Kuwait UniversityCollege of Engineering and Petroleum





ME417 CONTROL OF MECHANICAL SYSTEMS

PART II: CONTROLLER DESIGN USING ROOT-LOCUS

LECTURE 3: TRANSIENT RESPONSE DESIGN VIA GAIN ADJUSTMENT

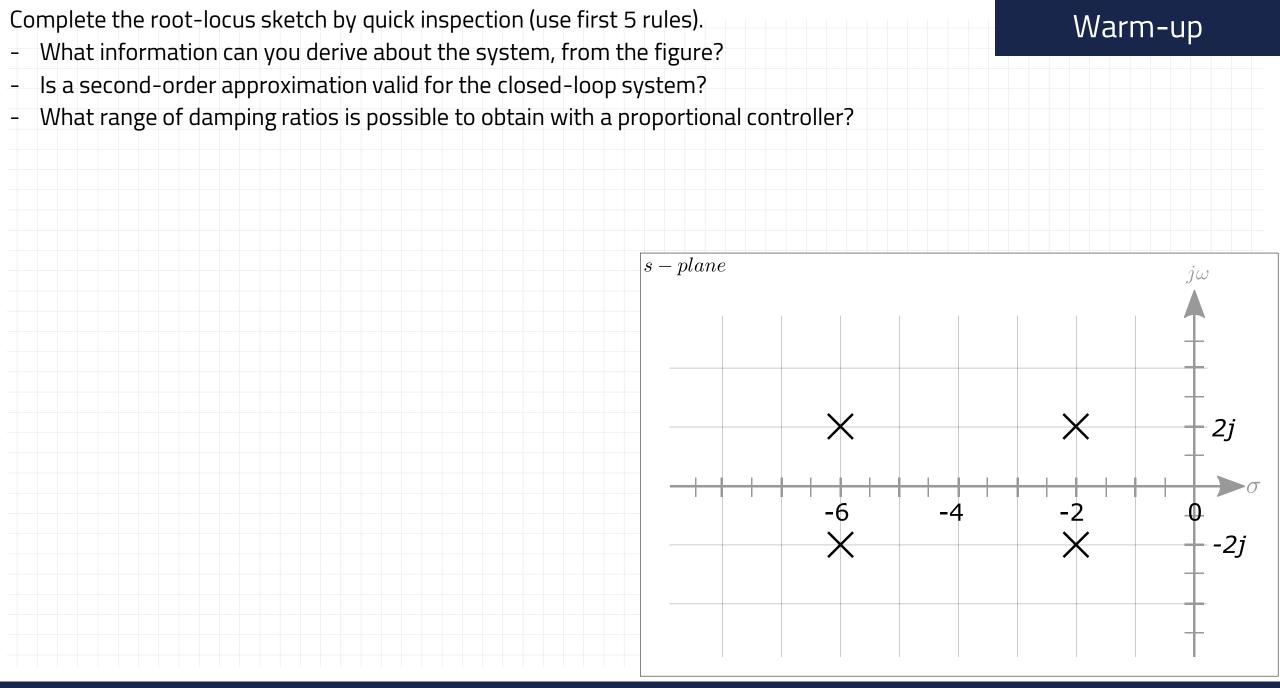
Summer 2020

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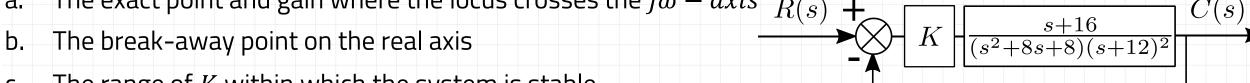
Lecture Plan

- Objectives:
 - Introduce Controller Design through Gain Adjustment
- Reading:
 - Nise: 8.6-8.7
- Practice problems included

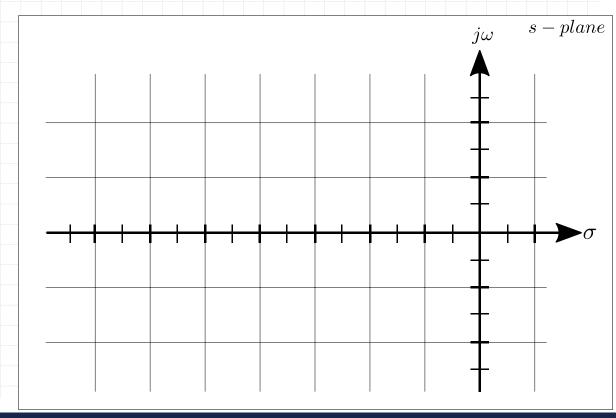




a. The exact point and gain where the locus crosses the $j\omega-axis$ R(s)



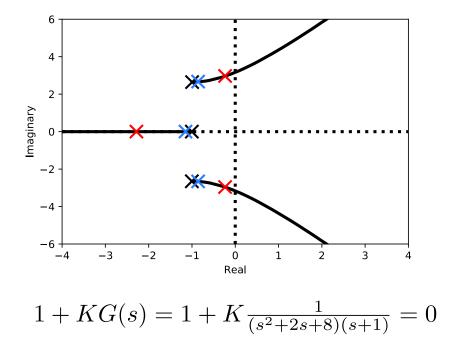
- c. The range of *K* within which the system is stable.
- d. Find the exact point and gain where the locus crosses the $\frac{4}{5}$ damping ratio line



Transient Response Design of Higher-Order Systems via Root-Locus

- Remember that the performance specification points: $T_s, T_r, T_p, \%OS$ were defined for a **general** second-order systems: For feedback systems, that is systems with two complex closed-loop poles and no closed-loop zeros.
- Under some conditions, we can justify a second-order approximation when dealing with higher order systems, or systems with closed-loop zeros:
 - 1. If higher order poles (the 3rd pole and higher) are further into the LHP than the dominant second-order pair of poles. (The "five times" rule of thumb)
 - 2. If closed-loop zeros, if present, are close enough to higher order closed-loop poles that results in their effect being diminished (cancelled).
 - 3. If closed-loop zeros, if present, and not cancelled by higher-order closed loop poles, are instead further into the LHP than the dominant second-order pair of poles.

- 1. If higher order poles (the 3rd pole and higher) are further into the LHP than the dominant second-order pair of poles. (The "five times" rule of thumb)
- For the system shown, note how increasing the gain moves the higher-order pole further into the LHP and brings the dominant complex pole pairs closer to the $j\omega \alpha xis$





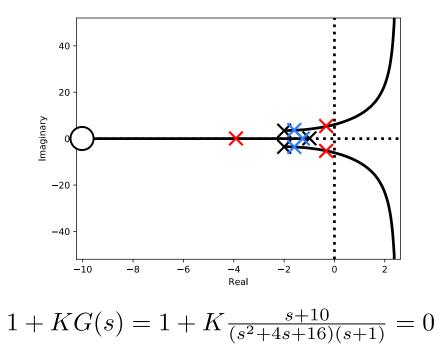
- ★ Closed-Loop poles w/ K=0.05
- ★ Closed-Loop poles w/ K=4



Step responses of CL system



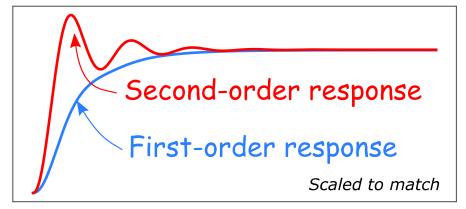
- 2. Closed-loop zeros, if present, are close enough to higher order closed-loop poles that results in their effect being diminished (cancelled).
- Case A: A finite zero is to the left of CL higher-order pole
- Increasing the gain not only pushes the higher-order pole further into the LHP, but rapidly cancels its effect due to pole-zero cancellation.



X Open-Loop poles

★ Closed-Loop poles w/ K=0.05

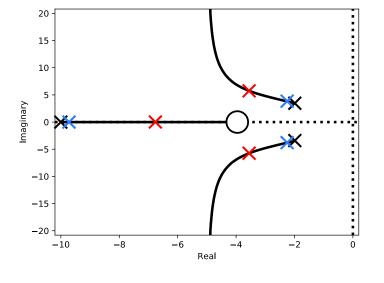
X Closed-Loop poles w/ K=4



Step responses of CL system



- 2. Closed-loop zeros, if present, are close enough to higher order closed-loop poles that results in their effect being diminished (cancelled).
- Case B: A finite zero is to the right of CL higher-order pole
- Increasing the gain brings the higher-order pole closer to the dominant poles, which should increase the order of the system response; however, note that at the same time, the effect of this higher-order pole is cancelled due to pole-zero cancellation.

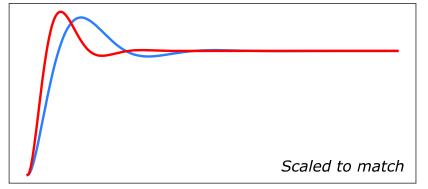


$$1 + KG(s) = 1 + K_{\frac{s+4}{(s^2+4s+16)(s+10)}} = 0$$



★ Closed-Loop poles w/ K=0.05

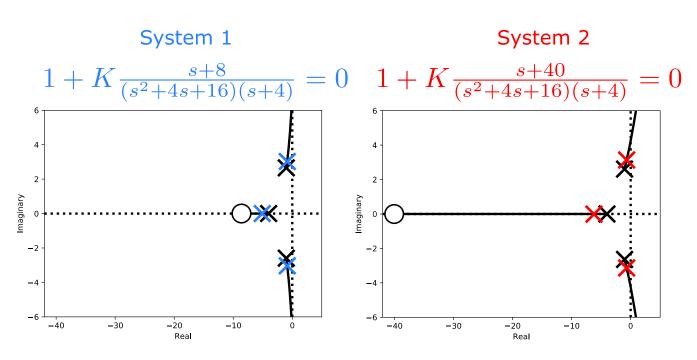
★ Closed-Loop poles w/ K=40



Step responses of CL system



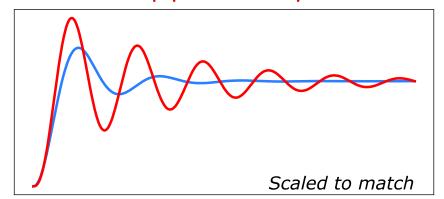
- 3. Closed-Loop zeros, if present, and not cancelled by higher-order closed loop poles, are instead further into the LHP than the dominant second-order pair of poles.
- This is the case where the zeros of the system are further into the LHP relative to the dominant closed-loop poles.



X Open-Loop poles

★ Closed-Loop poles for System 1 w/ K=1

★ Closed-Loop poles for System 2 w/ K=1



Step responses of CL system



Design a feedback system with a proportional controller, using the root-locus technique, for the dynamic system with the plant transfer function: $G_p(s) = \frac{(s+2)}{s(s+1)(s+8)}$

Example

$$G_p(s) = \frac{(s+2)}{s(s+1)(s+8)}$$

To yield a damped frequency of 15rad/s. Also estimate T_s , T_p , e_{ss} to unit ramp input. Justify your second-order approximation. Verify with MATLAB

