



NTNU – Trondheim  
Norwegian University of  
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## TTT4120 Digital Signal Processing Fall 2017

### Lecture: Laplace - Review

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## Contents and learning outcomes\*

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- Definition of Laplace transform and its existence
- Some properties of the Laplace transform
- System analysis and rational expressions

\* For more details, read the document [laplace\\_transform\\_final.pdf](#) which is available on ItsLearning

## Motivation

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- Linear time-invariant system:

$$\begin{array}{ccc}
 x(t) & \longrightarrow & \boxed{h(t)} \longrightarrow y(t) = h(t) * x(t) \\
 X(\Omega) & & Y(\Omega) = H(\Omega)X(\Omega)
 \end{array}$$

- What if  $h(t) = e^t u(t)$ ?
  - System is unstable,  $\int |h(t)| dt$  not finite
  - Fourier transform of  $h(t)$  does not exist
- Can we analyze such systems using a transform method while retaining the good properties of the Fourier transform?

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## Motivation...

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- Consider the continuous-time Fourier transform pair

$$X(\Omega) = \int_{-\infty}^{\infty} x(t) e^{-j\Omega t} dt$$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\Omega) e^{j\Omega t} d\Omega$$

- What if  $x(t) = e^t u(t)$ ?
  - Fourier transform of  $x(t)$  does not exist
- Can we analyze such signals using a transform method while retaining the good properties of the Fourier transform?
- Very similar to our development of the z-transform!

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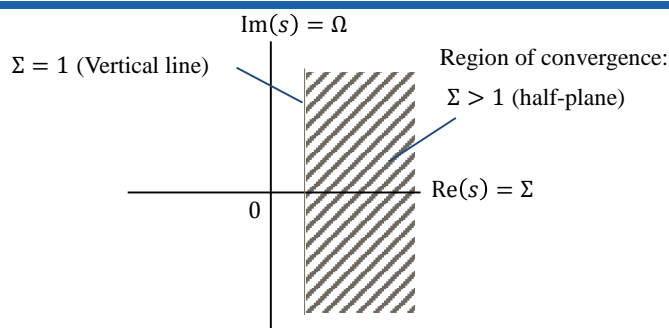
## Basic idea

- Capture the source of instability or inapplicability of the FT
- Apply the FT to the modified (captured) signal
- Example: Suppose we have signal  $x(t) = e^t u(t)$ 
  - Problem is due to the exponential growth
  - Capture the signal by multiplying it by a decaying exponential **stronger** than the growing one, i.e.,  $e^{-\Sigma t} x(t)$ ,  $\Sigma > 0$
  - What values of  $\Sigma$  allows for a FT of  $e^{-\Sigma t} x(t)$ ?

$$\begin{aligned} \int_{-\infty}^{\infty} x(t) e^{-\Sigma t} e^{-j\Omega t} dt &= \int_0^{\infty} e^t e^{-\overbrace{(\Sigma + j\Omega)}^s} t dt = \\ &= \frac{e^{(1-s)t}}{(1-s)} \Big|_{t=0}^{\infty} = \frac{1}{s-1}, \text{ for } \Sigma > 1 \end{aligned}$$

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## Basic idea...



- Convergence has only to do with  $\Sigma$  and not  $\Omega$
- We have more general transform of the signal  $x(t)$

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## Basic idea...

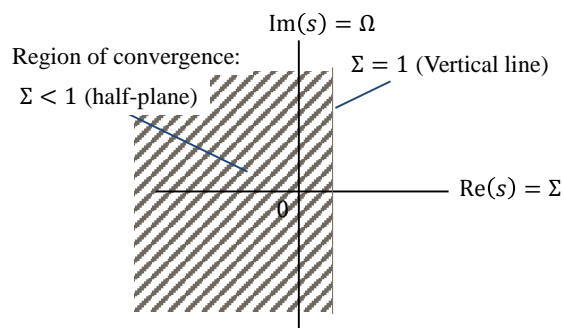
- Example: Suppose we have signal  $x(t) = -e^t u(-t)$ 
  - What values of  $\Sigma$  allows for a FT for  $e^{-\Sigma t} x(t)$ ?

$$\begin{aligned} \int_{-\infty}^{\infty} x(t) e^{-\Sigma t} e^{-j\Omega t} dt &= \int_{-\infty}^0 -e^t e^{-\overbrace{(\Sigma + j\Omega)}^s} t dt = \\ &= \left. \frac{-e^{(1-s)t}}{(1-s)} \right|_{t=-\infty}^0 = \frac{1}{s-1}, \text{ for } \Sigma < 1 \end{aligned}$$

- Same expression as before but different ROC

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## Basic idea...



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## Definition of Laplace transform

- The Laplace transform of a continuous-time signal  $x(t)$  is

$$X(s) = \mathcal{L}\{x(t)\} = \int_{-\infty}^{\infty} x(t)e^{-st} dt$$

- Notation:  $x(t) \xleftrightarrow{\mathcal{L}} X(s)$   $x(t) = \mathcal{L}^{-1}\{X(s)\}$
- Transforms  $x(t)$  into its complex-plane representation  $X(s)$
- Transform only exists whenever integral exists
- Region of convergence (ROC) of  $X(s)$  is the set of all values of  $s = \Sigma + j\Omega$  for which  $X(s)$  attains a finite value

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## Most important properties

- Linearity:

$$x_3(t) = a_1x_1(t) + a_2x_2(t) \xleftrightarrow{\mathcal{L}} X_3(s) = a_1X_1(s) + a_2X_2(s)$$

ROC of  $X_3(s)$  at least  $\mathcal{R}_{X_1} \cap \mathcal{R}_{X_2}$  **but can extend beyond** intersection

- Differentiation:

$$\frac{d^k x(t)}{dt^k} \xleftrightarrow{\mathcal{L}} s^k X(s) \text{ (for zero initial conditions)}$$

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## Comparison with z-transform

- ROC in z-transforms is a ring or a disc centered at zero
- ROC in Laplace transforms is a half plane or a strip parallel to the  $j\Omega$ -axis in the  $s$ -plane

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## System analysis

- Systems described by constant-coefficient differential equations

$$\begin{array}{ccc}
 x(t) & \longrightarrow & \boxed{h[n]} \longrightarrow \sum_{k=0}^N a_k \frac{d^k y(t)}{dt^k} = \sum_{k=0}^M b_k \frac{d^k x(t)}{dt^k} \\
 X(s) & & Y(s) = H(s)X(s)
 \end{array}$$

- Rational system function

$$H(s) = \frac{\sum_{k=0}^M b_k s^k}{\sum_{k=0}^N a_k s^k} = C \frac{\prod_{k=0}^M (s - z_k)}{\prod_{k=0}^N (s - p_k)}$$

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## System analysis

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- Frequency response:  $H(\Omega) = H(s)|_{s=j\Omega}$
- Stable system: imaginary axis part of region of convergence
- Causal system:
  - ROC:  $\text{Re}(s) > \max_k \text{Re}(p_k)$
  - Stable if all poles in left half plane
- Anti-causal system (nonrealizable):
  - ROC:  $\text{Re}(s) < \min_k \text{Re}(p_k)$
  - Stable if all poles in right half plane

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## Summary

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- Take-home messages
  - Concepts behind Laplace- and z-transforms similar
  - If you understand one, the other comes almost for free
- Reason for introducing Laplace in TTT4120 is its importance in IIR filter design

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