

## HW #4 Answers & Solutions

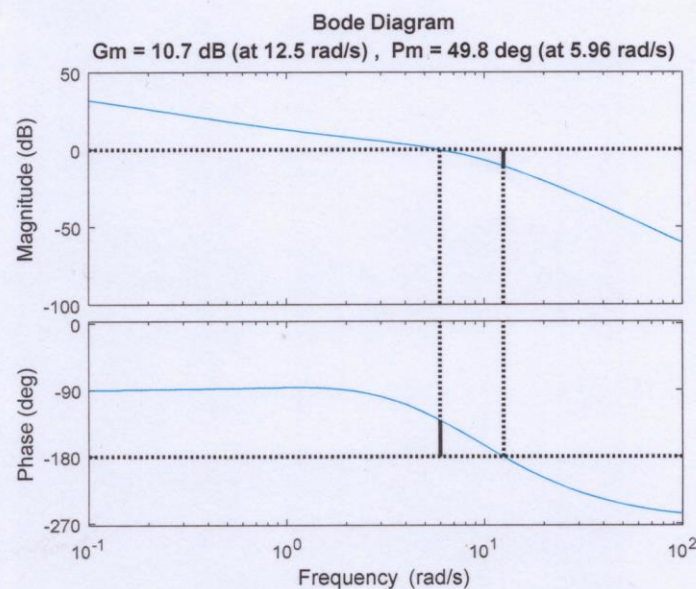
### Questions 1-2: Computer Assignments:

**Q1.** You should be able to verify that the gain margin of the system is  $\approx 15.5\text{dB}$  in three different ways (polar plot, Bode plot and root locus), by using the provided code in the question.

- Gain margin  $\approx 15.5\text{dB}$
- Phase margin  $\approx 90^\circ$ .

**Q2.**

a.



- Gain cross-over frequency = 5.96 rad/s.
- Phase cross-over frequency = 12.5 rad/s.
- Phase margin  $\approx 50^\circ$
- Gain margin = 10.7 dB

**b.** Desired phase margin is  $55^\circ$ . One can expect that a lead compensator with maximum phase ( $\phi_{max}$ ) of  $10^\circ$  ( $10 \approx \text{desired p.m.} - \text{actual p.m.} + 5^\circ$ ) should be sufficient to increase the phase margin to  $55^\circ$ . Unfortunately, since the phase of the system around the gain cross-over frequency changes too fast, it is not possible to achieve the desired phase margin with  $\phi_{max} = 10^\circ$ .

If you take the extra phase term added to be  $> 20^\circ$  instead of  $5^\circ$  (i.e.,  $\phi_{max} > 25$ ), you can achieve the desired phase margin.

**c.** The desired phase margin can be achieved by using a lag compensator  $G_c(s) \approx \frac{1+2.5s}{1+4s}$ .

Q3)

$$a) G(s) = \frac{0.2}{s^2(s+100)}$$

$$e_{ss} = 1/K_a, \quad K_a = \lim_{s \rightarrow 0} s^2 K G(s) = \lim_{s \rightarrow 0} \frac{K/5}{s+100} = \frac{K}{500}$$

$$e_{ss} < 0.25 \Rightarrow \frac{500}{K} < 0.25 \Rightarrow K > 2000, \quad K_{min} = 2000 //$$

$$b) G_{unc}(s) = K \cdot G(s) = \frac{400}{s^2 \cdot 100(1 + \frac{s}{100})} = \frac{4}{s^2(1 + \frac{s}{100})}$$

$$|G_{unc}(j\omega)| = \frac{4}{\omega^2(1 + (\frac{\omega}{100})^2)} \Big|_{\omega=\omega_g} = 1 \Rightarrow \omega_g \approx 2 \text{ rad/s}$$

$\approx 0$  for small  $\omega$

$$\angle G_{unc}(j\omega) = -\angle -\omega^2 - \angle 100 + j\omega$$

$$p.m. = \angle G_{unc}(j\omega_g) + 180 = -\angle -\omega_g^2 - \angle 100 + j\omega_g + 180$$

$$= -180 - \arctan \frac{2}{100} + 180 \approx 0 //$$

c)  $\angle G_{unc}(j\omega)$  is always below  $-180^\circ$ , therefore a phase-lag compensator cannot be used.

d) i) (Section: 2 & 2)

$$0 + \phi_m = p.m. desired + 5, \quad G_c(s) = \frac{1+Ts}{1+\alpha Ts}, \quad T > 0, \quad \alpha < 1$$

$$= 45 \Rightarrow \sin \phi_m = 0.71, \quad \alpha = \frac{1 - \sin \phi_m}{1 + \sin \phi_m} \approx 0.16, \quad \sqrt{\alpha} = 0.4$$

$\uparrow$  (Because  $1 < \omega_g < 10$ )

$$|G_{unc}(j\omega)|_{\omega=\omega_{gnew}} = \sqrt{\alpha} = 0.4 = \frac{4}{\omega_{gnew}^2 \sqrt{1 + \frac{\omega_{gnew}^2}{100^2}}} \Rightarrow 100 = \omega_{gnew}^4 + \frac{\omega_{gnew}^2}{100}$$

$$\omega_{gnew}^4 \approx 100 \Rightarrow \omega_{gnew} = \sqrt{10}$$

$$\omega_{gnew} = \omega_m = \frac{1}{\sqrt{\alpha} \cdot T} = T = \frac{1}{\sqrt{10} \cdot \sqrt{2}} = 0.79, \quad \alpha T = 0.16 \times 0.79 \approx 0.13 //$$

$$G_c(s) = \frac{1+Ts}{1+\alpha Ts} = \frac{1+0.79s}{1+0.13s} //$$



$$d) ii) (\text{Section: 3 \& 4}) \Rightarrow G_c(s) = \frac{1 + \alpha T s}{1 + T s}, \quad \alpha > 1, \quad T > 0$$

$$\phi_m = 45^\circ (= \text{p.m. desired} + 5 - 0)$$

$$\frac{1}{\alpha} = \frac{1 - \sin \phi_m}{1 + \sin \phi_m} \Rightarrow \alpha \approx 6.25 //$$

$$\left| G_{unc}(j\omega) \right| = \sqrt{\frac{1}{\alpha}} = 0.4 = \frac{4}{\omega_{gnw}^2 \sqrt{1 + \frac{\omega_{gnw}^2}{100^2}}} \Rightarrow 100 = \omega_{gnw}^4 + \frac{\omega_{gnw}^6}{100^2}$$

$\swarrow$   
neglect (Because  $1 \ll \omega_3 \ll 10$ )

$$\Rightarrow \omega_{gnw}^4 \approx 100 \Rightarrow \omega_{gnw} = \sqrt{10}$$

$$\omega_{gnw} = \omega_m = \frac{1}{\sqrt{\alpha}} \cdot \frac{1}{T} = 0.4 \times \frac{1}{T} = \omega_{gnw} = \sqrt{10} \Rightarrow T \approx 0.13 //$$

$$\alpha T \approx 0.81$$

$$G_c(s) = \frac{1 + \alpha T s}{1 + T s} = \frac{1 + 0.81s}{1 + 0.13s} //$$