Lecture 12 - Digital Circuits (I)

THE INVERTER

March 18, 2003

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- 1. Introduction to digital electronics: the inverter
- 2. NMOS inverter with resistor pull up

Reading assignment:

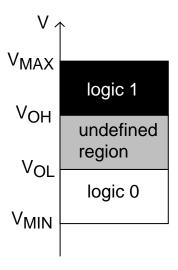
Howe and Sodini, Ch. 5, §§5.1-5.3.2

Key questions

- What are the key figures of merit of logic circuits?
- How can one make a simple inverter using a single MOSFET?

1. Introduction to digital electronics: the inverter

In digital electronics, digitally-encoded information is represented by means of two distinct voltage ranges:

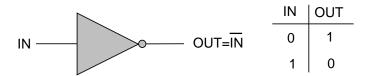


- logic 0: $V_{MIN} \le V \le V_{OL}$
- logic 1: $V_{OH} \le V \le V_{MAX}$
- undefined logic value: $V_{OL} \leq V \leq V_{OH}$.

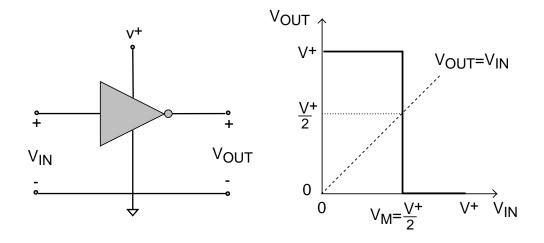
Logic operations are performed using logic gates.

Simplest logic operation of all: $inversion \Rightarrow inverter$

□ Ideal inverter:



Circuit representation and ideal transfer function:



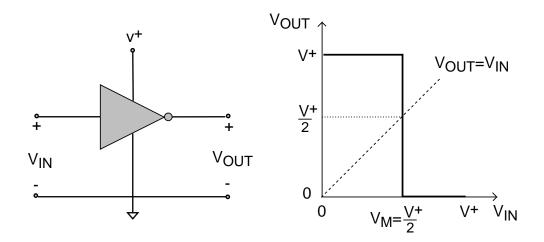
Define switching point or logic threshold:

$$V_M \equiv \text{ input voltage for which } V_{OUT} = V_{IN}$$

-for
$$0 \le V_{IN} \le V_{M} \Rightarrow V_{OUT} = V^{+}$$

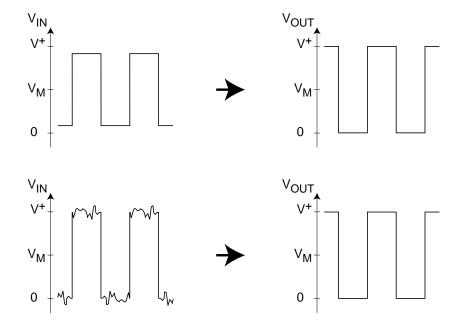
-for $V_{M} \le V_{IN} \le V^{+} \Rightarrow V_{OUT} = 0$

Key property of ideal inverter: signal regeneration

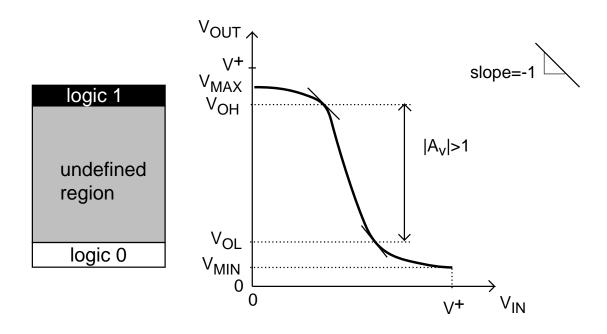


Ideal inverter returns well defined logical outputs (0 or V^+) even in the presence of considerable noise in V_{IN} (from voltage spikes, crosstalk, etc.)

 \Rightarrow signal is regenerated!



□ "Real" inverter:



In a real inverter, valid logic levels defined as follows:

• *logic 0*:

 $V_{MIN} \equiv \text{output voltage when } V_{IN} = V^+$

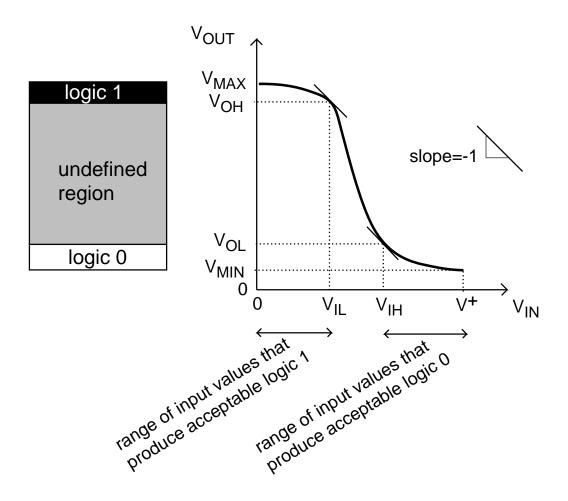
 $V_{OL} \equiv \text{smallest output voltage where slope=-1}$

• *logic* 1:

 $V_{OH} \equiv \text{largest output voltage where slope=-1}$

 $V_{MAX} \equiv \text{output voltage when } V_{IN} = 0$

Two other important voltages:

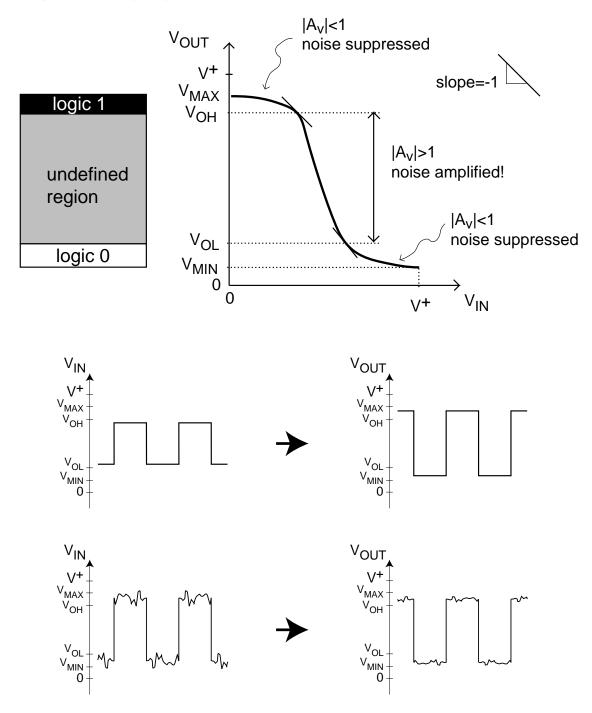


 $V_{IL} \equiv \text{smallest input voltage where slope=-1}$ $V_{IH} \equiv \text{highest input voltage where slope=-1}$

To have signal regeneration:

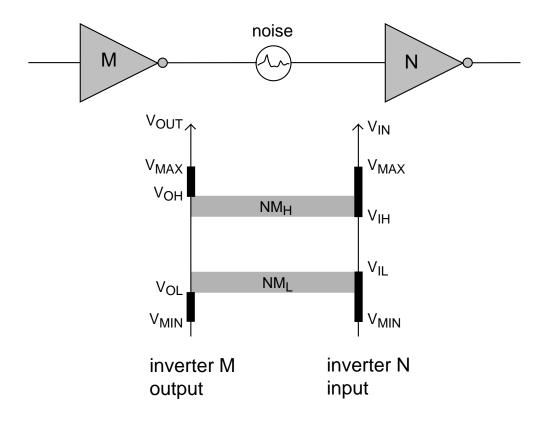
range of input values that produce acceptable logic output > range of valid logic values

Key to signal regeneration in inverter: $noise suppression \Rightarrow high voltage gain$



Quantify noise immunity through noise margins.

Chain of two inverters:



Define noise margins:

$$NM_H = V_{OH} - V_{IH}$$
 noise margin high $NM_L = V_{IL} - V_{OL}$ noise margin low

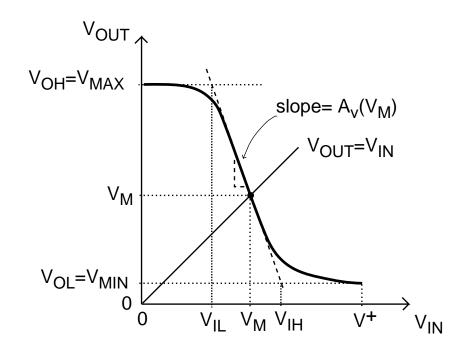
When signal is within noise margins:

- logic 1 output from first inverter interpreted as logic 1 input by second inverter
- logic 0 output from first inverter interpreted as logic 0 input by second inverter

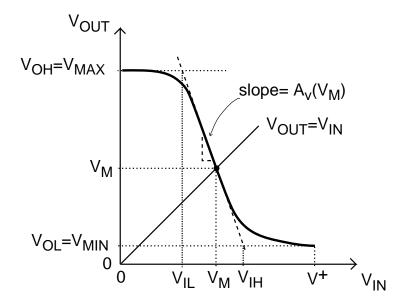
Simplifications for hand calculations

Hard to compute $A_v = -1$ points in transfer function.

Approximate calculation:



- Assume $V_{OL} \simeq V_{MIN}$ and $V_{OH} \simeq V_{MAX}$
- Trace tangent of transfer function at V_M (slope=small signal voltage gain at V_M)
- $V_{IL} \simeq \text{intersection of tangent with } V_{OUT} = V_{MAX}$
- $V_{IH} \simeq$ intersection of tangent with $V_{OUT} = V_{MIN}$
- to enhance noise margin: $|A_v(V_M)| \uparrow$



$$|A_v| \simeq \frac{V_{MAX} - V_M}{V_M - V_{IL}} \Rightarrow V_{IL} \simeq V_M - \frac{V_{MAX} - V_M}{|A_v|}$$

$$|A_v| \simeq \frac{V_M - V_{MIN}}{V_{IH} - V_M} \implies V_{IH} \simeq V_M (1 + \frac{1}{|A_v|}) - \frac{V_{MIN}}{|A_v|}$$

Then:

$$NM_L = V_{IL} - V_{OL} \simeq (V_{MAX} - V_{MIN}) - (V_{MAX} - V_M)(1 + \frac{1}{|A_v|})$$

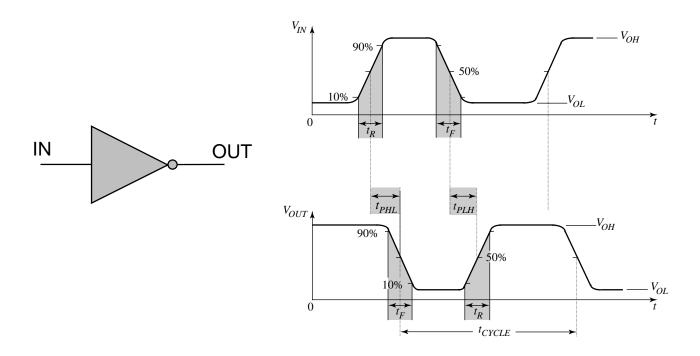
$$NM_H = V_{OH} - V_{IH} \simeq (V_{MAX} - V_{MIN}) - (V_M - V_{MIN})(1 + \frac{1}{|A_v|})$$

If $|A_v| \to \infty$:

$$NM_L \rightarrow V_M - V_{MIN}$$
 $NM_H \rightarrow V_{MAX} - V_M$

□ Transient characteristics

Look at inverter switching in the time domain:



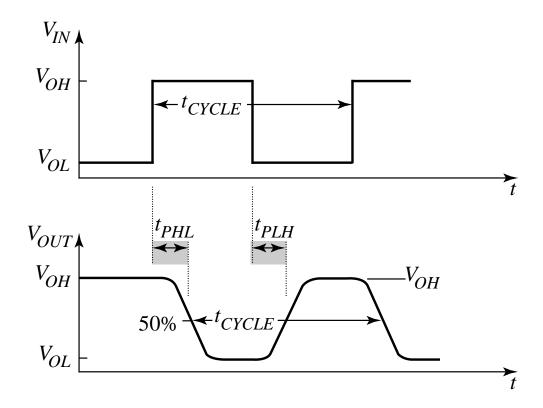
 $t_R \equiv rise \ time \ between 10\% \ and 90\% \ of total swing$ $t_F \equiv fall \ time \ between 90\% \ and 10\% \ of total swing$ $t_{PHL} \equiv propagation \ delay \ from \ high-to-low \ between 50\% \ points$

 $t_{PLH} \equiv propagation \ delay \ from \ low-to-high \ between 50\% \ points$

Propagation delay: $t_P = \frac{1}{2}(t_{PHL} + t_{PLH})$

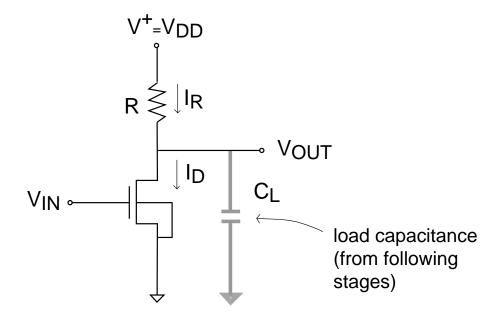
Propagation delay: simplification for hand calculations

- Input wavefunction = ideal square wave
- Propagation delay times = delay times to 50% point



- Hand calculations only approximate
- SPICE essential for accurate delay analysis

2. NMOS inverter with resistor pull up

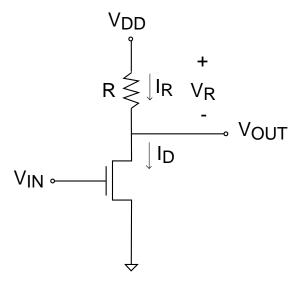


Features:

- $V_{BS} = 0$ (typically not shown)
- C_L summarizes capacitive loading of following stages (other logic gates, interconnect lines)

Basic operation:

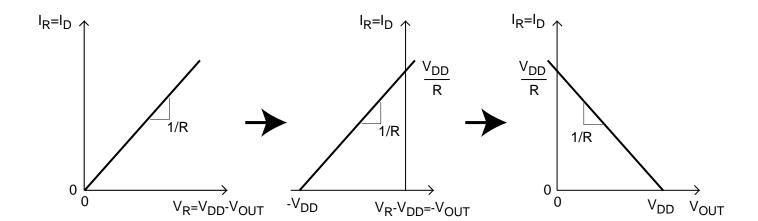
- if $V_{IN} < V_T$, MOSFET OFF $\Rightarrow V_{OUT} = V_{DD}$
- if $V_{IN} > V_T$, MOSFET ON $\Rightarrow V_{OUT}$ small (value set by resistor/nMOS divider)



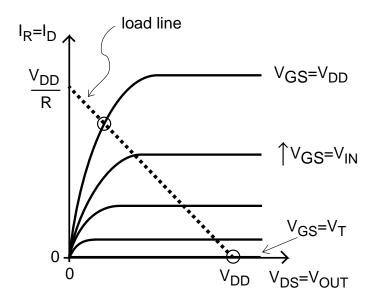
Transfer function obtained by solving:

$$I_R = I_D$$

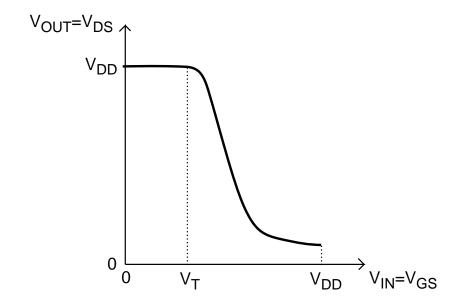
Can solve graphically: I-V characteristics of pull-up resistor on I_D vs. V_{OUT} transistor characteristics:



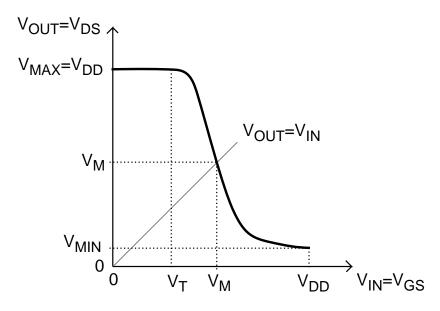
Overlap I-V characteristics of resistor pull-up on I-V characteristics of transistor:



Transfer function:



Logic levels:



For V_{MAX} , transistor is cut-off, $I_D = 0$:

$$V_{MAX} = V_{DD}$$

For V_{MIN} , transistor is in linear regime; solve:

$$I_D = \frac{W}{L} \mu_n C_{ox} (V_{DD} - \frac{V_{MIN}}{2} - V_T) V_{MIN} = I_R = \frac{V_{DD} - V_{MIN}}{R}$$

For V_M , transistor is in saturation; solve:

$$I_D = \frac{W}{2L} \mu_n C_{ox} (V_M - V_T)^2 = I_R = \frac{V_{DD} - V_M}{R}$$

Will continue next lecture with analysis of noise margin and dynamics...

Key conclusions

- Logic circuits must exhibit *noise margins* in which they are inmune to noise in input signal.
- Logic circuits must be *regenerative*: able to restore clean logic values even if input is noisy.
- Propagation delay: time for logic gate to perform its function.
- Concept of *load line*: graphical technique to visualize transfer characteristics of inverter.
- First-order solution (by hand) of inverter figures of merit easy if regimes of operation of transistor are correctly identified.
- For more accurate solutions, use SPICE (or other circuit CAD tool).