

Kuwait University

College of Engineering and Petroleum



جامعة الكويت
KUWAIT UNIVERSITY

ME417 CONTROL OF MECHANICAL SYSTEMS

PART II: CONTROLLER DESIGN USING ROOT-LOCUS

LECTURE 1: INTRODUCTION TO ROOT-LOCUS

Summer 2020

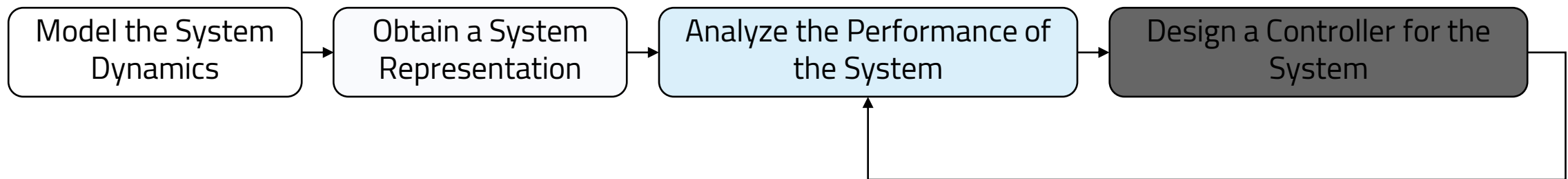
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Lecture Plan

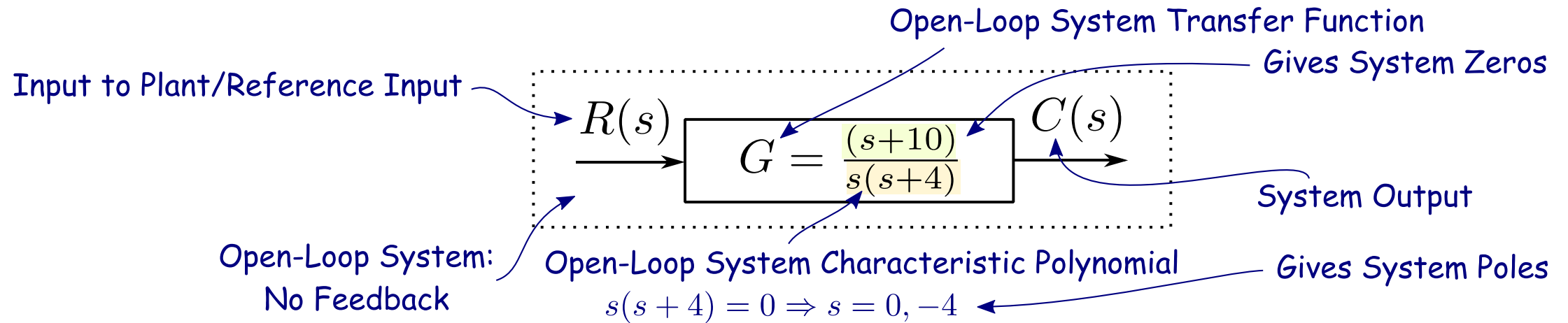
- Objectives:
 - Review the anatomy of a control system block diagram
 - Introduction to the concept of root-locus diagrams
 - Overview the properties of the root-locus
- Reading:
 - Nise: 8.1-8.3
- Practice problems are more applicable after the subsequent lectures.

Where we are

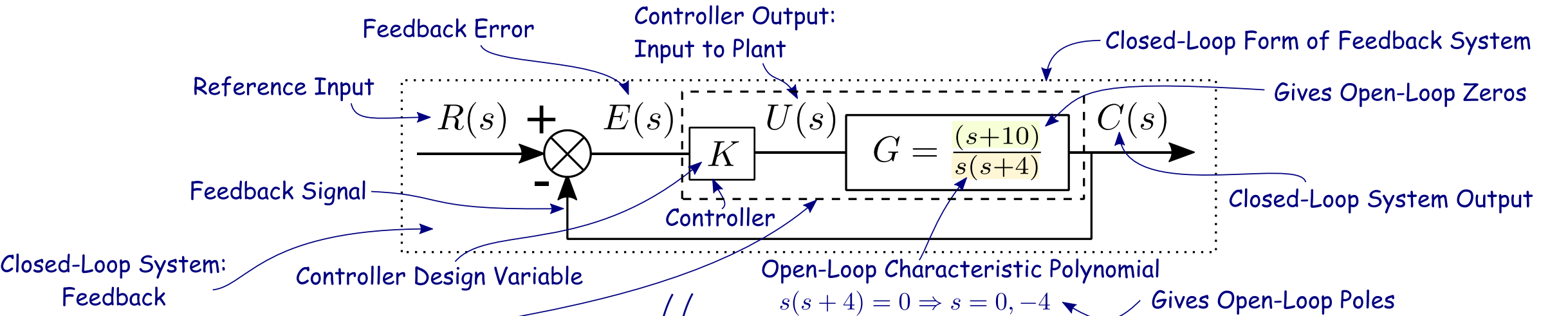
- We will now begin to introduce the first main technique of designing a controller.
- The design technique we will learn in this part is a graphical technique, it offers an alternative and qualitative method to understand the behavior of a dynamic system.
- It is also considered a way to represent the system and a method to analyze the performance (inherent to the design intent).



Review of Control System Block Diagram Anatomy: Open-Loop System

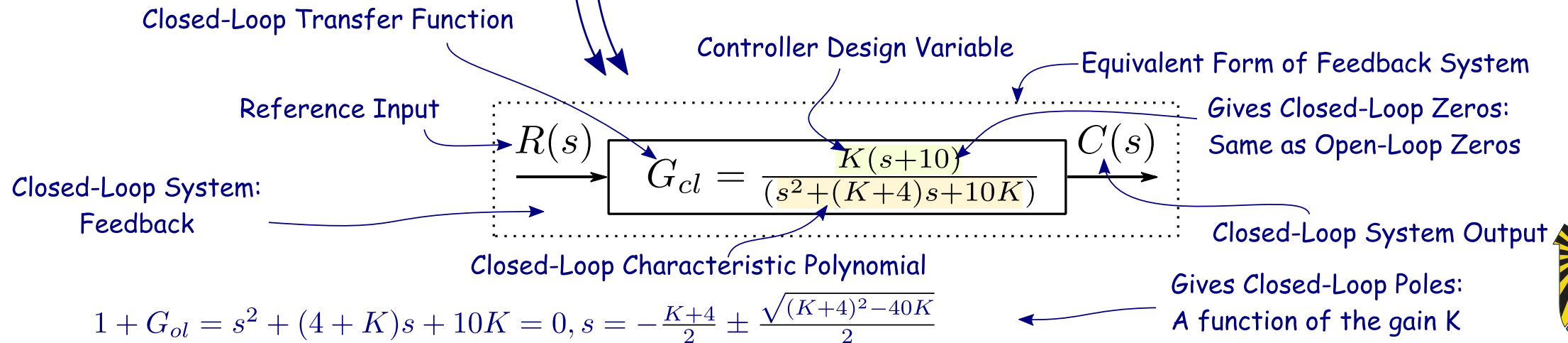


Review of Control System Block Diagram Anatomy: Unity Closed-Loop System



Open-Loop Transfer Function
a.k.a Forward Transfer Function: $G_{ol} = K \frac{(s+10)}{s(s+4)}$

via Block Diagram Reduction we convert the Closed-Loop form of a Feedback System, to an equivalent Open-Loop form



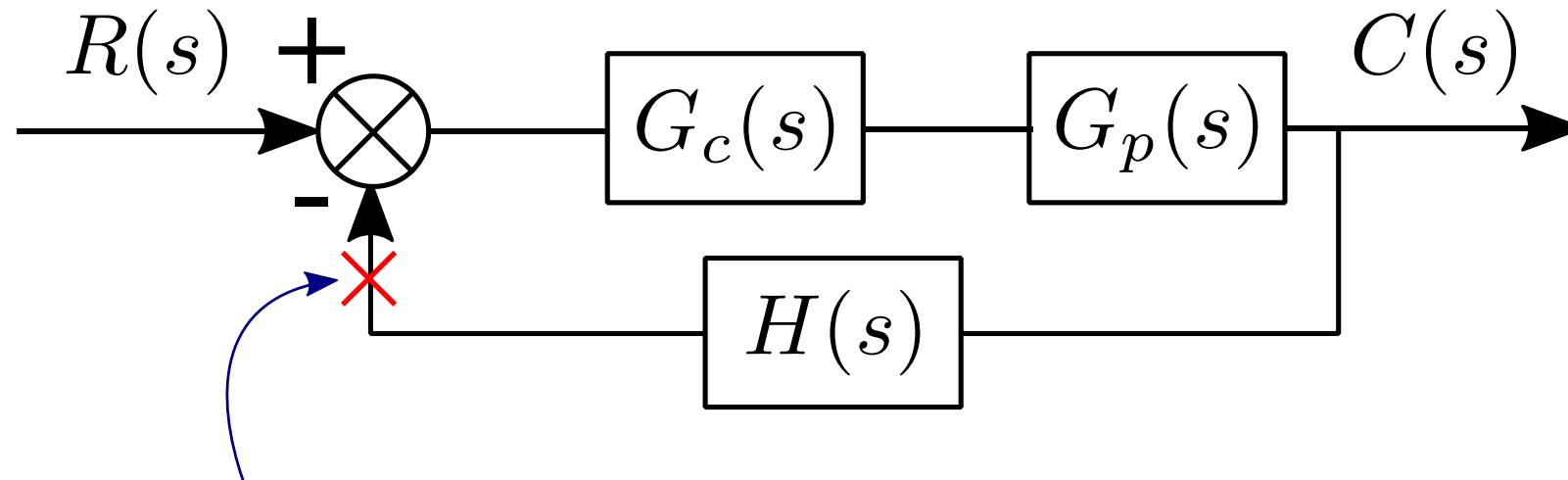
$$1 + G_{ol} = s^2 + (4 + K)s + 10K = 0, s = -\frac{K+4}{2} \pm \frac{\sqrt{(K+4)^2 - 40K}}{2}$$



Open-Loop Transfer Function of a Non-Unity Feedback System

- The open-loop transfer function of a **any** feedback system is obtained by terminating the feedback signal at just before the summation block. And multiplying all the blocks in series up to the termination point.
- For a Non-Unity Feedback System, the open-loop transfer function is:

$$G_{ol} = G_C(s)G_P(s)H(s)$$

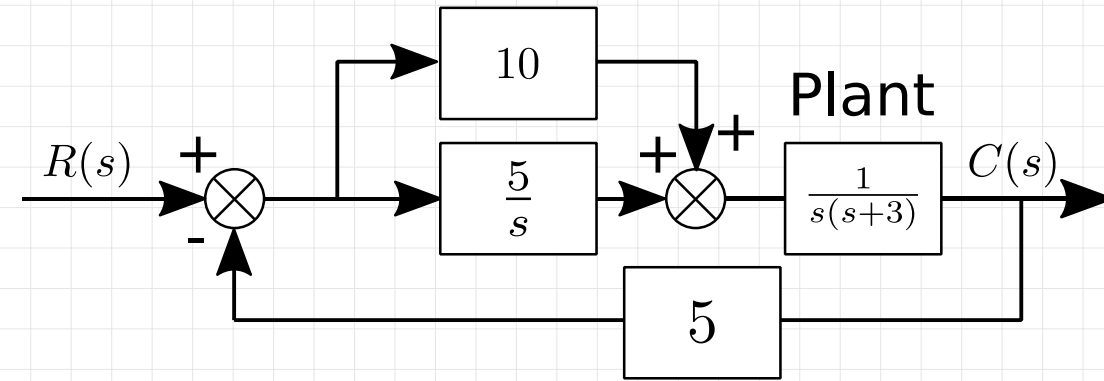


Terminate the signal here to "open the loop"



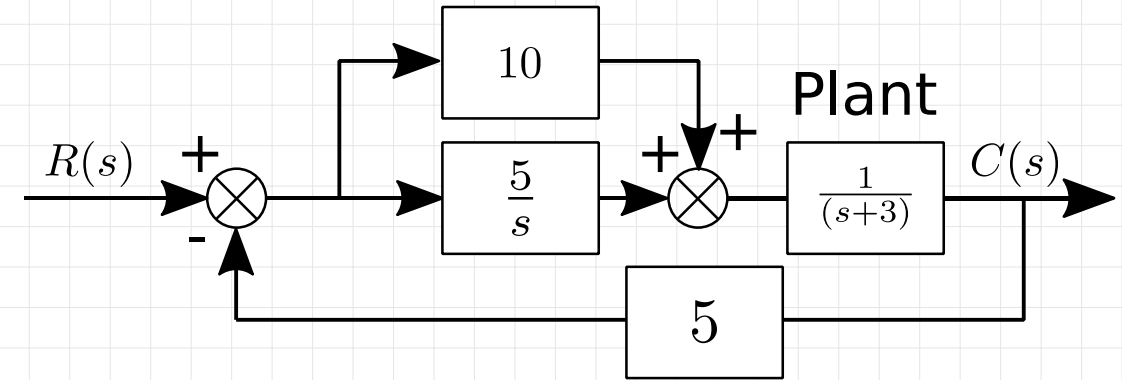
For the following unity feedback system. Derive the following:

- The plant transfer function
- The controller transfer function
- The open-loop transfer function without feedback
 - (The open-loop **system** T.F.)
- The open-loop transfer function with feedback
- The forward transfer function
- The closed-loop transfer function
- The closed-loop characteristic polynomial
- The input to the plant in terms of the reference.
- The controller design variables



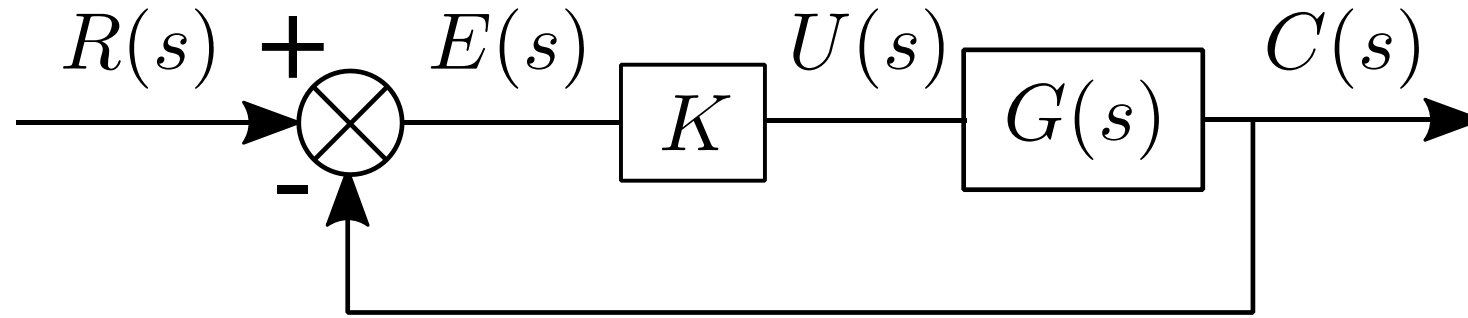
For the following unity feedback system, place the:

- Plant poles and zeros
- Open-loop poles and zeros
- Closed-loop poles and zeros

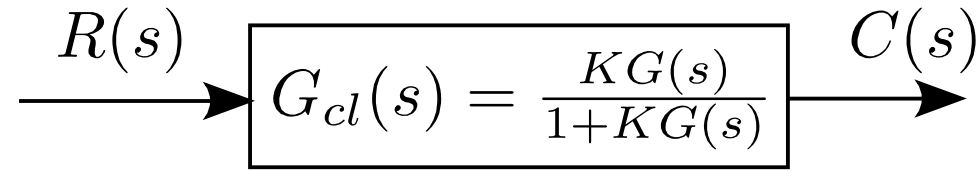


What is a Root-Locus

- Given a feedback control system of the form



- With the equivalent open-loop **form**:

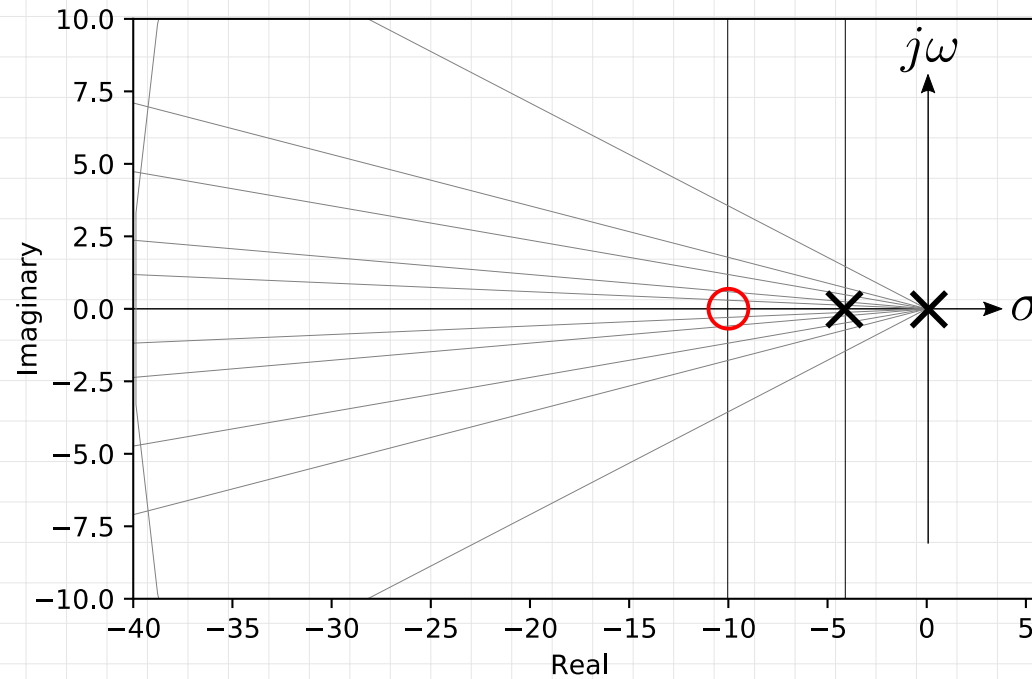


- The root-locus is the locus of the roots of the characteristic polynomial of the closed-loop transfer function: $1 + KG_{ol} = 0$, on the s -plane, as K goes from 0 to ∞



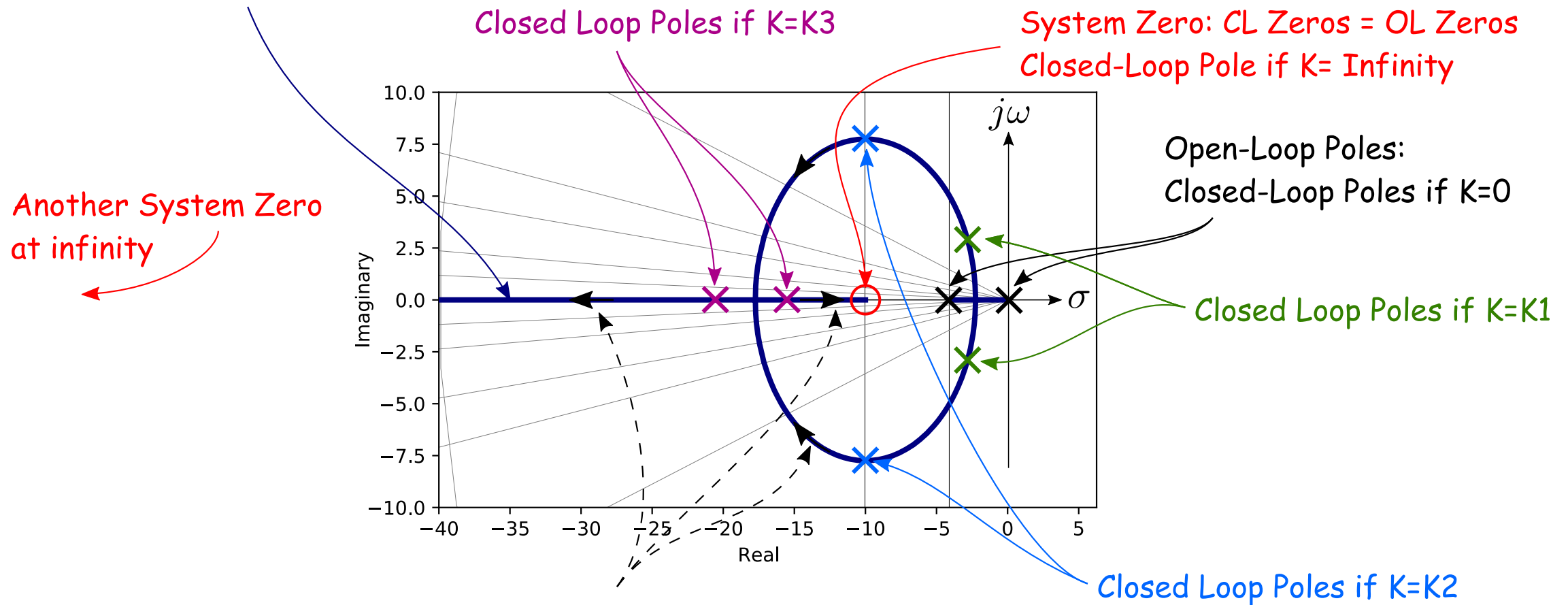
Closed-Loop System Representation on the S-Plane

$$G_{ol} = K \frac{(s+10)}{s(s+4)}$$



Closed-Loop System Representation on the S-Plane

The Locus of the Closed-Loop Poles as K goes from 0 to infinity (Root-Locus)



Closed-Loop Pole Location Motion as K increases:

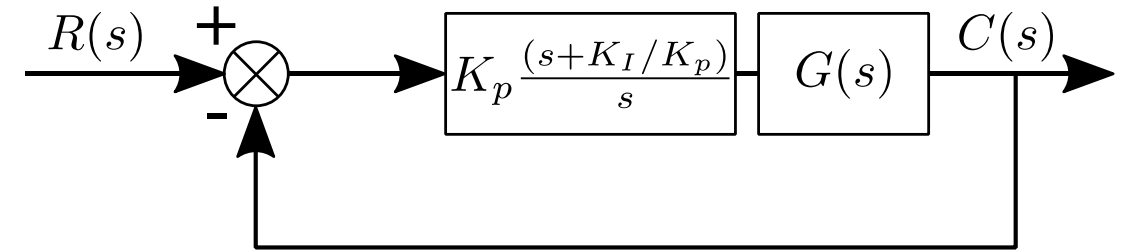
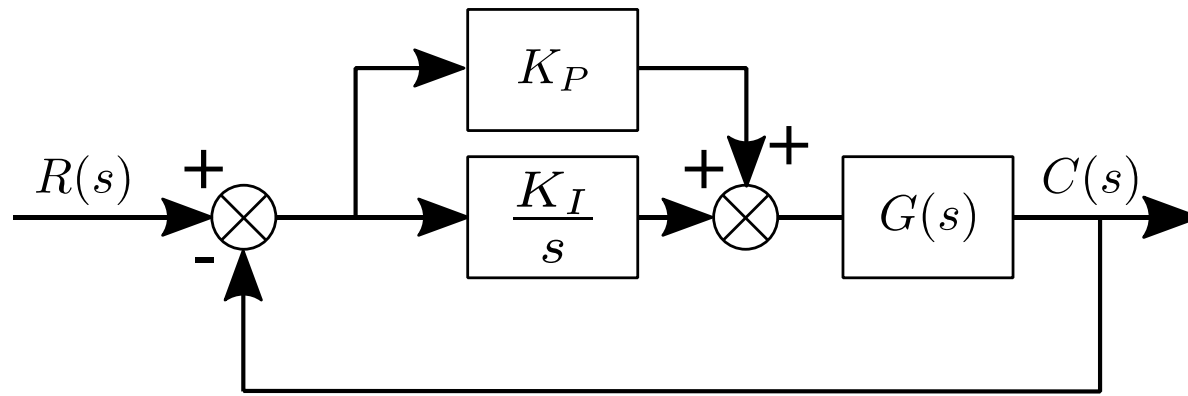
As K increases, CL Poles move away from OL poles and toward system zeros:

Here: $K_1 < K_2 < K_3$

$$G_{ol} = K \frac{(s+10)}{s(s+4)} \quad G_{cl} = \frac{K(s+10)}{(s^2 + (K+4)s + 10K)}$$

What if we have more than just one gain?

- The root locus is drawn for a single variable K , what if there is more than one gain in the feedback loop, as is the case when applying a PID controller?
- Take the **PI** Controller:



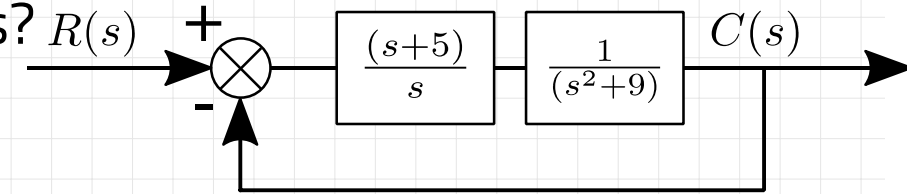
- We can factor out the proportional gain K_p as the root locus variable and fix the ratio K_I/K_p . The ratio $K_I/K_p = z_1$, determines the location of the added zero by the PI controller. The root-locus will become a variable of K_p
- Char. Poly.: $1 + K_p \frac{(s+z_1)G(s)}{s} = 0$



We wish to apply a PI feedback controller on a dc motor to control its speed. Find the values of the gain K_I , if $K_p = 1$, to achieve a settling time of 0.25s, when a step input of 10rad/s is applied.

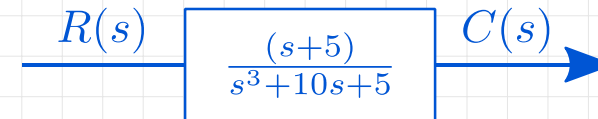


Given the following unity feedback system, determine the location of the open-loop poles, then compute the characteristic polynomial for the closed-loop system. How would you find the poles of the closed-loop sys?



The open-loop transfer function is: $G_{ol} = G_c G_p = \frac{(s+5)}{s(s^2+9)}$

The poles of the open loop transfer function are the roots of the open-loop characteristic polynomial:
 $s(s^2 + 9) = 0 \Rightarrow s = 0, \pm 3j$



The closed-loop transfer function is: $G_{cl} = \frac{G_c G_p}{1 + G_c G_p}$

The characteristic polynomial of the closed-loop system: $1 + G_c G_p = 0$

The poles of the closed-loop system are the roots of the closed-loop characteristic polynomial:

$$1 + \frac{(s+5)}{s(s^2+9)} = s^3 + 10s + 5 = 0$$

In order to find the poles of the closed loop system we need to factor a 3rd degree polynomial. This becomes hard, but fortunately, we can use the root-locus technique to treat such case.



Why design via Root-Locus

- What if we want to:
 1. Observe the effect of changing gain parameters on the system response
 - *Where would the poles of the closed-loop system be as we change the gain K*
 2. Observe the effect of adding dynamic compensation to the closed loop system
 - *What happens when we use a controller that adds poles and zeros to the closed-loop system?*
- Example: PI Controller, $G_c = K_p \frac{s+K_I/K_p}{s}$, adds a zero and a pole*
- 3. Examine the sensitivity and stability of a closed loop system
 - *How close are the closed-loop poles to stability?*
- 4. Design controllers for higher order systems
 - *It is hard to factor roots for polynomials of 3rd and higher order.*
- A graphical controller design **technique**, such as the root-locus, can help us



Complex Numbers and Vector Representation

- The Laplace Function $F(s)$, is a function of the complex variable s , but how do we evaluate the function at any s ?

- If we have the function in factored form

$$F(s) = \frac{\prod_{i=1}^m (s + z_i)}{\prod_{j=1}^n (s + p_j)} = \frac{(s + z_1)(s + z_2) \dots (s + z_m)}{(s + p_1)(s + p_2) \dots (s + p_n)}$$

Where \prod denotes product, m the number of zeros, n the number of poles

- The solution: $F(s) = M \angle \theta$

$$M = \frac{\prod(\text{zero vector lengths})}{\prod(\text{pole vector lengths})}$$

$$\theta = \sum \text{zero angles} - \sum \text{pole angles} = \sum_{i=1}^m \angle(s + z_i) - \sum_{j=1}^n \angle(s + p_j)$$



Evaluate, vectorially, the function $F(s) = \frac{(s+1)}{s(s+2)}$ at $s = -3 + j4$

To solve for this vectorially, we find the lengths and angles of all poles and zeros then add them

Zero: Length = $\sqrt{(-3 - (-1))^2 + (4 - 0)^2} = \sqrt{20}$, angle = $\tan^{-1}\left(\frac{4-0}{-3+1}\right) = 116^\circ$

Pole at 0: Length = $\sqrt{(-3 - (0))^2 + (4 - 0)^2} = \sqrt{25} = 5$, angle = $\tan^{-1}\left(\frac{4-0}{-3-0}\right) = 127^\circ$

Pole at -2: Length = $\sqrt{(-3 - (-2))^2 + (4 - 0)^2} = \sqrt{17}$, angle = $\tan^{-1}\left(\frac{4-0}{-3+2}\right) = 104^\circ$

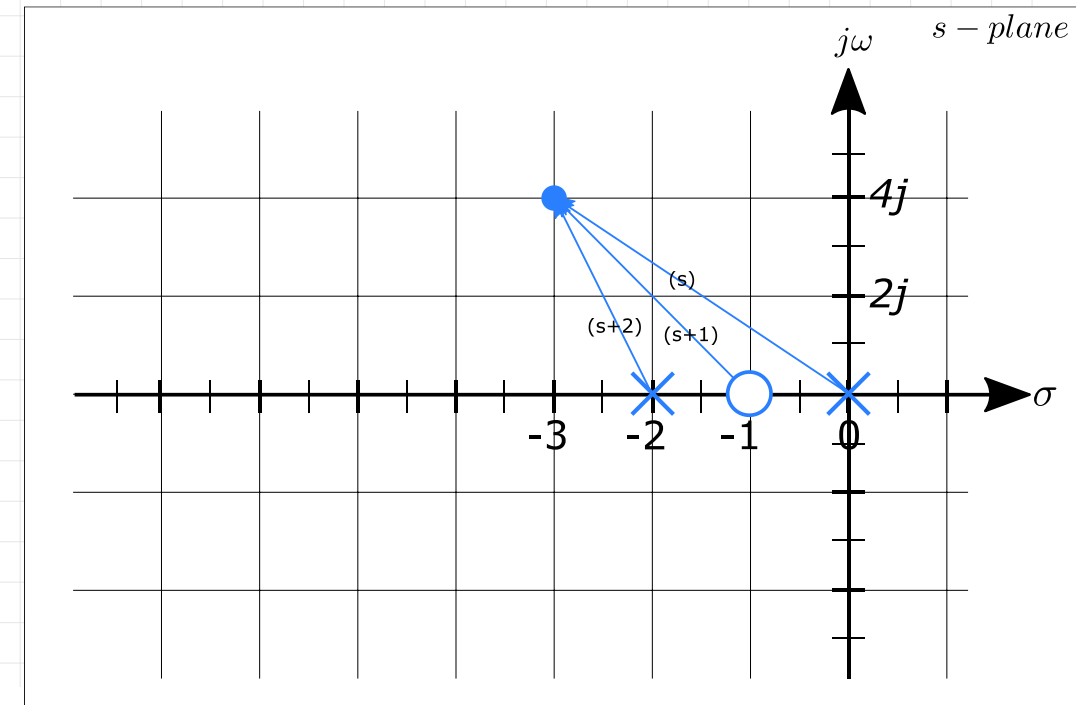
$$F(s = -3 + j4) = M\angle\theta = \frac{\prod(\text{zero vector lengths})}{\prod(\text{pole vector lengths})} \angle \sum \text{zero angles} - \sum \text{pole angles}$$

$$F(s = -3 + j4) = \left(\frac{\sqrt{20}}{5\sqrt{17}}\right) \angle 116^\circ - 127^\circ - 104^\circ = -114^\circ$$

$$F(s = -3 + j4) = 0.217 \angle -114^\circ$$

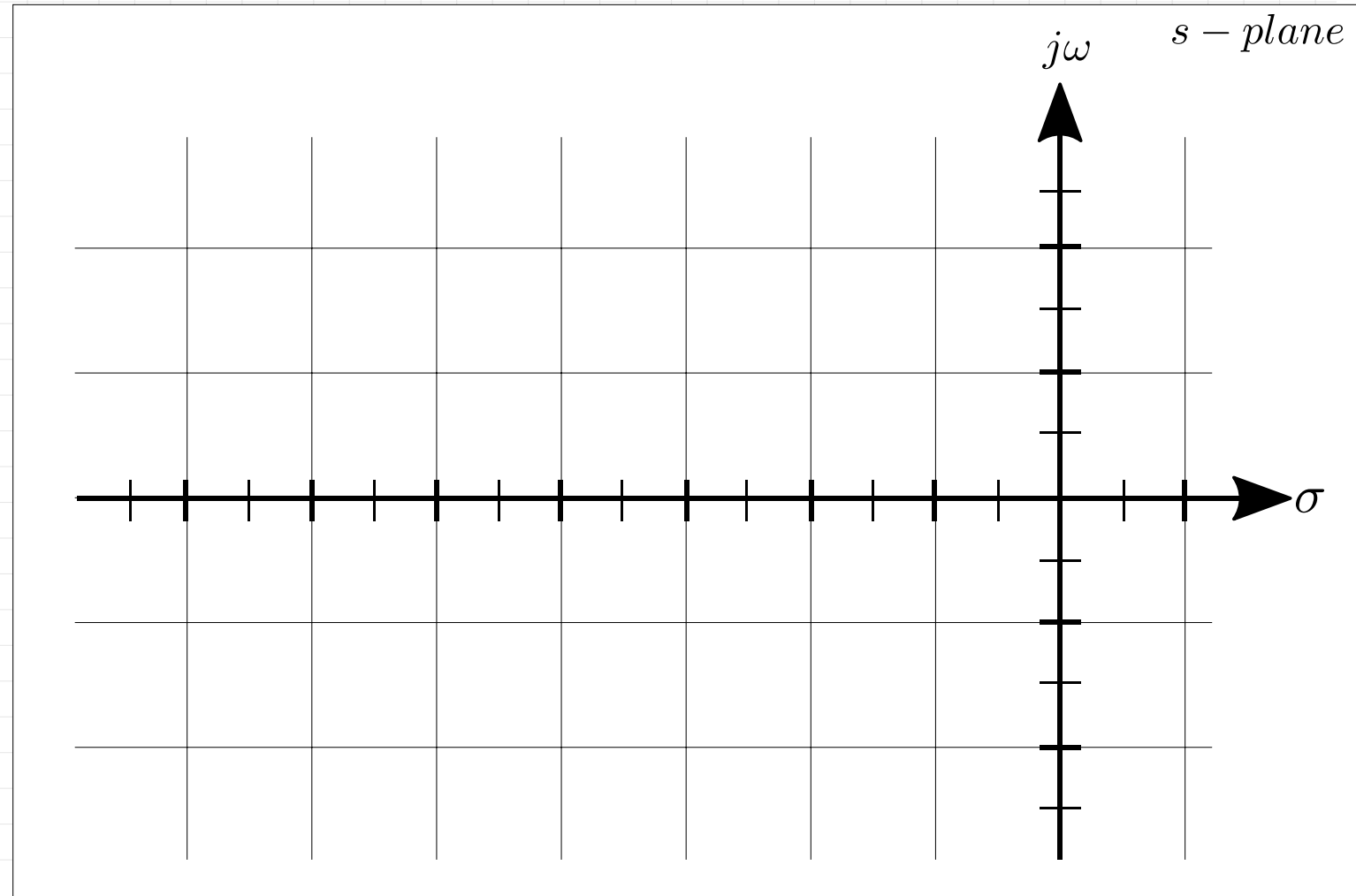
Verify in MATLAB using `evalfr()`

$$F = -0.089 - .197j$$



Evaluate, vectorially, the function $F(s) = \frac{(s+3)(s+1)}{s(s+2)}$ at $s = -2 + j$

Algebraically: $F = 0.4 - 0.8j$



Properties of the Root-Locus

- Given the closed-loop transfer function
 - $G_{cl} = \frac{KG(s)}{1+KG(s)H(s)}$: non-unity feedback
- A pole, s , exists when the characteristic polynomial becomes zero:

$$1 + KG(s)H(s) = 0,$$
$$KG(s)H(s) = -1 = 1\angle(2k + 1)180^\circ, \quad k = 0, \pm 1, \pm 2, \pm 3, \dots$$

Where -1 is represented in polar form as: $1\angle(2k + 1)180^\circ$

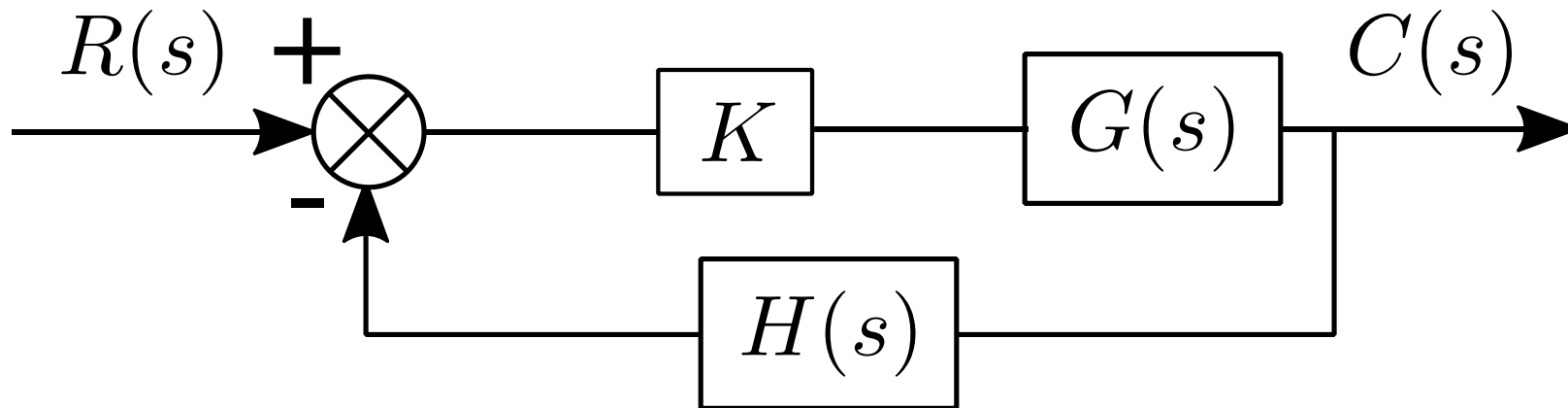
- Also, the magnitude: $|KG(s)H(s)| = 1$
 - And if we assume $K \geq 0$ strictly. Then $K = \frac{1}{|G(s)||H(s)|}$ (eq. 1)
- And angle: $\angle KG(s)H(s) = (2k + 1)180^\circ$ (eq. 2)
- In other words: for s to be a pole of the CL system it has to satisfy eq. 1 and eq. 2



Properties of the Root-Locus

- The root locus is the locus of pole locations in the s -plane for varying values of gain $K \geq 0$, in the closed loop transfer function $G_{cl} = \frac{KG(s)}{1+KG(s)H(s)}$, that satisfy the following two conditions:

1. Magnitude Condition: $K = \frac{1}{|G(s)||H(s)|}$
2. Angle Condition: $\angle KG(s)H(s) = (2k + 1)180^\circ$



For the feedback system shown, show that the point $s = -5 + 3.87j$, is on the root-locus and find the value of the gain K at this point

To find if a point lies on the root-locus, given a feedback system open-loop (forward) transfer function. We check to see if the Angle Condition is met:

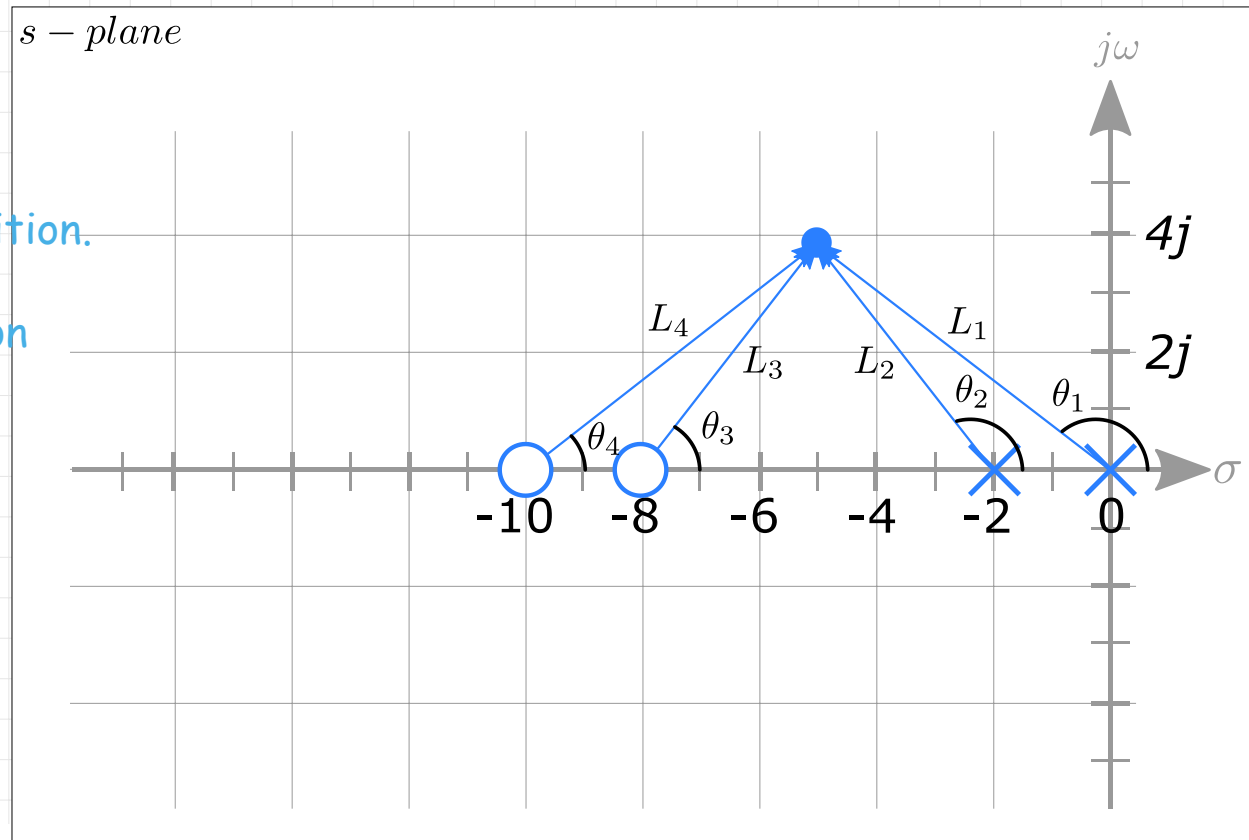
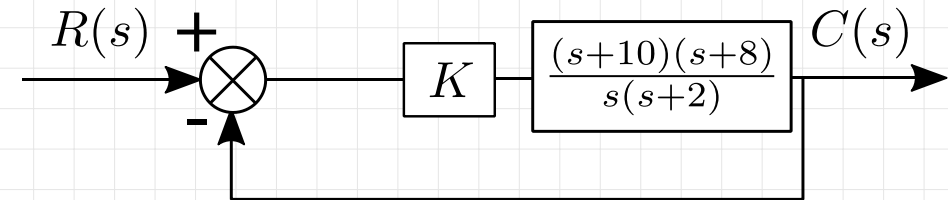
Angle Condition: $\angle \sum \text{zero angles} - \angle \sum \text{pole angles} = \angle |KG(s)| = \angle (2k + 1)180^\circ$

$$\begin{aligned} \angle |KG(s)| &= \theta_3 + \theta_4 - \theta_1 - \theta_2 \\ &= \text{atan}(3.87/3) + \text{atan}(3.87/5) - (180 - \text{atan}(3.87/5)) \\ &\quad - (180 - \text{atan}(3.87/3)) \approx 180 - 180 - 180 \approx -180^\circ \end{aligned}$$

The point is thus on the root locus, as it meets the angle condition.

To find the value of the gain, we check the magnitude condition

$$K = \frac{1}{|G(s)|} = \frac{\text{Pole Lengths}}{\text{Zero Lengths}} = \frac{L_1 L_2}{L_3 L_4} = 1$$



For the feedback system shown, show that the point $s = -15$, is on the root-locus and find the value of the gain K at this point

