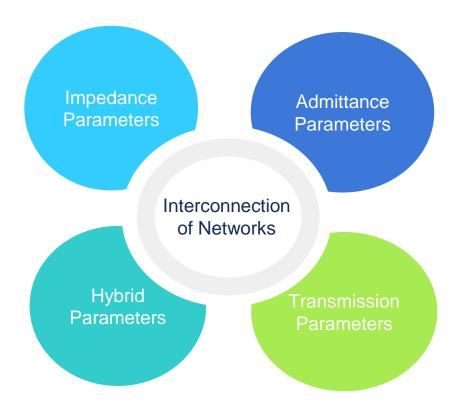
## EE 287 CIRCUIT THEORY

GIDEON ADOM-BAMFI

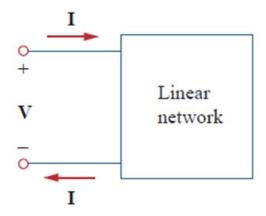
## What to expect?





#### Introduction to Two Port Networks

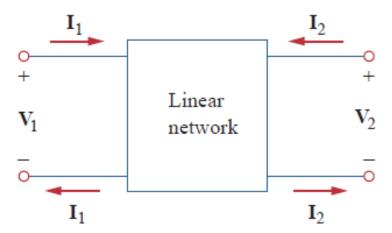
 A pair of terminals through which a current may enter or leave a network is known as a **PORT**.



- Two terminal devices or elements such as resistors, capacitors and inductors result in one-port networks.
- Most of the circuits we have dealt with so far are two-terminal or one-port circuits and can be modelled using Thevenin or Norton equivalent circuits.

#### Introduction to Two Port Networks II

- The majority of devices (op amps, transistors, transformers) and electric systems have two pairs of terminals.
- These devices are known as two-port networks.
- The standard configuration of a two-port network is as shown below:



## Why Two Port Networks?

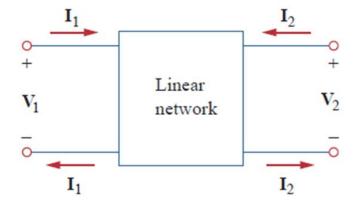
- A two-port network is an electrical network with two separate ports for input and output.
- Thus a two-port network has two terminal pairs acting as access points.

Our study of two-port networks is for at least two reasons:

- 1. Such networks are useful in communications, control systems, power systems and electronics.
- 2. Knowing the two-port parameters of a network or system enables us to treat it as 'black box' when embedded within a larger network.

#### Two Port Parameters

To characterize a two-port network requires that we relate the four port variables  $V_1, V_2, I_1$ , and  $I_2$ .



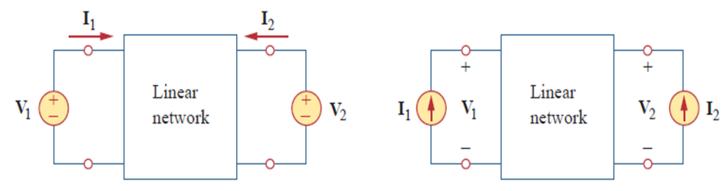
- Depending on which two of the four port variables are given, there exists different ways to describe the relationship between these variables.
- The relationship between voltages and currents are described in terms of quantities known as parameters.

#### Two Port Parameters

Our goal in this lecture is to learn how to find the characteristic parameters of a two-port network:

- Impedance Parameters
- Admittance Parameters
- Hybrid Parameters
- Transmission Parameters

A two-port network may be voltage-driven or current driven



- The impedance parameters are obtained by expressing the terminal voltages in terms of the terminal currents.
- Given currents  $I_1$  and  $I_2$ , voltages  $V_1$  and  $V_2$  are derived as:

$$V_1 = z_{11}I_1 + z_{12}I_2$$
  
 $V_2 = z_{21}I_1 + z_{22}I_2$ 

or in matrix form as:

$$\begin{bmatrix} \mathbf{V}_1 \\ \mathbf{V}_2 \end{bmatrix} = \begin{bmatrix} \mathbf{z}_{11} & \mathbf{z}_{12} \\ \mathbf{z}_{21} & \mathbf{z}_{22} \end{bmatrix} \begin{bmatrix} \mathbf{I}_1 \\ \mathbf{I}_2 \end{bmatrix}$$

where

The z terms are called the **impedance parameters** or simply **z parameters**, and have units of ohms.

The values of the parameters can be evaluated by setting

 $I_1 = 0$  (input port open-circuited)

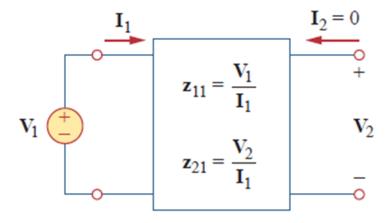
 $I_2 = 0$  (output port open-circuited)

$$z_{11} = \frac{V_1}{I_1} \Big|_{I_2 = 0} \qquad z_{12} = \frac{V_1}{I_2} \Big|_{I_1 = 0}$$

$$Z_{21} = \frac{V_2}{I_1} \Big|_{I_2 = 0} \qquad z_{22} = \frac{V_2}{I_2} \Big|_{I_1 = 0}$$

Since the z-parameters are obtained by open-circuiting the input or output port They are also called the **open-circuit impedance parameters.** 

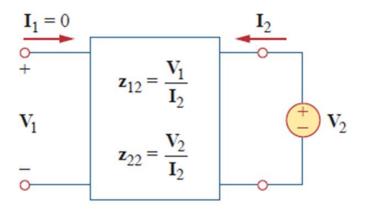
We obtain  $z_{11}$  and  $z_{21}$  by connecting a voltage  $V_1$  (or current source  $I_1$ ) to port 1 with port 2 open circuited and finding  $I_1$  and  $V_2$ 



We then get

$$\mathbf{z}_{11} = \frac{\mathbf{V}_1}{\mathbf{I}_1} \qquad \qquad \mathbf{z}_{21} = \frac{\mathbf{V}_2}{\mathbf{I}_2}$$

Similarly, we obtain  $z_{12}$  and  $z_{22}$  by connecting a voltage  $V_2$  (or a current source  $I_2$ ) to port 2 with port 1 open-circuited and finding  $I_2$  and  $V_1$ 

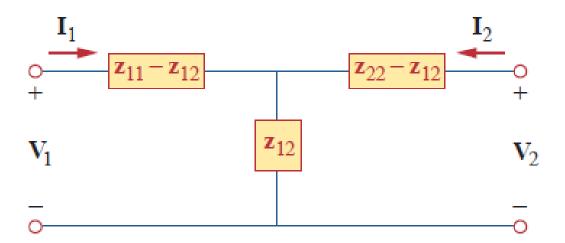


We then get

$$\mathbf{z}_{12} = \frac{\mathbf{V}_1}{\mathbf{I}_2} \qquad \qquad \mathbf{z}_{22} = \frac{\mathbf{V}_1}{\mathbf{I}_2}$$

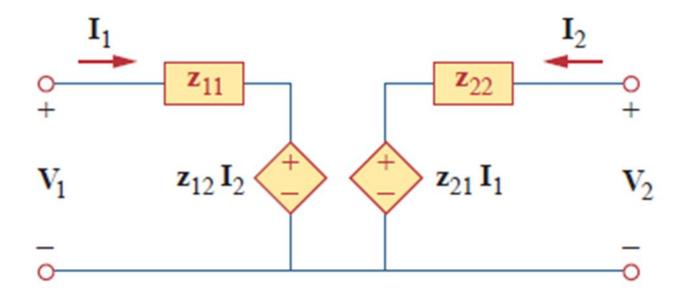
## T-Equivalent Circuit

- Once we know what the impedance parameters are, we can model the two-port network with an equivalent circuit.
- A reciprocal network  $(z_{12} = z_{21})$  can be replaced by the T-equivalent circuit below:



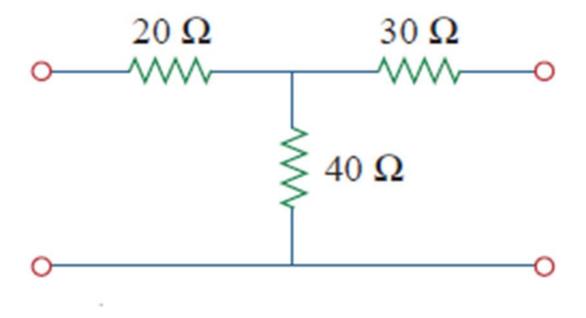
## T-Equivalent Circuit

• If the network is not reciprocal, a more general equivalent network is used:

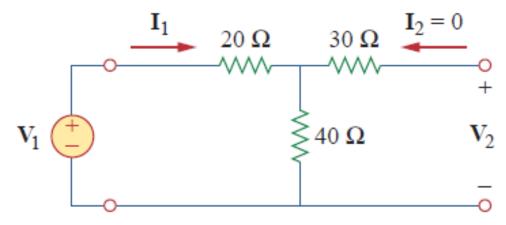


### Question 1

Determine the z parameters for the circuit below:



To determine  $z_{11}$  and  $z_{21}$ , we apply a voltage source  $V_1$  to the input port and leave the output open as shown below:



KVL for the first mesh yields:  $-V_1 + 20I_1 + 40I_1 = 0$ 

Then

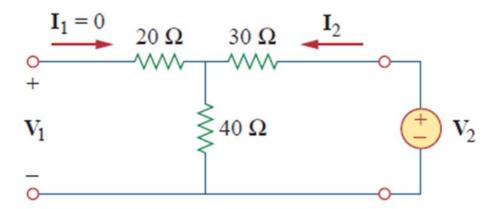
$$z_{11} = \frac{V_1}{I_1} = \frac{(20 + 40)I_1}{I_1} = 60 \Omega$$

That is  $z_{11}$  is the input impedance at port 1

The transfer impedance  $z_{21}$  is:

$$z_{21} = \frac{V_2}{I_1} = \frac{40I_1}{I_1} = 40 \Omega$$

To find  $z_{12}$  and  $z_{22}$ , we apply voltage source  $V_2$  to the output port and leave the input port open as shown below:

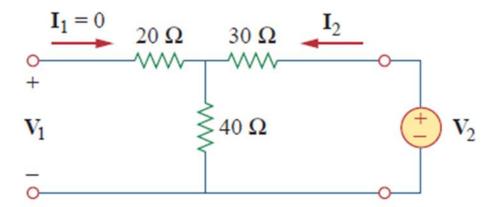


That is  $z_{11}$  is the input impedance at port 1

The transfer impedance  $z_{21}$  is:

$$z_{21} = \frac{V_2}{I_1} = \frac{40I_1}{I_1} = 40 \Omega$$

To find  $z_{12}$  and  $z_{22}$ , we apply voltage source  $V_2$  to the output port and leave the input port open as shown below:



Then

$$z_{12} = \frac{V_1}{I_2} = \frac{40I_2}{I_2} = 40 \Omega$$

and

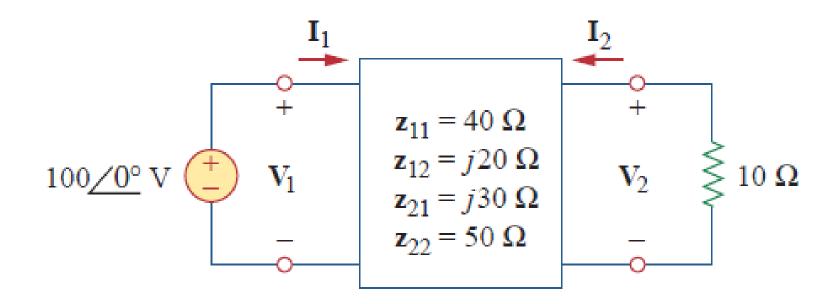
$$z_{22} = \frac{V_2}{I_2} = \frac{(30 + 40)I_2}{I_2} = 70 \Omega$$

Thus

$$[z] = \begin{bmatrix} 60\Omega & 40\Omega \\ 40\Omega & 70\Omega \end{bmatrix}$$

#### Question 2

Find I<sub>1</sub> and I<sub>2</sub> in the circuit below



We substitute the given z parameters in the matrix shown below:

$$\begin{bmatrix} \mathbf{V}_1 \\ \mathbf{V}_2 \end{bmatrix} = \begin{bmatrix} \mathbf{z}_{11} & \mathbf{z}_{12} \\ \mathbf{z}_{21} & \mathbf{z}_{22} \end{bmatrix} \begin{bmatrix} \mathbf{I}_1 \\ \mathbf{I}_2 \end{bmatrix} = [\mathbf{z}] \begin{bmatrix} \mathbf{I}_1 \\ \mathbf{I}_2 \end{bmatrix}$$

and get

$$V_1 = 40I_1 + j20I_2$$
  
 $V_2 = j30I_1 + 50I_2$ 

Since we are looking for  $I_1$  and  $I_2$ , we substitute

$$V_1 = 100 \angle 0^\circ$$
,  $V_2 = 10I_2$ 

Into the above the equations above and get the equations on the next slide.

The equation becomes

$$100 = 40I_1 + j20I_2 \tag{1}$$

and

$$-10I_2 = j30I_1 + 50I_2 \Rightarrow I_1 = j2I_2$$
 (2)

Substituting (2) into (1)

$$100 = j80I_2 + j20I_2$$
  $\Rightarrow$   $I_2 = \frac{100}{j100} = -j$ 

From (2)

$$I_1 = j2(-j) = 2$$

Thus

$$I_1 = 2A \angle 0^{\circ}, \ I_2 = 1A \angle -90^{\circ}$$

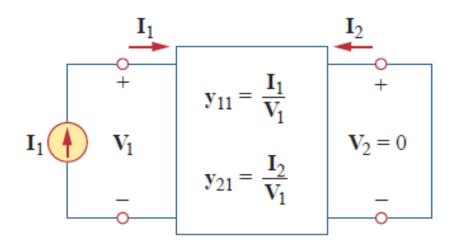
- Admittance parameters are obtained by expressing the terminal currents in terms of the terminal voltages.
- That is given voltages  $V_1$  and  $V_2$ , currents  $I_1$  and  $I_2$  are derived:

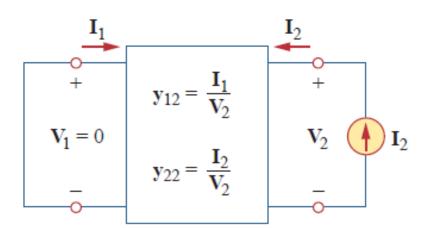
$$I_1 = y_{11}V_1 + y_{12}V_2$$
  
$$I_2 = y_{21}V_1 + y_{22}V_2$$

or in matrix form

$$\begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} y_{11} & y_{12} \\ y_{21} & y_{22} \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = [y] \begin{bmatrix} V_1 \\ V_2 \end{bmatrix}$$

- The y terms are known as the **admittance parameters** or y parameters and have units of 'siemens'.
- The values of the parameters can be obtained by setting  $V_1 = 0$  (input port short-circuited) or  $V_2 = 0$  (output port short-circuited)





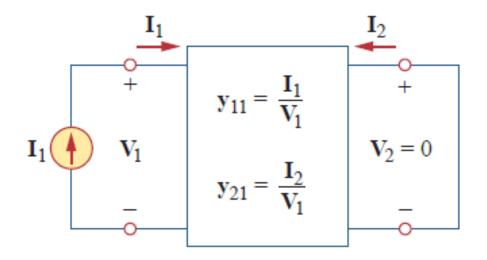
Thus

$$y_{11} = \frac{I_1}{V_1} | V_2 = 0$$
  $y_{12} = \frac{I_1}{V_2} | V_1 = 0$ 

$$y_{21} = \frac{I_2}{V_1} | V_2 = 0$$
  $y_{22} = \frac{I_2}{V_2} | V_1 = 0$ 

Since the y parameters are obtained by short-circuiting the input or out put port They are also called **short-circuiting admittance parameters.** 

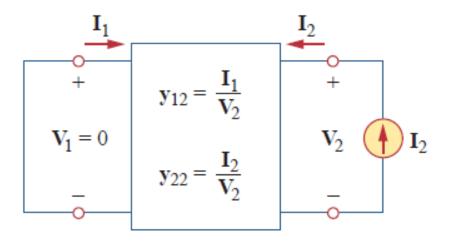
We obtain  $y_{11}$  and  $y_{21}$  by connecting a current  $I_1$  to port 1 and short-circuiting port 2



Find V<sub>1</sub> and I<sub>2</sub> and then calculate

$$y_{11} = \frac{I_1}{V_1}, \qquad y_{21} = \frac{I_2}{V_1}$$

Similarly, we obtain  $y_{12}$  and  $y_{22}$  by connecting a current source  $I_2$  to port 2 and short-circuiting port 1.

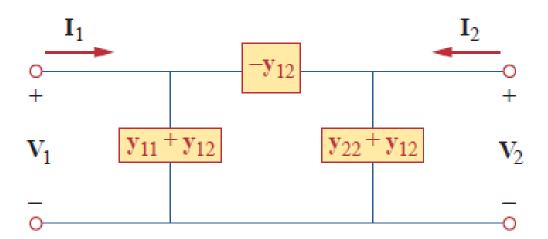


Find I<sub>1</sub> and V<sub>2</sub> and then calculate:

$$y_{12} = \frac{I_1}{V_2}, \qquad y_{22} = \frac{I_2}{V_2}$$

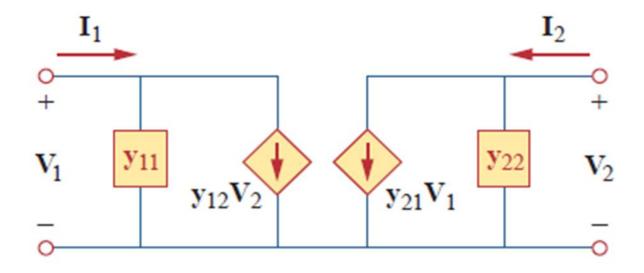
### π-Equivalent Circuit

- For a two-port network that is linear and has no dependent sources
- The transfer admittances are equal  $(y_{12} = y_{21})$
- A reciprocal network  $(y_{12} = y_{21})$  can be modelled by the  $\pi$ -equivalent circuit as shown below:



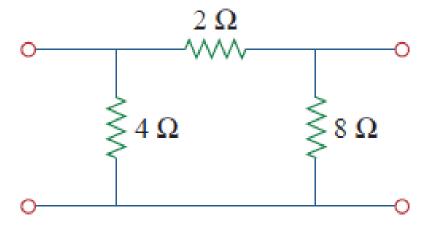
### π-Equivalent Circuit

• If the network is not reciprocal, a more general equivalent network is shown below:



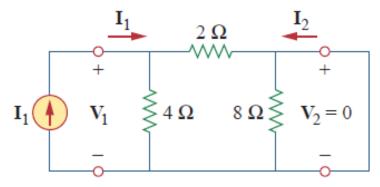
### Question 1

Obtain the y parameters for the  $\pi$  network shown below:



To find  $y_{11}$  and  $y_{21}$ 

Short-circuit the output port and connect a current source I<sub>1</sub> to the input port as shown below:



Since the 8- $\Omega$  resistor is short-circuited, the 2-  $\Omega$  resistor is in parallel with the 4-  $\Omega$  Hence

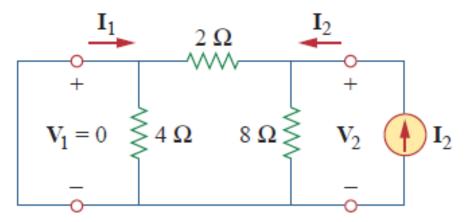
$$V_1 = I_1(4//2) = \frac{4}{3}I_1$$
  $y_{11} = \frac{I_1}{V_1} = \frac{I_1}{\frac{4}{3}I_1} = 0.75S$ 

By current division

$$-I_2 = \frac{4}{4+2}I_1 = \frac{2}{3}I_1 \quad y_{21} = \frac{I_2}{V_1} = \frac{-\frac{2}{3}I_1}{\frac{4}{3}I_1} = -0.5S$$

To get  $y_{12}$  and  $y_{22}$ 

Short-circuit the input port and connect a current source I<sub>2</sub> to the output port as shown below:



The 4- $\Omega$  resistor is short-circuited so the 2-  $\Omega$  and the 8-  $\Omega$  resistors are in parallel

$$V_2 = I_2(8//2) = \frac{8}{5}I_2$$
  $y_{22} = \frac{I_2}{V_2} = \frac{I_2}{\frac{8}{5}I_2} = \frac{5}{8} = 0.625S$ 

By current division

$$-I_1 = \frac{8}{8+4}I_2 = \frac{4}{5}I_2 \qquad \qquad y_{12} = \frac{I_1}{V_2} = \frac{-\frac{4}{5}I_2}{\frac{8}{5}I_2} = -0.5S$$

# 3 Hybrid Parameters

### **Hybrid Parameters**

- This set of parameters is based on making  $V_1$  and  $I_2$  the dependent variable
- That is given V<sub>2</sub> and I<sub>1</sub>, V<sub>1</sub> and I<sub>2</sub> are derived

Thus we obtain

$$V_1 = h_{11}I_1 + h_{12}V_2$$
$$I_2 = h_{21}I_1 + h_{22}V_2$$

Or in matrix form

$$\begin{bmatrix} \mathbf{V}_1 \\ \mathbf{I}_2 \end{bmatrix} = \begin{bmatrix} \mathbf{h}_{11} & \mathbf{h}_{12} \\ \mathbf{h}_{21} & \mathbf{h}_{22} \end{bmatrix} \begin{bmatrix} \mathbf{I}_1 \\ \mathbf{V}_2 \end{bmatrix} = [\mathbf{h}] \begin{bmatrix} \mathbf{I}_1 \\ \mathbf{V}_2 \end{bmatrix}$$

#### **Hybrid Parameters**

- The h terms are known as the hybrid parameters or simply the h parameters.
- The h parameters are very useful for describing electronic devices such as transistors.
- It is much easier to measure experimentally the h parameters of such devices than to measure the z or y parameters
- The values of the parameters are determined as:

$$h_{11} = \frac{V_1}{I_1} | V_2 = 0$$
  $h_{12} = \frac{V_1}{V_2} | I_1 = 0$ 

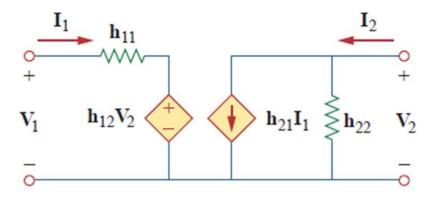
$$h_{21} = \frac{I_2}{I_1} | V_2 = 0$$
  $h_{22} = \frac{I_2}{V_2} | I_1 = 0$ 

#### Procedure for calculating the h parameters

- We apply a voltage or current source to the appropriate port.
- Short-circuit or open-circuit the other port, depending on the parameter of interest, and perform regular circuit analysis
- For reciprocal networks

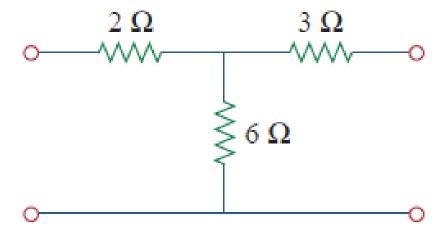
$$h_{12} = -h_{21}$$

The figure below shows the hybrid model of a two-port network



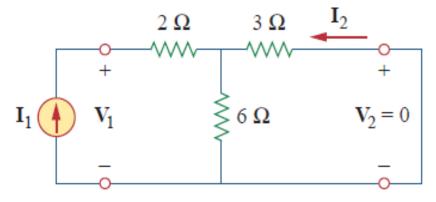
#### Question 1

Find the hybrid parameters for the two-port network shown below



To find  $h_{11}$  and  $h_{21}$ 

We short circuit the output port and connect a current source I<sub>1</sub> to the input port as shown below:



From the figure above

$$V_1 = I_1(2+3//6) = 4I_1$$

Hence

$$h_{11} = \frac{V_1}{I_1} = 4\Omega$$

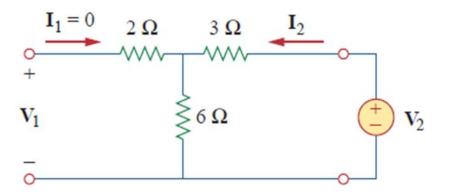
By current division

$$-I_2 = \frac{6}{6+3}I_1 = \frac{2}{3}I_1$$

Hence

$$h_{21} = \frac{I_2}{I_1} = -\frac{2}{3}$$

To obtain  $h_{12}$  and  $h_{22}$ , we open-circuit the input port and connect a voltage source  $V_2$  to the output port as shown in the figure below



By voltage division

$$V_1 = \frac{6}{6+3} V_2 = \frac{2}{3} V_2$$

Hence

$$h_{12} = \frac{V_1}{V_2} = \frac{2}{3}$$

$$V_2 = (3+6)I_2 = 9I_2$$

Thus

$$h_{22} = \frac{I_2}{V_2} = \frac{1}{9} S$$

- Another set of parameters relates the variables at the input port to those at the output port.
- The direction of the output current is reversed

Thus

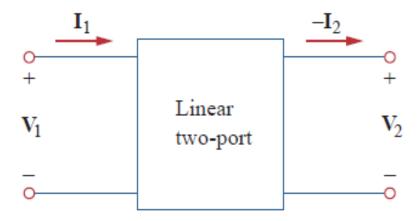
$$V_1 = AV_2 - BI_2$$
$$I_1 = CV_2 - DI_2$$

Or in matrix form

$$\begin{bmatrix} V_1 \\ I_1 \end{bmatrix} = \begin{bmatrix} A & B \\ C & D \end{bmatrix} \begin{bmatrix} V_2 \\ -I_2 \end{bmatrix} = [T] \begin{bmatrix} V_2 \\ -I_2 \end{bmatrix}$$

These equations relate the input variables ( $V_1$  and  $I_1$ ) to the output variables ( $V_2$  and  $-I_2$ )

Notice that in computing the transmission parameters, - I<sub>2</sub> is used rather than I<sub>2</sub>, because the current is considered to be leaving the network as shown in the circuit below:



- This is done merely for conventional reasons
- When you cascade two-ports (output to input) it is most logical to think of I<sub>2</sub> as leaving the two-port

- It is also customary in the power industry to consider  $I_2$  as leaving the two-port.
- The two-port parameters provide a measure of how a circuit transmits voltage and current from a source to a load.

- They are useful in the analysis of transmission lines.
- Also known as ABCD parameters and are used in the design of telephone, microwave networks and radars.

The transmission parameters are determined as:

$$A = \frac{V_1}{V_2} |_{I_2 = 0}$$
  $B = -\frac{V_1}{I_2} |_{V_2 = 0}$ 

$$C = \frac{I_1}{V_2} \Big|_{I_2 = 0}$$
  $D = -\frac{I_1}{I_2} \Big|_{V_2 = 0}$ 

Specifically,

A = Open-circuit voltage ratio

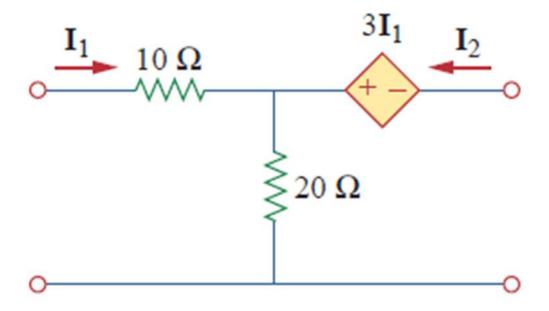
B = Negative short-circuit transfer impedance

C = Open-circuit transfer admittance

D = Negative short-circuit current ratio

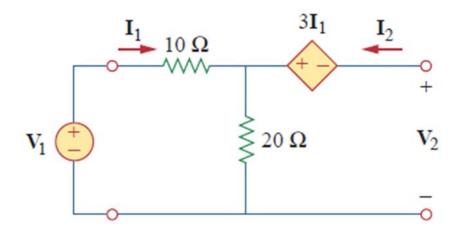
### Example 1

Find the transmission parameters for the two-port network in the figure shown below



To determine A and C

We leave the output port open as shown below so that  ${\rm I_2}=0$  and place a voltage source  ${\rm V_1}$  at the input port



we have

$$V_1 = (10 + 20)I_1 = 30I_1$$
 and  $V_2 = 20I_1 - 3I_1 = 17I_1$ 

Thus

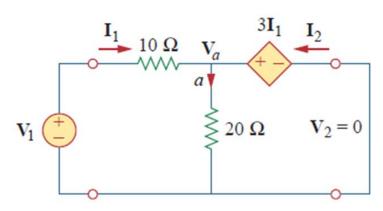
$$A = \frac{V_1}{V_2} = \frac{30I_1}{17I_1} = 1.765$$

$$C = \frac{I_1}{V_2} = \frac{I_1}{17I_1} = 0.0588S$$

To obtain B and D

We short-circuit the output port so that  $V_2=0$  and place a voltage source  $V_1$  at the input

port. See figure below



KCL at node a gives

$$\frac{V_1 - Va}{10} - \frac{V_a}{20} + I_2 = 0$$

But

$$V_a = 3I_1$$

$$V_a = 3I_1$$
 and  $I_1 = \frac{V_1 - V_2}{10}$ 

Combining these equations

$$V_a = 3I_1$$

$$V_a = 3I_1 \qquad V_1 = 13I_1$$

Substituting them in the KCL equation yields

$$I_1 - \frac{3I_1}{20} + I_2 = 0 \implies \frac{17}{20}I_1 = -I_2$$

Therefore

$$D = -\frac{I_1}{I_2} = \frac{20}{17} = 1.176$$

and

$$B = -\frac{V_1}{I_2} = \frac{-13I_1}{\left(-\frac{17}{20}\right)I_1} = 15.29\Omega$$

## 5 Interconnection of Networks

#### Interconnection of networks

- A large complex network may be divided into subnetworks for the purpose of analysis and design.
- The subnetworks are modelled as two-port networks interconnected to form the original network.
- The two-port networks may therefore be regarded as building blocks that can be interconnected to form a complex network.
- The interconnection can be in series, in parallel, or in cascade although the interconnected network can be described by any of the six parameters.
- A certain set of parameters may have a definite advantage.

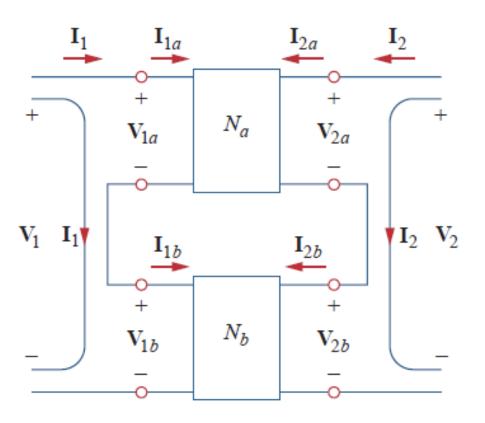
#### Interconnection of networks

#### For example,

- When the networks are in series, their individual z parameters add up to give the z parameters of the large network.
- When they are in parallel, their individual y parameters add up to give the y parameters of the larger network.
- When they are cascaded, their individual transmission parameters can be multiplied together to give the transmission parameters of the larger network.

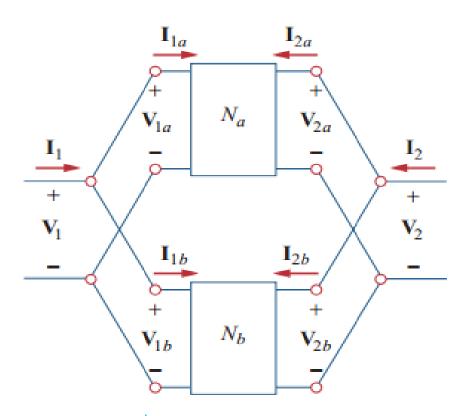
#### Series Connection of 2 Two-Port Networks

Consider the series connection of two two-port networks as shown below



#### Parallel Connection of 2 Two-Port Networks

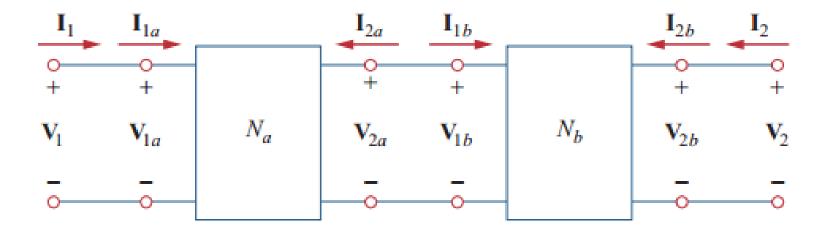
Consider the parallel connection of two two-port networks as shown below



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#### Cascade Connection of 2 Two-Port Networks

Consider the cascade connection of two two-port networks as shown below



# Thanks! Any questions?