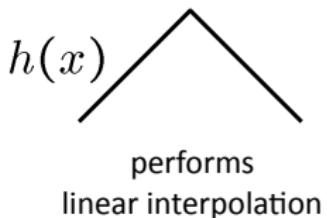
Image Interpolation



Source: B. Curless

Reconstruction filters

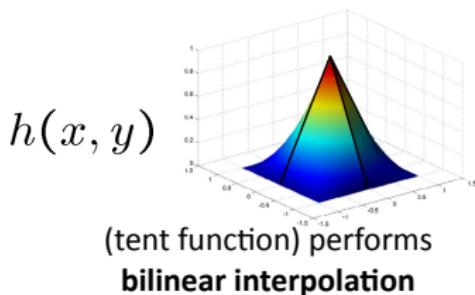
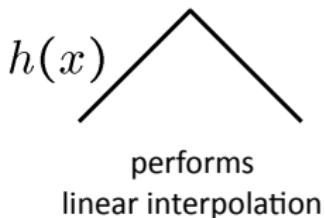
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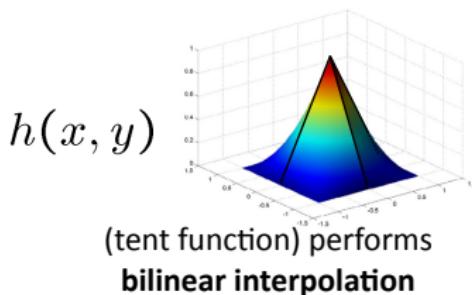
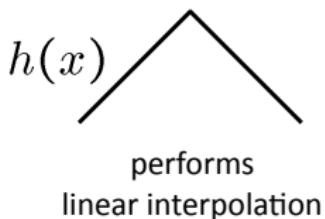
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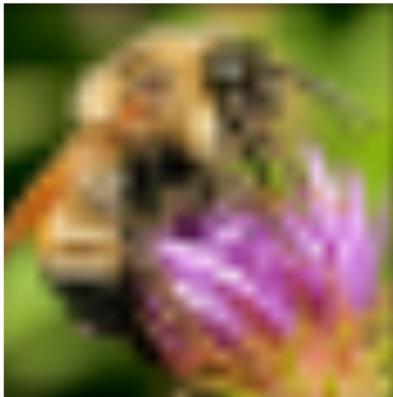
Original image



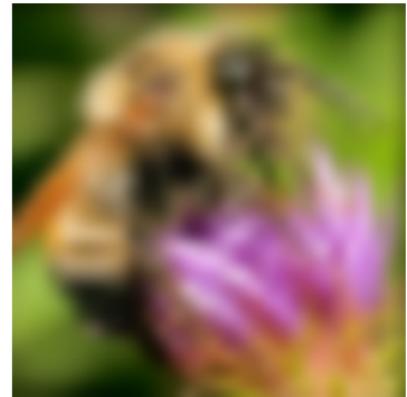
Interpolation results



Nearest-neighbor interpolation



Bilinear interpolation



Bicubic interpolation

[Source: N. Snavely]

Image Interpolation

What operation have we done?

Also used for *resampling*



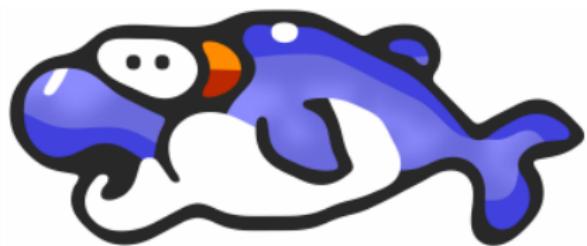
[Source: N. Snavely]

Depixelating Pixel Art

- Published by [Kopt et al., SIGGRAPH 2011]



Nearest-neighbor result (original: 40×16 pixels)



Our result

More Examples



When are Pyramids Useful?

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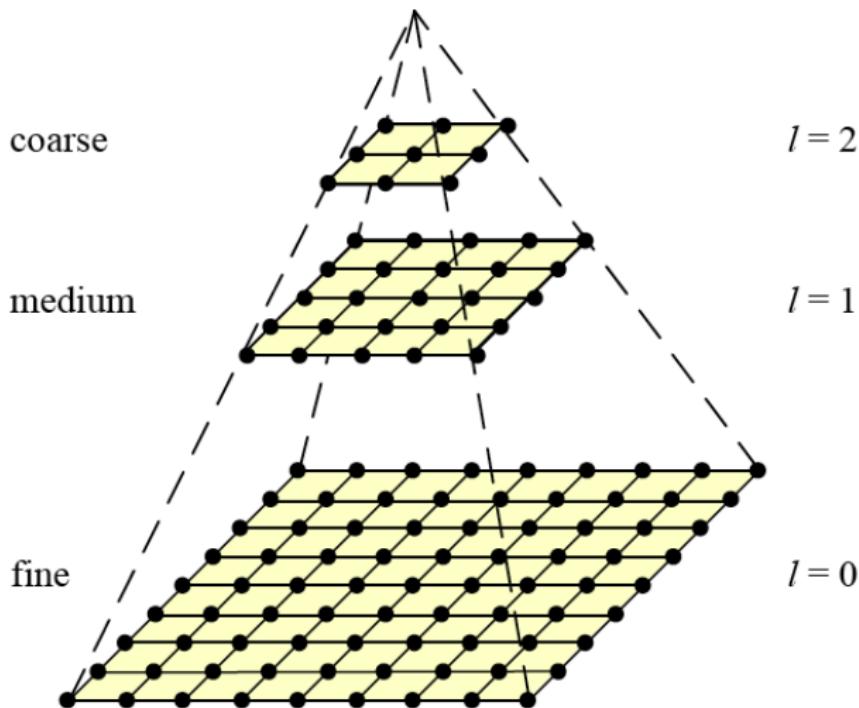
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Image Pyramid



[Source: R. Szeliski]

Interpolation and Decimation

- To **interpolate** (or upsample) an image to a higher resolution, we need to select an **interpolation kernel** with which to convolve the image

$$g(i,j) = \sum_{k,l} f(k,l)h(i - rk, j - rl)$$

with r the up-sampling rate.

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Multi-Resolution Representations

The most used one is the **Laplacian pyramid**:

- We first **blur** and **subsample** the original image by a factor of two and store this in the next level of the pyramid.
- Subtract then this low-pass version from the original to yield the **band-pass Laplacian image**.

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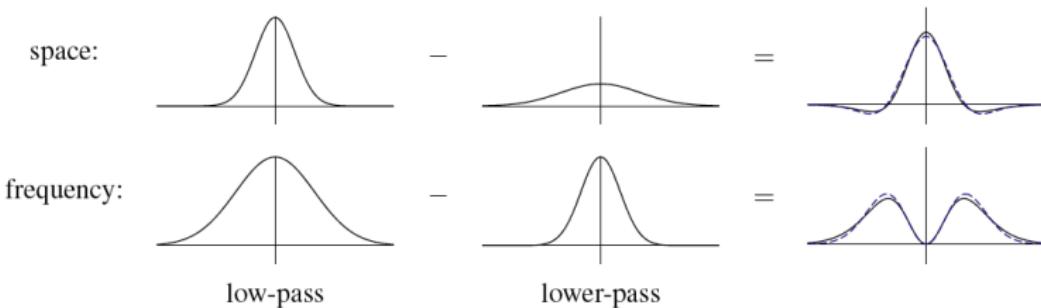
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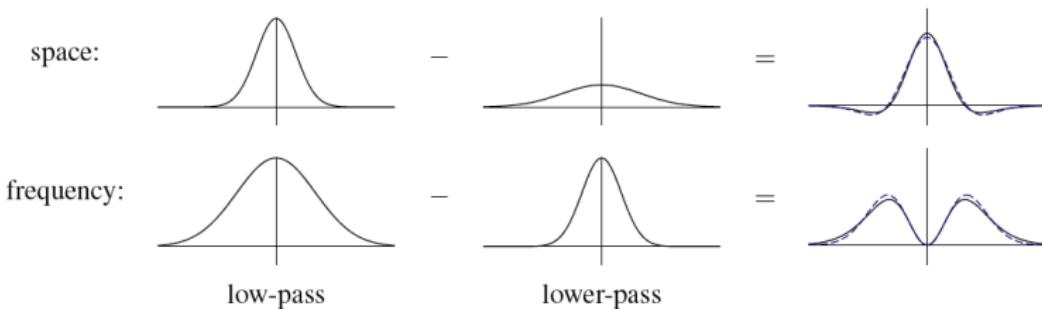


[Source: R. Szeliski]

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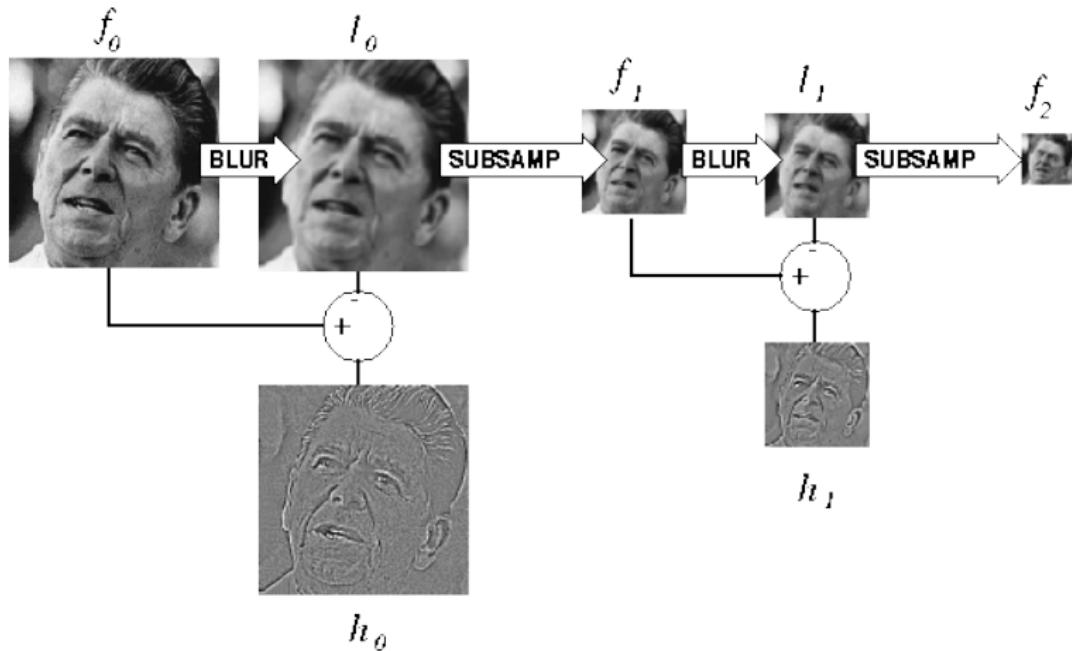
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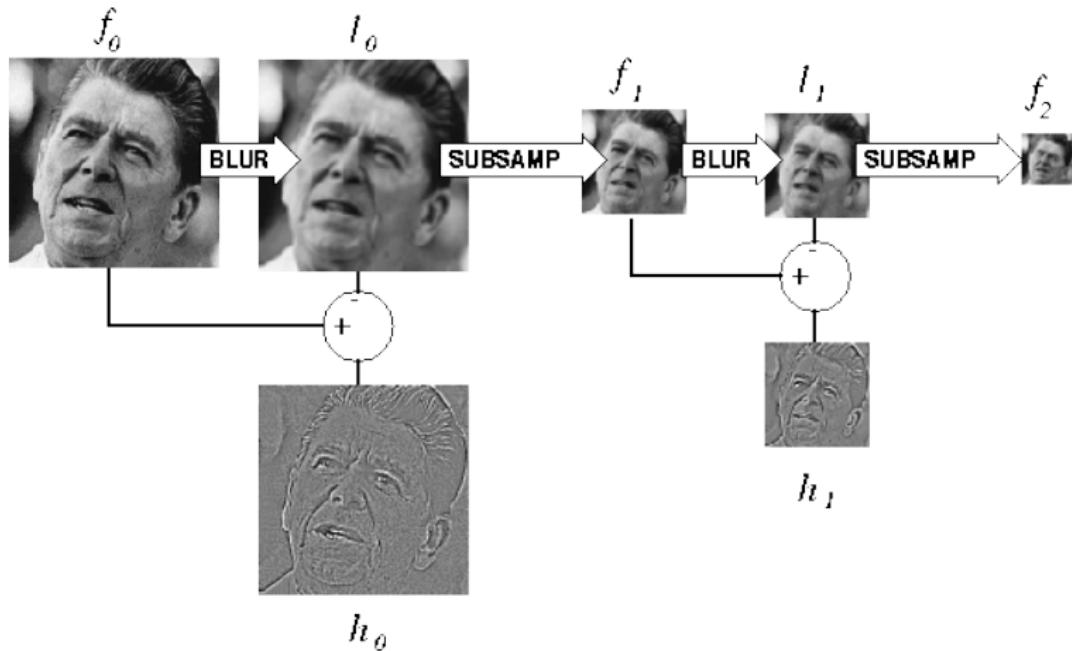
[Source: R. Szeliski]

Laplacian Pyramid Construction



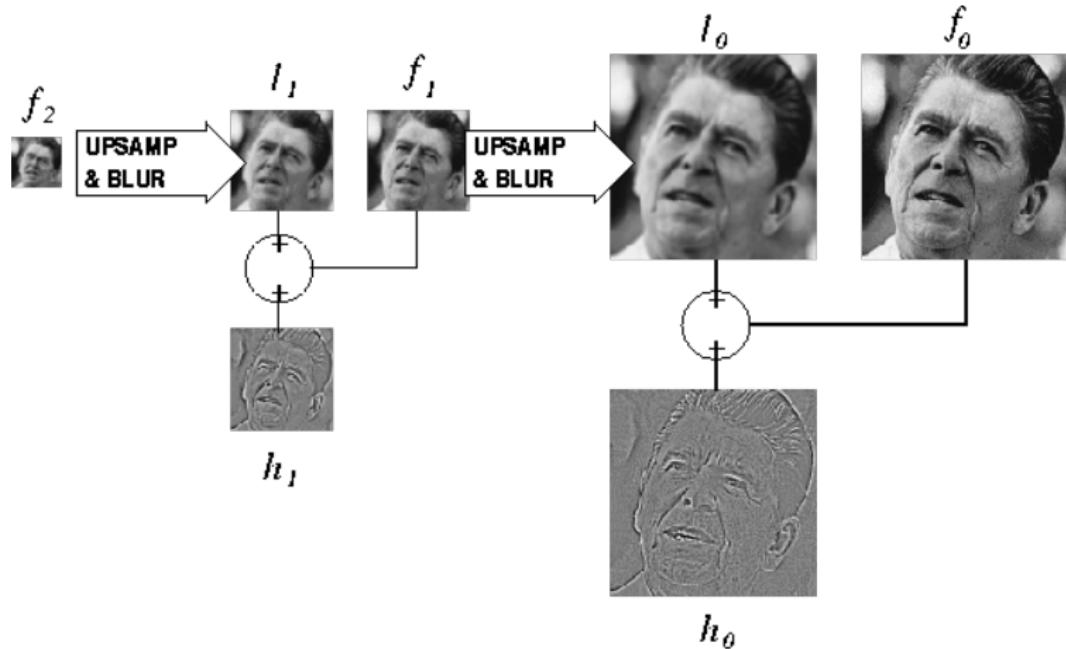
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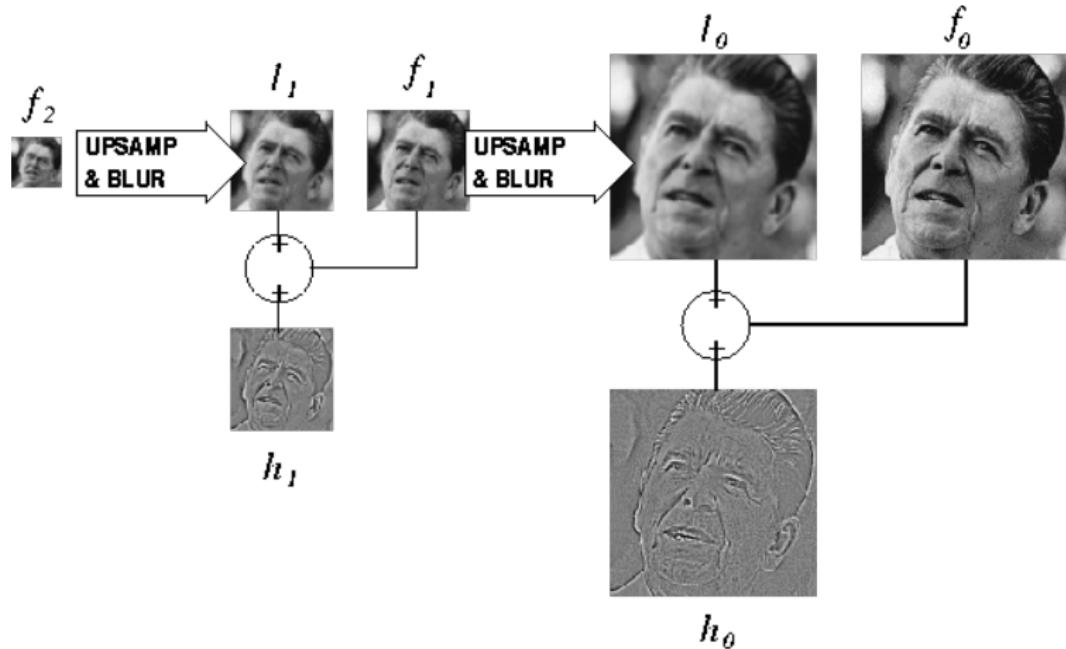
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Laplacian Pyramid Re-construction



- When is this useful?

Laplacian Pyramid Re-construction



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More Complex Filters

Steerable Filters

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Example of Steerable Filter

- 2D symmetric Gaussian with $\sigma = 1$ and assume constant is 1

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$$G_1^0 = \frac{\partial}{\partial x} \exp(-x^2 + y^2) = -2x \exp(-x^2 + y^2)$$

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$$G_1^\theta = \cos \theta G_1^0 + \sin \theta G_1^{90}$$

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- Because convolution is a linear operation, we can synthesize an image filtered at an arbitrary orientation by taking linear combinations of the images filtered with G_1^0 and G_1^{90}

$$\text{if } R_1^0 = G_1^0 * I \text{ and } R_1^{90} = G_1^{90} * I \text{ then } R_1^\theta = \cos \theta R_1^0 + \sin \theta R_1^{90}$$

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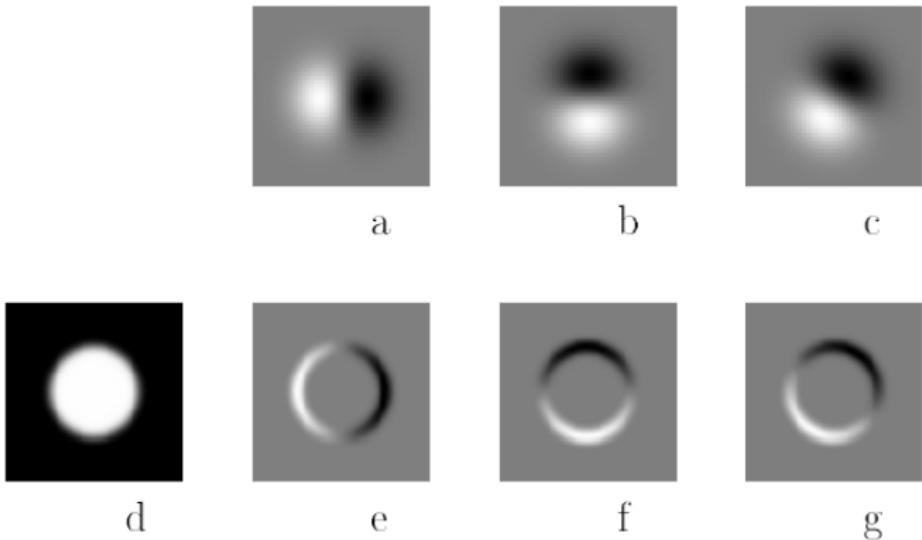


Figure 2-1: Example of steerable filters. (a) $G_1^{0^\circ}$, first derivative with respect to x (horizontal) of a Gaussian. (b) $G_1^{90^\circ}$, which is $G_1^{0^\circ}$, rotated by 90° . From a linear combination of these two filters, one can create G_1^θ , an arbitrary rotation of the first derivative of a Gaussian. (c) $G_1^{30^\circ}$, formed by $\frac{1}{2}G_1^{0^\circ} + \frac{\sqrt{3}}{2}G_1^{90^\circ}$. The same linear combinations used to synthesize G_1^θ from the basis filters will also synthesize the response of an image to G_1^θ from the responses of the image to the basis filters: (d) Image of circular disk. (e) $G_1^{0^\circ}$ (at a smaller scale than pictured above) convolved with the disk, (d). (f) $G_1^{90^\circ}$ convolved with (d). (g) $G_1^{30^\circ}$ convolved with (d), obtained from $\frac{1}{2}$ [image e] + $\frac{\sqrt{3}}{2}$ [image f].

[Source: W. Freeman 91]

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What about the second order derivative?

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$$G_{\hat{\mathbf{u}}\hat{\mathbf{u}}} = u^2 G_{xx} + 2uv G_{x,y} + v^2 G_{y,y}$$

with $\hat{\mathbf{u}} = (u, v)$

Other transformations

Integral Images

- If an image is going to be repeatedly convolved with different box filters, it is useful to compute the **summed area table**.
- It is the running sum of all the pixel values from the origin

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- To find the summed area (integral) inside a rectangle $[i_0, i_1] \times [j_0, j_1]$ we simply combine four samples from the summed area table.

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Example of Integral Images

| | | | | |
|---|----------|----------|----------|---|
| 3 | 2 | 7 | 2 | 3 |
| 1 | 5 | 1 | 3 | 4 |
| 5 | 1 | 3 | 5 | 1 |
| 4 | 3 | 2 | 1 | 6 |
| 2 | 4 | 1 | 4 | 8 |

(a) $S = 24$

| | | | | |
|----|-----------|-----------|----|----|
| 3 | 5 | 12 | 14 | 17 |
| 4 | 11 | 19 | 24 | 31 |
| 9 | 17 | 28 | 38 | 46 |
| 13 | 24 | 37 | 48 | 62 |
| 15 | 30 | 44 | 59 | 81 |

(b) $s = 28$

| | | | | |
|-----------|----|----|-----------|----|
| 3 | 5 | 12 | 14 | 17 |
| 4 | 11 | 19 | 24 | 31 |
| 9 | 17 | 28 | 38 | 46 |
| 13 | 24 | 37 | 48 | 62 |
| 15 | 30 | 44 | 59 | 81 |

(c) $S = 24$

Figure 3.17 Summed area tables: (a) original image; (b) summed area table; (c) computation of area sum. Each value in the summed area table $s(i, j)$ (red) is computed recursively from its three adjacent (blue) neighbors (3.31). Area sums S (green) are computed by combining the four values at the rectangle corners (purple) (3.32). Positive values are shown in **bold** and negative values in *italics*.

Non-linear filters: Median filter

- We have seen **linear filters**, i.e., their response to a sum of two signals is the same as the sum of the individual responses.

$$h \circ (f + g) = h \circ f + h \circ g$$

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$$w(i,j,k,l) = \exp \left(-\frac{(i-k)^2 + (j-l)^2}{2\sigma_d^2} - \frac{\|f(i,j) - f(k,l)\|^2}{2\sigma_r^2} \right)$$

composed of the **domain kernel** and the **range kernel**.

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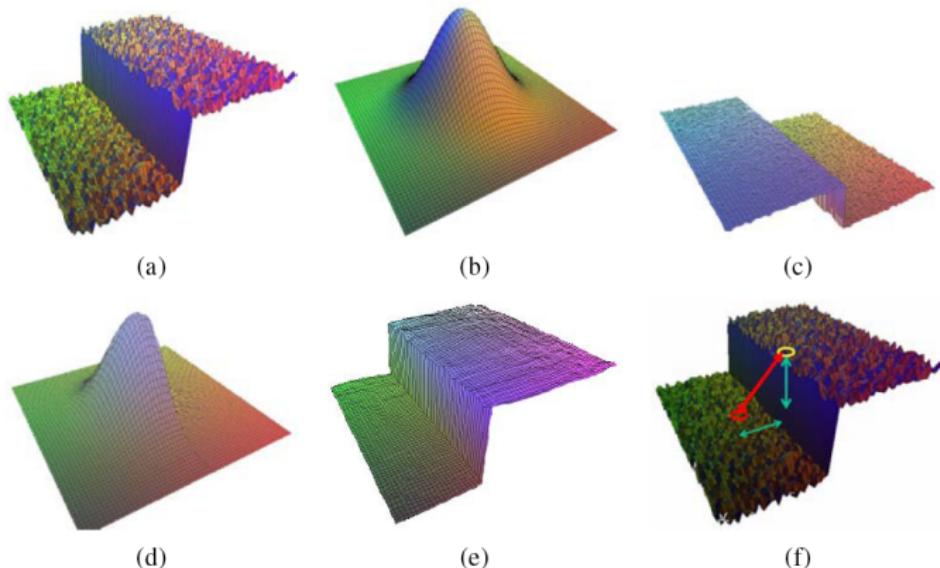


Figure: Bilateral filtering [Durand & Dorsey, 02]. (a) noisy step edge input. (b) domain filter (Gaussian). (c) range filter (similarity to center pixel value). (d) bilateral filter. (e) filtered step edge output. (f) 3D distance between pixels

[Source: R. Szeliski]

Distance Transform

- Useful to quickly precomputing the distance to a curve or a set of points.
- Let $d(k, l)$ be some distance metric between pixel offsets, e.g., Manhattan distance

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| | | | | | | | |
|---|---|---|---|---|---|---|---|
| 0 | 0 | 0 | 0 | 0 | 1 | 0 | 0 |
| 0 | 0 | 1 | 1 | 1 | 1 | 0 | 0 |
| 0 | 1 | 1 | 1 | 1 | 1 | 1 | 0 |
| 0 | 1 | 1 | 1 | 1 | 1 | 1 | 0 |
| 0 | 1 | 1 | 1 | 1 | 0 | 0 | 0 |
| 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

(a)

| | | | | | | | |
|---|---|---|---|---|---|---|---|
| 0 | 0 | 0 | 0 | 1 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 | 2 | 0 | 0 | 0 |
| 0 | 1 | 2 | 2 | 3 | 1 | 0 | 0 |
| 0 | 1 | 2 | 3 | | | | |
| | | | | | | | |
| | | | | | | | |
| | | | | | | | |
| | | | | | | | |

(b)

| | | | | | | | |
|---|---|---|---|---|---|---|---|
| 0 | 0 | 0 | 0 | 1 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 | 2 | 0 | 0 | 0 |
| 0 | 1 | 2 | 2 | 3 | 1 | 0 | 0 |
| 0 | 1 | 2 | 3 | 1 | 1 | 0 | 0 |
| 0 | 1 | 2 | 1 | 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

(c)

| | | | | | | | |
|---|---|---|---|---|---|---|---|
| 0 | 0 | 0 | 0 | 1 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 | 1 | 0 | 0 | 0 |
| 0 | 1 | 2 | 2 | 2 | 1 | 0 | 0 |
| 0 | 1 | 2 | 2 | 1 | 1 | 0 | 0 |
| 0 | 1 | 2 | 1 | 1 | 1 | 0 | 0 |
| 0 | 1 | 2 | 1 | 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

(d)

Figure: City block distance transform: (a) original binary image; (b) top to bottom (forward) raster sweep: green values are used to compute the orange value; (c) bottom to top (backward) raster sweep: green values are merged with old orange value; (d) final distance transform.

[Source: R. Szeliski]

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| | | | | | | |
|---|---|---|---|---|---|---|
| 0 | 0 | 0 | 0 | 1 | 0 | 0 |
| 0 | 0 | 1 | 1 | 1 | 0 | 0 |
| 0 | 1 | 1 | 1 | 1 | 1 | 0 |
| 0 | 1 | 1 | 1 | 1 | 1 | 0 |
| 0 | 1 | 1 | 1 | 0 | 0 | 0 |
| 0 | 0 | 1 | 0 | 0 | 0 | 0 |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 |

(a)

| | | | | | | |
|---|---|---|---|---|---|---|
| 0 | 0 | 0 | 0 | 1 | 0 | 0 |
| 0 | 0 | 1 | 1 | 2 | 0 | 0 |
| 0 | 1 | 2 | 2 | 3 | 1 | 0 |
| 0 | 1 | 2 | 3 | | | |
| | | | | | | |
| | | | | | | |
| | | | | | | |

(b)

| | | | | | | |
|---|---|---|---|---|---|---|
| 0 | 0 | 0 | 0 | 1 | 0 | 0 |
| 0 | 0 | 1 | 1 | 2 | 0 | 0 |
| 0 | 1 | 2 | 2 | 3 | 1 | 0 |
| 0 | 1 | 2 | 3 | 1 | 1 | 0 |
| 0 | 1 | 2 | 1 | 0 | 0 | 0 |
| 0 | 0 | 1 | 0 | 0 | 0 | 0 |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 |

(c)

| | | | | | | |
|---|---|---|---|---|---|---|
| 0 | 0 | 0 | 0 | 1 | 0 | 0 |
| 0 | 0 | 1 | 1 | 1 | 0 | 0 |
| 0 | 1 | 2 | 2 | 2 | 1 | 0 |
| 0 | 1 | 2 | 2 | 1 | 1 | 0 |
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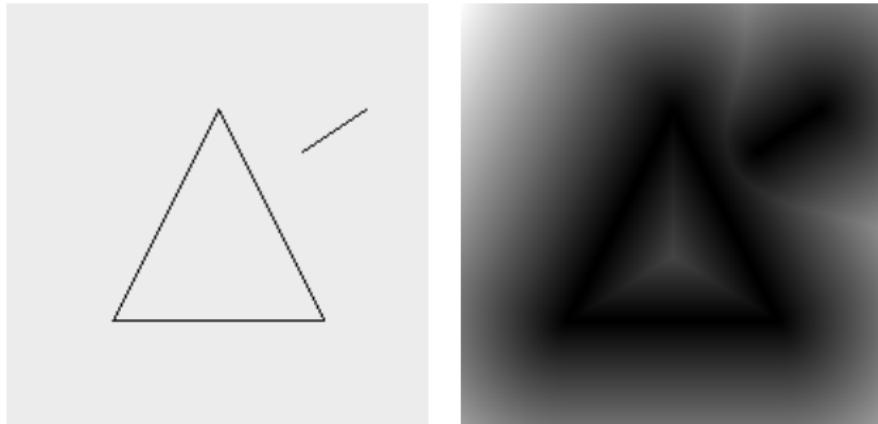
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Figure: City block distance transform: (a) original binary image; (b) top to bottom (forward) raster sweep: green values are used to compute the orange value; (c) bottom to top (backward) raster sweep: green values are merged with old orange value; (d) final distance transform.

[Source: R. Szeliski]

Example of Distance Transform

- More complicated in the Euclidean case.
- Example of a distance transform



- The ridges is the **skeleton** or **medial axis**.
- Extension: Signed distance transform.

[Source: P. Felzenszwalb]

Fourier Transform

- Fourier analysis could be used to analyze the frequency characteristics of various filters.
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- Pass a sinusoid of known frequency through the filter and to observe by how much it is attenuated

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Filtering and Fourier

- Convolution can be expressed as a weighted summation of shifted input signals (sinusoids); so it is just a single sinusoid at that frequency.

$$o(x) = h(x) * s(x) = A \sin(\omega x + \phi_o)$$

A is the **gain** or **magnitude** of the filter, while the phase difference $\Delta\phi = \phi_o - \phi_i$ is the **shift** or **phase**

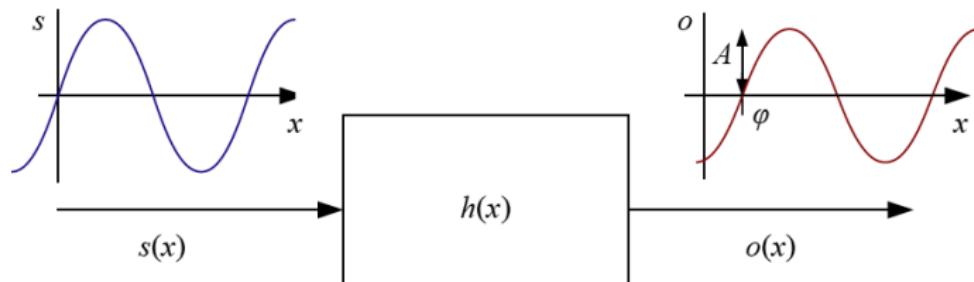


Figure 3.24 The Fourier Transform as the response of a filter $h(x)$ to an input sinusoid $s(x) = e^{j\omega x}$ yielding an output sinusoid $o(x) = h(x) * s(x) = A e^{j\omega x + \phi}$.

Complex notation

- The sinusoid is express as $s(x) = e^{j\omega x} = \cos \omega x + j \sin \omega x$ and the filter sinusoid as

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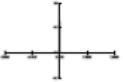
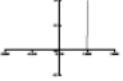
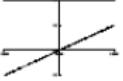
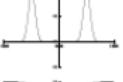
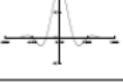
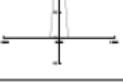
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Properties Fourier Transform

| Property | Signal | Transform |
|--------------------|---------------------|--|
| superposition | $f_1(x) + f_2(x)$ | $F_1(\omega) + F_2(\omega)$ |
| shift | $f(x - x_0)$ | $F(\omega)e^{-j\omega x_0}$ |
| reversal | $f(-x)$ | $F^*(\omega)$ |
| convolution | $f(x) * h(x)$ | $F(\omega)H(\omega)$ |
| correlation | $f(x) \otimes h(x)$ | $F(\omega)H^*(\omega)$ |
| multiplication | $f(x)h(x)$ | $F(\omega) * H(\omega)$ |
| differentiation | $f'(x)$ | $j\omega F(\omega)$ |
| domain scaling | $f(ax)$ | $1/aF(\omega/a)$ |
| real images | $f(x) = f^*(x)$ | $\Leftrightarrow F(\omega) = F(-\omega)$ |
| Parseval's Theorem | $\sum_x [f(x)]^2$ | $= \sum_\omega [F(\omega)]^2$ |

[Source: R. Szeliski]

| Name | Signal | | Transform |
|-----------------------|---|---|---|
| impulse |  | $\delta(x)$ | \Leftrightarrow 1  |
| shifted impulse |  | $\delta(x - u)$ | $\Leftrightarrow e^{-j\omega u}$  |
| box filter |  | $\text{box}(x/a)$ | $\Leftrightarrow a \text{sinc}(a\omega)$  |
| tent |  | $\text{tent}(x/a)$ | $\Leftrightarrow a \text{sinc}^2(a\omega)$  |
| Gaussian |  | $G(x; \sigma)$ | $\Leftrightarrow \frac{\sqrt{2\pi}}{\sigma} G(\omega; \sigma^{-1})$  |
| Laplacian of Gaussian |  | $(\frac{x^2}{\sigma^4} - \frac{1}{\sigma^2})G(x; \sigma)$ | $\Leftrightarrow -\frac{\sqrt{2\pi}}{\sigma} \omega^2 G(\omega; \sigma^{-1})$  |
| Gabor |  | $\cos(\omega_0 x)G(x; \sigma)$ | $\Leftrightarrow \frac{\sqrt{2\pi}}{\sigma} G(\omega \pm \omega_0; \sigma^{-1})$  |
| unsharp mask |  | $(1 + \gamma)\delta(x) - \gamma G(x; \sigma)$ | $\Leftrightarrow \frac{(1 + \gamma)}{\sigma} - \frac{\sqrt{2\pi}\gamma}{\sigma} G(\omega; \sigma^{-1})$  |
| windowed sinc |  | $\frac{\text{rcos}(x/(aW))}{\text{sinc}(x/a)}$ | \Leftrightarrow (see Figure 3.29)  |

[Source: R. Szeliski]

| Name | Kernel | Transform | Plot |
|----------|------------------------------------|---|------|
| box-3 | $\frac{1}{3} [1 \ 1 \ 1]$ | $\frac{1}{3}(1 + 2 \cos \omega)$ | |
| box-5 | $\frac{1}{5} [1 \ 1 \ 1 \ 1 \ 1]$ | $\frac{1}{5}(1 + 2 \cos \omega + 2 \cos 2\omega)$ | |
| linear | $\frac{1}{4} [1 \ 2 \ 1]$ | $\frac{1}{2}(1 + \cos \omega)$ | |
| binomial | $\frac{1}{16} [1 \ 4 \ 6 \ 4 \ 1]$ | $\frac{1}{4}(1 + \cos \omega)^2$ | |
| Sobel | $\frac{1}{2} [-1 \ 0 \ 1]$ | $\sin \omega$ | |
| corner | $\frac{1}{2} [-1 \ 2 \ -1]$ | $\frac{1}{2}(1 - \cos \omega)$ | |

[Source: R. Szeliski]

2D Fourier Transform

- Same as 1D, but in 2D. Now the sinusoid is

$$s(x, y) = \sin(\omega_x x + \omega_y y)$$

- The 2D Fourier in continuous domain is then

$$H(\omega_x, \omega_y) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} h(x, y) e^{-j(\omega_x x + \omega_y y)} dx dy$$

and in the discrete domain

$$H(k_x, k_y) = \frac{1}{MN} \sum_{x=0}^{M-1} \sum_{y=0}^{N-1} h(x, y) e^{-2\pi j \frac{k_x x + k_y y}{MN}}$$

where M and N are the width and height of the image.

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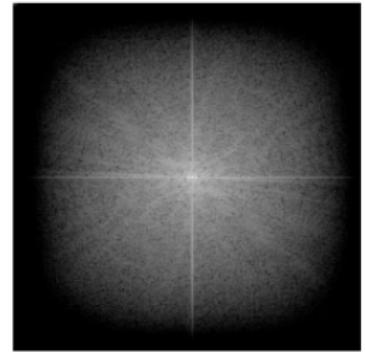
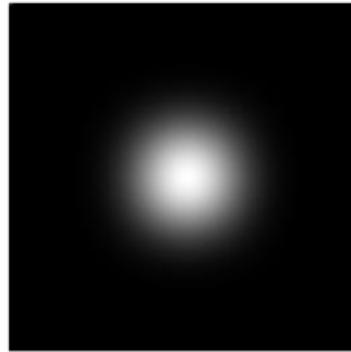
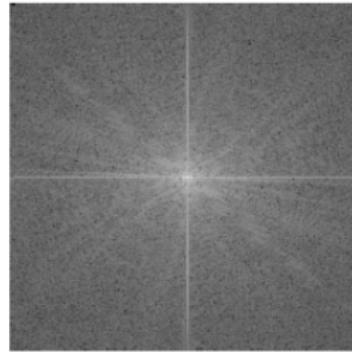
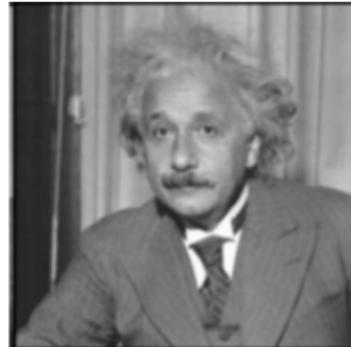
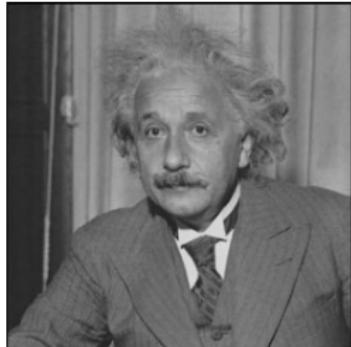
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Example of 2D Fourier Transform



[Source: A. Jepson]

Next class ... image features