

Digital modulation

Lecture 4

Digital modulation

- **Digital modulation** is the **process** of **changing** one of the **characteristics** of an analog signal based on the information in digital data.
 - ✓ This is also known as “**Digital-to-Analog Conversion**”
- Topics discussed in this section:
 - **Aspects of Digital-to-Analog Conversion**
 - **Amplitude Shift Keying**
 - **Frequency Shift Keying**
 - **Phase Shift Keying**
 - **Quadrature Amplitude Modulation**

Digital modulation

- **Digital modulation** is the **process** of **changing** one of the **characteristics** of an analog signal based on the information in digital data.

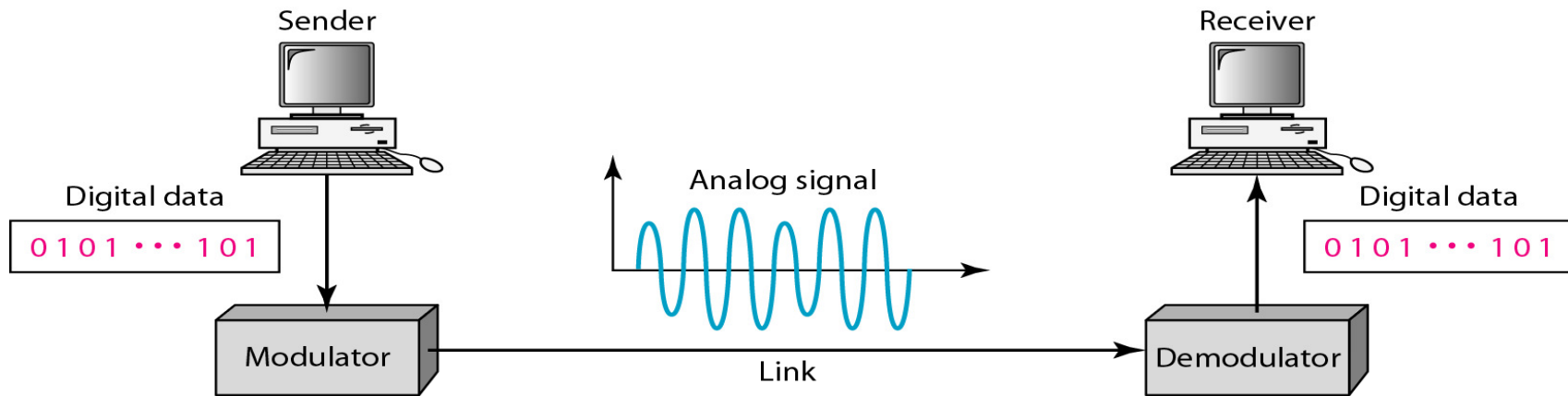
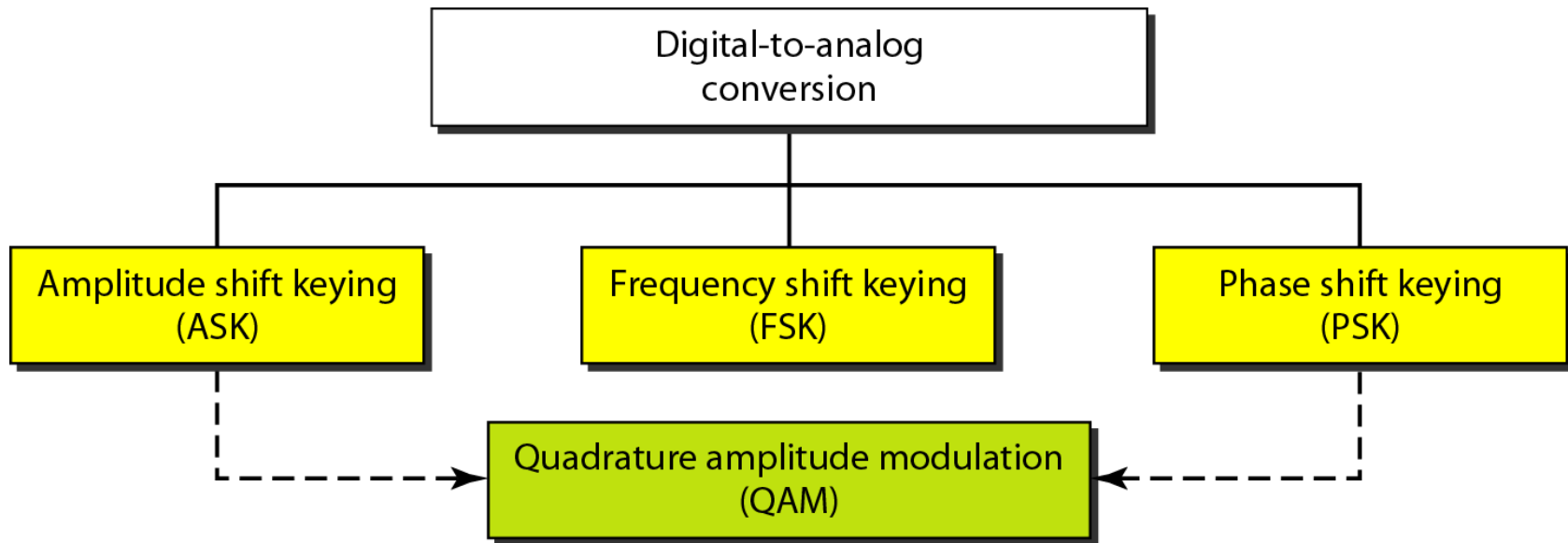


Fig. 1: Digital modulation

- **Type of various Digital modulation techniques are:**
 - Amplitude Shift Keying
 - Frequency Shift Keying
 - Phase Shift Keying
 - Quadrature Amplitude Modulation

Digital-to-Analog Conversion

- Types of Digital-to-Analog conversion:



Digital-to-Analog Conversion

- Some Aspects of Digital-to-Analog Conversion:
 - Bit Rate: Bit rate is the number of bits per second.
 - Baud Rate: Baud rate is the number of signal elements per second.
- In the analog transmission of digital data, the baud rate is less than or equal to the bit rate.

Digital-to-Analog Conversion

- **Example 1:**

An analog signal carries **4 bits** per **signal element**. If **1000 signal elements** are sent per second, find the **bit rate**.

- **Solution:**

In this case, $r = 4$, $S = 1000$, and N is unknown. We can find the value of N from

$$S = N \times \frac{1}{r} \quad \text{or} \quad N = S \times r = 1000 \times 4 = 4000 \text{ bps}$$

- **Example 2:**

An analog signal has a **bit rate** of **8000 bps** and a **baud rate** of **1000 baud**. How many **data elements** are carried by each signal element? How many **signal elements** do we need?

- **Solution:**

In this example, $S = 1000$, $N = 8000$, and r and L are unknown. We find first the value of r and then the value of L .

$$\begin{aligned} S = N \times \frac{1}{r} & \quad \rightarrow \quad r = \frac{N}{S} = \frac{8000}{1000} = 8 \text{ bits/ baud} \\ r = \log_2 L & \quad \rightarrow \quad L = 2^r = 2^8 = 256 \end{aligned}$$

Amplitude Shift Keying (ASK)

- **Amplitude Shift Keying (ASK):**

In Amplitude Shift Keying, the **amplitude** of the carrier signal is **varied** to create signal elements. Both **frequency** and **phase** remain **constant** while the **amplitude** changes.

- **Binary amplitude shift keying:**

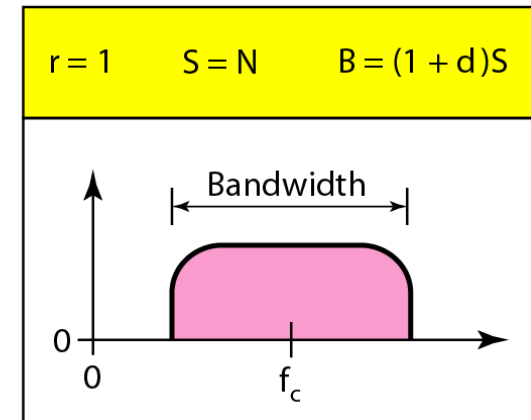
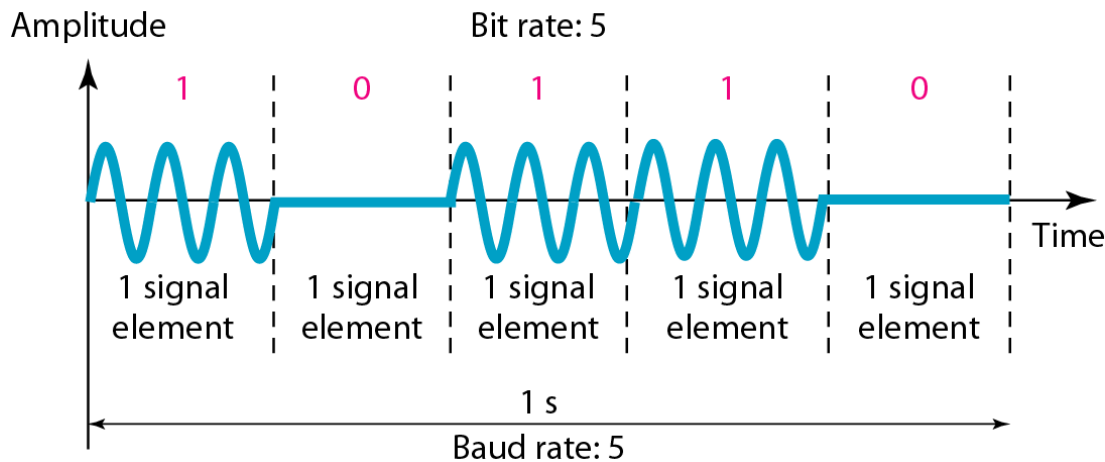


Fig. 3: Binary amplitude shift keying

Amplitude Shift Keying (ASK)

- ASK is implemented by **changing** the amplitude of a carrier signal to reflect amplitude levels in the digital signal.
- For example: a digital “**1**” could not affect the signal, whereas a digital “**0**” would, by making it zero.
- The **line encoding** will determine the values of the analog waveform to reflect the digital data being carried.
- The **bandwidth B** of ASK is proportional to the signal rate **S** .

$$B = (1+d)*S$$

- “ **d** ” is due to **modulation and filtering**, lies between **0 and 1** .

Implementation of Binary ASK

- Implementation of binary ASK:

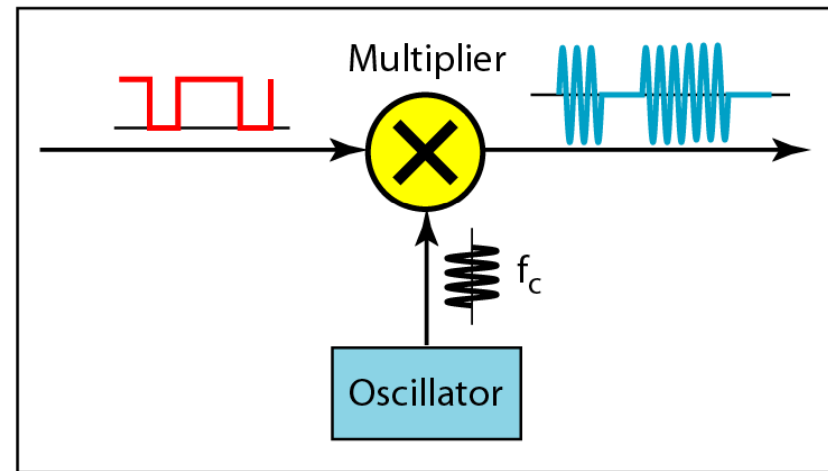
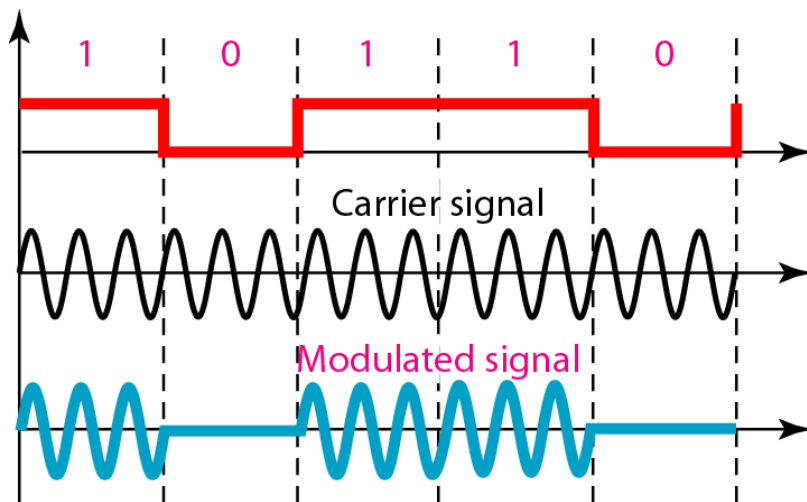


Fig. 4: Implementation of Binary ASK

ASK

- **demodulation:** only the presence or absence of a sinusoid in a given time interval needs to be determined
- **advantage:** simplicity
- **disadvantage:** ASK is very susceptible to noise interference – noise usually (only) affects the amplitude, therefore ASK is the modulation technique most affected by noise
- **application:** ASK is used to transmit digital data over optical fiber

ASK

- Example 3:

We have an available bandwidth of **100 kHz** which spans from **200 to 300 kHz**. What are the **carrier frequency** and the **bit rate** if we modulated our data by using **ASK** with **$d = 1$** ?

- Solution:

The middle of the bandwidth is located at **250 kHz**. This means that our carrier frequency can be at **$f_c = 250 \text{ kHz}$** . We can use the formula for bandwidth to find the bit rate (**with $d = 1$ and $r = 1$**).

$$B = (1 + d) \times S = 2 \times N \times \frac{1}{r} = 2 \times N = 100 \text{ kHz} \quad \rightarrow \quad N = 50 \text{ kbps}$$

ASK

- Example 4:

In data communications, we normally use **full-duplex** links with communication in both directions. We need to **divide** the bandwidth into two with **two carrier frequencies**, as shown in **Figure 5**. The figure shows the **positions** of **two carrier frequencies** and the **bandwidths**. The available bandwidth for each direction is now **50 kHz**, which **leaves** us with a data rate of **25 kbps** in each direction.

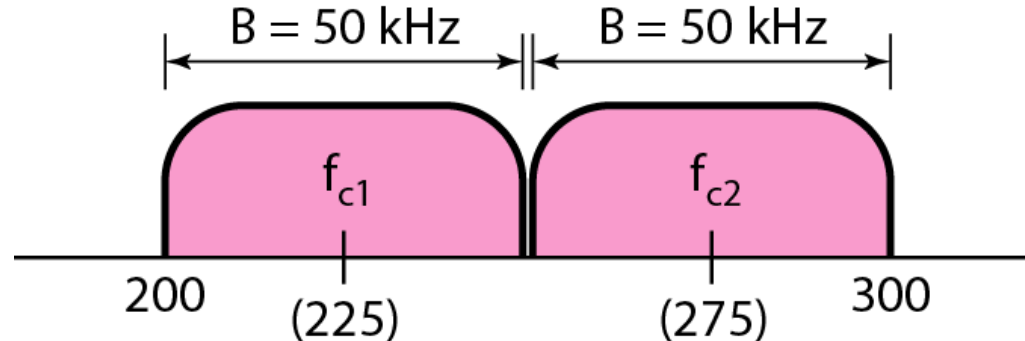


Fig. 5: Bandwidth of full-duplex ASK used in Example 4.

ASK

- **Multilevel ASK:**

The above discussion uses only **two amplitude levels**. We can have **multilevel ASK** in which there are more than two levels.

We can use **4, 8, 16, or more different amplitudes** for the signal and modulate the data using **2, 3, 4, or more bits at a time**.

In these cases, **$r=2$, $r=3$, $r=4$, and so on**. It is implemented with **QAM**.

Frequency Shift Keying (FSK)

- **Frequency Shift Keying (FSK):**

In Frequency Shift Keying (FSK), the **frequency** of the carrier signal is **varied** to **represent** digital data.

The **frequency** of the modulated signal is **constant** for the duration of **one signal element**, but **changes** for the next element if the **data element changes**.

Both **peak amplitude** and **phase** remain **constant** for all signal elements.

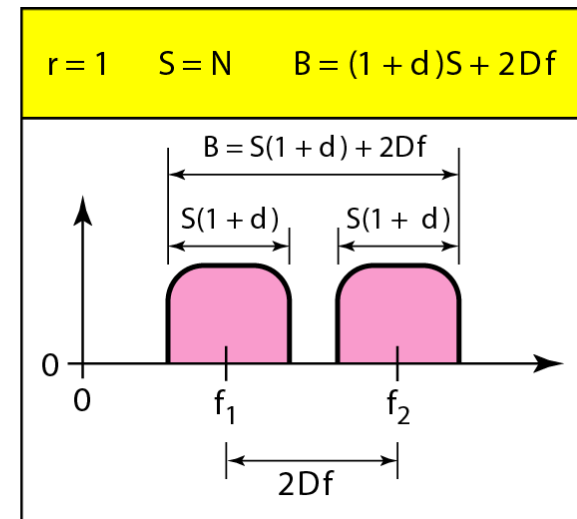
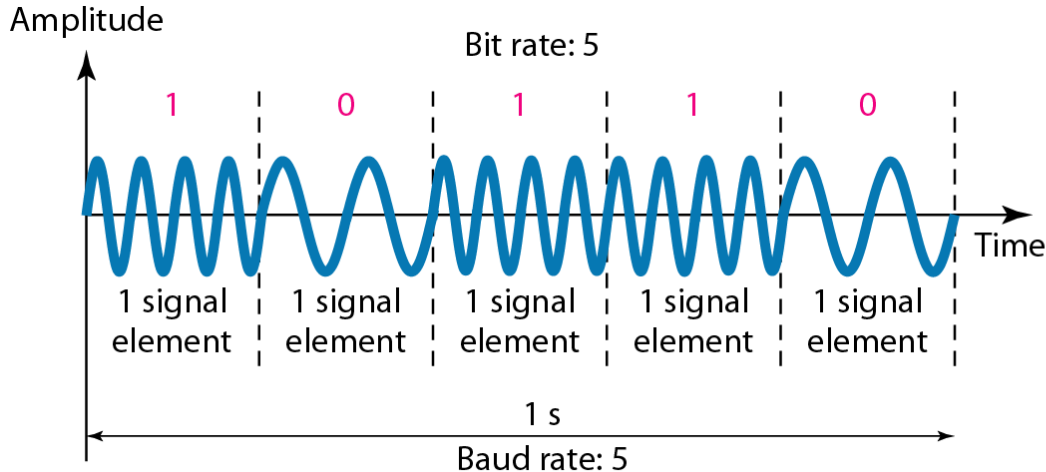


Fig. 6: Binary frequency shift keying

Frequency Shift Keying (FSK)

- The digital data stream changes the frequency of the carrier signal, f_c .
- For example, a “1” could be represented by $f_1 = f_c + \Delta f$, and a “0” could be represented by $f_2 = f_c - \Delta f$.
- If the difference between the two frequencies (f_1 and f_2) is $2\Delta f$, then the required **BW B** will be:

$$B = (1+d) \times S + 2\Delta f$$

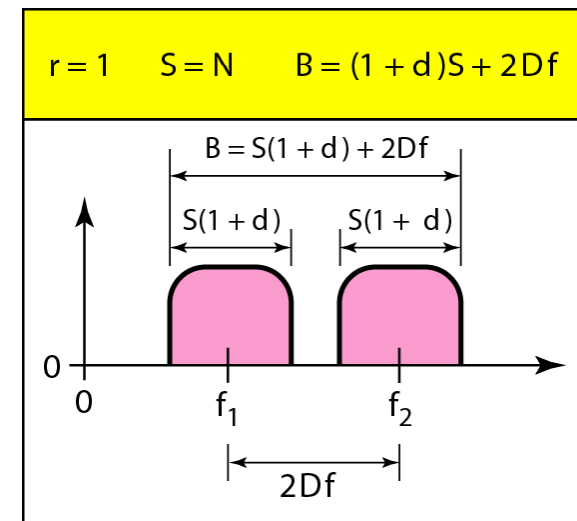
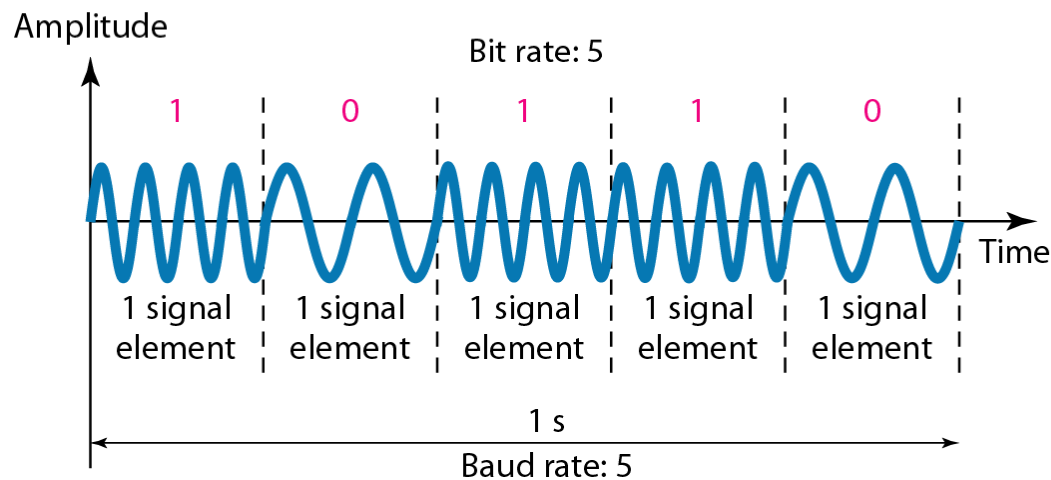


Fig. 6: Binary frequency shift keying

Implementation of binary FSK

- Implementation of binary FSK:

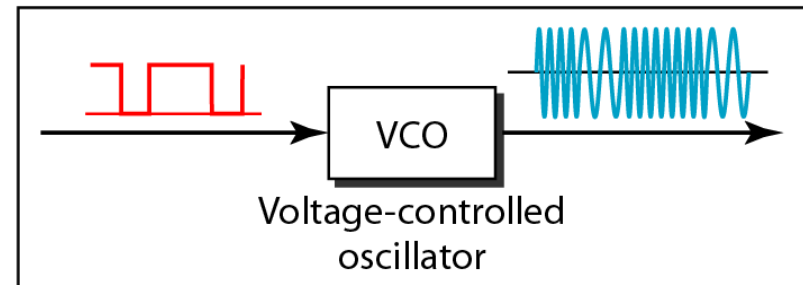
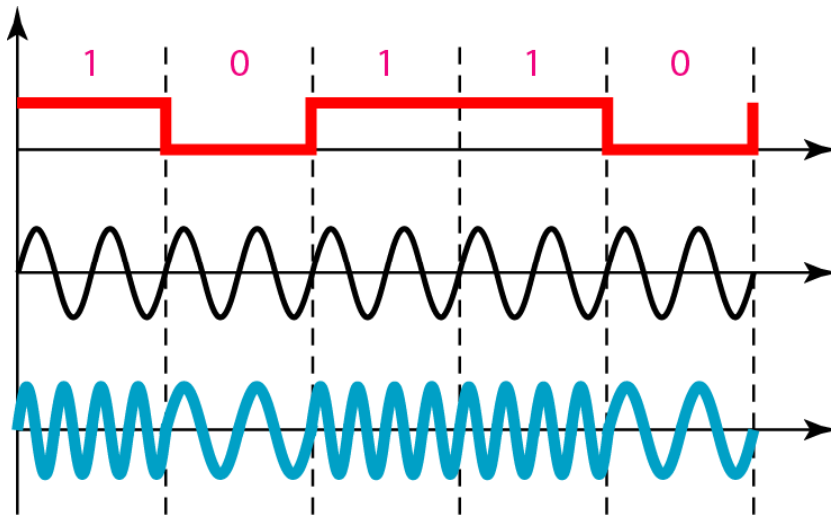


Fig. 7: Implementation of Binary FSK

Frequency Shift Keying (FSK)

- **demodulation:** demodulator must be able to determine which of two possible frequencies is present at a given time
- **advantage:** FSK is less susceptible to errors than ASK – receiver looks for specific frequency changes over a number of intervals, so voltage (noise) spikes can be ignored
- **disadvantage:** FSK spectrum is 2 x ASK spectrum
- **application:** over voice lines, in high-freq. radio transmission, etc.

Frequency Shift Keying (FSK)

- Example 5:

We have an available bandwidth of **100 kHz** which spans from **200 to 300 kHz**. What are the **carrier frequency** and the **bit rate** if we modulated our data by using **FSK** with **$d = 1$** ?

- Solution:

This problem is similar to **Example 3**, but we are modulating by using **FSK**. The **midpoint** of the band is at **250 kHz**. We choose **$2\Delta f$** to be **50 kHz**; this means

$$B = (1 + d) \times S + 2\Delta f = 100 \quad \rightarrow \quad 2S = 50 \text{ kHz} \quad S = 25 \text{ kbaud} \quad N = 25 \text{ kbps}$$

FSK

- **Multilevel FSK:**

In Multilevel FSK modulation (MFSK), we can use **more than two frequencies**.

For example, we can use **four** different frequencies ***f1, f2, f3 and f4*** to send **2 bits** at a time.

To send **3 bits** at a time, we can use ***eight frequencies*** and so on.

For the proper operation of the modulator and demodulator, it can be shown that the minimum value of **$2\Delta f$** need to be equal to **S** . The bandwidth with **$d=0$** is,

$$B = (1+d) \times S + (L-1)2\Delta f = L \times S$$

FSK

- **Example 6:**

We need to send data **3 bits** at a time at a bit rate of **3 Mbps**. The carrier frequency is **10 MHz**. Calculate the **number of levels (different frequencies)**, the **baud rate**, and the **bandwidth**.

- **Solution:**

We can have $L = 2^3 = 8$. The baud rate is $S = 3 \text{ Mbps}/3 = 1 \text{ Mbaud}$. This means that the carrier frequencies must be **1 MHz apart** ($2\Delta f = 1 \text{ MHz}$). The bandwidth is $B = 8 \times 1 \text{ MHz} = 8 \text{ MHz}$. Figure 8 shows the allocation of frequencies and bandwidth.

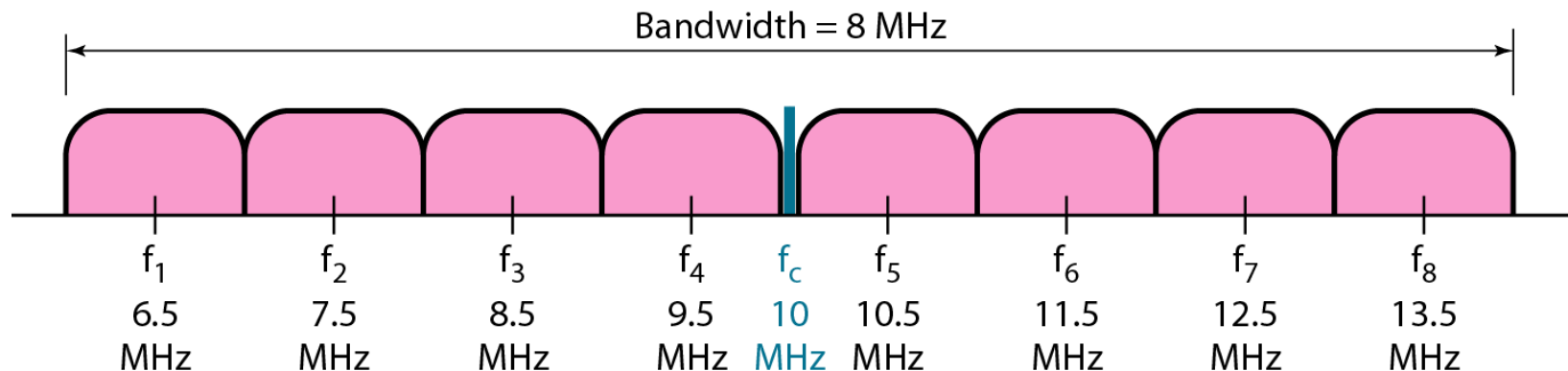


Fig. 8: Bandwidth of MFSK used in example 6.

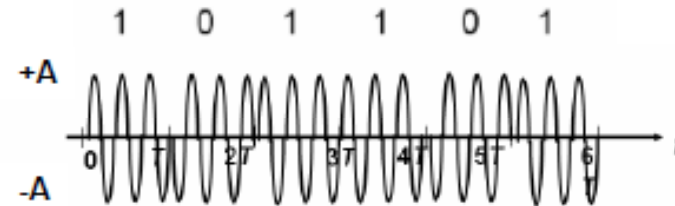
Phase Shift Keying (PSK)

PSK – phase of carrier signal is varied to represent binary 1 or 0

- peak amplitude & freq. remain constant during each bit interval
- example: binary 1 = 0° phase, binary 0 = 180° (π rad) phase
 \Rightarrow **PSK is equivalent to multiplying carrier signal by +1 when the information is 1, and by -1 when the information is 0**

2-PSK, or
Binary PSK,
since only 2
different phases
are used.

$$s(t) = \begin{cases} A \cos(2\pi f_c t), & \text{binary 1} \\ A \cos(2\pi f_c t + \pi), & \text{binary 0} \end{cases}$$
$$s(t) = \begin{cases} A \cos(2\pi f_c t), & \text{binary 1} \\ -A \cos(2\pi f_c t), & \text{binary 0} \end{cases}$$



- **demodulation**: demodulator must determine the phase of received sinusoid with respect to some reference phase
- **advantage**:
 - PSK is less susceptible to errors than ASK, while it requires/occupies the same bandwidth as ASK
 - more efficient use of bandwidth (higher data-rate) are possible, compared to FSK !!!
- **disadvantage**: more complex signal detection / recovery process, than in ASK and FSK

Phase Shift Keying (PSK)

- Implementation of BPSK:

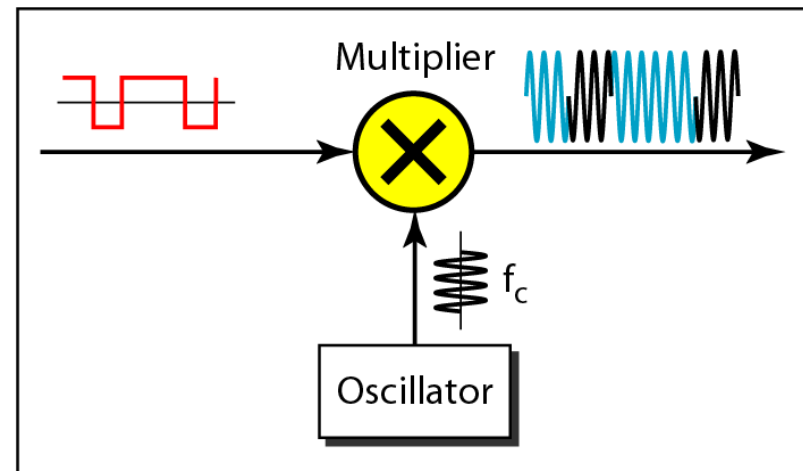
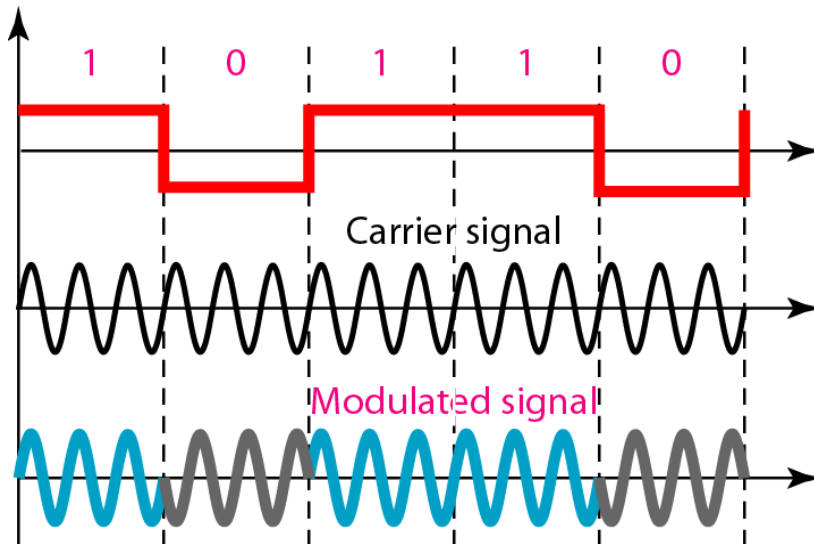
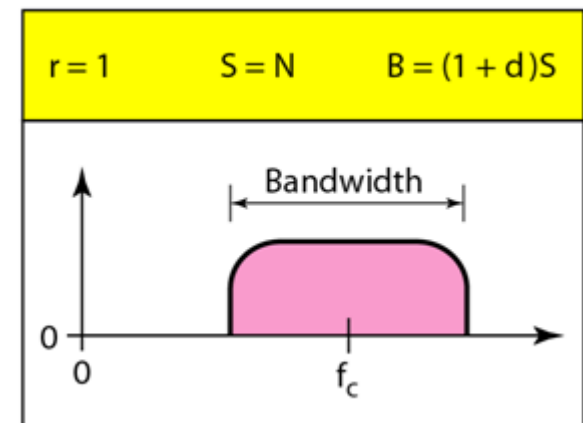


Fig. 10: Implementation of BPSK

The **bandwidth** requirement:

$$B = (1+d) \times S$$



Phase Shift Keying (PSK)

PSK Detection

$$\cos^2 A = \frac{1}{2}(1 + \cos 2A)$$

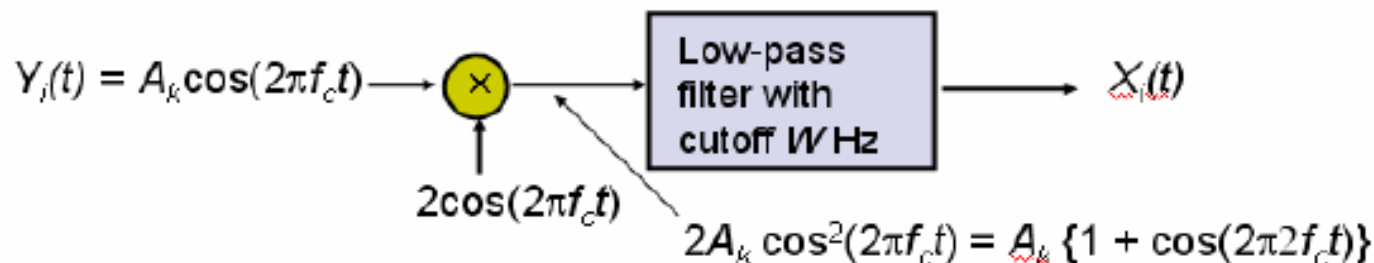
- multiply the received / modulated signal $\pm A \cos(2\pi f_c t)$ by $2 \cos(2\pi f_c t)$

- resulting signal

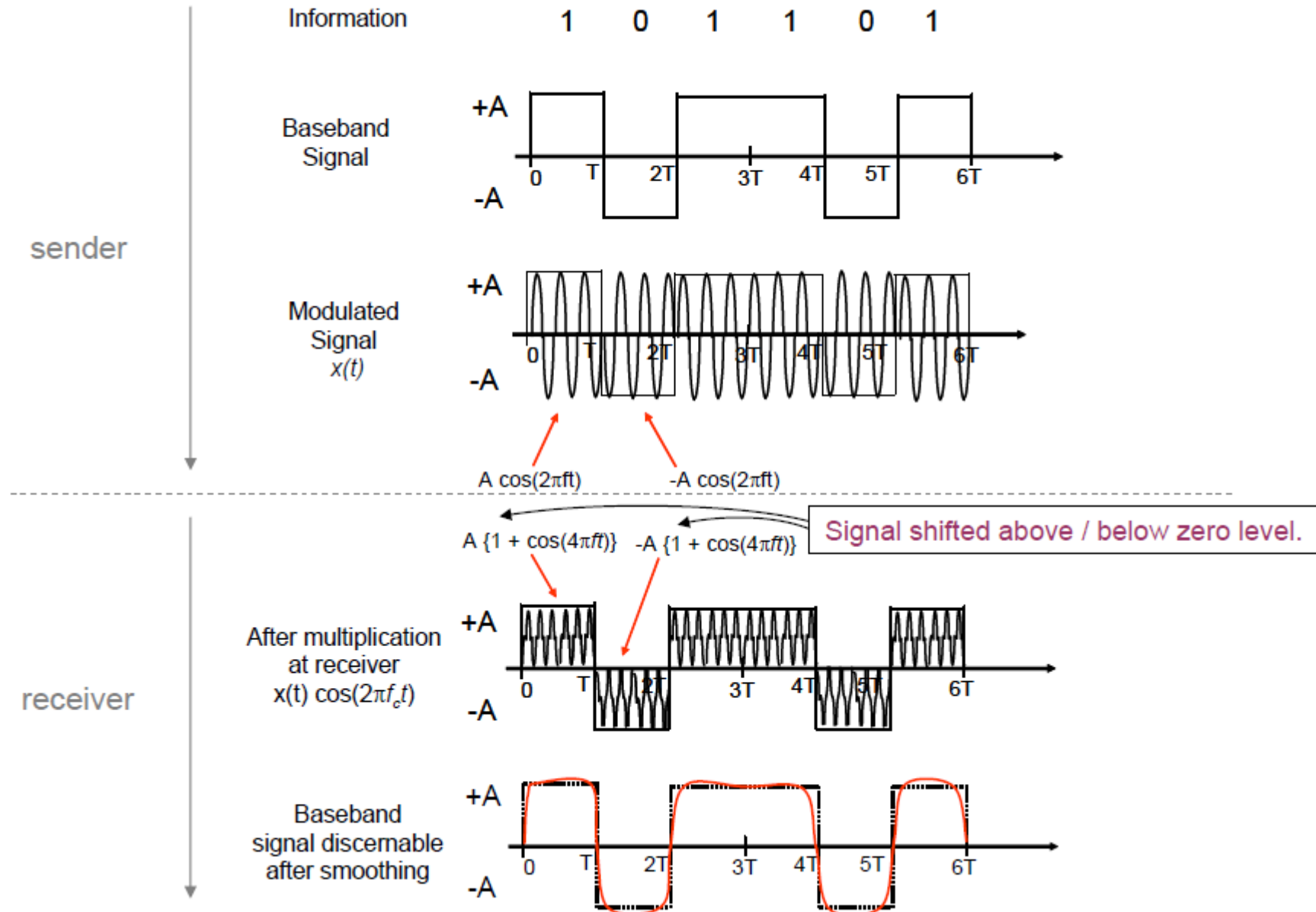
$$2A \cos^2(2\pi f_c t) = A[1 + \cos(4\pi f_c t)], \quad \text{binary 1}$$

$$-2A \cos^2(2\pi f_c t) = -A[1 + \cos(4\pi f_c t)], \quad \text{binary 0}$$

- by removing the oscillatory part with a low-pass filter, the original baseband signal (i.e. the original binary sequence) can be easily determined



Phase Shift Keying (PSK)



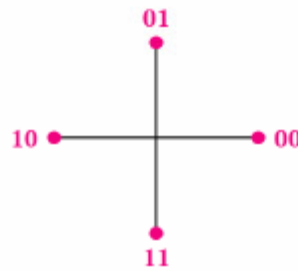
Quadrature PSK(QPSK)

QPSK = 4-PSK – PSK that uses phase shifts of $90^\circ = \pi/2$ rad \Rightarrow 4 different signals generated, each representing 2 bits

$$s(t) = \begin{cases} A\cos(2\pi f_c t), & \text{binary 00} \\ A\cos(2\pi f_c t + \frac{\pi}{2}), & \text{binary 01} \\ A\cos(2\pi f_c t + \pi), & \text{binary 10} \\ A\cos(2\pi f_c t + \frac{3\pi}{2}), & \text{binary 11} \end{cases}$$

Dibit	Phase
00	0
01	90
10	180
11	270

Dibit
(2 bits)



Constellation diagram

- **advantage:** higher data rate than in PSK (2 bits per bit interval), while bandwidth occupancy remains the same
- 4-PSK can easily be extended to 8-PSK, i.e. n-PSK
- however, higher rate PSK schemes are limited by the ability of equipment to distinguish small differences in phase

QPSK and its implementation

- QPSK and its implementation:

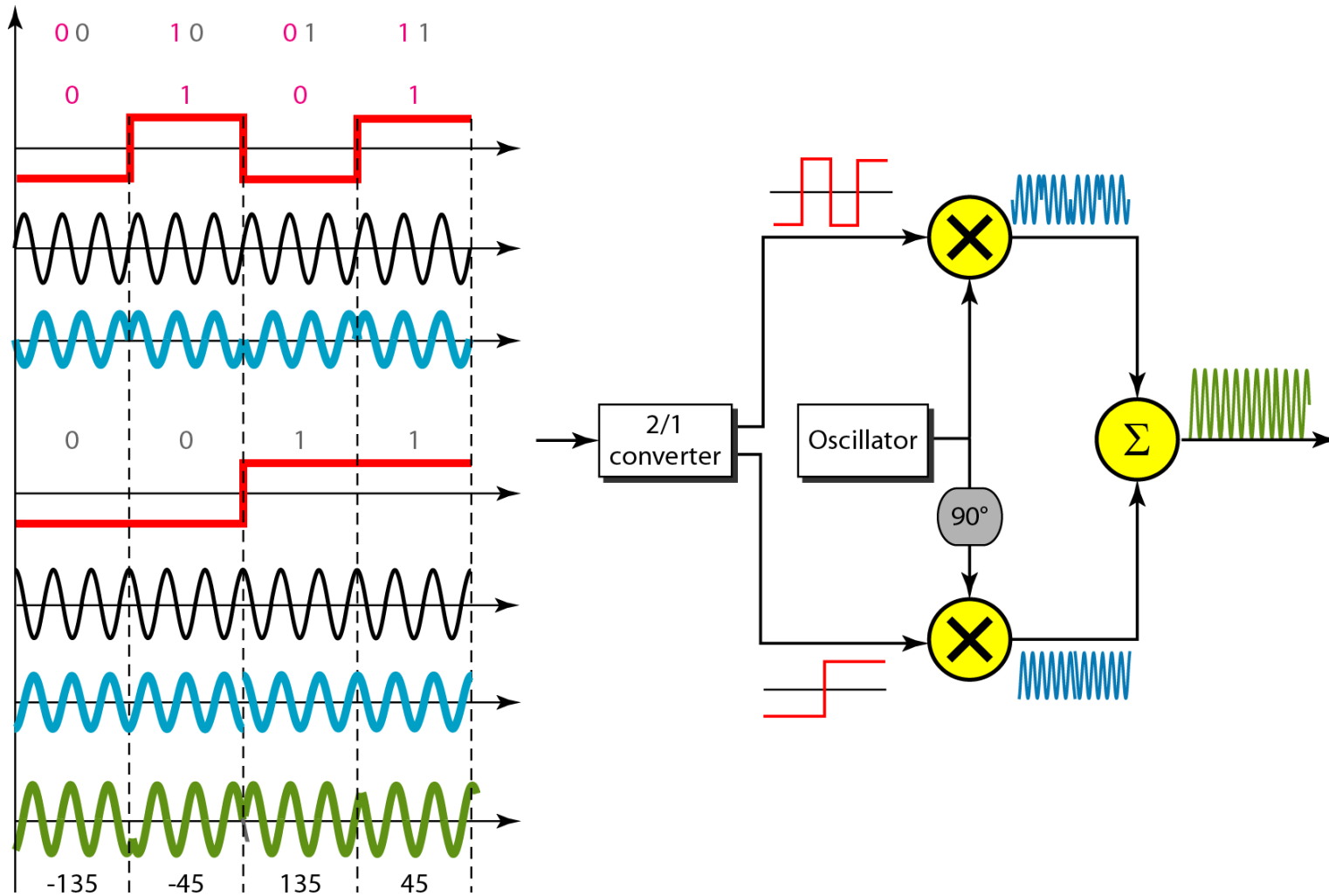


Fig. 11: Implementation of QPSK

Quadrature PSK(QPSK)

- Example 7:

Find the bandwidth for a signal transmitting at **12 Mbps** for **QPSK**. The value of **$d = 0$** .

- Solution:

For QPSK, **2 bits** is carried by one signal element. This means that **$r = 2$** . So the signal rate (baud rate) is **$S = N \times (1/r) = 6 \text{ Mbaud}$** . With a value of **$d = 0$** , we have **$B = S = 6 \text{ MHz}$** .

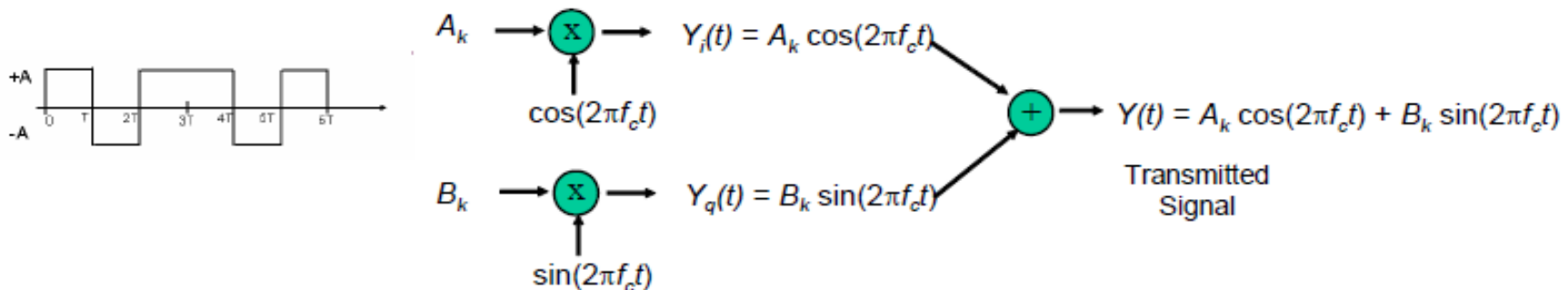
Quadrature Amplitude Modulation (QAM)

Quadrature Amplitude Modulation (QAM) – uses “two-dimensional” signalling

- original information stream is split into two sequences that consist of odd and even symbols, e.g. B_k and A_k

1	0	1	1	0	1	...
1	-1	1	1	-1	1	...
B_1	A_1	B_2	A_2	B_3	A_3	...

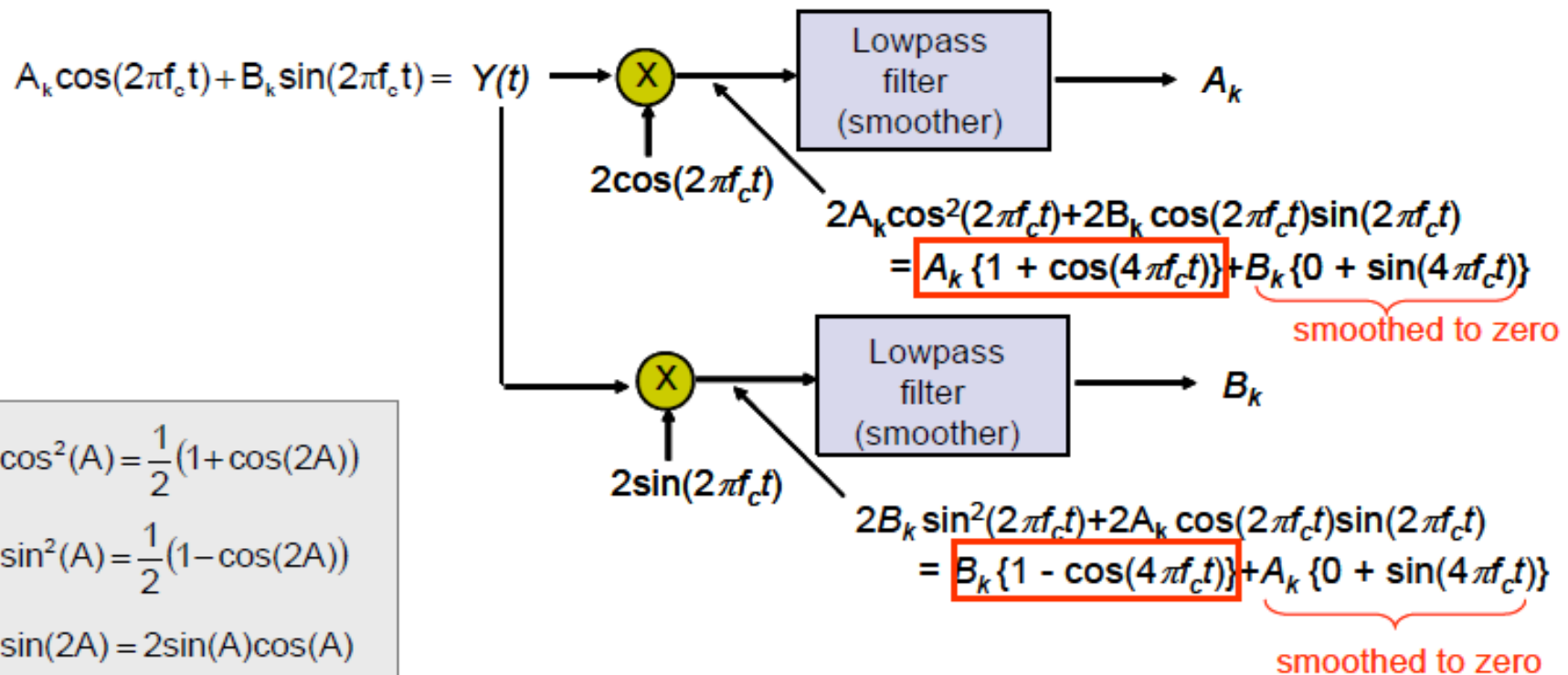
- A_k sequence (**in-phase comp.**) is modulated by $\cos(2\pi f_c t)$
 B_k sequence (**quadrature-phase comp.**) is modulated by $\sin(2\pi f_c t)$
- composite signal $A_k \cos(2\pi f_c t) + B_k \sin(2\pi f_c t)$ is sent through the channel



- advantage:** data rate = 2 bits per bit-interval!

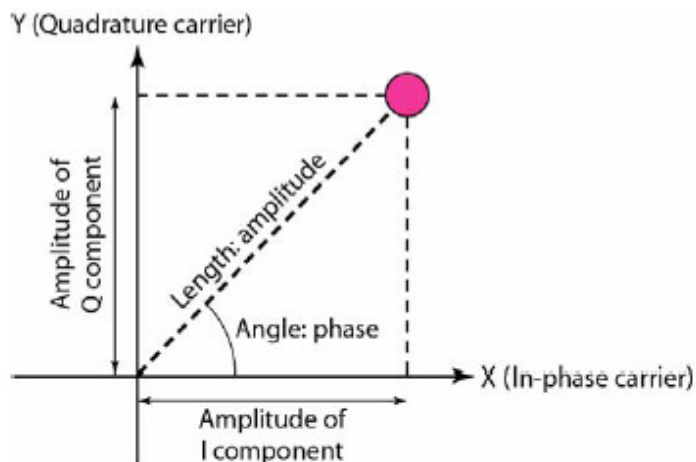
Quadrature Amplitude Modulation (QAM)

- QAM Demodulation**
- by multiplying $Y(t)$ by $2 \cdot \cos(2\pi f_c t)$ and then low-pass filtering the resultant signal, sequence A_k is obtained
 - by multiplying $Y(t)$ by $2 \cdot \sin(2\pi f_c t)$ and then low-pass filtering the resultant signal, sequence B_k is obtained



Signal Constellation

Constellation Diagram – used to represent possible symbols that may be selected by a given modulation scheme as points in 2-D plane



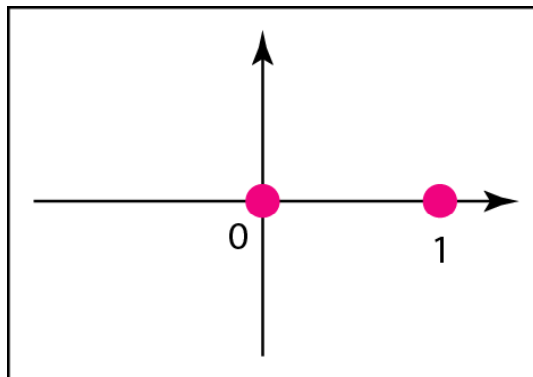
- X-axis is related to in-phase carrier: $\cos(\omega_c t)$
 - the projection of the point on the X-axis defines the peak amplitude of the in-phase component
- Y-axis is related to quadrature carrier: $\sin(\omega_c t)$
 - the projection of the point on the Y-axis defines the peak amplitude of the quadrature component
- the length of line that connects the point to the origin is the peak amplitude of the signal element (combination of X & Y components)
- the angle the line makes with the X-axis is the phase of the signal element

Constellation Diagrams

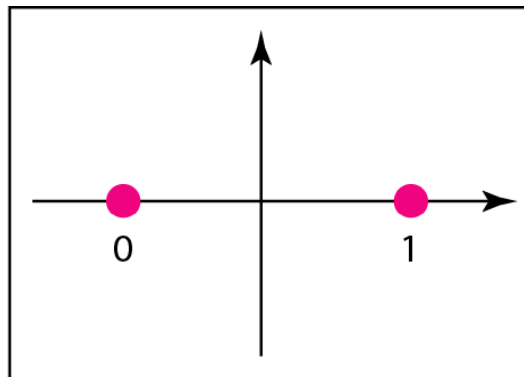
- **Example 8:**

Show the **constellation** diagrams for an **ASK (OOK)**, **BPSK**, and **QPSK** signals.

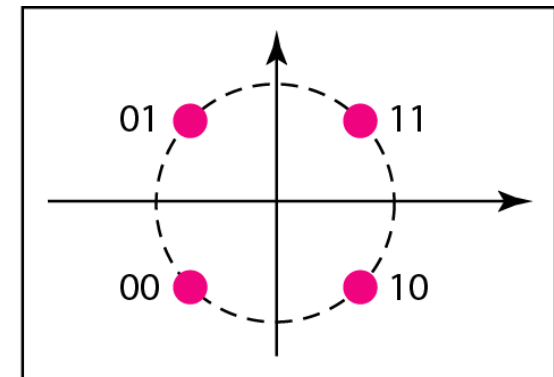
- **Solution:**



a. ASK (OOK)



b. BPSK

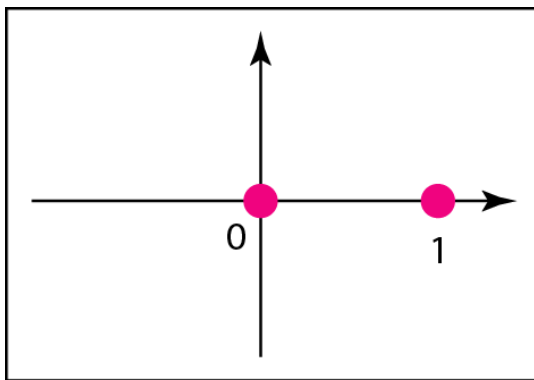


c. QPSK

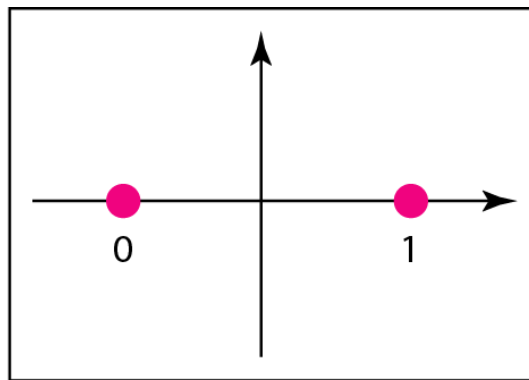
- For ASK, we are using only an in-phase carrier. Therefore, the **two points** should be on the **X axis**. **Binary 0** has an amplitude of **0V**; **binary 1** has an amplitude of **1 V**(for example). The points are located at the origin and at **1 unit**.

Constellation Diagrams

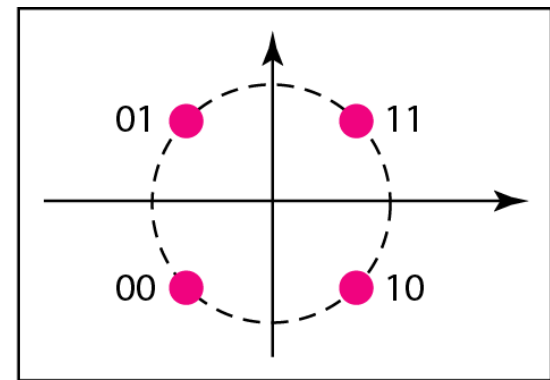
- **BPSK** creates two different signal elements, **one** with amplitude **1 V and in phase** and the **other** with amplitude **-1 V and 180° out of phase**.
- **QPSK** uses two carriers, one **in-phase** and the **other quadrature**. The point representing **11** is made of **two combined signal elements**, **both** with an amplitude of **1V**. **One element** is **represented** by an **in-phase** carrier, the other element by a **quadrature** carrier.



a. ASK (OOK)



b. BPSK



c. QPSK

Constellation Diagrams

- **Quadrature Amplitude Modulation (QAM):**

The **idea** of using two carriers, one in-phase and the other quadrature, **with different amplitude levels** for **each** carrier is the concept behind **quadrature amplitude modulation(QAM)**.

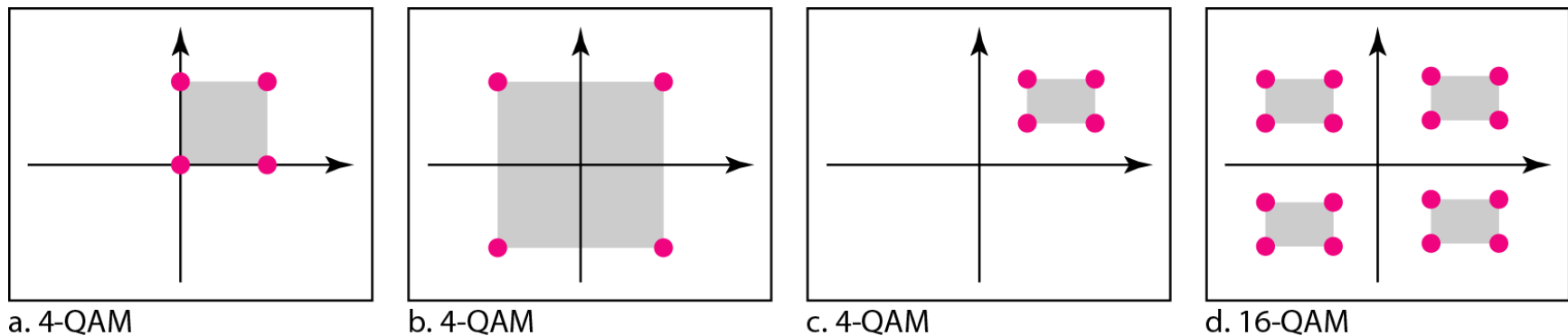


Fig. 13: Constellation diagrams for some QAMs

Continuous-phase frequency-shift keying (CPFSK)

- **Continuous-phase frequency-shift keying (CPFSK)** is a commonly used variation of frequency-shift keying (FSK), which is itself a special case of analog frequency modulation.
 - FSK is a method of modulating digital data onto a sinusoidal carrier wave, encoding the information present in the data to variations in the carrier's instantaneous frequency **between one of two frequencies** (referred to as the space frequency and mark frequency).
 - In general, a standard **FSK signal does not have continuous phase**, as the modulated waveform switches instantaneously between two sinusoids with different frequencies.
- As the name suggests, the phase of a CPFSK is in fact continuous;
 - this attribute is desirable for signals that are to be transmitted over **bandlimited channel**, as **discontinuities in a signal introduce wideband frequency components**.
 - In addition, some classes of amplifiers exhibit **nonlinear behavior** when driven **with nearly discontinuous signals**; this could have undesired effects on the shape of the transmitted signal.

Minimum Shift Keying (MSK)

- **Minimum Shift Keying (MSK):**

- **Minimum shift keying (MSK)** is a special type of **continuous phase-frequency shift keying (CPFSK)** with a **modulation index of 0.5**.
- A modulation index of **0.5 corresponds** to the minimum frequency spacing that **allows** two FSK signals to be **coherently orthogonal**, and the name minimum shift keying **implies** the minimum frequency separation (i.e. bandwidth) that **allows orthogonal detection**.

- **Gaussian Minimum Shift keying (GMSK):**

- Gaussian Minimum Shift Keying (GMSK) is a modification of MSK (i.e. CPFSK with $m = 1/2$).
- A filter used to reduce the **bandwidth** of a **baseband pulse train** prior to modulation is called a **pre-modulation filter**.
- The Gaussian pre-modulation filter smoothens the phase trajectory of the MSK signal thus limiting the instantaneous frequency variations. The result is an **FM modulated signal** with a much **narrower bandwidth**. This bandwidth reduction does not come for free since the pre-modulation filter smears the individual pulses in pulse train. As a consequence of this smearing in time, adjacent pulses interfere with each other generating what is commonly called **inter-symbol interference or ISI**. In the applications where GMSK is used, the trade-off between power efficiency and bandwidth efficiency is well worth the cost.

Shannon's Theorem

- **Shannon's Theorem:**

Shannon's Theorem gives an upper bound to the capacity of a link, in bits per second (bps), as a function of the available bandwidth and the **signal-to-noise ratio** of the link.

The Theorem can be stated as:

$$C = B * \log_2(1 + S/N)$$

where **C** is the **achievable channel capacity**, **B** is the **bandwidth** of the line, **S** is the **average signal power** and **N** is the **average noise power**.

The **signal-to-noise ratio (S/N)** is usually expressed in **decibels (dB)** given by the formula:

$$10 * \log_{10}(S/N)$$

so for example a signal-to-noise ratio of **1000** is commonly expressed as

•
$$10 * \log_{10}(1000) = 30 \text{ dB}$$

END