

Attitude Control of a 3-DoF Quadrotor Platform using a Linear Quadratic Integral Differential Game Approach

Ali Baniasad, Alireza Sharifi, Reza Pordal, Hadi Nobahari

Attitude Control of a 3-DoF Quadrotor Platform using a Linear Quadratic Integral Differential Game Approach

Abstract—In this study, a linear quadratic integral differential game approach is applied to regulate and track the Euler angles for a quadrotor experimental platform using two players. One produces commands for each channel of the quadrotor and another generates the worst disturbance based on the mini-maximization of a quadratic criterion with integral action. For this purpose, first, the attitude dynamics of the platform are modeled and its parameters are identified based on the Nonlinear Least Squares Trust-Region Reflective method. The performance of the proposed controller is evaluated for regulation and tracking problems. The ability of the controller is also examined in the disturbance rejection. Moreover, the influence of uncertainty modeling is studied on the obtained results. Then, performance of the proposed controller is compared with the classic Proportional Integral Derivative, Linear Quadratic Regulator, and Linear Quadratic Integral Regulator. The results demonstrate the effectiveness of the Game Theory on the Linear Quadratic Regulator approach when the input disturbance occurs.

Index Terms—Linear Quadratic controller, Differential Game Theory, Quadrotor, 3-DoF Experimental Platform, Attitude Control.

I. INTRODUCTION

QUADROTORs are a type of Vertical Unmanned Aerial Vehicle (VUAV), that have various applications such as investigation, strategic operation, optical sensing, and entertainment. The safe flight of the quadrotor in the presence of disturbances relies on precise control. A crucial subsystem of a control system for the quadrotor is the Attitude Control System (ACS). To regulate the quadrotor attitude, a Proportional Integral Derivative (PID) controller is utilized in [1], [2]. Due to the nonlinearity dynamics of the quadrotor, the PID strategy is not effective in the presence of disturbance and modeling error. To provide a faster control command in facing the modeling error and reduce the disturbance effect in the attitude control, the model-based control strategies including nonlinear control, intelligent control, optimal control, and robust control have been implemented on the ACS of the quadrotor.

Synergetic Control [3], Feedback Linearization (FBL) [4], and Sliding Mode Control (SMC) [5], which are a group of the nonlinear control category, have been utilized to regulate the Euler angles (roll, pitch, and yaw angles) of the quadrotor. To control the attitude of the quadrotor intelligently, the controller strategies such as reinforcement learning [6], iterative learning [7], machine learning [8], and fuzzy logic [9] have been implemented. Moreover, to produce the optimal control commands, the controller strategies including Linear Quadratic Gaussian (LQG) [10], Linear Quadratic Regulator (LQR) [11], Linear Quadratic Integral Regulator (LQIR) [12],

and Model Predictive Controller (MPC) approaches [13] have been implemented on a quadrotor.

In the robust control strategies, H_∞ , μ -synthesis, and Linear Quadratic Regulator Differential Game (LQR-DG) [14] have also been utilized to stabilize the Euler angles of the quadrotor using the mini-maximization of a quadratic criterion including control effort and regulation performance in a worst-case scenario. In the LQR-DG controller [15], the control signals are produced using a pursuit-evasion game, one player tracks the best command, and the second player generates the input disturbance. For eliminating the steady-state error, the LQR-DG controller with integral action, called LQIR-DG, is implemented on a model of the ship [16].

In this paper, an LQIR-DG method is implemented real-time on 3-DoF experimental platform of the quadrotor to produce the robust control commands, i.e. rotational velocity of the quadrotor. To this end, first, the experimental platform of the quadrotor is modeled using the Newton-Euler formulation and its linear state-space form is derived. Then, the parameters of the quadrotor are estimated by matching experimental data with results from the model simulation. In the next step, the proposed controller is implemented on the Arduino Mega2560 board using the embedded coder platform in MATLAB and its performance is investigated in regulation and tracking problems. Moreover, the rejection capability of the input disturbance and modeling error is tested. Finally, a comparison is also performed between the results of classical PID, LQR, and LQIR with the proposed method. The results demonstrate that this method has an excellent performance in the attitude control of the quadrotor platform. A demo video of the results is available online here.

This research is organized as follows: the problem statement is presented in section II. In section III, the dynamic platform is modeled. Then, the presented controller architecture is denoted in section IV. The numerical results and a conclusion are represented in sections V and VI, respectively.

mds ??????

August 26, 2015 ??????

II. PROBLEM STATEMENT

The experimental quadrotor platform rotates freely with rotational velocity ($\Omega_i, i = 1, 2, 3, 4$) about its roll, pitch, and yaw axes, according to Figure 1. The angular velocities in the body frame (p, q, r) and the Euler angles (ϕ, θ, ψ) are measured using an Attitude Heading Reference System (AHRS). The measured states are utilized in the structure of the proposed controller to stabilize the quadrotor platform. The graphical

abstract of the LQIR-DG controller strategy is depicted in Figure 2.



Fig. 1: 3-DoF Quadrotor platform.

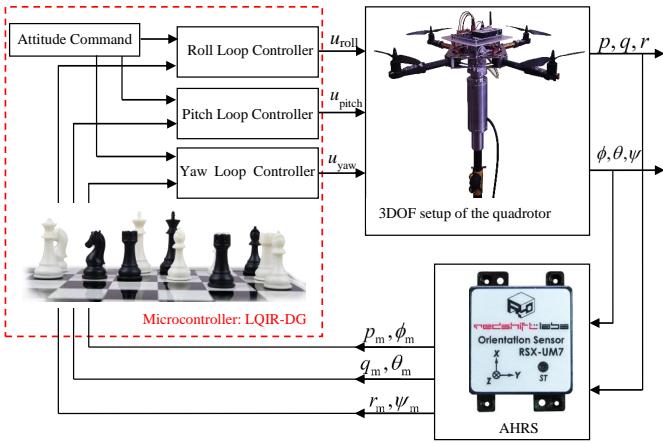


Fig. 2: Graphical abstract of the LQIR-DG controller.

III. MODEL OF THE QUADROTOR PLATFORM

Here, the quadrotor platform is modeled as nonlinear. Then, a state-space model and a linear model are developed for control purposes to be utilized in the controller strategy. Finally, a nonlinear identification method is applied to identify the parameters of the quadrotor.

A. Quadrotor Configuration

According to Figure 3, the 3-DoF quadrotor schematic is including four rotors rotating the z_B axis in the body frame with a rotational velocity, Ω_i ($i = 1, 2, 3, 4$). To eliminate the yawing moment, rotors (2, 4) and (1, 3) rotate clockwise and counter clockwise, respectively.

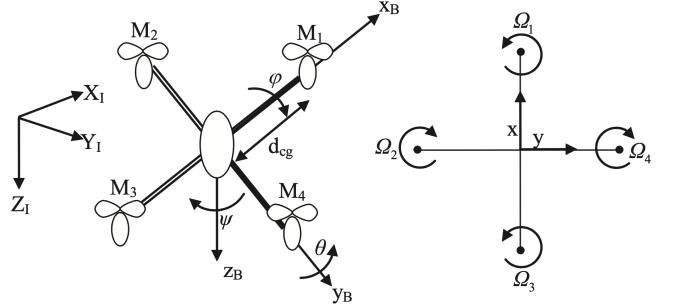


Fig. 3: Quadrotor configuration.

B. Dynamic Modeling of the Quadrotor Platform

Here, according to Newton-Euler, the model of the quadrotor platform is presented as follows [17], [18]:

$$\dot{p} = \frac{I_{yy} - I_{zz}}{I_{xx}} qr + q \frac{I_{rotor}}{I_{xx}} \Omega_{c,r} + \frac{b d_{cg} (\Omega_{c,2}^2 - \Omega_{c,4}^2)}{I_{xx}} + \frac{d_{roll}}{I_{xx}} \quad (1)$$

$$\dot{q} = \frac{I_{zz} - I_{xx}}{I_{yy}} rp + p \frac{I_{rotor}}{I_{xx}} \Omega_{c,r} + \frac{b d_{cg} (\Omega_{c,1}^2 - \Omega_{c,3}^2)}{I_{yy}} + \frac{d_{pitch}}{I_{yy}} \quad (2)$$

$$\dot{r} = \frac{I_{xx} - I_{yy}}{I_{zz}} pq + \frac{d (\Omega_{c,1}^2 - \Omega_{c,2}^2 + \Omega_{c,3}^2 - \Omega_{c,4}^2)}{I_{zz}} + \frac{d_{yaw}}{I_{zz}} \quad (3)$$

where $\Omega_{c,i}$ ($i = 1, 2, 3, 4$) is the rotational velocity, computed as

$$\Omega_{c,1}^2 = \Omega_{mean}^2 + \frac{1}{2b d_{cg}} u_{pitch} + \frac{1}{4d} u_{yaw} \quad (4)$$

$$\Omega_{c,2}^2 = \Omega_{mean}^2 + \frac{1}{2b d_{cg}} u_{roll} - \frac{1}{4d} u_{yaw} \quad (5)$$

$$\Omega_{c,3}^2 = \Omega_{mean}^2 - \frac{1}{2b d_{cg}} u_{pitch} + \frac{1}{4d} u_{yaw} \quad (6)$$

$$\Omega_{c,4}^2 = \Omega_{mean}^2 - \frac{1}{2b d_{cg}} u_{roll} - \frac{1}{4d} u_{yaw} \quad (7)$$

In the above equation, Ω_{mean} is the rotational velocity of the rotors. Also, d_{cg} , d , and b represent the distance between the rotors and the gravity center, drag factor, and thrust factor, respectively. d_{roll} , d_{pitch} , and d_{yaw} denote the disturbances produced in the body coordinate frame. Additionally, u_{roll} , u_{pitch} , and u_{yaw} are control commands generated by the LQIR-DG controller. I_{rotor} is rotor inertia, and I_{xx} , I_{yy} , and I_{zz} are the moments of inertia. Euler angle rates are also determined from angular body rates as follows:

$$\begin{bmatrix} \dot{\phi} \\ \dot{\theta} \\ \dot{\psi} \end{bmatrix} = \begin{bmatrix} 1 & \sin(\phi) \tan(\theta) & \cos(\phi) \tan(\theta) \\ 0 & \cos(\phi) & -\sin(\phi) \\ 0 & \sin(\phi)/\cos(\theta) & \cos(\phi)/\cos(\theta) \end{bmatrix} \begin{bmatrix} p \\ q \\ r \end{bmatrix} \quad (8)$$

The residual rotor velocity, denoted by $\Omega_{c,r}$, is calculated as follows:

$$\Omega_{c,r} = -\Omega_{c,1} + \Omega_{c,2} - \Omega_{c,3} + \Omega_{c,4} \quad (9)$$

C. State-Space Formulation

By defining $\mathbf{x}_{\text{roll}} = [x_1 \ x_2]^T = [p \ \phi]^T$, $\mathbf{x}_{\text{pitch}} = [x_3 \ x_4]^T = [q \ \theta]^T$, and $\mathbf{x}_{\text{yaw}} = [x_5 \ x_6]^T = [r \ \psi]^T$, the formulation of the quadrotor platform is presented as follows:

$$\dot{x}_1 = \Gamma_1 x_3 x_5 + \Gamma_2 x_3 \Omega_r + \Gamma_3 b d_{\text{cg}} (\Omega_{c,1}^2 - \Omega_{c,3}^2) + \Gamma_3 d_{\text{roll}} \quad (10)$$

$$\dot{x}_2 = x_1 + (x_3 \sin(x_2) + x_3 \cos(x_2)) \tan(x_4) \quad (11)$$

$$\dot{x}_3 = \Gamma_4 x_1 x_5 - \Gamma_5 x_1 \Omega_r + \Gamma_6 b d_{\text{cg}} (\Omega_{c,2}^2 - \Omega_{c,4}^2) + \Gamma_6 d_{\text{pitch}} \quad (12)$$

$$\dot{x}_4 = x_3 \cos(x_2) - x_5 \sin(x_2) \quad (13)$$

$$\dot{x}_5 = \Gamma_7 x_1 x_3 + \Gamma_8 d (\Omega_{c,1}^2 - \Omega_{c,2}^2 + \Omega_{c,3}^2 - \Omega_{c,4}^2) + \Gamma_8 d_{\text{yaw}} \quad (14)$$

$$\dot{x}_6 = (x_3 \sin(x_4) + x_5 \cos(x_2)) / \cos(x_4) \quad (15)$$

where $\Gamma_i (i = 1, \dots, 8)$ is defined as

$$\begin{aligned} \Gamma_1 &= \frac{I_{yy} - I_{zz}}{I_{xx}}, & \Gamma_2 &= \frac{I_{\text{rotor}}}{I_{xx}}, & \Gamma_3 &= \frac{1}{I_{xx}} \\ \Gamma_4 &= \frac{I_{zz} - I_{xx}}{I_{yy}}, & \Gamma_5 &= \frac{I_{\text{rotor}}}{I_{xx}}, & \Gamma_6 &= \frac{1}{I_{yy}} \\ \Gamma_7 &= \frac{I_{xx} - I_{yy}}{I_{zz}}, & \Gamma_8 &= \frac{1}{I_{zz}} \end{aligned} \quad (16)$$

The measurement vector, obtained from the AHRS, is presented as follows:

$$\mathbf{z} = [p \ q \ r \ \phi \ \theta \ \psi]^T + \boldsymbol{\nu} \quad (17)$$

where $\boldsymbol{\nu}$ is a Gaussian white noise. Moreover, the superscripts T indicate the transpose notation.

D. Linear Model

By defining $\dot{\mathbf{x}} = [\dot{x}_{\text{roll}} \ \dot{x}_{\text{pitch}} \ \dot{x}_{\text{yaw}}]^T$, the linear model of the quadrotor platform represented about the equilibrium points ($\mathbf{x}_e^* = 0$ and $\mathbf{u}_e^* = 0$) as

$$\dot{\mathbf{x}} = \mathbf{A} \mathbf{x} + \mathbf{B} (\mathbf{u} + \mathbf{d}) \quad (18)$$

\mathbf{A} is the dynamic system matrix, denoted as

$$\mathbf{A} = \begin{bmatrix} \mathbf{A}_{\text{roll}} & \mathbf{0} & \mathbf{0} \\ \mathbf{0} & \mathbf{A}_{\text{pitch}} & \mathbf{0} \\ \mathbf{0} & \mathbf{0} & \mathbf{A}_{\text{yaw}} \end{bmatrix} \quad (19)$$

$\mathbf{A}_{\text{roll}} = \mathbf{A}_{\text{pitch}} = \mathbf{A}_{\text{yaw}} = \begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix}$. Also, \mathbf{B} is the input matrix defined as

$$\mathbf{B} = \begin{bmatrix} \mathbf{B}_{\text{roll}} & \mathbf{0} & \mathbf{0} \\ \mathbf{0} & \mathbf{B}_{\text{pitch}} & \mathbf{0} \\ \mathbf{0} & \mathbf{0} & \mathbf{B}_{\text{yaw}} \end{bmatrix} \quad (20)$$

where $\mathbf{B}_{\text{roll}} = \begin{bmatrix} \frac{1}{I_{xx}} & 0 \end{bmatrix}^T$, $\mathbf{B}_{\text{pitch}} = \begin{bmatrix} \frac{1}{I_{yy}} & 0 \end{bmatrix}^T$, and $\mathbf{B}_{\text{yaw}} = \begin{bmatrix} \frac{1}{I_{zz}} & 0 \end{bmatrix}^T$.

E. Identification of the Platform Parameters

In this section, the Nonlinear Least Squares (NLS) algorithm is utilized for estimating the model parameters (Γ) of the 3-DoF experimental platform using experimental data. This technique is based on the Trust-Region Reflective (TRR) method, which finds the best values for Γ by minimizing a cost function, defined as

$$\min_{\Gamma} (\| e(\Gamma) \|^2) = \min_{\Gamma} \left(\sum_{j=1}^n (\mathbf{z}_j - \tilde{\mathbf{z}}_j)(\mathbf{z}_j - \tilde{\mathbf{z}}_j)^T \right) \quad (21)$$

where \mathbf{z} and $\tilde{\mathbf{z}}$ are the experimental and simulated output signals when the same input signals are applied. Moreover, n is the number of scenarios. To find a vector Γ , the optimization process performs until convergence is achieved. The structure of the identification approach is illustrated in figure 4.

IV. LQIR-DG CONTROLLER STRUCTURE

First, the augmented states of the quadrotor platform, including the states and their integrals are selected to use in the structure of the LQIR-DG controller for eliminating the steady-state errors. Then, the design methodology of the controller structure is introduced to produce the best commands for the 3-DoF quadrotor platform.

A. Augmented States

To augment an integral action into the control strategy architecture, the augmented states are defined as $\mathbf{x}_a = [\mathbf{x} \ \int \mathbf{x}]^T$. Then, the quadrotor platform model, utilized in the controller structure, is presented as

$$\dot{\mathbf{x}}_a = \begin{bmatrix} \mathbf{A} & \mathbf{0} \\ \mathbf{I} & \mathbf{0} \end{bmatrix} \mathbf{x}_a + \begin{bmatrix} \mathbf{B} \\ \mathbf{0} \end{bmatrix} (\mathbf{u} + \mathbf{d}) \quad (22)$$

The notation \mathbf{I} denotes the identity matrix.

B. LQIR-DG Control Scheme with Integral Action

In the proposed controller scheme, two fundamental players are selected in accordance with the game theory approach. The primary player determines the control commands, while another player generates the worst possible disturbance. To achieve the primary objective, the first player minimizes the following cost function but the other player maximizes it:

$$\min_u \max_d J(\mathbf{x}_{a_i}, d_i, u_i) = \min_d \max_u \int_0^{t_f} \left(\mathbf{x}_{a_i}^T \mathbf{Q}_i \mathbf{x}_{a_i} + u_i^T R u_i - d_i^T R_d d_i \right) dt \quad (23)$$

where t_f is the stop time and i -index denotes the roll, pitch, and yaw channels of the quadrotor. \mathbf{Q}_i , R_d , and R are weight coefficients of the cost function. By solving the above problem, the optimal control command is computed as follows [19]:

$$u_i = -\mathbf{K}_i \mathbf{x}_{a_i} \quad (24)$$

Moreover, the worst disturbance is obtained as

$$d_i = \mathbf{K}_{d_i} \mathbf{x}_{a_i} \quad (25)$$

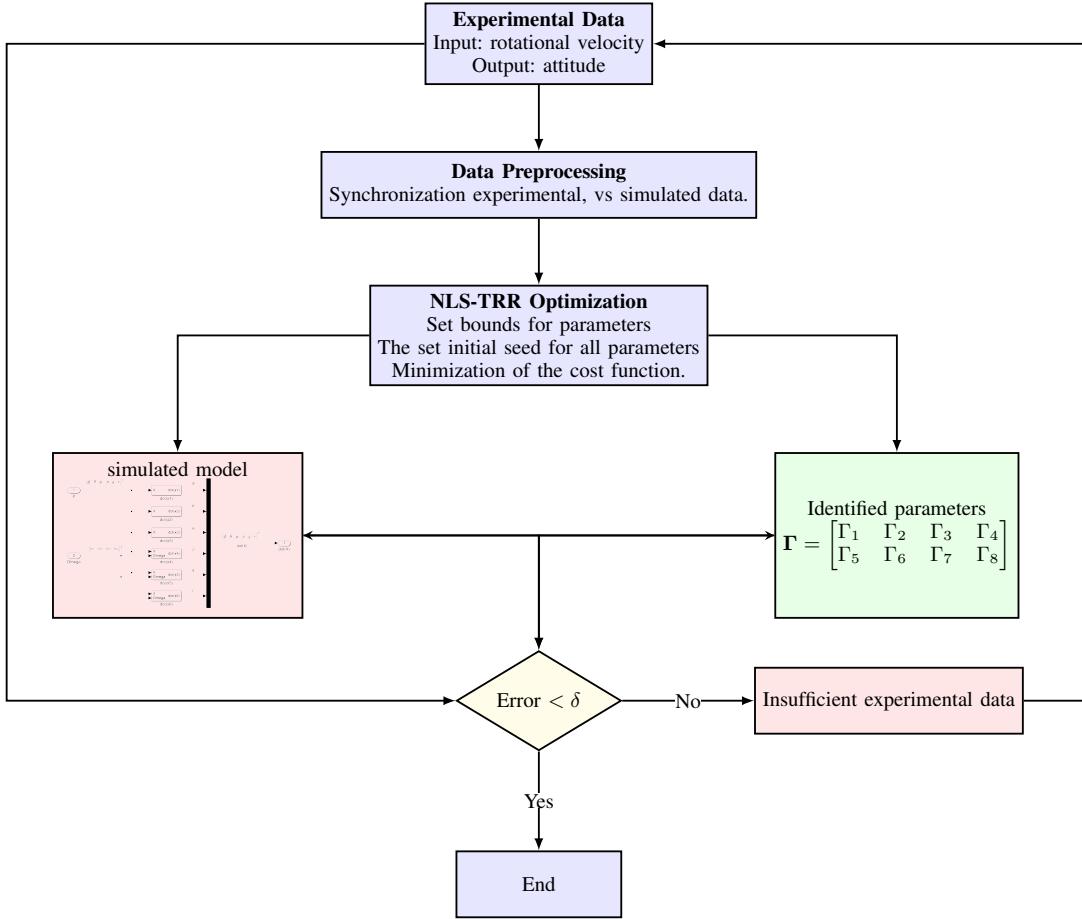


Fig. 4: Structure of TRRLS identification approach.

Here, \mathbf{K}_{d_i} and \mathbf{K}_i are gain values defined as follows:

$$\mathbf{K}_{d_i} = R_d^{-1} \mathbf{B}_{a_{d_i}}^T \mathbf{P}_{a_{d_i}} \quad (26)$$

$$\mathbf{K}_i = R^{-1} \mathbf{B}_{a_i}^T \mathbf{P}_{a_i} \quad (27)$$

\mathbf{P}_{a_i} and $\mathbf{P}_{a_{d_i}}$ satisfy

$$-\mathbf{A}_a^T \mathbf{P}_{a_{d_i}} - \mathbf{Q}_i - \mathbf{P}_{a_{d_i}} \mathbf{A}_a + \mathbf{P}_{a_{d_i}} \mathbf{S}_{a_i} \mathbf{P}_{a_i} + \mathbf{P}_{a_{d_i}} \mathbf{S}_{a_{d_i}} \mathbf{P}_{a_{d_i}} = \mathbf{0} \quad (28)$$

$$-\mathbf{A}_a^T \mathbf{P}_{a_i} - \mathbf{Q}_i - \mathbf{P}_{a_i} \mathbf{A}_a + \mathbf{P}_{a_i} \mathbf{S}_{a_{d_i}} \mathbf{P}_{a_{d_i}} + \mathbf{P}_{a_i} \mathbf{S}_{a_i} \mathbf{P}_{a_i} = \mathbf{0} \quad (29)$$

where $\mathbf{S}_{a_i} = \mathbf{B}_{a_i} R^{-1} \mathbf{B}_{a_i}^T$ and $\mathbf{S}_{a_{d_i}} = \mathbf{B}_{a_{d_i}} R_d^{-1} \mathbf{B}_{a_{d_i}}^T$.

V. RESULTS

The results of the parameter identification and the LQIR-DG Controller for the quadrotor platform are presented. First, the quadrotor parameters are estimated based on the NLS method. Then, performance of the LQIR-DG structure is evaluated. Tables I and II present the quadrotor and LQIR-DG parameters, respectively.

TABLE II: LQIR-DG controller parameters

Channel	Weighting Matrix	Values
Roll	\mathbf{Q}_{roll}	$\text{diag}([0.02, 65.96, 83.04, 0.00])$
Pitch	$\mathbf{Q}_{\text{pitch}}$	$\text{diag}([435.01, 262.60, 262.60, 0.00])$
Yaw	\mathbf{Q}_{yaw}	$\text{diag}([4 \times 10^{-4}, 0.00, 0.133, 0])$
-	R	1
-	R_d	1.2764

A. Identification of the 3-DoF quadrotor platform model

As described in section III-C, the parameters of the quadrotor platform, denoted by $\Gamma_i (i = 1, \dots, 8)$, are identified using the NLS-TRR algorithm. To increase the accuracy of parameter identification, three scenarios are considered according to Table III. In the first scenario, depicted in Figure 5, the quadrotor rotates about only one axis (roll, pitch, or yaw axes) to identify the parameters Γ_3 , Γ_6 , and Γ_8 . In the second scenario, according to Figure 6, the parameters Γ_2 and Γ_5 are estimated by rotating the experimental platform around its roll and pitch axes simultaneously. Finally, Figure 7 displays the results of the third scenario including the estimation of the parameters Γ_1 , Γ_4 , and Γ_7 for the UAV model, when the platform freely rotates around three axes. After the termination condition is met, the optimal values of the quadrotor parameters are computed and denoted in Table IV. These results illustrate that

TABLE I: Quadrotor parameters

Parameter	Unit	Value	Parameter	Unit	Value
d_{cg}	m	0.2	I_{xx}	$\text{kg} \cdot \text{m}^2$	0.02839
d	$\text{N} \cdot \text{m} \cdot \text{sec}^2 / \text{rad}^2$	3.2×10^{-6}	I_{yy}	$\text{kg} \cdot \text{m}^2$	0.03066
b	$\text{N} \cdot \text{sec}^2 / \text{rad}^2$	3.13×10^{-5}	I_{zz}	$\text{kg} \cdot \text{m}^2$	0.0439
Ω_{mean}	rpm	2000	I_{rotor}	$\text{kg} \cdot \text{m}^2$	4.4398×10^{-5}

the outputs of the simulation results for the quadrotor model are consistent with reality.

TABLE IV: True values of the quadrotor parameters.

Parameter	Value	Parameter	Value
Γ_1	-0.9622	Γ_5	3.6441×10^{-4}
Γ_2	-0.0154	Γ_6	7.5395×10^{-5}
Γ_3	5.4716×10^{-5}	Γ_7	0.1308
Γ_4	1.0457	Γ_8	4.3753×10^{-5}

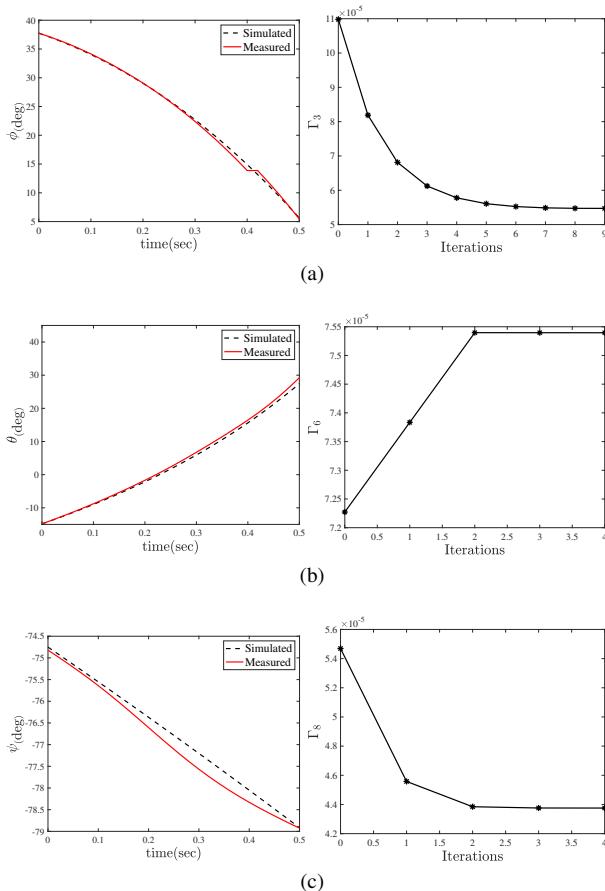


Fig. 5: Identification process results when the quadrotor rotates about only one axis: (a) identification of Γ_3 in free roll motion. (b) identification of Γ_6 in free pitch motion. (c) identification of Γ_8 in free yaw motion.

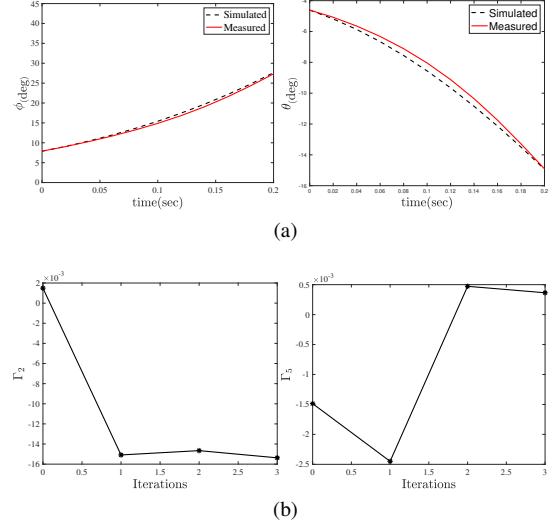


Fig. 6: Identification process results when the quadrotor rotates about its roll and pitch axes: (a) comparison of simulation and experimental results. (b) identification of Γ_2 and Γ_5 .

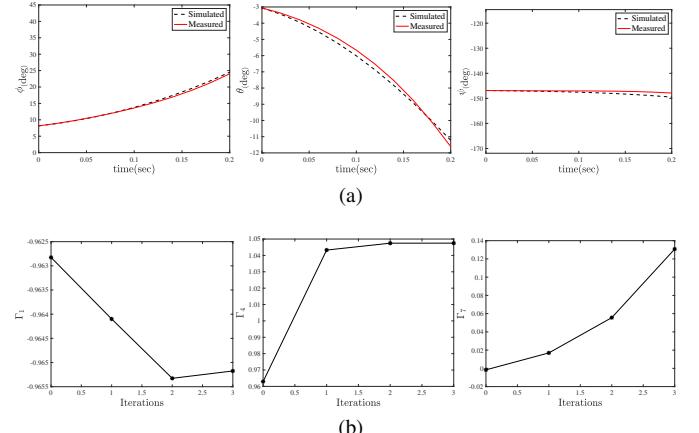


Fig. 7: Identification process results when the quadrotor rotates about its roll, pitch, and yaw axes: (a) comparison of simulation and experimental results. (b) identification of Γ_1 , Γ_4 and Γ_7 parameters.

B. Evaluation of LQIR-DG Performance

In this section, the LQIR-DG controller algorithm is evaluated in three scenarios i) regulation and tracking problems, ii) disturbance rejection, and iii) impact of model uncertainty. Finally, a comparison of the proposed controller is performed with a PID controller and variants of the LQR controller. The PID controller parameters are presented in Table V.

TABLE III: Scenarios for identification of quadrotor parameters.

Scenario	Description	Initial Condition (deg)			Rotational Velocity Commands (rpm)			
		ϕ_0	θ_0	ψ_0	Ω_1	Ω_2	Ω_3	Ω_4
I	roll free	38	-	-	2000	2000	2000	3400
	pitch free	-	-15	-	3700	2000	2000	2000
	yaw free	-	-	-75	2000	3300	2000	3300
II	roll & pitch free	8	-5	-	1700	3800	2400	1700
III	roll, pitch, & yaw free	8	-3	-146	1700	3800	2400	1700

TABLE V: PID controller parameters

Channel	K_p	K_i	K_d
roll	18	6	9
pitch	22	15	16

1) *Investigating of the Regulation and Tracking Problems:* The results of the proposed approach are presented for tracking the desired roll and pitch angles in Figures 8 and 9. Figure 8 (a) compares the desired and output signals, i.e., the Euler angles during the regulation problem. Moreover, Figure 8 (b) compares the desired square wave signals with a frequency of 0.02 Hz and an amplitude of 20 degrees with the output signals, when the quadrotor platform freely rotates around roll and pitch simultaneously. Figures 9 (a) and (b) show the rotational velocity commands of the quadrotor in the regulation and tracking problems, respectively. These results demonstrate that the roll and pitch angles are accurately controlled by the proposed approach.

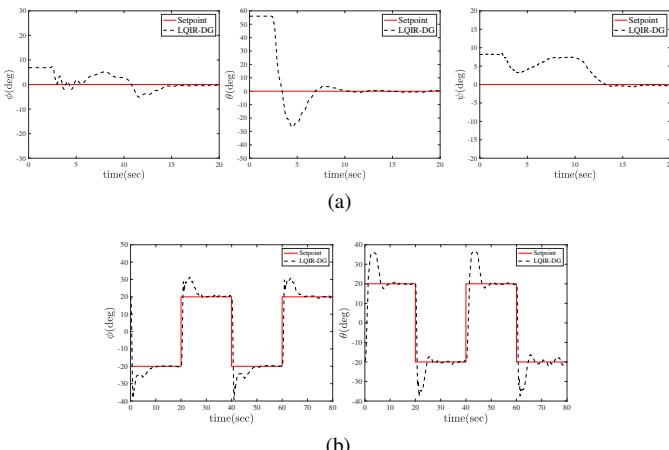


Fig. 8: Comparison of actual roll and pitch angles with the desired values in (a) regulation and (b) tracking problems.

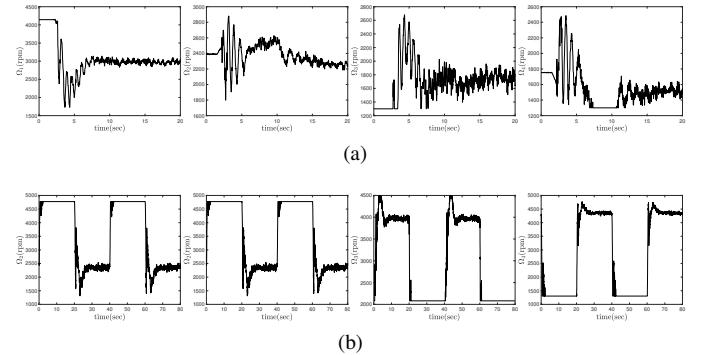


Fig. 9: Rotational velocity commands in (a) regulation and (b) tracking problems.

2) *Investigating the Disturbance Rejection:* Here, the effect of the input disturbance is investigated on the performance of the proposed controller. The input disturbance, d_{Ω_i} , is considered as a change in the command of the rotational velocity, modeled as

$$d_{\Omega_1} = d_{\Omega_2} = -d_{\Omega_3} = -d_{\Omega_4} = \begin{cases} 500 \text{ rpm} & 20 < t < 60 \\ 0 & \text{otherwise} \end{cases} \quad (30)$$

Figure 10 illustrates the roll and pitch angles in the regulation problem, when the input disturbance occurs. These results indicate that the proposed controller can stabilize the quadrotor platform in the presence of input disturbance.

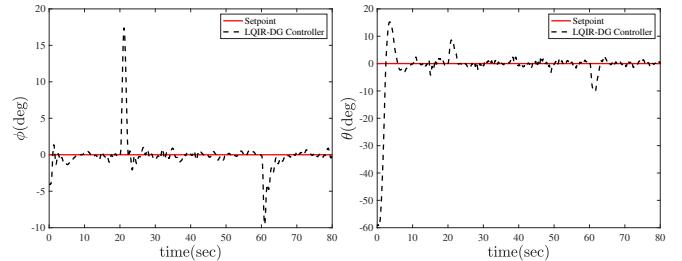


Fig. 10: Comparison of actual roll and pitch angles with the desired, when the input disturbance occurs.

3) *Investigating the Impact of Modeling Uncertainty:* The effect of the modeling uncertainty is investigated on the performance of the proposed controller. To achieve this, 50 and 100 grams weights are added to the roll and pitch axes, respectively, as shown in Figure 11. Figure 12 (a) compares the desired and the actual roll angle and Figure 12 (b) shows the desired and the actual pitch angle, when the uncertainty of

moments of inertia is present. Moreover, Figure 12 (c) shows the rotational velocity commands of the experimental platform, when the model uncertainty is applied. The implementation results show that the platform outputs converge to the desired values in the presence of the modeling uncertainty.

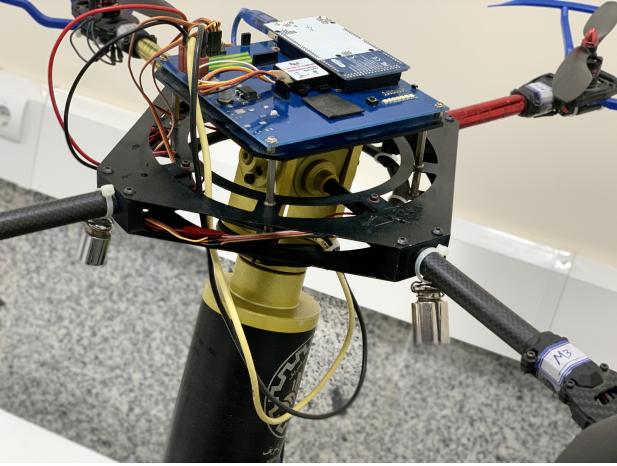


Fig. 11: Quadrotor 3-DoF platform with added weights.

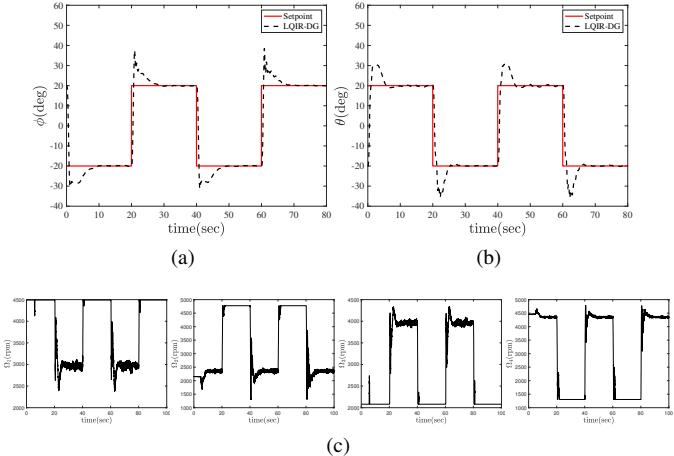


Fig. 12: Comparison of actual roll and pitch angles with desired values, when the modeling uncertainty is present.

4) Comparison with the Control Strategies: Figure 13 compares the LQIR-DG controller performance with the PID controller and variant of the LQR strategies such as the LQR and LQIR. Moreover, the box plot of all controllers is plotted in Figure 14 for the cost function, introduced in equation (23). The median of Root Mean Square Error (RMSE) is shown in the crossline in the box plot. These results indicate that the proposed controller is able to provide rapid convergence and excellent transient response relative to other controllers for attitude control of the experimental platform.

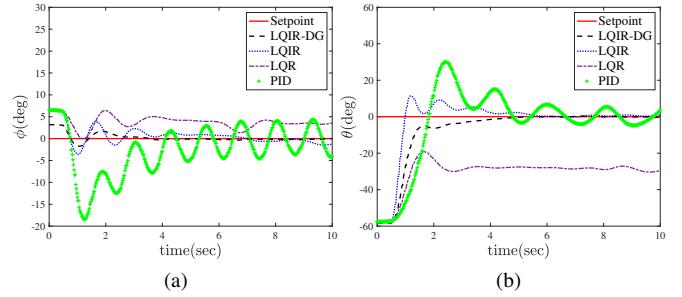


Fig. 13: Comparison of LQIR-DG structure with the variant of LQR and PID in regulation problem: (a) roll angle (b) pitch angle.

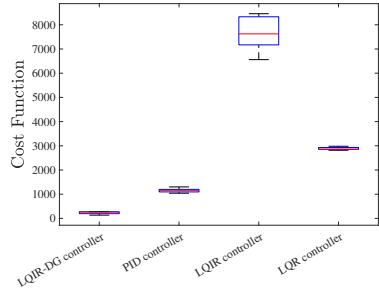


Fig. 14: Box plot of LQIR-DG, LQR, LQIR, and PID controllers.

VI. CONCLUSION

In this paper, the linear quadratic integral differential game approach, was used in real-time for attitude control of the platform quadrotor. For the implementation of the controller structure, an accurate dynamic model was considered for the experimental platform. Then, the model parameters were identified using the NSL method. For evaluation of the proposed method, the regulation and tracking proposed were successfully performed. Moreover, the ability of the proposed method was investigated in the rejection of the input disturbance and modeling error in the experimental platform. Finally, a comparison was also performed between the results of classical PID, LQR, and LQIR with the proposed method. The implementation results illustrated the excellent performance of the LQIR controller based on the game theory approach in attitude control for the quadrotor platform.

REFERENCES

- [1] A. Abdul Salam and I. Ibraheem, "Nonlinear pid controller design for a 6-dof uav quadrotor system," *Engineering Science and Technology, an International Journal*, vol. 22, 03 2019.
- [2] H. Bolandi, M. Rezaei, R. Mohsenipour, H. Nemati, and S. Smailzadeh, "Attitude control of a quadrotor with optimized pid controller," *Intelligent Control and Automation*, vol. 04, pp. 342–349, 01 2013.
- [3] K. Chara, A. Yassine, F. Srairi, and K. Mokhtari, "A robust synergetic controller for quadrotor obstacle avoidance using bézier curve versus b-spline trajectory generation," *Intelligent Service Robotics*, vol. 15, 03 2022.
- [4] A. Aboudonia, A. El-Badawy, and R. Rashad, "Disturbance observer-based feedback linearization control of an unmanned quadrotor helicopter," *Proceedings of the Institution of Mechanical Engineers Part I Journal of Systems and Control Engineering*, vol. 230, 07 2016.
- [5] H. Wang and M. Chen, "Sliding mode attitude control for a quadrotor micro unmanned aircraft vehicle using disturbance observer," in *Proceedings of 2014 IEEE Chinese Guidance, Navigation and Control Conference*, 2014.
- [6] X. Lin, J. Liu, Y. Yu, and C. Sun, "Event-triggered reinforcement learning control for the quadrotor uav with actuator saturation," *Neurocomputing*, vol. 415, pp. 135–145, 2020. [Online]. Available: <https://www.sciencedirect.com/science/article/pii/S0925231220311504>
- [7] H. A. Foudeh, P. Luk, and J. Whidborne, "Application of norm optimal iterative learning control to quadrotor unmanned aerial vehicle for monitoring overhead power system," *Energies*, vol. 13, no. 12, 2020. [Online]. Available: <https://www.mdpi.com/1996-1073/13/12/3223>
- [8] S. Chen, Y. Li, Y. Lou, K. Lin, and X. Wu, "Learning real-time dynamic responsive gap-traversing policy for quadrotors with safety-aware exploration," *IEEE Transactions on Intelligent Vehicles*, pp. 1–14, 2022.
- [9] H. S. Kim, K. Lee, and Y. H. Joo, "Decentralized sampled-data fuzzy controller design for a vtol uav," *Journal of the Franklin Institute*, vol. 358, no. 3, pp. 1888–1914, 2021. [Online]. Available: <https://www.sciencedirect.com/science/article/pii/S0016003220308383>
- [10] A. Zulu and S. John, "A review of control algorithms for autonomous quadrotors," *Open Journal of Applied Sciences*, vol. 04, pp. 547–556, 01 2014.
- [11] F. Ahmad, P. Kumar, A. Bhandari, and P. P. Patil, "Simulation of the quadcopter dynamics with lqr based control," *Materials Today: Proceedings*, vol. 24, pp. 326–332, 2020, international Conference on Advances in Materials and Manufacturing Applications, IConAMMA 2018, 16th -18th August, 2018, India. [Online]. Available: <https://www.sciencedirect.com/science/article/pii/S2214785320329047>
- [12] B. Anjali, V. A., and N. J L, "Simulation and analysis of integral lqr controller for inner control loop design of a fixed wing micro aerial vehicle (mav)," *Procedia Technology*, vol. 25, pp. 76–83, 12 2016.
- [13] R. Xue, L. Dai, D. Huo, H. Xie, Z. Sun, and Y. Xia, "Compound tracking control based on mpc for quadrotors with disturbances," *Journal of the Franklin Institute*, vol. 359, no. 15, pp. 7992–8013, 2022. [Online]. Available: <https://www.sciencedirect.com/science/article/pii/S0016003222005336>
- [14] H. Nobahari, A. Baniasad, and A. Sharifi, "Linear quadratic integral differential game applied to the real-time control of a quadrotor experimental setup," in *2022 10th RSI International Conference on Robotics and Mechatronics (ICRoM)*, 2022, pp. 578–583.
- [15] V. Y. Glizer and V. Turetsky, "Linear-quadratic pursuit-evasion game with zero-order players dynamics and terminal constraint for the evader," *IFAC-PapersOnLine*, vol. 48, no. 25, pp. 22–27, 2015, 16th IFAC Workshop on Control Applications of Optimization CAO 2015. [Online]. Available: <https://www.sciencedirect.com/science/article/pii/S2405896315023113>
- [16] Z. Zwierzewicz, "On the ship course-keeping control system design by using robust and adaptive control," in *2014 19th International Conference on Methods and Models in Automation and Robotics (MMAR)*, 2014, pp. 189–194.
- [17] S. Bouabdallah and R. Siegwart, "Full control of a quadrotor," in *2007 IEEE/RSJ International Conference on Intelligent Robots and Systems*, 2007, pp. 153–158.
- [18] S. Bouabdallah, "Design and control of quadrotors with application to autonomous flying," 01 2007.
- [19] J. Engwerda, "Linear quadratic games: An overview," *Macroeconomics, WorkingPaper*, 2006, subsequently published in *Advances in Dynamic Games and their Applications (book)*, 2009 Pagination: 32.



Hadi Nobahari received his B.S., M.S. and Ph.D. in aerospace engineering, flight dynamic and control division, from Sharif University of Technology in 1998, 2000 and 2006, respectively. He is an Associate Professor in the Department of Aerospace Engineering of Sharif University of Technology. His research interests include novel applications of heuristic algorithms in aerospace, intelligent guidance and control systems, cooperative guidance and navigation systems.



Alireza Sharifi was born in Iran in 1988. He received his M.S. degree in aerospace engineering from Sharif University of Technology, Tehran, Iran, in 2013. He is currently a Ph.D. candidate majored in aerospace engineering in Sharif University of Technology. His research interests are artificial intelligence, swarm filter and nonlinear estimation.operative guidance and navigation systems.



Ali Baniasad always say khairol omor ouseteha.



Alireza Sharifi