

$$\frac{Y(s)}{R(s)} = \frac{G_c G_p}{1 + G_c G_p}$$

$$\frac{Y(s)}{T_d(s)} = \frac{G_p}{1 + G_c G_p}$$

$$\frac{Y(s)}{N(s)} = \frac{-G_c G_p}{1 + G_c G_p}$$

- The only way to limit noise is to have good sensors.

Define Sensitivity:

$$S(s) = \frac{1}{1 + G_c G_p}$$

Loop Transmission :  $L(s) = G_c G_p$

$$\Rightarrow S(s) = \frac{1}{1 + L}$$

Complimentary Sensitivity Function

$$C(s) = \frac{L}{1 + L}$$

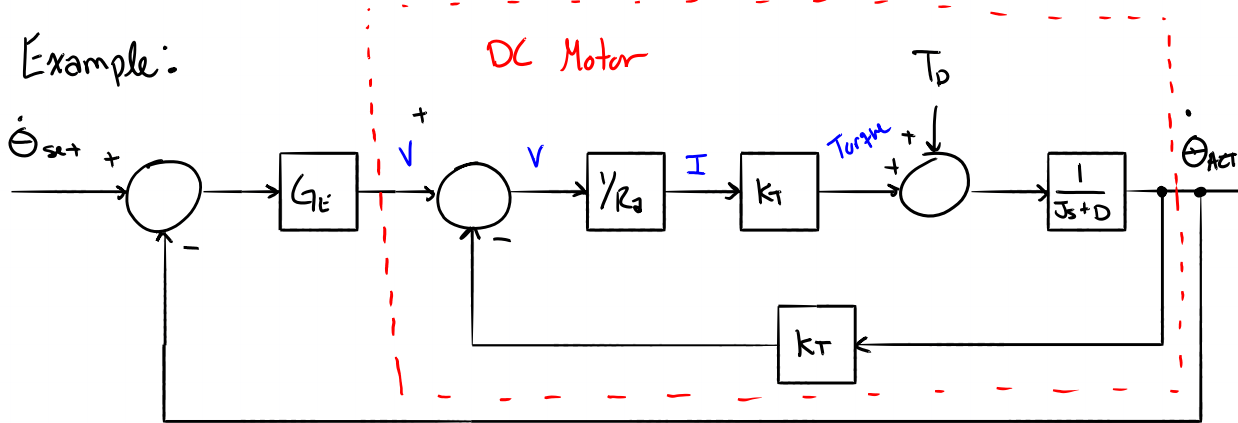
$$\Rightarrow C(s) + S(s) = 1$$

$$C(s) + R(s) = 1$$

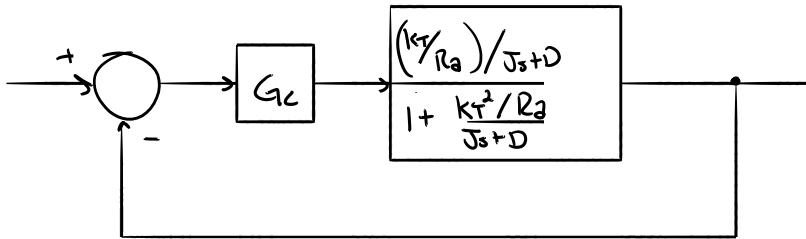
As  $S(s) \rightarrow 1$  the system becomes more sensitive.

IF  $C(s) \approx 1 \therefore \text{Sensitivity} \approx 0$

Example:



Set  $T_D = 0$

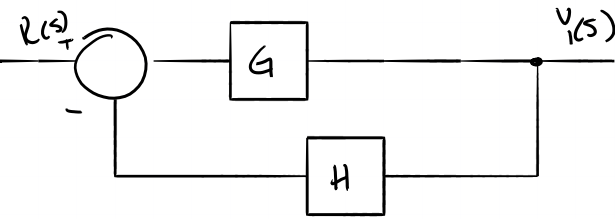


$$\Rightarrow \frac{\dot{\Theta}_{act}}{\dot{\Theta}_{set}} = \frac{G_c \frac{k_T/R_a}{Js + D}}{1 + \frac{k_T^2/R_a}{Js + D} + \frac{G_c k_T}{R_a}}$$

$$\left. \frac{\dot{\Theta}_{act}}{\dot{\Theta}_{set}} \right|_{\text{Steady State}} =$$

→ Replace  $s$  w/  $0$ . Because  $s$ 's are derivatives we are setting the equation @ steady state.

# ■ Sensitivity Math



$$S = \frac{\left( \frac{\Delta T(s)}{T(s)} \right)}{\left( \frac{\Delta G(s)}{G(s)} \right)}$$

$\left( \frac{\Delta T(s)}{T(s)} \right)$  ~ Fractional change  $T(s)$   
 Insensitive is close to 0 for this function.  
 $\left( \frac{\Delta G(s)}{G(s)} \right)$  ~ Fractional change of  $G(s)$

$$S = \lim_{\Delta T(s), \Delta G(s) \rightarrow 0} \frac{\Delta T(s) / T(s)}{\Delta G(s) / G(s)} = \frac{\delta T(s)}{\delta G(s)} \frac{G(s)}{T(s)} = \frac{\delta \ln T}{\delta \ln G}$$

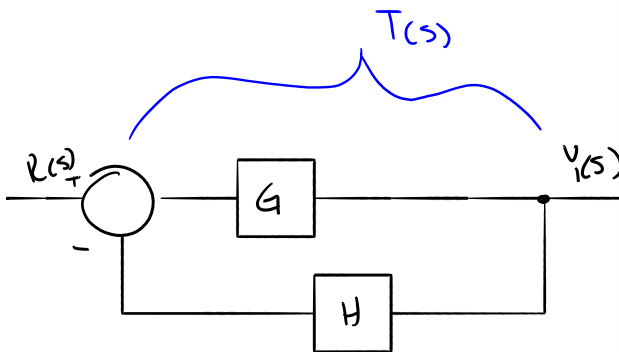
$$S_\alpha^T = S_G^T S_\alpha^G = \frac{\partial T}{\partial G} \frac{G}{T} \frac{\partial G}{\partial \alpha} \frac{\alpha}{G}$$

$$\delta(\ln A) = \frac{dA}{A}$$

↳ Sensitivity of parameter  $\alpha$  in a system with a transfer function  $T$ .

Manipulation used.

Example



$$G(s) = \frac{K}{\frac{s}{10} + 1}, \quad H=1$$

$$\frac{Y(s)}{R(s)} = T(s) = \frac{G(s)}{1 + G(s)}$$

$$S_G^T = \frac{\partial T}{\partial G} \frac{G}{T}$$

Using Quotient Rule

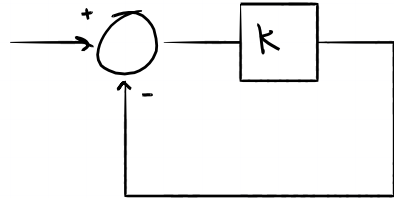
$$\frac{\partial T}{\partial G} = \frac{\partial}{\partial G} \left( \frac{G(s)}{1+G(s)} \right) = \frac{(1+G(s))(1) - G(s)1}{(1+G(s))^2}$$

$$\Rightarrow \frac{\partial T}{\partial G} = \frac{1}{[1+G(s)]^2}$$

$$\Rightarrow S_G^T = \frac{1}{[1+G(s)]^2} \cdot G(s) \cdot \frac{1+G(s)}{G(s)} = \frac{1}{1+G(s)}$$

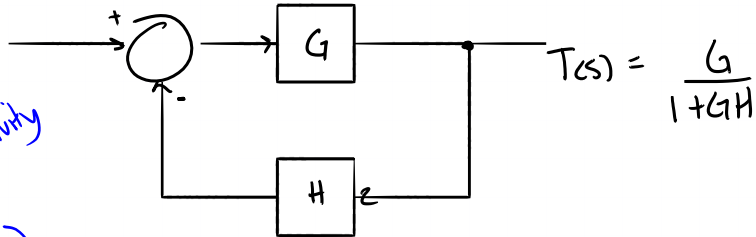
Example

$$T(s) = \frac{K}{1+K}$$



Example

1 to 1 sensitivity

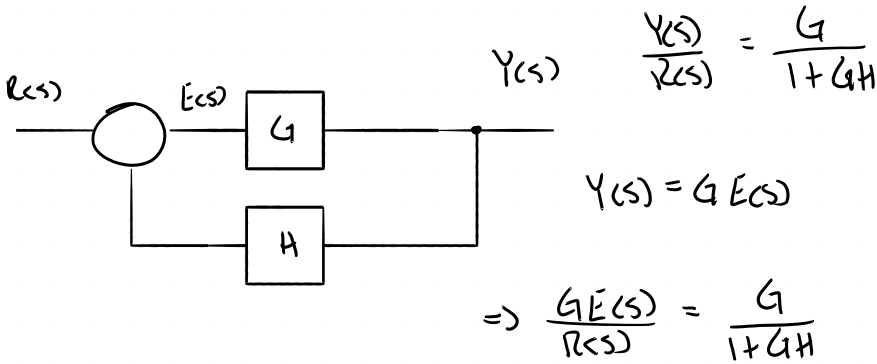


$$S_H^T = \frac{\partial T}{\partial H} \frac{H}{T} = \frac{-GH}{1+GH}$$

← This says if you have variation then the sensor will reflect that back to the input.

\* This is why it is important to have a good sensor because if you don't (as we can see in the sensitivity) the sensor will make your system do whatever it wants.

## Steady State Error



$$\Rightarrow \frac{E(s)}{R(s)} = \frac{1}{1+GH}$$

System error due to any input

Example

$$G = \frac{1}{s(s+1)} \quad H(s) = 1$$

$$\frac{E(s)}{R(s)} = \frac{1}{1 + \frac{1}{s(s+1)}} = \frac{s(s+1)}{s^2 + s + 1}$$

If  $R(s)$  is a unit step function  $R(s) = \frac{1}{s}$

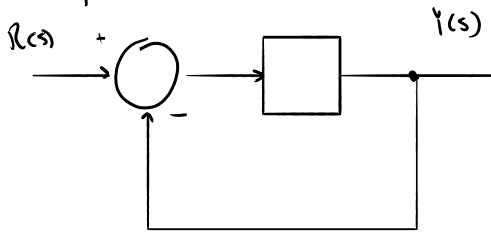
$$E(s) = \frac{1}{s} \frac{s(s+1)}{s^2 + s + 1} = \frac{s+1}{s^2 + s + 1}$$

Final value Theorem

$$\lim_{t \rightarrow \infty} e(t) = \lim_{s \rightarrow 0} s E(s)$$

$$\lim_{s \rightarrow 0} s \left[ \frac{s+1}{s^2+s+1} \right] = 0$$

Example



$$\text{Let } G(s) = \frac{1}{s+1}$$

↓

$$G(s) = \frac{k}{s+1}$$

$$\frac{E(s)}{R(s)} = \frac{1}{1 + \frac{1}{s+1}} = \frac{s+1}{s+2}$$

$$\lim_{t \rightarrow \infty} e(t) = \lim_{s \rightarrow 0} s E(s) \quad \text{lets assume } R(s) = 1/s$$

$$\Rightarrow \lim_{s \rightarrow 0} s \left[ \frac{1}{s} \frac{s+1}{s+2} \right] = \underline{\underline{\frac{1}{2}}}$$

This means that this system has a finite steady state error.

\* Integrators accumulate their inputs.

The number of free integrators define the system type.

↳ Free integrators are those  $s$  values that can be factored out by themselves.

Number of free integrators set the steady error output based on the input type.

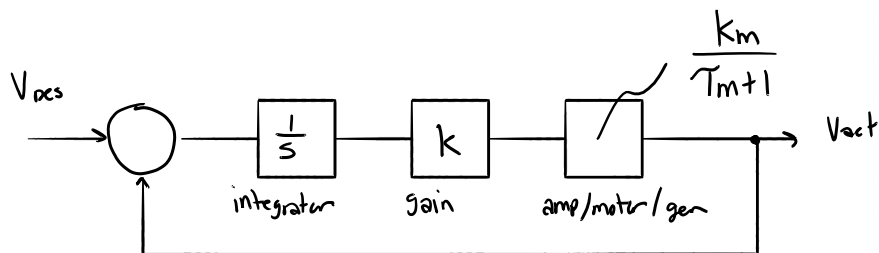
Suppose we have a system

$$G = \frac{1}{s(s+1)}$$

↑ open loop

↑ Free integrator

There is 0 steady state error for a step input in feedback.



finite steady state error for ramp input

$\infty$  steady state error for parabolic input.

2 Free integrators:

0 ss. error for step

0 ss error for ramp

finite ss error for parabola

\* Triple free integrators are difficult to control.

■ Position Error constant

$$R(s) = \frac{1}{s}$$

$$K_p \equiv \lim_{s \rightarrow 0} G = \lim_{s \rightarrow 0} \frac{k n(s)}{s^l d(s)} = \begin{cases} \frac{k(n(0))}{d(0)} & l=0 \\ \infty & \text{for } l>0 \end{cases}$$

↑ defined as

$$e(\infty) = \frac{1}{1+K_p}$$