

## EECE571D 2025W1 Assignment 1 (100 points)

Due: October 24, 2025

A three-phase two-level VSC is connected to a constant DC voltage source on its DC-side, and its AC-side is connected to an infinite bus through an LR filter. The VSC system has the following specifications:

- Grid voltage: 580 V (line-to-line, RMS)
- Grid frequency: 60 Hz
- DC-link voltage: 1000 V
- Rated power: 100 kVA
- Modulation strategy: Sinusoidal PWM
- Control mode: GFL PQ control
- Switching frequency: 3.06 kHz

For all subsequent questions, you may use any design procedure to obtain your controller parameters. Bode plots, root locus, MATLAB SISOTOOL, Nyquist plot, state-space designs are all accepted, but trial-and-error methods are not acceptable.

You do not need to design the outer PQ controller loop: instead, divide your power set-points by grid voltage to obtain inner current controller references.

You can simulate the VSC system in either MATLAB/Simulink, PSCAD, or PLECS. Converter latency and higher-order harmonics can be neglected.

1. **[10 points]** Design the LR filter values for the VSC, using the procedure developed in class. Then use these values in all subsequent questions.
2. **[40 points]** Design PI current controller and PLL controller for the VSC, using the net control block diagram models developed in class.
  - a. Show your detailed design procedure for current controller (e.g., provide bode plot/root locus plot of your system, any calculations you have done, and/or any code, and explain why you have chosen the final controller parameters) and provide the final PI controller design in the format of  $C(s) = k_p + \frac{k_i}{s}$
  - b. Show your detailed design procedure for PLL controller (e.g., provide bode plot/root locus plot of your system, any calculations you have done, and/or any code, and explain why you have chosen the final controller parameters) and provide the final PI controller design in the format of  $H(s) = k_{pPLL} + \frac{k_{iPLL}}{s}$
  - c. Simulate the GFL VSC with your controllers: In the beginning, set both  $P^{ref}$  and  $Q^{ref}$  to 0, and subsequently, step up  $P^{ref}$  to 100 kW. Plot the following and briefly comment on the results (make sure you plot the waveforms for a total of 6 cycles, i.e., 2 cycles before step change and 4 cycles after step change)

- i.  $i_d$  and  $i_d^{ref}$  on the same plot
  - ii.  $i_q$  and  $i_q^{ref}$  on the same plot
  - iii. 3-phase AC currents on the same plot
  - iv. Measured real and reactive powers at the PCC on the same plot
- d. Repeat part (c) with a step change of 100 kVAr in reactive power exchange
3. **[40 points]** Repeat question (2) by placing second order low-pass filters on the  $i_{dq}$  measurements. The filter has a unity gain,  $\omega_n = 5000 \text{ rad/s}$ ,  $\zeta = 0.8$ , with an expression of
- $$L(s) = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$
- Keep in mind that there is no need to (i) re-design LR filter and (ii) re-design PLL controller again for this question.
4. **[10 points]** For the 3-phase system shown in class (gnd— $V_t$ —L—R— $V_s$ —gnd), we showed that  $[V_{s\alpha\beta 0}] = R[i_{\alpha\beta 0}] + L \frac{d}{dt}[i_{\alpha\beta 0}] + [V_{t\alpha\beta 0}]$ , where R and L are scalars. How, if at all, does this formulation change if  $V_s$  is unbalanced?

**General advice:** Don't over-think and over-complicate yourself. Follow the standard procedures you've learned, and everything should be fine. Ask me whenever you run into difficulties.