# Tutorial 6 Discrete Time Signals and Fourier Series

2014/1

## Sampling of Signals

• Suppose we have an analog signal and we wish to create a discrete version of it.

When the rate is not quick enough

when the rate is too fast

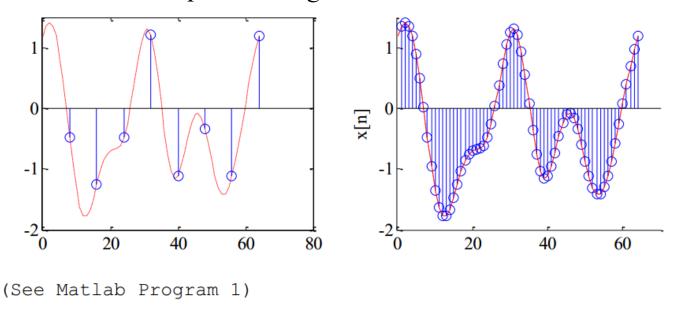


Figure 1 - Continuous and a discrete signal

• There is an optimum sampling rate which captures enough information without overdoing it.

## Sampling of Signals

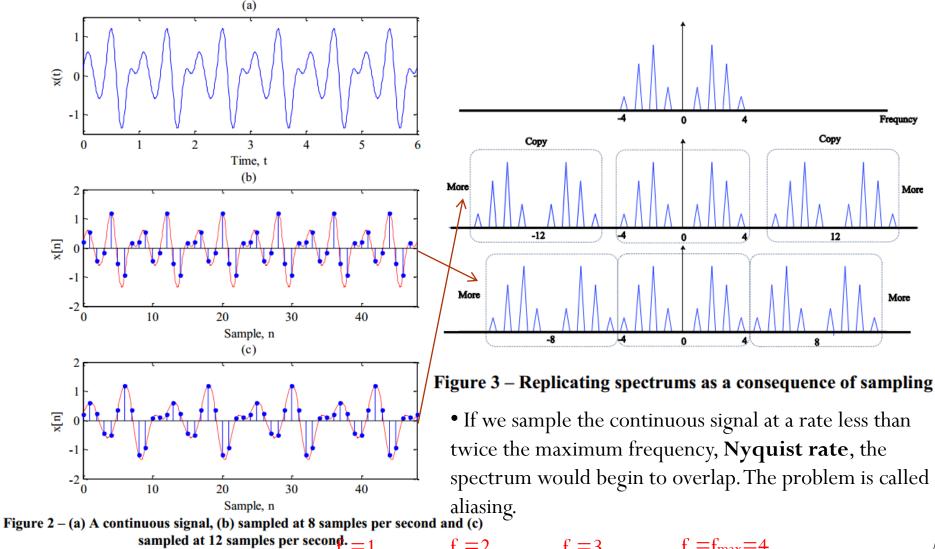
- Sampling Theorem
  - For any analog signal containing among its frequency contents a maximum frequency of fmax, the underlying signal can be represented faithfully by N equally spaced samples, provided the sampling rate is at least two times fmax samples per second.
- A Maximum Sampling Period  $T_s = \frac{1}{2f_{\text{max}}}$  seconds
- Sampling Frequency  $F_s = \frac{1}{T_s} = 2f_{\text{max}}$  samples/second

$$f_{\text{max}} = \frac{1}{2T_s} Hz$$

$$\omega_{\text{max}} = 2\pi f_{\text{max}} = \frac{\pi}{T_{\text{s}}} radians / \sec$$

## Sampling of Signals

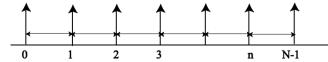
(See Matlab Program 2)



 $x(t) = .25\sin(2\pi t) + .7\cos(4\pi t) - .5\cos(6\pi t) + .15\sin(8\pi t)$ 

## Specifying a Discrete Signal

- If a continuous signal is referred to as x(t) then a discrete sampled signal is written as  $x(kT_s)$ 
  - K=0,  $\pm 1$ ,  $\pm 2$ ,...: the sample number



• T<sub>s</sub>: the sampling period

Figure 4 – Discrete signal samples

- To create a discrete signal from a continuous signal
  - 1. the continuous signal is multiplied by an impulse train of the sampling period.
  - 2. multiply the sampled signal once again by an impulse train, point by point.

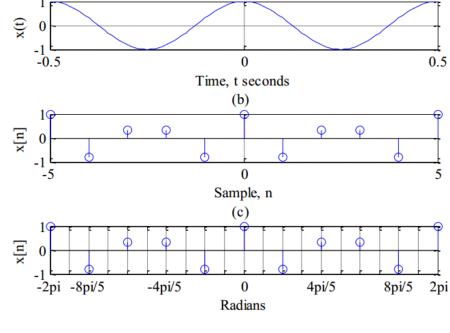
    "combing" the signal to create

$$x(kT_s) = x(t) \times \sum_{k=-\infty}^{\infty} \underbrace{\delta(t - kT_s)}^{\text{a discrete signal.}}_{\text{continuous}}$$

$$x[k] = \sum_{k=-\infty}^{\infty} x(kT_s)\delta(t - kT_s)$$
 (discrete)

## Specifying a Discrete Signal

- Two ways to specify a sampled signal
  - By sample numbers. (Figure 5(b))
  - By its phase. (Figure 5(c))
- Example  $x(t) = \sin(2\pi f_0 t), f_0 = 1.$   $x[-10] = \sin[2\pi(-10 \div 5)] = 0$   $x[-9] = \sin[2\pi(-9 \div 5)] = \sin[-3.6\pi] = .951$   $x[-8] = \sin[2\pi(-8 \div 5)] = \sin[-3.2\pi] = .588$   $x[-7] = \sin[2\pi(-6 \div 5)] = \sin[-2.4\pi] = -.951$   $x[-6] = \sin[2\pi(-5 \div 5)] = \sin[-2\pi] = 0$  $x[-5] = \sin[2\pi(-4 \div 5)] = \sin[-1.6\pi] = .951$



(a)

 $Figure \ 5-Sampling \ of \ a \ continuous \ signal$ 

(Matlab Program 3)

## Parameters of a Discrete Signal

- To create a discrete signal, we sample this signal with sample time of Ts. The Sampling Period Ts is independent of the fundamental period To of the continuous signal and no larger than the Nyquist threshold ½ fmax.
- If we know the signal frequency fo and the sampling frequency Fs, we can also write the signal this way replacing Ts with 1/Fs.

$$x[k] = \sin \left[ \left( \frac{2\pi f_0}{F_s} \right) k \right]$$

• Digital Frequency (phase advance)
$$\Omega_0 = \frac{2\pi f_0}{F_s} \frac{radians \times cycles / \sec ond}{sample / \sec ond} \rightarrow \frac{radians}{sample}$$

- each sample moves the signal by the value of digital frequency.
- Ex. a signal of frequency 10 Hz sampling with a sampling frequency Fs = 30 Hz, then its digital frequency  $\Omega_0 = \frac{2\pi \times 10}{30} = 2\pi/3$

## Parameters of a Discrete Signal

- K<sub>0</sub>: a ratio of the sampling frequency to the fundamental frequency or the maximum frequency.
  - ex.  $K_0 = \frac{F_s}{f_0} = \frac{30}{10} = 3$
- If there are Ko samples in one period, then Ko times the digital frequency  $\hat{\Omega}$  must equal  $2\pi$ . We define the period of the signal in samples as:  $K_0\Omega_0 = 2\pi$  or  $K_0 = \frac{2\pi}{\Omega_0}$ .

  • The fundamental frequency is given by  $\Omega_0 = \frac{2\pi}{K_0}$ .
- Suppose the underlying signal is periodic, the sampled discrete signal is also periodic if and only if N is an integer.

$$N=mK_0=mrac{T_0}{T}=mrac{F_s}{f}$$
 fo

• ex. A signal with fundamental frequency of 5 which is sampled at a rate of 20, will have a fundamental period of 4 and as such this sampling rate would result in a periodic discrete signal.

- The definition of a periodic signal for a discrete signal is  $x[k] = x[k+K_0]$ 
  - Ex. Show that the periodicity of a discrete sinusoid  $\sin(\Omega n)$ .  $\sin(k\Omega) = \sin(\Omega(k+K_0)) = \sin(k\Omega+K_0\Omega)$

only if  $K_{\circ}\Omega = 2\pi$ , the equation holds.

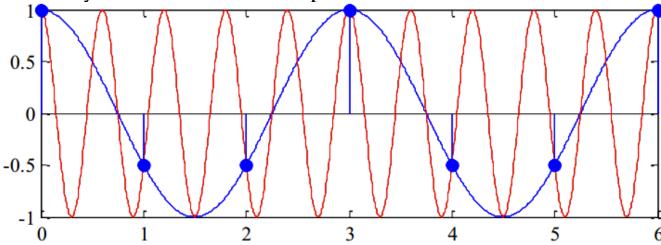


Figure 6 – A discrete signal can represent any number of continuous signals.

• Example 3-1

What is the digital frequency of this signal? What is its period?  $x[k] = \cos \left[ \frac{2\pi}{3} k + \frac{\pi}{3} \right]$ 

- The digital frequency  $\Omega_0 = 2\pi/3$
- Its period  $K_0 = 2\pi/(2\pi/3) = 3$

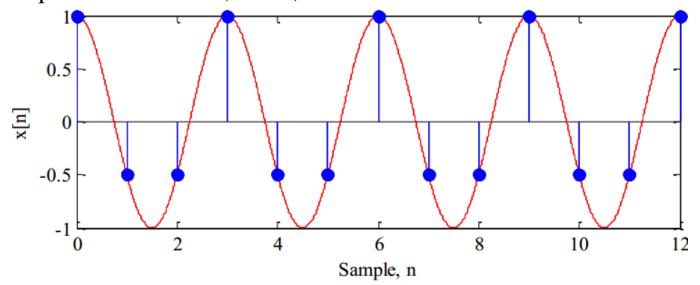


Figure 7 – Discrete signal of example 3-1

• Example 3-2

What is the period of this discrete signal? Is it periodic?

$$f[k] = \sin\left(\frac{3\pi k}{4} + \frac{\pi}{4}\right)^{\kappa}$$

- The digital frequency  $\Omega_0 = 3\pi/4$
- The period of the signal is:  $3\pi m = m4 \rightarrow K_0 = 8$ , m=2

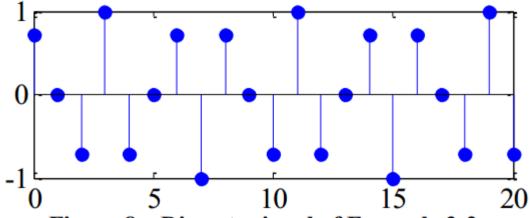


Figure 8 – Discrete signal of Example 3-2

• Example 3-3

Is this discrete signal  $f[n] = \sin[.5n + \pi]$  periodic?

- The fundamental frequency is 0.5.
- $\bullet K_0 = \frac{2\pi}{\Omega_0} k = 12\pi k$
- The continuous signal is periodic but the discrete samples of it is not because  $K_0$  is not a ratio of integers.

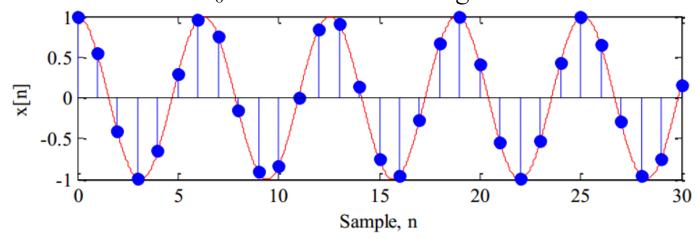


Figure 9 – Non-periodic discrete signal of example 3-3.

(Matlab Program 6)

- A Fourier series representation for discrete time signals
  - The <u>discrete</u> complex exponential is written by replacing t with k. We can write this in terms of the digital frequency as:  $e^{j\omega t} \quad \text{continuous signal}$
  - In the discrete case, the harmonic relationship is based on phase:

$$\Omega + 2\pi k \rightarrow \Omega$$

• Every time increases by  $2\pi$ , we get a new complex exponential given by:

$$e^{j(\Omega+2\pi)k} = e^{j\Omega n} \underbrace{e^{j2\pi k}}_{e^{j2\pi k}} = e^{j\Omega k}$$
$$e^{j2\pi k} = \cos(2\pi k) - \underbrace{j\sin(2\pi k)}_{=0} = \cos(2\pi k) = 1$$

- Example 3-4
  - Show harmonics of the exponential  $e^{j\frac{2\pi}{3}t}$  if it is being sampled with sampling period of 0.25 seconds.
    - We can write the exponential in discrete form by replacing t with  $kT_s=k/4$

$$y[k] = e^{j\frac{2\pi}{12}k}$$

• Let's plot the real part of this signals along with its next two harmonics, which are  $e^{\int \left(\frac{2\pi}{12}+2\pi\right)^k}$  and  $e^{\int \left(\frac{2\pi}{12}+4\pi\right)^k}$ 

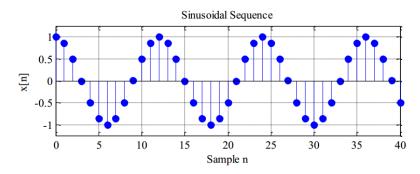


Figure 10 – Three discrete harmonic signals

- Let's see what happens as digital frequency  $\Omega$  is varied just within the 0 to  $2\pi$  range.
  - within the 0 to  $2\pi$  range.

     Take the signal  $x[k] = e^{j\frac{2\pi}{6}k}$ . Its digital frequency is equal  $2\pi/6$  and its period  $K_0$  is equal to 6.
  - We will increase the frequency of this signal in 6 steps.

$$\phi_0 = 2\pi (n = 0) / 6 = 0$$

$$\phi_1 = 2\pi (n = 1) / 6 = 2\pi / 6$$

$$\phi_2 = 2\pi (n = 2) / 6 = 4\pi / 6$$

$$\vdots$$

$$\phi_5 = 2\pi (n = 5) / 6 = 10\pi / 6$$

• The variable n steps from 0 to  $K_0$ -1. There are N harmonics, and we index them with letter n. Index k remains the index of the sample.

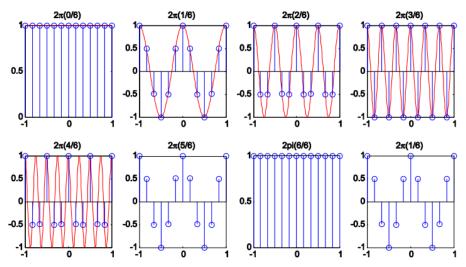


Figure 11 – Discrete signals in between harmonic frequencies

(Matlab Program 8)

- The analog signals are harmonic along the frequency axis whenever  $\omega_k = k\omega_0$ .
- Discrete signals are harmonic in between these values when specified in terms of phase.
- if we can do the orthogonality test, and we find that they are indeed harmonic to each other.

$$\sum_{n=0}^{\infty} \phi_n \, \phi_2^* = 0$$

- Key ideas about discrete signals
  - 1. We do not know the underlying analog signal nor its frequency. All we know is that if the sampling frequency is Fs, then we can from a discrete signal unambiguously extract signals of only of frequency half as much.
  - 2. A discrete signal is defined by its digital frequency. The units of digital frequency are in radians per sample. We can think of it as being defined over a circle from 0 to  $2\pi$  (or  $-\pi$  to  $+\pi$ )
  - 3. A discrete signal of frequency  $\Omega_0$  is exactly the same as all its harmonics when  $\Omega_0 \pm 2\pi k$  for all k.
  - 4. There are only N distinct discrete—time complex exponential signals that are harmonically related for any given period N. K<sub>0</sub> is the smallest such number and called the fundamental period.

• If  $\Omega_0 = \frac{2\pi}{5}$ , we can write the discrete harmonic complex exponentials as:

$$e^{-j\Omega_{0}^{n}k},$$

$$e^{-j\left(\frac{2\pi}{n}\times 0\right)k}, e^{-j\left(\frac{2\pi}{n}\times 1\right)k}, e^{-j\left(\frac{2\pi}{n}\times 2\right)k}, e^{-j\left(\frac{2\pi}{n}\times 3\right)k}, \cdots, e^{-j\left(\frac{2\pi}{n}\times (K_{0}^{-1})\right)k}$$

- The index n is used to indicate the harmonics. The index k is the time sample.
- The discrete-time representation of the signal is written as the weighted sum of these.  $x[k] = \sum_{k=0}^{K_0-1} C_n e^{j(n\Omega_0)k}$

$$x[k] = \sum_{k=0}^{K_0-1} C_n e^{j(n\Omega_0)k}$$

• The complex coefficients,  $C_0, C_1, \dots, C_{(K_0-1)}$  are given as

$$C_n = \frac{1}{K_0} \sum_{n=0}^{K_0 - 1} x[k] e^{-j(n\Omega_0)k}$$

- we show that the DTFS coefficients of the nth harmonic are exactly the same as the coefficient for a harmonic that is an integer multiple of mK<sub>0</sub> samples away so that:  $C_n = C_{(n+mK_0)}$ 
  - Here m is an integer. The nth coefficient is equal to

$$C_n = \sum_{k=0}^{N-1} x[n]e^{-j n\Omega_0 k}$$

• The  $(n+mK_0)$  coefficient is given by

$$C_{(n+mK_0)} = \sum_{k=0}^{N-1} x[k] e^{-j(n+mK_0)\Omega_0 k} = \sum_{k=0}^{N-1} x[k] e^{-jn\Omega_0 k} e^{-jmK_0\Omega_0 k} \\ e^{-jmK_0\Omega_0 k} = e^{-jm2\pi k} = 1$$

• So we have:  $C_{(n+mK_0)} = \sum_{k=0}^{N-1} x[k]e^{-jn\Omega_0 k} = C_n$ 

• Example 3-5

Find the discrete time Fourier series coefficients of this signal.

$$x[k] = 1 + \sin\left(\frac{2\pi}{10}k\right)$$

• The fundamental period  $K_0 = 10$ 

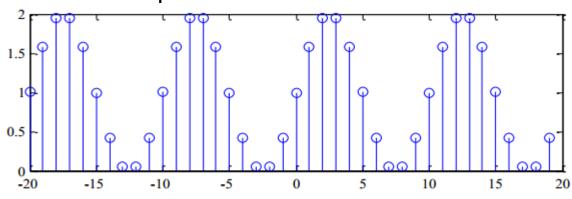


Figure 12 – Signal of example 3-5

• the Euler equivalent expression for this signal

$$x[k] = 1 + \frac{1}{2j} e^{j\left(\frac{2\pi}{10}\right)^k} - \frac{1}{2j} e^{-j\left(\frac{2\pi}{10}\right)^k}$$

- Example 3-5
  - $C_n = \frac{1}{K_0} \sum_{k=0}^{K_0 1} x[k] e^{-j(n\Omega_0)k} = \frac{1}{10} \sum_{k=0}^{9} x[k] e^{-j(\frac{2\pi}{10}n)k}$
  - $\begin{array}{c} \bullet \\ x[k] = [1.0000, \, 1.5878, \, 1.9511, \, 1.9511, \, 1.5878, \, 1.0000, \, 0.4122, \, 0.0489, \, 0.0489, \, 0.4122] \end{array}$
  - $C_n = C_0 = \frac{1}{10} \sum_{k=0}^9 x[k] \underbrace{e^{-j \ o \, \Omega_0 \ k}}_{10} = \frac{1}{10} \sum_{k=0}^9 x[k] = 1$
  - $\begin{array}{c} \bullet \ \, C_1 = \frac{1}{10} \sum_{k=0}^9 x[k] e^{-j \, 1 \cdot \Omega_0 \, k} = \frac{1}{2j} \\ C_{-1} = \frac{1}{10} \sum_{k=0}^9 x[k] e^{j \, 1 \cdot \Omega_0 \, k} = -\frac{1}{2j} \end{array}$
  - The rest of the coefficients from C<sub>2</sub> to C<sub>9</sub> are zero.
  - the coefficients repeat after  $C_9$  so that  $C_{(1+9k)} = C_1$ .

• Example 3-6

Compute the DTFSC of this discrete signal.

$$x[k] = 0.5 + 0.25 \cos\left(\frac{2\pi}{5}k\right) - 0.6 \sin\left(\frac{2\pi}{4}k\right)$$

- The period, K<sub>0</sub> of the cosine is 5 and the period, K<sub>0</sub> of sine is 4. Period of the whole signal is 20 (least common multiple).
- The fundamental frequency  $\Omega_0 = \frac{2\pi}{20} = \frac{\pi}{10}$ •  $C_n = \frac{1}{K_0} \sum_{n=0}^{K_0-1} x[k] e^{-j(n\Omega_0)k}$

• 
$$C_n = \frac{1}{K_0} \sum_{n=0}^{K_0 - 1} x[k] e^{-j(n\Omega_0)k}$$
  
 $= \frac{1}{20} \sum_{n=0}^{19} x[k] e^{-j(\frac{\pi}{10}n)k} C_0 = C_n = 0.5$ 

- $x[k] = \frac{1}{20} \left[ x[0]e^{-j(2\pi/5)} \right] C_4 = .125 C_{-4} = .125$  $C_5 = .3j C_{-5} = -.3j$
- $x[k] = 0.5 + 0.125 \left( e^{\left(\frac{2\pi}{5}k\right)} + e^{-\left(\frac{2\pi}{5}k\right)} \right) + 0.3j \left( e^{\left(\frac{2\pi}{4}k\right)} e^{-\left(\frac{2\pi}{4}k\right)} \right)$

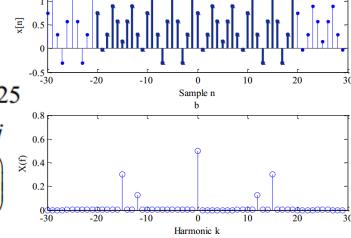


Figure 13- Signal of example 3-6

• Example 3-7

Compute the DTFS of this periodic discrete signal. The signal repeats with period 4 and has two impulses of amplitude 2 and 1.

• The fundamental frequency  $\Omega_0 = \frac{2\pi}{4} = \frac{\pi}{2}$ 

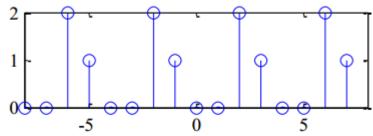


Figure 14 – Signal of example 3-7

• We write the expression for the DTFSC  $C_n = \sum_{k=0}^{3} x[k]e^{-j\left(\frac{\pi}{2}n\right)k}$ 

$$e^{-j\left(\frac{\pi}{2}\right)nk} = \left[\cos\left(\frac{\pi}{2}\right) - j\sin\left(\frac{\pi}{2}\right)\right]^{kn} = (-j)^{nk}$$

- Example 3-7
  - Now substitute this into the DTFSC equation and calculate the coefficients, knowing there are only n = 4 harmonics in the signal because the number of harmonics are equal to the fundamental period of the signal.

$$C_n = \frac{1}{4} \sum_{k=0}^{3} x[k] (-j)^{kn}$$

$$\begin{array}{ll} \bullet & C_0 = \frac{1}{4} \ 2 + 1 \ = \frac{3}{4} & for \ n = 0, \, k = 0, 1, 2, 3 \\ C_1 = \frac{1}{4} (2 - j 1) = \frac{1}{2} - \frac{j}{4} & for \ n = 1, \, k = 0, 1, 2, 3 \\ C_2 = \frac{1}{4} \ 2 - 1 \ = \frac{1}{4} & for \ n = 2, \, k = 0, 1, 2, 3 \\ C_3 = \frac{1}{4} \ 2 + j 1 \ = \frac{1}{2} + \frac{j}{4} & for \ n = 2, \, k = 0, 1, 2, 3 \end{array}$$

- Example 3-7
  - We can setup the DTFSC equation in matrix form by setting the basic exponential to a constant and then writing it in terms of two variables, the index n and k.

$$e^{j\Omega_0}=W \ e^{-jn\Omega_0 k}=W^{-nk}$$

$$e^{-J^{m_{0}n}} = W^{-m_{0}}$$
 
$$C_{n} = \frac{1}{K_{0}} x[k] \begin{bmatrix} W^{-0 \times 0} & W^{-1 \times 0} & W^{-2 \times 0} & W^{-3 \times 0} \\ W^{-0 \times 1} & W^{-1 \times 1} & W^{-2 \times 1} & W^{-3 \times 1} \\ W^{-0 \times 2} & W^{-1 \times 2} & W^{-2 \times 2} & W^{-3 \times 2} \\ W^{-0 \times 3} & W^{-1 \times 3} & W^{-2 \times 3} & W^{-3 \times 3} \end{bmatrix}_{n \times k}$$

• We used Matlab to compute the coefficients.

$$\begin{bmatrix} 2 & 1 & 0 & 0 \end{bmatrix} \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & 1 \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{bmatrix} = \begin{bmatrix} .75 & .5 - .25j & .25 & .5 + .25j \end{bmatrix}$$

• Example 3-8

Find the discrete-time Fourier series coefficients of this signal.

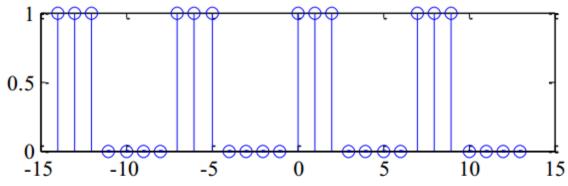


Figure 15 – Signal of example 3-8 with N = 3. K0 = 7

$$\bullet \ \, C_n = \frac{1}{K_0} \sum_{k=-N}^N 1 \cdot e^{-jn\Omega_0 k} \, + \frac{1}{K_0} \sum_{k=N+1}^{K_0-N-1} 0 \cdot e^{-jn\Omega_0 k} = \frac{1}{K_0} \left[ e^{-jn\Omega_0 N} \, \frac{1 - e^{-jn\Omega_0 (2N+1)}}{1 - e^{-jn\Omega_0}} \right]$$

$$\quad \bullet \ \, C_{\scriptscriptstyle n} = \frac{1}{K_{\scriptscriptstyle 0}} \bigg[ \frac{\sin \ Nn\pi \ / \ K_{\scriptscriptstyle 0}}{\sin \ n\pi \ / \ K_{\scriptscriptstyle 0}} \bigg]$$

• Example 3-8

Matlab program 12

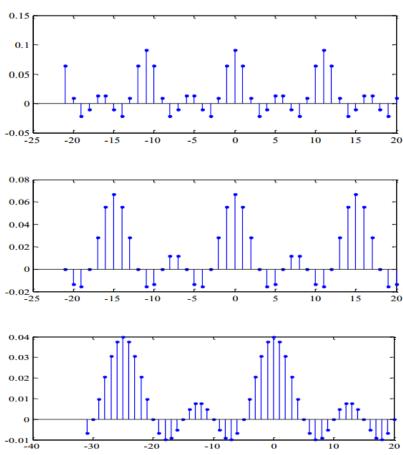


Figure 16 – Coefficients of the periodic pulses, (a) with N=3, K0=7, (b) N=3, K0=15, (c) N=3, K0=2

## Summary

- A discrete signal can be created by sampling a continuous signal with an impulse train of desired sampling frequency.
- The sampling frequency should be greater the two times the highest frequency in the signal of interest.
- The fundamental period of a discrete signal, given by K<sub>0</sub> must be an integer for the signal to be periodic.
- The fundamental discrete frequency of the signal, given by  $\Omega_{^0}$  is equal to  $2\pi/K_{^0}$
- The period of a digital frequency is an integer multiple of. Harmonic discrete frequencies vary by integer multiple of  $2\pi$ , such that  $\Omega$  and  $\Omega+2\pi k$  are harmonic and identical.
- Because discrete harmonic frequencies are identical, we cannot use them to represent a discrete signal.

## Summary

- In stead we divide the range from 0 to  $2\pi$  by  $K_0$  and use these digital frequencies as the basis set.
- Hence there are only  $N = K_0$  harmonics available to represent a discrete signal. The Fourier analysis is limited to these N harmonics.
- Beyond the  $2\pi$  range of harmonic frequencies, the discrete-time Fourier series coefficients, (DTFSC) repeat.
- In contract, the continuous-time signal coefficients are periodic and do not repeat.
- Sometimes we can solve the coefficients using closed from solutions but in a majority of the cases, matrix methods are used to find the coefficients of a signal.
- Matrix method is easy to setup but is computationally intensive.

```
%Program Chapter 3 - Program 1
f0=1;
Fs = 32;
Ts1 = 1/Fs;
t = 0: Ts1: 2;
clf:
figure(1) % heavy sample
xt = cos(2*pi*t) - .3 + .6*sin(3*pi*t+.5) + .5*cos(4*pi*t) -
.3*\cos(5*pi*t+.25);
ylabel('x[n]');
xlabel('Sample');
hold on
plot(t/Ts1,xt,'-. r');
n = 0: 2*Fs;
xn1=cos(2*pi*n*Ts1) - .3 + .6*sin(3*pi*n*Ts1+.5) + .5*cos(4*pi*n*Ts1) -
.3*\cos(5*pi*n*Ts1+.25);
stem(n, xn1);
hold off
figure(2) % light sample
plot(t/Ts1,xt,'-. r');
hold on
n = 0: 2*4;
Ts1 = 1/4;
xn1=cos(2*pi*n*Ts1) - .3 + .6*sin(3*pi*n*Ts1+.5) + .5*cos(4*pi*n*Ts1) -
.3*\cos(5*pi*n*Ts1+.25);
stem(n*8, xn1);
ylabel('x(t)');
xlabel('Sample');
hold off
```

%Chapter 3 - Program 2

```
axis([ 0 48 -2 2]);
t = 0: .01: 6;
                                                    hold off
x = .25*sin(2*pi*1*t) + .7*cos(2*pi*2*t) -
.5*\cos(2*pi*3*t) + .15*\sin(2*pi*4*t);
clf;
                                                    figure (3)
figure(1);
                                                    n2 = 0: fs2*6-1;
                                                    fs2 = 12;
plot(t, x)
                                                    xn12 = .25*sin(2*pi*1*n2/fs2)+.7*cos(2*pi*2*n2/fs2)-
title('(a)')
                                                     .5*\cos(2*pi*3*n2/fs2)+.15*\sin(2*pi*4*n2/fs2);
ylabel('x(t)')
                                                    plot(t*12, x , '--r')
xlabel('Time, t')
                                                    ylabel('x[n]')
                                                    xlabel('Sample, n')
figure (2);
                                                    title('(c)')
n = 0: 47;
                                                    hold on
fs1 = 8:
                                                    stem(n2, xn12, '.')
xn8 = .25*sin(2*pi*1*n/fs1)+.7*cos(2*pi*2*n/fs1)-
                                                    axis([ 0 48 -2 2]);
.5*\cos(2*pi*3*n/fs1)+.15*\sin(2*pi*4*n/fs1);
                                                    hold off
plot(t*8,x, '--r')
ylabel('x[n]')
                                                    figure (4);
xlabel('Sample, n')
                                                    n = 0: 47;
title('(b)')
                                                    clf;
hold on
                                                    xnd = (1/48) * fft (xn8);
stem(n, xn8, '.')
                                                    xnd2 = abs(fftshift(xnd));
                                                    plot(n, xnd2)
```

```
title('(c) ');
%Chapter 3 - Program 3
                            xlabel('Radians');
t = -.5: .01: .5;
                            vlabel('x[n]');
v1 = cos(4*pi*t);
                            axis([-2*pi 2*pi -1. 1.])
clf;
                            % Define x-ticks and their labels..
                            set(gca, 'xTick', -2*pi: pi/5: 2*pi)
                            set(gca,'xTickLabel',{'-2pi', '', '-8pi/5', '', '', '-4pi/5', '',
subplot(3,1,1)
                            '', '', '0', '', '', '4pi/5', '', '', '', '8pi/5', '', '2pi'})
plot(t, y1)
grid;
title('(a) ');
xlabel('Time, t seconds');
vlabel('x(t)');
axis:
subplot(3,1,2)
n = -5:5;
y2 = cos(4*pi*n*.2);
stem(n, y2)
grid;
title('(b) ');
xlabel('Sample, n');
vlabel('x[n]');
subplot(3,1,3)
%n = -5:5;
n2 = -2*pi: 2*pi/5: 2*pi;
v2 = cos(2*n2);
stem(n2, y2)
axis([-2*pi 2*pi -1 1]);
grid;
```

```
%Chapter 3 - Program 4

f0=2;
Fs = 6;
t = 0: .001: 1;
n = 0: Fs*t;
n2 = 0: Fs

xt1=cos(2*f0*pi*t);
y = cos(2*f0*pi*n2/Fs)
xt2= cos(2*5*pi*f0*t);
figure(1)

plot(t*Fs, xt1, t*Fs, xt2, 'r')
hold on
stem(n2, y, 'filled')
```

```
%Chapter 3 - Program 5
f0=1;
Fs = 3;
t = 0: .001: 4;
n = 0: Fs*t;
n2 = 0: Fs*4
xt1=cos(2*f0*pi*t);
y = cos(2*f0*pi*n2/Fs)
figure(1)
grid;
plot(t*Fs, xt1, 'r')
xlabel('Sample, n')
ylabel('x[n]')
hold on
stem(n2, y, 'filled')
```

```
%Chapter 3 - Program 6
f0=.5/pi;
Fs = 1;
t = 0: .001: 30;
n2 = 0: Fs*30
xt1=cos(2*f0*pi*t);
y = \cos(2*f0*pi*n2/Fs)
 figure(1)
grid
plot(t*Fs, xt1, 'r')
xlabel('Sample, n')
ylabel('x[n]')
hold on
stem(n2, y, 'filled')
```

```
%Chapter 3 - Program 7
n = 0:40;
w = 2*pi/12;
phase = 0;
A = 1.0;
HShift = 2; %change this (even numbers only) to see effect of shift
x = A*cos((w+(HShift*pi))*n - phase);
clf;
stem(n,x, 'filled'); % Plot the generated sequence
axis([0 40 -1.25     1.25]);
grid;
title('Sinusoidal Sequence');
xlabel('Sample n');
ylabel('x[n]');
axis;
```

axis([-12.5 12.5 -1.1 1.1]);

subplot(3,4,7);stem(n,real(Phi6n),'Marker','.');xlabel('7')

subplot(3,4,8);stem(n,real(Phi7n),'Marker','.');xlabel('8')

subplot(3,4,9);stem(n,real(Phi8n),'Marker','.');xlabel('9')

subplot(3,4,10);stem(n,real(Phi9n),'Marker','.');xlabel('10')

subplot(3,4,11);stem(n,real(Phi10n),'Marker','.');xlabel('11')

subplot(3,4,12);stem(n,real(Phi11n),'Marker','.');xlabel('12')

Phi6n=exp(j\*w0\*k\*n);

Phi7n=exp(j\*w0\*k\*n);

Phi8n=exp(j\*w0\*k\*n);

Phi9n=exp(j\*w0\*k\*n);

Phil0n=exp(j\*w0\*k\*n);

Philln=exp(j\*w0\*k\*n);

k=6:

k=7:

k=8:

k=9:

k=10;

k=11;

```
%Chapter 3 - Program 8

Phi3n=exp(j*w0*k*n);
subplot(3,4,4); stem(n,real(Phi3n),'Marker','.'); xlabel('4')
axis([-12.5 12.5 -1.1 1.1]);
k=4;
Phi4n=exp(j*w0*k*n);
subplot(3,4,5); stem(n,real(Phi4n),'Marker','.'); xlabel('5')
axis([-12.5 12.5 -1.1 1.1]);
k=5;
Phi5n=exp(j*w0*k*n);
subplot(3,4,6); stem(n,real(Phi5n),'Marker','.'); xlabel('6')
```

k=0;

k=1;

hold on

hold off

k=2:

Phi0n=exp(j\*w0\*k\*n);

stem(n, real(PhiOn), 'Marker', '.'); xlabel('1')

subplot(3,4,2);stem(n,real(Philn),'Marker','.');xlabel('2')

subplot(3,4,3); stem(n,real(Phi2n),'Marker','.'); xlabel('3')

subplot (3,4,1);

t= -10: 1/18: 10;

plot(t, cos(w0\*k\*t))

Philn=exp(j\*w0\*k\*n);

plot(t, cos(w0\*k\*t))

Phi2n=exp(j\*w0\*k\*n);

t = -10: 1/18: 10;

axis([-12.5 12.5 -1.1 1.1]);

axis([-12.5 12.5 -1.1 1.1]);

axis([-12.5 12.5 -1.1 1.1]);

```
% Chapter 3 - Program 9
nmin = -10;
nmax = 9;
ND = abs(nmin) + nmax + 1;
n = nmin: nmax;
x1 = 0.5 + 0.25*\cos(2*pi*n/5) - 0.6*\sin(2*pi*n/4);
clf
subplot(2,1,1)
stem(n,x1, '.');
pt = sum(x1.^2)*1/20
x1
title('a')
ylabel('x[n]')
xlabel('Sample n')
xnd = (1/ND)*dft(x1, ND);
subplot(2,1,2)
pt = sum(x1.^2)*1/20
\times 1
title('a')
ylabel('x[n]')
xlabel('Sample n')
xnd = (1/ND)*dft(x1, ND);
subplot(2,1,2)
```

```
%Chapter 3 - Program 10
%Figure 12
a0 = [0 0 0 0]
d = [.2.71.1.9.5]
n = 0: length(xom)-1;
N = 256;
figure(1)
stem(n, xom)
title('(d)')
figure (2)
X = fft(xom, N);
plot(abs(fftshift(X)))
w = 6*pi * (0:(N-1)) / N;
w2 = fftshift(w);
plot(w2)
w3 = unwrap(w2 - 2*pi);
plot(w3)
plot(w3, abs(fftshift(X)))
xlabel('radians')
plot(w3/pi, abs(fftshift(X)))
xlabel('radians / \pi')
```

```
% Chapter 3 - Program 12
N = 5;
K0 = 11;
n = -21:20;
coff = (1/K0) *diric(n*2*pi/K0, N);
stem(n, coff, '.')
```