Digital Signal Processing for Music

Part 23: Source Coding

alexander lerch

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source coding introduction 1/3

intro ●00



typical audio bit rates

$$16 \, \text{bit} \cdot 44100 \, \text{sps} \cdot 2 \, \text{chan} = 1411.2 \, \text{kbps}$$

 $24 \, \text{bit} \cdot 192000 \, \text{sps} \cdot 5 \, \text{chan} = 23040 \, \text{kbps}$

- reasons for bit rate reduction
 - economical reasons: cheaper transmission/storage
 - technical reasons: restricted storage/transmission bandwidth
- applications for source coding
 - Internet: streaming, distribution, peer-2-peer, VoIP, ...
 - Media: DVD-V/A, . . .
 - Portable Devices: MP3-Player, cell phones, Mini-Disc, . . .
 - Broadcasting: (Digital) Radio, TV, ...
 - Cinema: DD, DTS, SDDS
 - .

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source coding introduction 1/3

intro ●00

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source coding introduction 2/3

intro

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How can the bitrate be reduced



intro

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I lossless:

remove redundant information (unnecessary to reconstruct the signal)

- entropy coding
- (linear predictive coding)

source coding introduction 2/3

intro



How can the bitrate be reduced



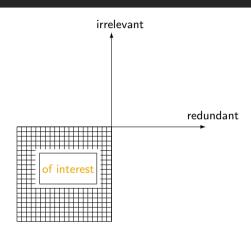
- entropy coding
- (linear predictive coding)
- 2 lossy: remove irrelevant information (not "missed" by the recipient)
 - waveform coding
 - perceptual coding



source coding introduction 3/3

intro 000





note: words to be transmitted are referred to as symbols

information content

The less frequent a symbol, the higher its information content, self-information, surprisal.

$$I_n = \log_2\left(\frac{1}{p_n}\right)$$

entropy

The entropy is the *Expected Value* of the information content. It is the *theoretic minimum of bits* required for transmission.

$$H = \sum_{n=0}^{N-1} p_n \cdot I_n$$

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fundamentals: information content and entropy examples

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dice:
$$p_n = \frac{1}{6}$$

$$I_n = \log_2\left(\frac{1}{p_n}\right) = 2.58 \text{ bit}$$

$$H = 6 \cdot \frac{1}{6} \cdot 2.58 \text{ bit}$$

fundamentals: information content and entropy examples

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■ imperfect dice: $p_1 = \frac{1}{2}, \ p_{2...6} = \frac{1}{10}$

fundamentals: information content and entropy examples

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■ imperfect dice:
$$p_1 = \frac{1}{2}, \ p_{2...6} = \frac{1}{10}$$

$$I_1 = \log_2(2) = 1 \text{ bit}$$
 $I_{2...6} = \log_2(10) = 3.32 \text{ bit}$
 $H = \frac{1}{2} \cdot 1 + \frac{5}{10} \cdot 3.32 = 2.16 \text{ bit}$

idea: use shorter words for frequent symbols

■ 3 possible symbols

symbol	probability	word
А	p = 0.5	
В	p = 0.25	
C	p = 0.25	

entropy

$$H = \sum_{n=0}^{N-1} p_n \log_2 \left(\frac{1}{p_n} \right) = 1.5$$

- $lue{}$ transmit the following group of symbols: ABCA
 ightarrow 010110
- required bits:

$$\frac{transmitted\ bits}{transmitted\ symbols} = \frac{6}{4} = 1.5$$

⇒ optimal transmission

source coding entropy coding: example 1

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symbol	probability	word
Α	p = 0.5	0
В	p = 0.25	10
C	p = 0.25	11

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entropy coding: example 1

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symbol	probability	word
Α	p = 0.7	
В	p = 0.2	
C	ho=0.1	

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⇒ *non*-optimal transmission

source coding entropy coding: example 2

■ 3 possible symbols

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A	p = 0.7	0
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source coding entropy coding: example 2

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source coding entropy coding: example 2

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source coding

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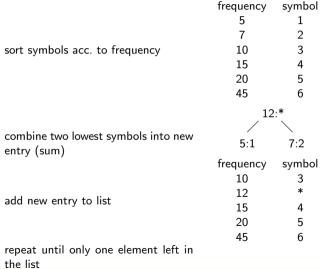
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 \Rightarrow non-optimal transmission

huffman coding: tree construction 1/2

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huffman coding: tree construction 2/2



5:1 7:2 10:3 15:4 20:5 45:6

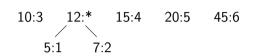
source coding huffman coding: tree construction 2/2

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5:1 7:2 10:3 15:4 20:5 45:6

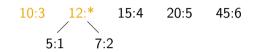
source coding huffman coding: tree construction 2/2

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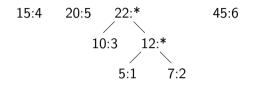
huffman coding: tree construction 2/2





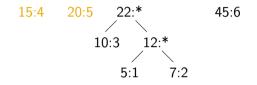
huffman coding: tree construction 2/2





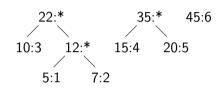
huffman coding: tree construction 2/2





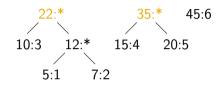
source coding huffman coding: tree construction 2/2

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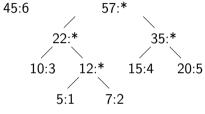
huffman coding: tree construction 2/2





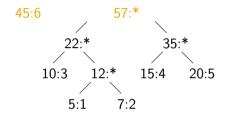
huffman coding: tree construction 2/2





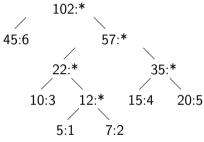
huffman coding: tree construction 2/2





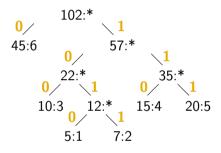
huffman coding: tree construction 2/2





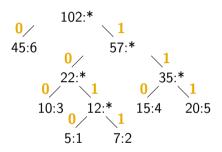
source coding huffman coding: tree construction 2/2





source coding huffman coding: tree construction 2/2

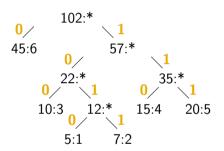
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frequency	symbol	code
5	1	1010
7	2	1011
10	3	100
15	4	110
20	5	111
45	6	0

source coding huffman coding: tree construction 2/2



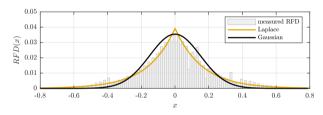


frequency	symbol	code
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note: no code is prefix of another code!

■ Symbole: 2^w

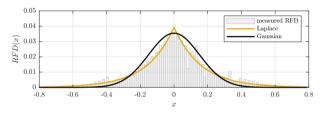
■ PDF indicates probability per symbol



matlab source: plotRfd

■ Symbole: 2^w

■ PDF indicates probability per symbol





■ Huffman coding is only optimal if $p_n = \frac{1}{2^k}$

- alternative: arithmetic coding
 - allows other probability distributions
 - encodes the whole sequence in one fractional number $0.0 \le f < 1.0$
 - principle:
 - 1 assume initial interval of [0,1[
 - 2 assign interval segments to all symbols, e.g. A = [0, 0.7], B = [0.7, 0.9], C = [0.9, 1]
 - 3 select interval based on current symbol
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source coding

arithmetic coding: example 1/2

sequence ABCA,
$$p_A = 0.6$$
, $p_B = 0.2$, $p_C = 0.1$, $p_T = 0.1$,

$$A = [0, 0.6], B = [0.6, 0.8], C = [0.8, 0.9], T = [0.9, 1]$$

- **decoding** 0.463:
 - 1 $0.463 \in \text{segment } 1 (\rightarrow A),$
 - ▶ set interval [0, 0.6] → bounds: 0, 0.36, 0.48, 0.54, 0.6
 - 2 0.463 ∈ segment 2 (\to *B*),
 - ▶ set interval $[0.36, 0.48[\rightarrow bounds: 0.36, 0.432, 0.456, 0.468, 0.48]$
 - 3 $0.463 \in \text{segment } 3 (\rightarrow C)$
 - ightharpoonup set interval $[0.456, 0.468] \rightarrow \text{bounds}$: 0.456, 0.4632, 0.4656, 0.4668, 0.468
 - 4 $0.463 \in \text{segment } 1 (\rightarrow A)$,
 - ▶ set interval [0.456, 0.4632] → bounds: 0.456, 0.46032, 0.46176, 0.46248, 0.4632
 - **5** $0.463 \in \text{segment 4} (\rightarrow \text{terminate})$

source coding

arithmetic coding: example 1/2

sequence ABCA,
$$p_{\rm A}=0.6, p_{\rm B}=0.2, p_{\rm C}=0.1, p_{\rm T}=0.1,$$
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arithmetic coding: example 1/2

metic coding: example
$$1/2$$
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 - 5 $0.463 \in \text{segment 4} (\rightarrow \text{terminate})$

source coding

arithmetic coding: example 1/2

sequence ABCA,
$$p_{\rm A}=0.6, p_{\rm B}=0.2, p_{\rm C}=0.1, p_{\rm T}=0.1,$$
 $A=[0,0.6[,B=[0.6,0.8[,C=[0.8,0.9[,T=[0.9,1[$

- **decoding** 0.463:
 - 1 $0.463 \in \text{segment } 1 (\rightarrow A)$.
 - ▶ set interval [0, 0.6] → bounds: 0, 0.36, 0.48, 0.54, 0.6
 - 2 $0.463 \in \text{segment } 2 (\rightarrow B)$,
 - ▶ set interval [0.36, 0.48] → bounds: 0.36, 0.432, 0.456, 0.468, 0.48
 - 3 $0.463 \in \text{segment } 3 (\rightarrow C)$,
 - ▶ set interval [0.456, 0.468] → bounds: 0.456, 0.4632, 0.4656, 0.4668, 0.468
 - 4 $0.463 \in \text{segment } 1 (\rightarrow A)$,
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arithmetic coding: example 2/2

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 - **5** select segment 4, set interval to [0.46248, 0.4632
 - 6 choose value from last segment (e.g., 0.463) and transmit

source coding

arithmetic coding: example 2/2

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idea: use preceding samples to estimate/predict future samples.

■ estimate the signal ×

$$\hat{x}(i) = \sum_{j=1}^{\mathcal{O}} b_j \cdot x(i-j)$$

prediction quality is measured by power of prediction error

$$e_{P}(i) = x(i) - \hat{x}(i)$$

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source coding

fundamentals: linear prediction — first order prediction 1/2

- **prediction** $\hat{x}(i) = b_1 \cdot x(i-1)$
- prediction error

$$\sigma_e^2 = \mathcal{E}\left\{ (x(i) - b_1 x(i-1))^2 \right\}$$

$$= \sigma_x^2 + b_1^2 \sigma_x^2 - 2b_1 r_{xx}(1)$$

$$= \left(1 + b_1^2 - 2b_1 \rho_{xx}(1) \right) \sigma_x^2$$

• optimum coefficient: $\frac{\partial \sigma_e^2}{\partial b_1} = 0$

$$2b_1\sigma_x^2 - 2\rho_{xx}(1)\sigma_x^2 = 0$$
$$b_1 = \rho_{xx}$$

minimum prediction error power

$$\sigma_e^2 = (1 + b_1^2 - 2b_1\rho_{xx}(1)) \sigma_x^2$$

$$= (1 + \rho_{xx}(1)^2 - 2\rho_{xx}(1)\rho_{xx}(1)) \sigma_x^2$$

$$= (1 - \rho_{xx}(1))\sigma_x^2$$

source coding

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fundamentals: linear prediction — first order prediction 2/2



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observations:

- power of prediction error always smaller or equal the power of the signal
- question: when is it equal to the signal?
- **special case**: $b_1 = 1$

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fundamentals: linear prediction — first order prediction 2/2



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fundamentals: linear prediction — prediction coefficients

Georgia Center for Music Tech (Technology Callege at Design

- prediction gain depends on
 - predictor coefficients b_j
 - signal
 - optimal coefficients can be derived by finding minimum of prediction error

$$\frac{\partial \sigma_e^2}{\partial b_j} = 0$$

⇒ (without derivation)

$$r_{\mathsf{xx}}(\eta) = \sum_{i=1}^{\mathcal{O}} b_{j,\mathrm{opt}} \cdot r_{\mathsf{xx}}(\eta - j), \quad 1 \leq \eta \leq \mathcal{O}$$

$$\mathbf{r}_{xx} = \mathbf{R}_{xx} \cdot \mathbf{b}_{opt}$$

 $\mathbf{b}_{opt} = \mathbf{R}_{xx}^{-1} \cdot \mathbf{r}_{xx}$

fundamentals: linear prediction — prediction coefficients

- Georgia Center for Music
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fundamentals: linear prediction — summary



predictor length

- rule of thumb: the longer the predictor, the better the prediction
- can range from 10 coefficients to hundreds

predictor coefficient updates

• better signal adaptation if coefficients are updated block-by-block

input signals

- white noise/random processes cannot be predicted
- periodic signals may theoretically be perfectly predicted

fundamentals: linear prediction — summary



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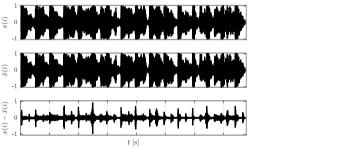
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fundamentals: linear prediction — audio example

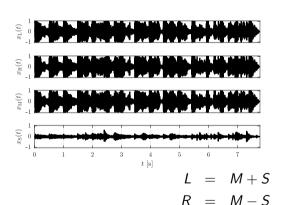
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order: 20



((





$$\blacktriangleleft M = \frac{L+R}{2}$$



- bitrate can be reduced by removing removing redundancy and/or irrelevance
- removing redundancy:
 - entropy coding: transmit frequent symbols with shorter codes
 - linear prediction: transmit diff signal plus predictor coefficients
- removing irrelevance:
 - reduce quantization wordlength
 - see slides below