CENG 384 - Signals and Systems for Computer Engineers Spring 2018-2019

Written Assignment 4

Sen, Ali e2264661@ceng.metu.edu.tr Sahin, Ismail e2264653@ceng.metu.edu.tr

June 2, 2019

- 1. (a) y[n](-3/4)y[n-1] + (1/8)y[n-2] = 2x[n]
 - (b) $H(e^{jw}) = Y(e^{jw})/X(e^{jw})$ applying the fourier transform to both sides of difference equation by using the linearity and time-shifting properties, we obtain the expression

$$\begin{split} H(e^{jw}) &= Y(e^{jw})/X(e^{jw}) = 2/(1-(3/4)e^{-jw}+(1/8)e^{-2jw}) \\ H(e^{jw}) &= 2/((1-(1/2)e^{-jw})(1-(1/4)e^{-jw})) = A/(1-(1/2)e^{-jw}) + B/(1-(1/4)e^{-jw}) \ A+B=2\\ (-A/4)-(B/2) &= 0\\ A=4\\ B=-2\\ H(e^{jw}) &= 4/(1-(1/2)e^{-jw}) - 2/(1-(1/4)e^{-jw}) \end{split}$$

- (c) The impulse response of this system can be found from FT table by using frequency response $h[n] = (4(1/2)^n 2(1/4)^n)u[n]$
- (d) $x[n] = (1/4)^n u[n]$ $X[e^{jw}] = 1/(1 - (1/4)e^{-jw})$ $Y(e^{jw}) = H(e^{jw})X(e^{jw}) = 2/((1 - (1/2)e^{-jw})(1 - (1/4)e^{-jw}))1/(1 - (1/4)e^{-jw})$ $= 2/((1 - (1/2)e^{-jw})(1 - (1/4)e^{-jw})^2)$ $Y(e^{jw}) = A/(1 - (1/4)e^{-jw}) + B/((1 - (1/4)e^{-jw})^2) + C/(1 - (1/2)e^{-jw})$ from this equation we find A = -4 B = -2 C = 8 $Y(e^{jw}) = -4/(1 - (1/4)e^{-jw}) - 2/((1 - (1/4)e^{-jw})^2) + 8/(1 - (1/2)e^{-jw})$ and from FT table $y[n] = (-4(1/4)^n - 2(n+1)(1/4)^n + 8(1/2)^n)u[n]$
- 2. $h_1[n] = (1/3)^n u[n]$ $H_1[e^{jw}] = 1/(1 - (1/3)e^{-jw}) = 3/(3 - e^{-jw})$ H_2 can be found from H-H₁ $H_2(e^{jw}) = (-12 + 5e^{-jw})/(12 - 7e^{jw} + e^{-2jw}) + 3/(e^{-jw} - 3)$ we can rewrite $(12 - 7e^{jw} + e^{-2jw})$ as $(e^{-jw} - 4)(e^{-jw} - 3)$ and when we multiply 3 with $e^{-jw} - 4$ we can make the summation

$$\begin{split} &H_2(e^{jw}) = (-24 + 8e^{-jw})/(e^{-jw} - 4)(e^{-jw} - 3) \\ &= A/(e^{-jw} - 4) + B/(e^{-jw} - 3) \\ &A + B = 8 \\ &-4B - 3A = -24 \\ &B = 0 \\ &A = 8 \\ &H_2(e^{jw}) = 8/(e^{-jw} - 4) = 2 * (1/1 - (1/4)e^{-jw}) \\ &\text{and from FT table we get} \\ &h_2[n] = -2(1/4)^n u[n] \end{split}$$

3. (a) $f_1 = 2\pi/2\pi$ Hz $f_2 = 3\pi/2\pi$ Hz $f_1 = n_1 * f_0$

$$\begin{split} f_2 &= n_2 * f_0 \\ f_1/n1 &= f_0 = f_2/n_2 \\ f_1/f_2 &= n_1/n_2 \\ 1/(3/2) &= 2\pi/3\pi \\ f_0 &= 2/3 \\ T_0 &= 3/2 \\ w &= 2\pi/T_0 \\ w &= 4\pi/3 \\ W &> |w| \\ \text{From table 4.2 in textbook we know the fourier transform of x(t) is} \\ 1 + \pi[\delta(w-3\pi) - \delta(w+3\pi)] \end{split}$$

(b) The Nyquist rate will be equal to twice the highest frequency in the signal

so $w_N = 2*w_{max}$ $w_N = 2*3\pi$ $w_N = 6\pi$ and period for sampling is $w_{max} = 2\pi/T$ $T = 2\pi/w_{max}$ T = 2/3

(c) $x_p(t) = \sum_{-\infty}^{\infty} x(kT)\delta(t-kT)$ From the multiplication property, we know that $x_p(jw) = (1/2\pi) \int_{-\infty}^{\infty} X(j\theta)P(j(w-\theta))d\theta$ $P(jw) = (2\pi/T) \sum_{-\infty}^{\infty} \delta(w-kw_0)$ $X_p(jw) = (1/T) \sum_{-\infty}^{\infty} X(j(w-kw_s))$ $w_s = 2\pi/(2/3)$ $= (2/3) \sum_{-\infty}^{\infty} X(j(4\pi/3 - k3\pi))$

4. (a)
$$T = \frac{2\pi}{\omega_s} = 2$$

$$X_p(j\omega) = \frac{1}{T} \sum_{\forall k} X(j(\omega - k\omega_s))$$

$$X_d(e^{j\omega}) = X_p(j\frac{\omega}{T})$$

$$X_d(e^{j\omega}) = \begin{cases} \frac{2\pi}{\pi}\omega & \text{if } |x| \leq \frac{\pi}{2} \\ 0 & \text{otherwise} \end{cases}, X_d(e^{j\omega}) = X_d(e^{j(\omega + T)})$$

(b) From discrete time Fourier transform table, we know that

$$\begin{split} e^{j\omega_0n} &\longleftrightarrow 2\pi \sum_{\forall k} \delta(\omega - \omega_0 - 2\pi k) \\ h[n] &= cos\pi n = \frac{1}{2} (e^{j\pi n} + e^{-j\pi n}) \\ \text{Then, } H(e^{j\omega}) &= \frac{1}{2} (2\pi \sum_{\forall k} \delta(\omega - \pi - 2\pi k) + 2\pi \sum_{\forall k} \delta(\omega + \pi - 2\pi k)) \\ &= \pi (\sum_{\forall k} \delta(\omega - \pi - 2\pi k) + \delta(\omega + \pi - 2\pi k)) \end{split}$$

(c)
$$Y_d(e^{jw}) = X_d(e^{jw}) * H_d(e^{jw})$$

 $H_d(e^{jw}) = j(w/T), \quad w < \pi$