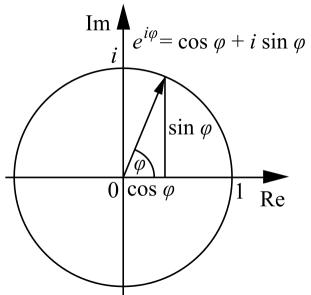
# Tutorial 5 Fourier Analysis Made easy – Part 2

2013/12/25 Merry Christmas

• Euler's formula  $e^{i\varphi} = \cos \varphi + i \sin \varphi$ 



• Complex Exponential  $e^{j\omega t} = \cos \omega t + j\sin \omega t$ , also called a Cisoid

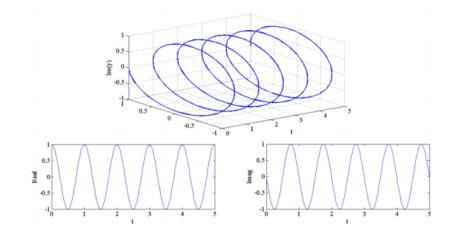


Figure 1 -  $e^{j\omega t}$  is a helix. Its projection on the real axis is a cosine and on the imaginary axis is sine.

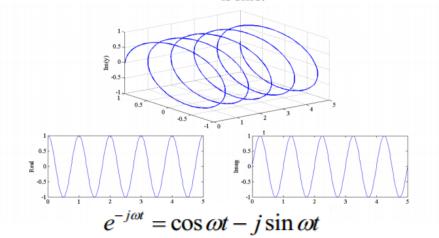
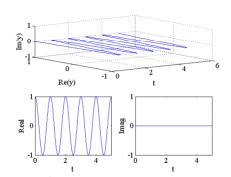


Figure  $2 - e^{-j\omega t}$  is a helix rotating in the opposite direction. The projections on the real axis is a cosine and on imaginary axis is a negative sine.

$$\sin \omega t = \frac{1}{2j} \left( (\cos \omega t + j \sin \omega t) - (\cos \omega t - j \sin \omega t) \right)$$
$$= \frac{1}{2j} \left( e^{j\omega t} - e^{-j\omega t} \right)$$

• 
$$\cos \omega t = \frac{1}{2} ((\cos \omega t + j \sin \omega t) + (\cos \omega t - j \sin \omega t))$$
  
=  $\frac{1}{2} (e^{j\omega t} + e^{-j\omega t})$ 



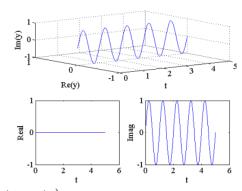


Figure 3 – Plotting  $\left(e^{j\omega t}+e^{-j\omega t}\right)/2$  gives a a cosine wave with no projection on the imaginary Figure 4 – Plotting  $\left(e^{j\omega t}-e^{-j\omega t}\right)/2$  gives us a sine wave with no projection on the real axis.

• Taylor Series Representation of exponential *e*<sup>x</sup>

$$\cos(x) = 1 - \frac{x^2}{2!} + \frac{x^4}{4!} - \frac{x^6}{6!} + \cdots$$

$$\sin(x) = x - \frac{x^3}{3!} + \frac{x^5}{5!} - \frac{x^7}{7!} + \cdots$$

$$e^x = 1 + x + \frac{x^2}{2!} + \frac{x^3}{3!} + \frac{x^4}{4!} + \cdots$$

• 
$$e^{j\theta} = 1 + j\theta + \frac{(j\theta)^2}{2!} + \frac{(j\theta)^3}{3!} + \frac{(j\theta)^4}{4!} + \frac{(j\theta)^5}{5!} + \cdots$$
  
 $= 1 + j\theta - \frac{\theta^2}{2!} - \frac{j\theta^3}{3!} + \frac{\theta^4}{4!} + \frac{j\theta^5}{3!} - \frac{\theta^2}{6!} - \frac{j\theta^7}{7!} + \cdots$   
 $= 1 - \frac{\theta^2}{2!} + \frac{\theta^4}{4!} - \frac{\theta^2}{6!} + \cdots$  This is a cosine.  
 $+ j\theta - \frac{j\theta^3}{3!} + \frac{j\theta^5}{3!} - \frac{j\theta^7}{7!} + \cdots$  This is  $j$  times sine.

• 
$$e^{j\pi/2} = \cos(\pi/2) + j\sin(\pi/2)$$
  $e^{j3\pi/2} = \cos(3\pi/2) + j\sin(3\pi/2)$   
 $= 0 + j(1)$   $= 0 + j(-1)$   
 $= j$   $= -j$   
 $e^{j\pi} = \cos(\pi) + j\sin(\pi) = -1$   
 $e^{j\pi} + 1 = 0$   
•  $x(t) = A\cos(\omega t + \theta)$   
 $= \frac{A}{2}e^{j(\omega t + \theta)} + \frac{A}{2}e^{-j(\omega t + \theta)}$   
 $= \frac{A}{2}e^{j\theta}e^{j\omega t} + \frac{A}{2}e^{-j\theta}e^{-j\omega t} = Q_{+}e^{j\omega t} + Q_{-}e^{-j\omega t}$   $Q_{+} = \frac{A}{2}e^{j\theta}$  :complex number  $= A/2\left(\cos(\omega t + \theta) + j\sin(\omega t + \theta) + \cos(\omega t + \theta) - j\sin(\omega t + \theta)\right)$   
 $= A\cos(\omega t + \theta)$ 

• Back to the Fourier Series...

$$f(t) = a_0 + \sum_{n=1}^{N} a_n \cos(\omega_n t) + \sum_{n=1}^{N} b_n \sin(\omega_n t)$$

$$a_0 = \frac{1}{T} \int_{0}^{T} f(t) dt \quad , \quad a_n = \frac{2}{T} \int_{0}^{T} f(t) \cos n\omega t \ dt \quad , \quad b_n = \frac{2}{T} \int_{0}^{T} f(t) \sin n\omega t \ dt$$

$$f(t) = a_0 + \sum_{n=1}^{N} \frac{a_n}{2} (e^{jn\omega t} + e^{-jn\omega t}) + \sum_{n=1}^{N} \frac{b_n}{2j} (e^{jn\omega t} - e^{-jn\omega t})$$

$$a_n = \frac{2}{T} \int_0^T f(t) \frac{1}{2} (e^{jn\omega t} + e^{jn\omega t}) dt , b_n = \frac{2}{T} \int_0^T f(t) \frac{1}{2j} (e^{jn\omega t} - e^{-jn\omega t}) dt$$

$$f(t) = a_0 + \sum_{n=1}^{\infty} \frac{1}{2} (a_n - jb_n) e^{jn\omega t} + \sum_{n=1}^{\infty} \frac{1}{2} (a_n + jb_n) e^{-jn\omega t}$$

$$a_0 = \frac{1}{2} \int_0^T f(t) e^{jn\omega t} dt + \frac{1}{2} \int_0^T f(t) e^{-jn\omega t} dt$$

$$a_{n} = \frac{1}{T} \int_{0}^{T} f(t)e^{jn\omega t}dt + \frac{1}{T} \int_{0}^{T} f(t)e^{-jn\omega t}dt$$

$$f(t) = g_{0}^{T} + \sum_{n=1}^{\infty} A_{n}e^{jn\omega t} + \sum_{n=1}^{\infty} B_{n}e^{-jn\omega t}$$

$$A_{n} = \frac{1}{T} \int_{0}^{T} f(t)e^{jn\omega t}dt \qquad B_{n} = \frac{1}{T} \int_{0}^{T} f(t)e^{-jn\omega t}dt$$

$$f(t) = \sum_{n=0}^{\infty} A_n e^{jn\omega t} + \sum_{n=0}^{\infty} B_n e^{-jn\omega t} \text{ (without a dc offset)}$$

$$f(t) = \sum_{n=-\infty}^{\infty} C_n e^{jn\omega t}$$

$$C_n = \frac{1}{2T} \int_{-T}^{T} f(t)e^{-jn\omega t} dt \text{, where } C_n = A_n + jB_n$$

$$C_n = \sqrt{A_n^2 + B_n^2}$$

$$\phi_n = \tan^{-1} \left( \frac{B_n}{A_n} \right)$$

## Magnitude And Power Spectrum

- Amplitude Spectrum the plot of coefficients we draw from the Fourier series coefficients.
- Power Spectrum converted from the amplitude spectrum by the Parseval's relationship

$$P = \frac{1}{T} \int_{0}^{T} |x(t)|^{2} dt \xrightarrow{\text{In time domain}} P = \sum_{k=-\infty}^{\infty} |x_{k}|^{2}$$

- For a discrete sinusoid of amplitude A, its average power is equal to  $P_{avg} = A^2/2$
- For a constant signal, then its average power is just A^2. In frequency domain, we can calculate it by

$$P_{avg} = a_0^2 + \sum_{n=1}^{\infty} \left( \frac{1}{2} a_n^2 + \frac{1}{2} b_n^2 \right) \xrightarrow{\text{In complex domain}} = \sum_{n=-\infty}^{\infty} C_n C_{-n}$$
$$\left| C_0 \right|^2 = \frac{1}{2} \left| a_0 \right|^2, \quad \left| C_n \right|^2 = \left| C_{-n} \right|^2 = \frac{1}{4} \left( a_n^2 + b_n^2 \right)$$

• Example 1 — Computer complex coefficients of a cosine wave  $f(t) = A\cos\omega t = \frac{A}{2}e^{j\omega t} + \frac{A}{2}e^{-j\omega t}$ 

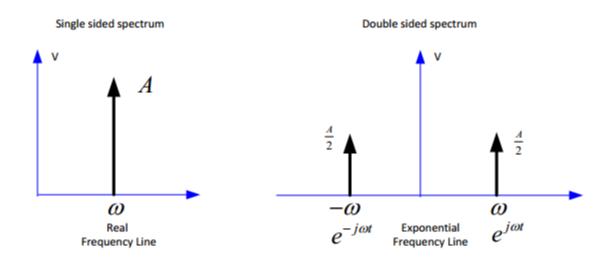


Figure 5 – Amplitude spectrum of  $A\cos\omega t$ 

• Example 2 – Computer complex coefficients of a sine wave  $f(t) = A\sin \omega t = \frac{A}{2i}e^{j\omega t} - \frac{A}{2i}e^{-j\omega t}$ 

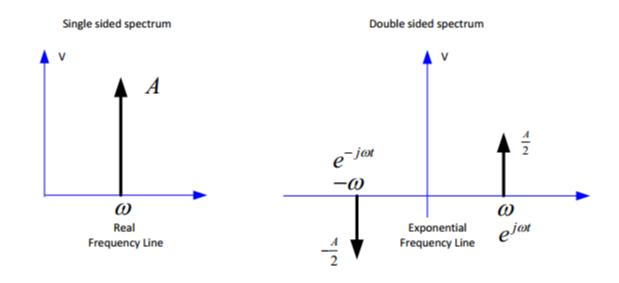
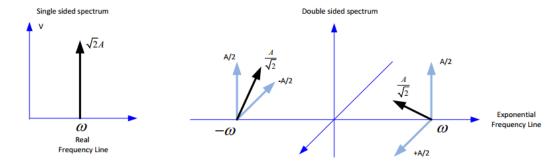


Figure 6 – Amplitude spectrum of  $A \sin \omega t$ 

• Example 3 – Computer coefficients of

$$f(t) = A(\cos \omega t + \sin \omega t) = \frac{A}{2}e^{j\omega t} + \frac{A}{2}e^{-j\omega t} + \frac{A}{2j}e^{j\omega t} - \frac{A}{2j}e^{-j\omega t}$$



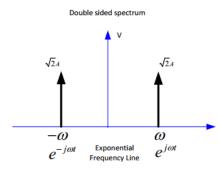


Figure 7 – Amplitude spectrum of  $A \sin \omega t + A \cos \omega t$ 

• Example 4 – Computer coefficients of

$$f(t) = A(\cos w + j\sin wt) = \frac{A}{2}e^{j\omega t} + \frac{A}{2}e^{-j\omega t} + \frac{A}{2}e^{j\omega t} - \frac{A}{2}e^{-j\omega t} = Ae^{j\omega t}$$

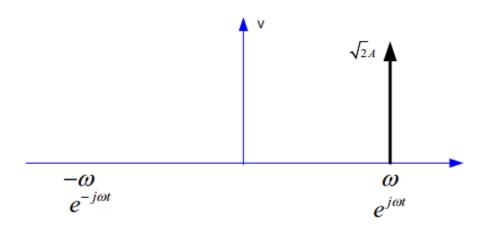


Figure 8 – Double-sided spectrum of  $A\cos\omega t + jA\sin\omega t$ 

Example 5 – Computer coefficients of

$$f(t) = A\cos(\omega = 0)t = \frac{A}{2}e^{j\omega t} + \frac{A}{2}e^{-j\omega t} = A$$

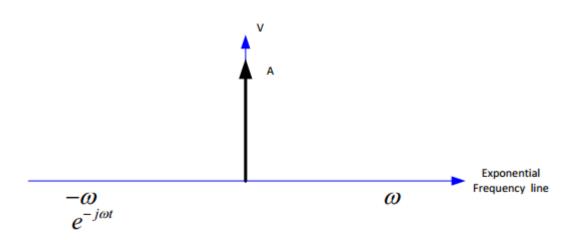


Figure 9 – Double-sided spectrum of A

• Example 6 – Computer coefficients of

$$x(t) = 2\cos^{2}(\omega t) = 2\left(\frac{e^{j\omega t} + e^{-j\omega t}}{2}\right)^{2} = 1 + \frac{1}{2}e^{j2\omega t} + \frac{1}{2}e^{-j2\omega t}$$
DC

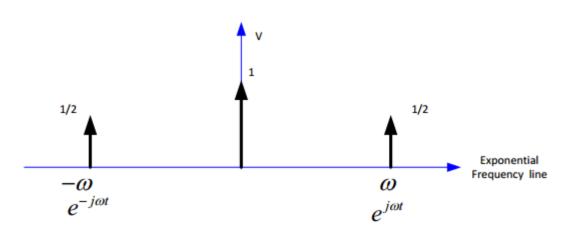


Figure 10 – Double-sided amplitude spectrum of  $2\cos^2(\omega t)$ 

• Example 7 – Computer coefficients of  $x(t) = 2\cos(\omega t)\cos(2\omega t) = \cos(\omega t) + \cos(3\omega t) = \frac{1}{2}e^{j\omega t} + \frac{1}{2}e^{-j\omega t} + \frac{1}{2}e^{-j3\omega t} + \frac{1}{2}e^{-j3\omega t}$ 

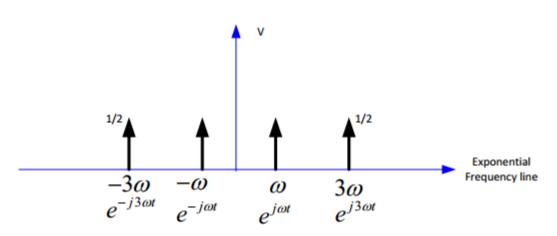


Figure 11 – Double-sided amplitude spectrum of  $2\cos(\omega t)\cos(2\omega t)$ 

 Example 8 – Computer complex coefficients and power spectrum of a real signal

$$f(t) = .8\cos 2\pi t - .6\sin 2\pi t + .8\cos 4\pi t + .3\sin 7\pi t$$

$$= \frac{0.8}{2}e^{j2\pi t} + \frac{0.8}{2}e^{-j2\pi t} - \frac{0.6}{2j}e^{j2\pi t} + \frac{0.6}{2j}e^{-j2\pi t} + \frac{0.8}{2}e^{j4\pi t} + \frac{0.8}{2}e^{-j4\pi t} + \frac{0.3}{2j}e^{j7\pi t} - \frac{0.3}{2j}e^{-j7\pi t}$$

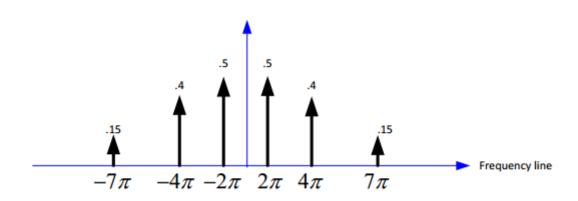


Figure 12 - Two-sided spectrum

• Example 9 – Computer complex coefficients of a real signal with variable phase

$$x(t) = 3 + 6\cos(4\pi t + 2) + j4\sin(4\pi t + 3) - j6\sin(10\pi t + 1.5)$$

$$= 3 + \left(3e^{4\pi t}e^{2} + 3e^{-4\pi t}e^{-2}\right) + \left(2e^{4\pi t}e^{3} - 2e^{-4\pi t}e^{-3}\right) + \left(3e^{10\pi t}e^{1.5} - 3e^{-10\pi t}e^{-1.5}\right)$$

$$= 3 + e^{j4\pi t}\left(3e^{2j} + 2e^{3j}\right) + e^{-j4\pi t}\left(3e^{-2j} - 2e^{-3j}\right) + 3e^{j10\pi t}\left(e^{j1.5}\right) + 3e^{-j10\pi t}\left(e^{-j1.5}\right)$$

$$= \sqrt{\left(3\cos(2) + 2\cos(3)\right)^{2} + \left(3\sin(2) + 2\sin(3)\right)^{2}}$$

$$= 6.16e^{j4\pi t}$$

$$e^{-j4\pi t}\left(3e^{-2j} - 2e^{-3j}\right)$$

$$= \sqrt{\left(3\cos(2) - 2\cos(3)\right)^{2} + \left(3\sin(2) - 2\sin(3)\right)^{2}}$$

$$= 3.46e^{-j4\pi t}$$
Figure 13 - Two-sided spectrum of a complex signal

Figure 13 - Two-sided spectrum of a complex signal

- Converting forms
  - Rule
    - 1. A rectangular form z = x + jy then its polar form is equal to

$$M \angle \theta = \sqrt{x^2 + y^2} \angle \tan^{-1}(y/x)$$
$$= \sqrt{x^2 + y^2} \angle (\tan^{-1}(y/x) + \pi) \quad \text{if } x < 0$$

- 2. Given a polar form  $z = M \angle \theta$  then its rectangular form is given by  $x + jy = M \cos \theta + jM \sin \theta$
- Example 1: Convert  $z = \angle$  927 to rectangular form
  - Real part =  $5\cos(.927) = 3$
  - Imaginary part =  $5\sin(.927)$ =4
  - Z = 3+j4

• Example 2: Convert z = -1-j to rectangular form

$$M = \sqrt{(-1)^2 + (-j)^2} = \sqrt{2}$$

$$\theta = \arctan\left(\frac{y}{x}\right) = \arctan\left(\frac{y}{x}\right) + \pi \quad \text{If } x < 0 = \arctan\left(\frac{-1}{-1}\right) + \pi = 3\pi/4$$

$$z = \sqrt{2} \angle 3\pi/4$$

• Example 3: Convert z = 1+j to polar form

$$Z = \sqrt{1^2 + j^2} = \sqrt{2}$$
 and  $\theta = \arctan(1) = \pi/4 = .785$ 

- Adding and Multiplying
  - add in rectangular form, multiply in polar.
  - Rule
    - 1. Given Z1 = a + jb and Z2 = c + jd then Z1+Z2 = (a+c)+j(b+d)
    - 2. Given  $Z_1 = M_1 \angle \theta_1$  and  $Z_2 = M_2 \angle \theta_2$  then  $Z_3 = Z_1 * Z_2 = M_1 M_2 \angle (\theta_1 + \theta_2)$
  - Example 1: Add  $Z1=\sqrt{2}\angle.785$  and  $Z2=5\angle.927$ 
    - Convert both to rectangular form
    - Z1=1+j and Z2=3+4j
    - Z3=(1+3)+j(1+4)=4+j5
  - Example 2: Multiply Z1=1+j and Z2=3+4j
    - Z1 = $\sqrt{2} \angle .785*Z2 = 5 \angle .927 = 5 \sqrt{2} \angle 1.71$
  - Example 3: Divide Z1=1+j and Z2=3+4j
    - $z_2 = 5 \angle .927 \div z_1 = \sqrt{2} \angle .785$
    - $z_3 = \frac{z_2}{z_1} = \frac{5}{\sqrt{2}} \angle (.927 .785)$

- Conjugation
  - The conjugate for a complex number z, is given by  $z^* = x jy$
  - $e^{j\omega t}$  is the complex conjugate of  $e^{-j\omega t}$

• 
$$z = M \angle \theta$$
,  $z^* = M \angle -\theta$ 

- $|z|^2 = zz^*$  (used to compute the power of the signal)
- $|z| = \frac{1}{2}(z + z^*)$  (used to computer the magnitude of the signal)

## Matlab Programs

Matlab Program no. 1 for plotting the complex exponential

```
%comexp produces the complex exponential diagram in Chapter 1.
%Try changing target function to see effect on signal.
t = 0:0.01:5;
y=5*exp(-(j*(2*pi))*t); % change this equation for different cases
subplot(2,2,1);
plot3(t,real(y),imag(y));
grid
xlabel('t'),ylabel('Re(y)'),zlabel('Im(y)');
title('3-D plot of a Complex Exponential');
subplot(2,2,3),plot(t,real(y)),xlabel('t'),
ylabel('Magnitude'),title('Re(y(t))');
subplot(2,2,4),plot(t,imag(y)),xlabel('t'),
ylabel('Angle'),title('Im(y(t))');
```

## Matlab Programs

```
Matlab Program no. 2 – Compute double sided DFT of a signal
clc; close all; clear all;
%Generate the signal
fs=128; % Sampling rate should be at least 16 times higher frequency to get a good picture.
N=1024; % FFT size.
t=0:1/fs:((N/fs)-(1/fs)); %Time it takes to creat N points.
x=3+6*\cos(8*pi*t+2)+j*8*\sin(8*pi*t+3)-j*6*\sin(30*pi*t+1); %The target signal.
figure(1);
plot(t, x); %Plot signal, all points.
%fplot('x', [0,1]) % want to plot the first 1 second only.
xlabel('Time (Hz)');ylabel('Signal in time domain ');
figure(2);
xf1=abs(fft(x))/N;% Compute the Double sided amplitude spectrum
xf=fftshift(xf1);
P=xf.*xf; %compute the power spectrum
% Map the frequency bin to the frequency (hz)
f=LINSPACE(0, fs, N);
f3=[-fs/2:fs/2:1024];%fk=k fs/N where k=0,1,2,...N-1
f2 = linspace(-fs/2, fs/2, 1024);
% now we will plot the DFT spectrums
plot(f2,xf);grid
xlabel('Frequency (Hz)');ylabel('Double-sided Amplitude Spectrum (DFT) ');
```