## **Transient State Analysis**

## 2.Effect of $\xi$ on nature of response:

## (i) Time Response of the Second Order (2<sup>nd</sup> Order) System for Unit Step Input:

Consider the unit step signal as an input to the first order system.

So, r(t)=u(t) and therefore, R(s)=1/s

From 2<sup>nd</sup> order standard block diagram, we can write C(s) as,

$$C(s) = \left(\frac{\omega_n^2}{s^2 + 2\xi\omega_n s + \omega_n^2}\right) R(s) \dots (2)$$

$$C(s) = \left(\frac{\omega_n^2}{s^2 + 2\xi\omega_n s + \omega_n^2}\right) \left(\frac{1}{s}\right) = \frac{A}{s} + \frac{Bs + C}{s^2 + 2\xi\omega_n s + \omega_n^2}$$

Do partial fraction of C(s).

$$C(s) = \frac{\omega_n^2}{s(s^2 + 2\xi\omega_n s + \omega^2)_n} = \frac{A}{s} + \frac{Bs + C}{s^2 + 2\xi\omega_n s + \omega^2_n}$$

$$\Rightarrow \frac{\omega_{n}^{2}}{s(s^{2}+2\xi\omega_{n}s+\omega_{n}^{2})} = \frac{A(s^{2}+2\xi\omega_{n}s+\omega_{n}^{2}) + (Bs+C)s}{s(s^{2}+2\xi\omega_{n}s+\omega_{n}^{2})}$$

$$\Rightarrow \omega_{n}^{2} = (A+B)s^{2} + (2A\xi\omega_{n}+C)s + A\omega_{n}^{2}$$

$$\Rightarrow \omega_{n}^{2} = (A+B)s^{2} + (2A\xi\omega_{n}+C)s + A\omega_{n}^{2}$$

$$\Rightarrow B=-1 \Rightarrow c=-2\xi\omega_{n} \Rightarrow A=1$$

$$\Rightarrow C(s) = \frac{1}{s} - \frac{s+2\xi\omega_{n}}{s^{2}+2\xi\omega_{n}s+\omega_{n}^{2}}$$

$$= \frac{1}{s} - \frac{s+2\xi\omega_{n}}{(s+\xi\omega_{n})^{2}+\omega_{n}^{2}}$$

$$= \frac{1}{s} - \frac{s+2\xi\omega_{n}}{(s+\xi\omega_{n})^{2}+\omega_{n}^{2}}$$

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$$\Rightarrow C(s) = \frac{1}{s} - \frac{s+\xi\omega_{n}}{(s+\xi\omega_{n})^{2}+\omega_{n}^{2}} - \frac{\xi\omega_{n}}{(s+\xi\omega_{n})^{2}+\omega_{n}^{2}}$$

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$$\Rightarrow C(s) = \frac{1}{s} - \frac{s+\xi\omega_{n}}{(s+\xi\omega_{n})^{2}+\omega_{n}^{2}} - \frac{\xi\omega_{n}}{(s+\xi\omega_{n})^{2}+\omega_{n}^{2}} - \frac{(s+\xi\omega_{n})^{2}+\omega_{n}^{2}}{(s+\xi\omega_{n})^{2}+\omega_{n}^{2}} - \frac{(s+\xi\omega_{n})^{2}+\omega_{n}^{2}}{(s+\xi\omega_{n})^{2}+$$

Taking Laplace inverse transform: -

$$c(t) = 1 - \frac{\epsilon}{\epsilon} \frac{\omega_n t}{\omega_n t} \cos \omega_n t - \frac{\epsilon}{\epsilon} \frac{\omega_n}{\omega_n t} \cdot \frac{\epsilon}{\epsilon} \cos \omega_n t + \frac{\epsilon}{\epsilon} \frac{\omega_n t}{\omega_n t} \cdot \frac{\epsilon}{\epsilon} \cos \omega_n t + \frac{\epsilon}{\epsilon} \frac{\omega_n t}{\epsilon} \cdot \frac{\epsilon}{\epsilon} \cos \omega_n t + \frac{\epsilon}{\epsilon} \frac{\omega_n t}{\epsilon} \cdot \frac{\epsilon}{\epsilon} \cos \omega_n t + \frac{\epsilon}{\epsilon} \frac{\omega_n t}{\epsilon} \cdot \frac{\epsilon}{\epsilon} \cos \omega_n t + \frac{\epsilon}{\epsilon} \frac{\omega_n t}{\epsilon} \cdot \frac{\epsilon}{\epsilon} \cos \omega_n t + \frac{\epsilon}{\epsilon} \frac{\omega_n t}{\epsilon} \cdot \frac{\epsilon}{\epsilon} \cos \omega_n t + \frac{\epsilon}{\epsilon} \frac{\omega_n t}{\epsilon} \cdot \frac{\epsilon}{\epsilon} \cos \omega_n t + \frac{\epsilon}{\epsilon} \frac{\omega_n t}{\epsilon} \cdot \frac{\omega_n t}{\epsilon} = \frac{\epsilon}{\epsilon} \frac{\omega_n t}{\epsilon} \cdot \frac{\omega_n t}{\epsilon$$

Now, Put the values of  $\xi$  and  $\omega_d$  in the expression of c(t), we get

$$c(t) = 1 - \frac{e^{t\omega_n t}}{|T-\xi|} \sin\left(\omega_n |T-\xi|\right) t + t \sin\left(\frac{1}{t}\right)$$

$$t + t \cos(t) + t \cos(t)$$

$$t + t \cos(t)$$

the error signal for the system
$$e(t) = r(t) - c(t)$$

$$= x - x + \frac{e^{\omega_n t}}{\sqrt{1-\xi^2}} \sin(\omega_n \sqrt{1-\xi^2})t + te^{-1} \frac{\sqrt{1-\xi^2}}{\xi_n}$$

$$= \frac{e(t)}{\sqrt{1-\xi^2}} \cdot \sin(\omega_n \sqrt{1-\xi^2})t + te^{-1} \frac{\sqrt{1-\xi^2}}{\xi_n}$$

$$= \frac{e(t)}{\sqrt{1-\xi^2}} \cdot \sin(\omega_n \sqrt{1-\xi^2})t + te^{-1} \frac{\sqrt{1-\xi^2}}{\xi_n}$$
The steady state error is
$$= ess = \lim_{t \to \infty} e(t)$$

$$= ess = \lim_{t \to \infty} e(t)$$
therefore, at steady state there is no error between input and output.

**Note:-** As the time response of  $2^{nd}$  order system is influenced by  $\xi$ , therefore, there are four possible cases for positively damped systems  $(\xi > 0)$ . (system will be stable or marginal stable system)

0< \$< 1: under damped \$=1: Critically damped \$>1: over damped \$=0: undamped

Also, there are three possible cases for negatively damped systems ( $\xi$  < 0). (system will be unstable)

## (a) Case I: Underdamped Case $(0 < \xi < 1)$

In this case, the response is given below:

The response is oscillatory with oscillating frequency with but decreasing amplitude due to exponential term Elent.

This type of response is called underdamped response.

Steady state value = 1.

Percrott is given by, e(t) = \frac{\overline{e}}{\overline{11-\overline{e}^2}} \sin(\overline{\psi}) \tag{1+\overline{\psi}})

The constant is \( T = \frac{\overline{e}}{\overline{e}} \overline{\overline{e}}.

Note: > the damped frequency always less than the undamped frequency because of factor &c.

i.e., wax wn [: ad:wnJI-&i]

Note:> As & increased, the response becomes progressively less oscillatory till it becomes critically damped (just non-oscillatory) for &= 1 and becomes overdamped for &>1.

## (b) Case II: Undamped Case $(\xi = 0)$

Note:- Since there is no time damping and therefore

- Oscillations never die out with time.
- ➤ Amplitude of oscillation = constant around steady-state.
- ➤ This response is known as undamped response.
- ➤ There is no loss of energy.

## (c) Case III: Critically damped Case $(\xi = 1)$

The time response at 
$$\varepsilon_{e}=1$$
 will be:

$$C(t) = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon_{e}\omega_{h}t}}{\sqrt{1-\varepsilon_{e}^{2}}} \right\} = \lim_{\varepsilon \to 1} \left\{ 1 - \frac{e^{-\varepsilon$$

→ fore &=1, oscillations in output response are just dissappearced. This type of response is called as <u>Critically</u> damped response.

-> charcacteristic eq?: s2+2EWns+Wn2=0

-> for &==1; The roots are - Wn, - Wh.

-> System is Absolute stable.

## (d) Case IV: Overdamped Case $(\xi > 1)$

The time response is given by

$$C(s) = \frac{1}{s} \cdot \frac{\omega_0^2}{\omega_0^2}$$

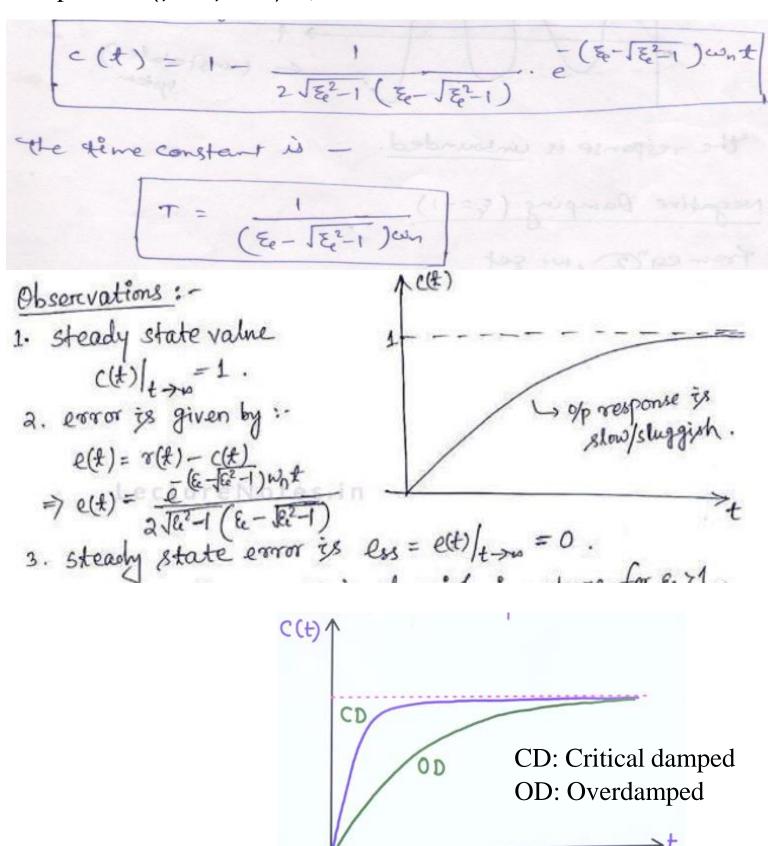
$$C(s) = \frac{1}{s} \cdot \frac{\omega_0^2}{s^2 + 2\epsilon \omega_0 s + \omega_0^2} = \frac{1}{s} \cdot \frac{\omega_0^2}{(s + \epsilon \omega_0)^2 - \omega_0^2(\epsilon^2 - 1)}$$

$$\Rightarrow C(s) = \frac{1}{s} \cdot \frac{\omega_0^2}{(s + \epsilon \omega_0)^2 - \omega_0^2} = \frac{1}{s} + \frac{1}{s} \cdot \frac{1}{s} \cdot \frac{1}{s} \cdot \frac{1}{s}$$

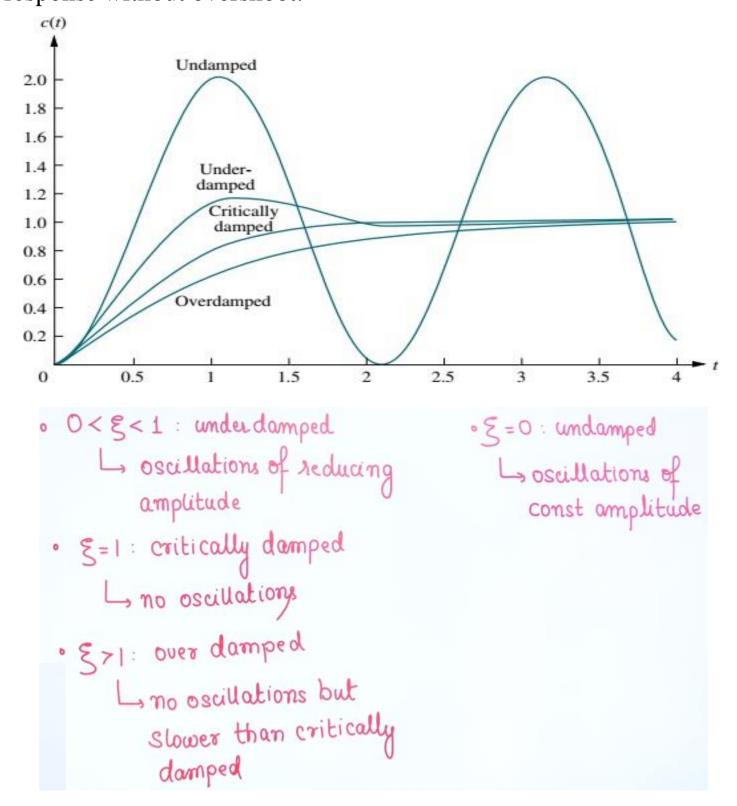
$$\Rightarrow C(s) = \frac{\omega_0^2}{s} \cdot \frac{\omega_0^2}{(s + \epsilon \omega_0)^2 - \omega_0^2} = \frac{\omega_0^2}{(s + \epsilon \omega_0)^2 - \omega_0^2(\epsilon^2 - 1)} = \frac{1}{s} \cdot \frac{1}{s} \cdot$$

Since, here  $\xi > 1$ , then  $T_1 \ll T_2$ .

As a result the first exponential term decaying much faster than the other exponential term. So, for time response neglect the term having the pole at  $-(\xi + \sqrt{\xi^2 - 1})\omega_n$ .

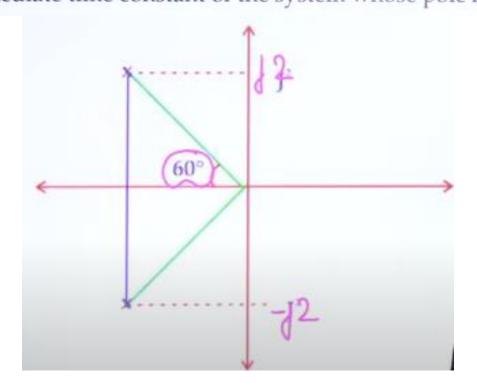


- Finally, the step response for the four cases of damping discussed in the above section are superimposed in figure below.
- Notice that the critically damped caes is the division between the overdamped cases and the underdamped cases and is the fastest response without overshoot.



Ques 1:

Calculate time constant of the system whose pole zero diagram is given.



Solution: This is underdamped system because from the given diagram, it can be seen that both the poles are lying left half of s-plane and both are conjugate complex (2<sup>nd</sup> and 3<sup>rd</sup> quadrant)

$$\cos 60 = \xi \Rightarrow \xi = 0.5$$

$$2 = \omega_n \sqrt{1 - \xi^2}$$

$$\omega_n = 2 / \sqrt{0.75} = 4 / \sqrt{3}$$

$$\text{under damped system}$$

$$T = 1 / (\text{pole}) = 4 / \sqrt{3} / 2 = 0.866$$

$$\text{Re(pole)} = 4 / \sqrt{3} / 2 = 0.866$$

# Ques 2: The transfer function of a system is given as $\frac{100}{s^2+20s+100}$ , This system is

- (a) An over damped system
- (b) An under damped system
- (c) A critically damped system
- (d) An unstable system

$$8^2 + 2\xi \omega_n s + \omega_n^2 = 0$$

$$\omega_n^2 = 100$$

$$\omega_n = 10$$

$$2\xi \omega_n = 20$$

$$\xi = 1 : \text{Critically damped}$$

#### Ques 3:

A unity negative feedback system has an open loop transfer function  $G(s) = \frac{K}{s(s+10)}$ . The gain K for the system to have a damping ratio of 0.25 is \_\_\_\_\_.

#### Solution:

$$T(s) = \frac{K}{s^2 + \log + K}$$

$$\lambda^2 + 2 \delta \omega_n + \omega_n^2 = 0$$

$$k = \omega_n^2$$

$$2 \times 0.25 \times \omega_n = 10$$

$$\omega_n = 20$$

$$k = \omega_n^2 = 400$$

$$k = \omega_n^2$$

$$2 \delta \omega_n = 10$$

#### Ques 4:

The open loop transfer function of a unity feedback control system is given by  $G(s) = \frac{K}{s(s+1)}$ . If the system becomes critically damped, then the system gain 'K' tends to become ......

#### Solution:

$$T(s) = \frac{k}{s^2 + s + k}$$

$$s^2 + 2\xi \omega_n s + \omega_n^2 = 0$$

$$k = \omega_n^2$$

$$2\xi \omega_n = 1$$

$$2\omega_{n} = 1$$

$$\omega_{n} = 0.5$$

$$K = \omega_{n}^{2} = 0.25$$

#### Ques 5:

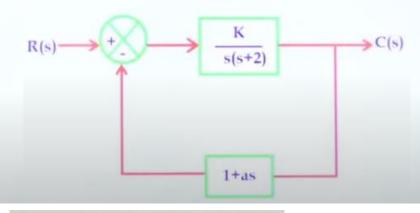
For the system shown in figure with a damping ratio  $\xi$  of 0.7 and an undamped natural frequency  $\omega_n$  of 4 rad /sec, the values of K and a are

(A) 
$$K = 4$$
,  $a = 0.35$ 

(B) 
$$K = 8$$
,  $a = 0.455$ 

(C) 
$$K = 16$$
,  $a = 0.225$ 

(D) 
$$K = 64$$
,  $a = 0.9$ 



#### Solution: (C)

$$T(S) = \frac{s(s+2)}{s(s+2)}$$

$$\frac{1+\frac{K}{s(s+2)}(1+as)}{K}$$

$$\frac{-\frac{K}{s^2+2s+1c+Kas}}{K}$$

$$\frac{-\frac{K}{s^2+(2+Ka)s+kc}}{s^2+(2+Ka)s+kc}$$

$$2 = \omega_n = 2 + Ka - \omega_r$$
 $4 \omega_n^2 = K$ 
 $3 K = 4^2 = 16$ 
 $2 \times 0.7 \times 46 = 2 + 16a$ 

$$\frac{1}{100} = \frac{1.4 \times 4 - 2210}{100}$$

$$= \frac{5.6 - 2}{100}$$

$$= \frac{3.6}{100}$$

$$= \frac{3.6}{100}$$

$$= \frac{3.6}{100}$$

$$= \frac{3.6}{100}$$

Quest- for the unity feedback system having 
$$G(S) = \frac{k}{S(ST+2)}$$
,

Find the fallowing.—

if the factor by which the gain k should be multiplied to increase the damping ratio from 0.15 to 0.6.

(ii) The factor by which the time constant 7 should be multiplied to reduce the damping ratio from 0.8 to 0.4.

Solowing there,  $G(S) = \frac{k}{S(ST+2)}$  and  $H(S) = 1$ 

$$\frac{c(S)}{R(S)} = \frac{G(S)}{1+G(S)\cdot H(S)} = \frac{k}{S^2T+2S+k} = \frac{k/T}{S^2+\frac{L}{T}\cdot S+k/T}$$

Now  $a_n^2 = \frac{k}{T}$  :  $a_n = \sqrt{\frac{k}{T}}$ 

and  $2 \xi a_n = \frac{2}{T}$  :  $\xi = \frac{2}{T\cdot 2}a_n = \frac{1}{TkT}$ 

i) Let  $\xi_1 = \frac{1}{Tk_1T}$ 

if  $\xi_2 = \frac{1}{Tk_2T}$  for  $\xi_3 = 0.15$  and  $\xi_4 = 0.6$  respectively

$$\frac{\xi_1}{\xi_2} = \frac{1}{Tk_2T} = \frac{1}{0.15} = \frac{1}{4}$$

$$\frac{k_2}{R_1} = \frac{1}{16} = \frac{1}{16}$$
 $\frac{k_2}{R_1} = \frac{1}{16} = \frac{1}{16}$ 

the gain must be multiplied by factor 1/16 to increase the damping ratio 0.15 to 0.6.

(ii) Let 
$$\xi_1 = \frac{1}{\sqrt{T_1 K}}$$
 and  $\xi_2 = \frac{1}{\sqrt{T_2 K}}$  for  $\xi_1 = 0.8$  and  $\xi_2 = 0.4$  respectively

$$\xi_1 = \sqrt{\frac{T_2 K}{T_1 K}} = \frac{0.8}{0.4} = 2$$

The time constant of must be multiplied by factor 4 to reduce the damping ratio from 0.8 to 0.4.

## **Assignment:**

- 1. Time Response of the Second Order (2<sup>nd</sup> Order) System for unit Impulse Input.
- 2. Time Response of the Second Order (2<sup>nd</sup> Order) System for unit Ramp Input.