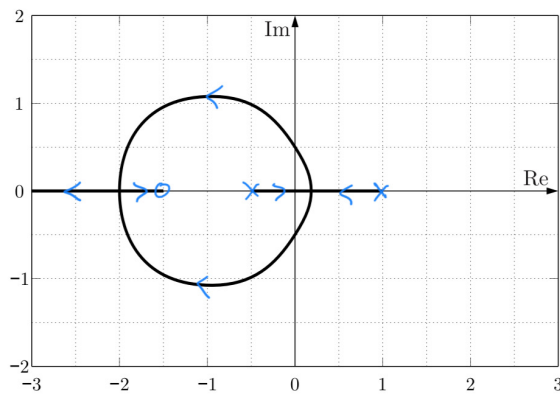


Exam 02-Control Engineering

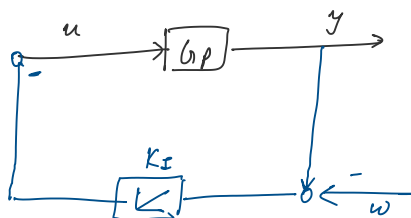
Monday, 10. February 2020 15:45

1.
a

b From figure,
 $s_{N1} = -1.5$
 $s_{p1} = -0.5$ $s_{p2} = 1$
 Now,

$$\begin{aligned}
 G_0(s) &= 1 \cdot \frac{(s+1.5)}{(s+0.5)(s-1)} \\
 &= \frac{1}{2} \left\{ \frac{2s+3}{(s+0.5)(s-1)} \right\} \\
 &= \frac{1}{2} \frac{(s+0.5) + (s-1) + 3.5}{(s+0.5)(s-1)} \\
 &= \frac{1}{2} \left\{ \frac{1}{s-1} + \frac{1}{s+0.5} + \frac{3.5}{(s+0.5)(s-1)} \right\} \\
 &= \frac{1}{2} \left\{ \frac{-1}{1-s} + \frac{2}{1+2s} + \frac{-7}{(1+2s)(1-s)} \right\}
 \end{aligned}$$

$$\begin{aligned}
 g(t) &= \mathcal{L}^{-1}\{G_0(s)\}(t) \\
 &= \frac{1}{2} \left\{ +1e^t + \frac{2}{2}e^{-\frac{t}{2}} + \frac{-7}{3}(e^{-\frac{t}{2}} - e^t) \right\} \\
 &= \frac{1}{2} \left\{ e^t \left(\frac{10}{3} \right) + e^{-\frac{t}{2}} \left(\frac{-4}{3} \right) \right\} \\
 g(t) &= \frac{5}{3}e^t - \frac{2}{3}e^{-\frac{t}{2}}
 \end{aligned}$$

2.
a

b The $G_I(j\omega)$ controller always show integrating behaviour. \therefore It is preferable for stationary accuracy

From Bode plot:
 $\alpha_{-} > \alpha_{+}$

∴ $G_{PD}(j\omega)^{PD}$ is preferable w.r.t phase margin

u) ∴ General form of $G_p = \frac{1}{T_1 T_2 s^2 + (T_2 + T_1)s + 1}$

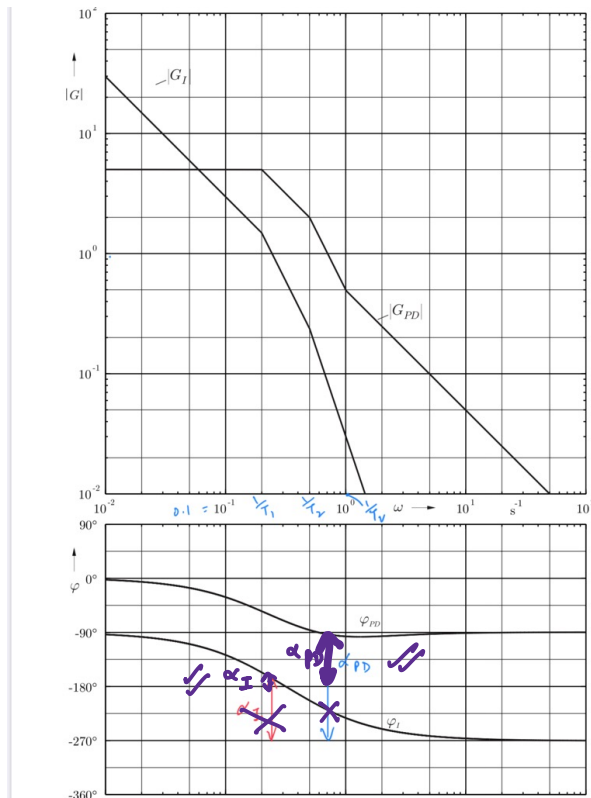
with slope changes of -1 at $\frac{1}{T_1}$ and $\frac{1}{T_2}$

From BODE plot and comparing it w/ given $G_p(j\omega)$

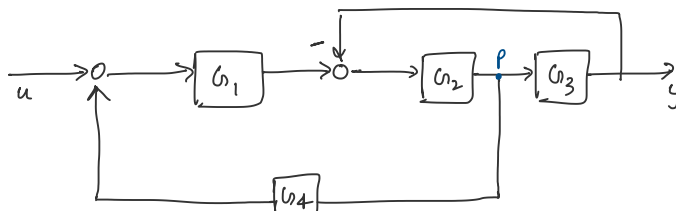
$$\frac{1}{T_1} = 0.2 \text{ sec}^{-1} \Rightarrow T_1 = 5 \text{ sec}$$

$$\frac{1}{T_2} = 0.5 \text{ sec}^{-1} \Rightarrow T_2 = 2 \text{ sec}$$

∴ $a_2 = T_1 T_2 = 10 \text{ sec}^2$ ✓
 $a_1 = (T_1 + T_2) = 7 \text{ sec}$ ✓



3



$$\{(U + G_4 P) G_1 - Y\} G_2 G_3 = Y$$

also, $Y = G_3 P \Rightarrow P = \frac{Y}{G_3}$

$$\text{Now, } \{(U + \frac{G_4 Y}{G_3}) G_1 - Y\} G_2 G_3 = Y$$

$$((\quad G_3 \quad / \quad) \quad)$$

$$Y(1 + G_2 G_3 - G_1 G_2 G_4) = U G_2 G_3 G_1$$

$$G(s) = \frac{Y(s)}{U(s)} = \frac{G_2 G_3 G_1}{1 + G_2 G_3 - G_1 G_2 G_4}$$

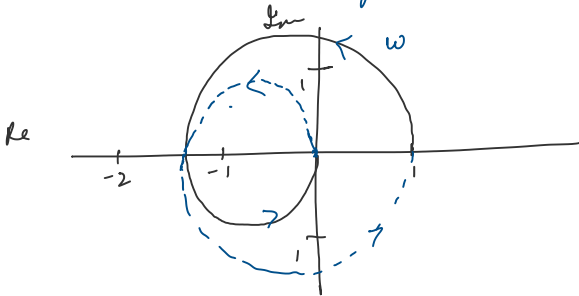
$$b. \quad G(s) = \frac{G_2 G_3}{1 + G_2 G_3}$$

here, $G_0(s) = G_2 G_3$. (say)

Now,

$\therefore G_2$ and G_3 both have one pole in right half plane.

\therefore Total pole in right half plane $= 2 = p$



\therefore 2 rounds anticlockwise about -1

$\therefore m = -2$, no. of revolutions of C' around -1 (opposite mathematically to the direction i.e. clockwise)

Now, $m = n - p$

$$n = m + p = -2 + 2 = 0$$

\therefore there are 0 zeros in $N(s)$ inside C (right s halfplane)

or 0 no. of poles of $G_2(s)$ in right s half plane. or C'

and, $G_2(s)$ is stable if $n \rightarrow 0$

\therefore In our case $G(s)$ is stable

4)

a) $|sI - A| = 0$ For characteristic eqⁿ

$$\Rightarrow \begin{vmatrix} s+1 & 5 \\ -1 & s-3 \end{vmatrix} = 0$$

$$\Rightarrow s^2 + s - 3s - 3 + 5 = 0$$

$$s^2 - 2s + 2 = 0$$

$$(s-1)^2 = 0 \quad s = \frac{2 \pm \sqrt{4-0}}{2} = 1 \pm i$$

$$s = 1$$

\therefore poles lie in right half plane

\therefore system is ~~stable~~ unstable.

$$b) \quad s_1 = -1 + j \quad s_2 = -1 - j$$

Characteristic eqⁿ

$$\begin{aligned} (s - s_1)(s - s_2) &= 0 \\ \{s - (-1 + j)\} \{s - (-1 - j)\} &= 0 \\ \{s + 1 - j\} \{s + 1 + j\} &= 0 \\ (s + 1)^2 - j^2 &= 0 \\ s^2 + 1 + 2s + 1 &= 0 \\ s^2 + 2s + 2 &= 0 \quad \text{--- (1)} \end{aligned}$$

Now, For given value of $U = -KX$

$$\begin{aligned} \dot{X} &= AX + B - KX \\ &= (A - BK)X \\ &= \left\{ \begin{pmatrix} -1 & -5 \\ 1 & 3 \end{pmatrix} - \begin{pmatrix} k_1 & k_2 \\ 0 & 0 \end{pmatrix} \right\} X \\ &= \underbrace{\begin{pmatrix} -1 - k_1 & -5 - k_2 \\ 1 & 3 \end{pmatrix}}_{A_k} X \end{aligned}$$

Now, Characteristic eqⁿ

$$\begin{aligned} |sI - A_k| &= 0 \\ \begin{vmatrix} s + 1 + k_1 & 5 + k_2 \\ -1 & s - 3 \end{vmatrix} &= 0 \end{aligned}$$

$$s^2 + s(1 + k_1) - 3s - 3(1 + k_1) + (5 + k_2) = 0$$

$$s^2 + s(k_1 - 2) + (2 - 3k_1 + k_2) = 0 \quad \text{--- (2)}$$

Comparing eq (1) and (2)

$$\begin{aligned} 2 &= k_1 - 2 \\ \boxed{4} &= k_1 \end{aligned}$$

$$\begin{aligned} 2 - 3 \cdot 4 + k_2 &= 2 \\ \boxed{k_2} &= 12 \end{aligned}$$