

Ch.3 *Fourier Series Representation of Periodic signals*

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Part IV *Fourier Series and LTI Systems*

Outline

- Fourier Series and LTI Systems
- Filtering

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Fourier Series and LTI Systems

- Recall: eigenfunction

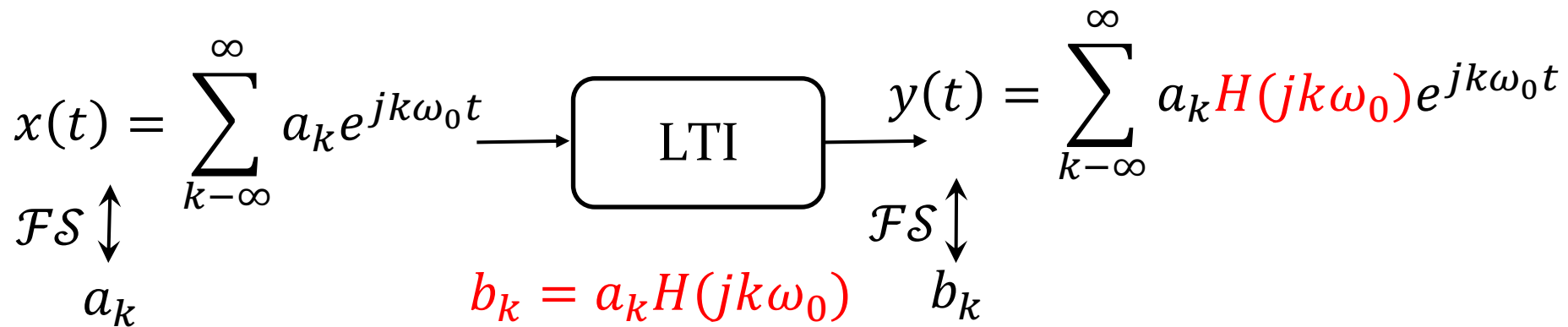
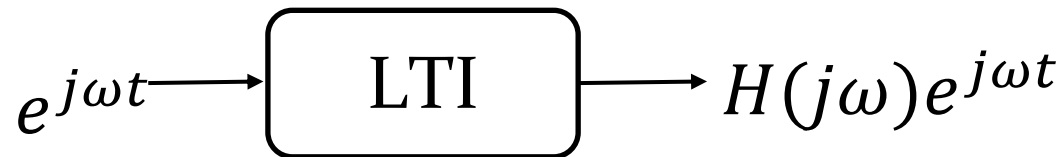
$$\begin{array}{ll} e^{st} \longrightarrow \boxed{\text{LTI}} \longrightarrow H(s)e^{st} & H(s) = \int_{-\infty}^{\infty} h(\tau)e^{-s\tau}d\tau \\ z^n \longrightarrow \boxed{\text{LTI}} \longrightarrow H[z]z^n & H[z] = \sum_{k=-\infty}^{\infty} h[k]z^{-k} \end{array}$$

- $H(s)$ and $H[z]$ are referred to as "**system function**" (transfer functions).
- In this chapter, we consider:
 - Continuous-time: $\text{Re}\{s\} = 0 \rightarrow s = j\omega$
 - Discrete-time: $|z| = 1 \rightarrow z = e^{j\omega}$

Fourier Series and LTI Systems

- Frequency response for CT system: $H(j\omega)$

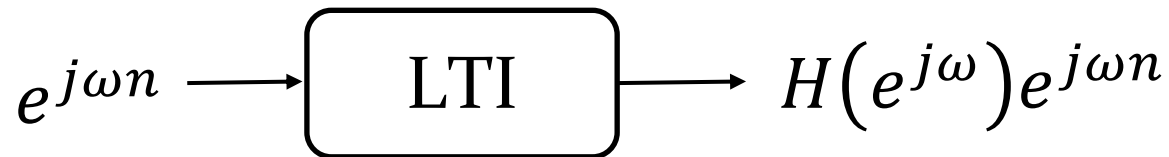
$$H(s) = \int_{-\infty}^{\infty} h(\tau) e^{-s\tau} d\tau \xRightarrow{s=j\omega} H(j\omega) = \int_{-\infty}^{\infty} h(\tau) e^{-j\omega\tau} d\tau$$



Fourier Series and LTI Systems

- Frequency response for DT system: $H(e^{j\omega})$

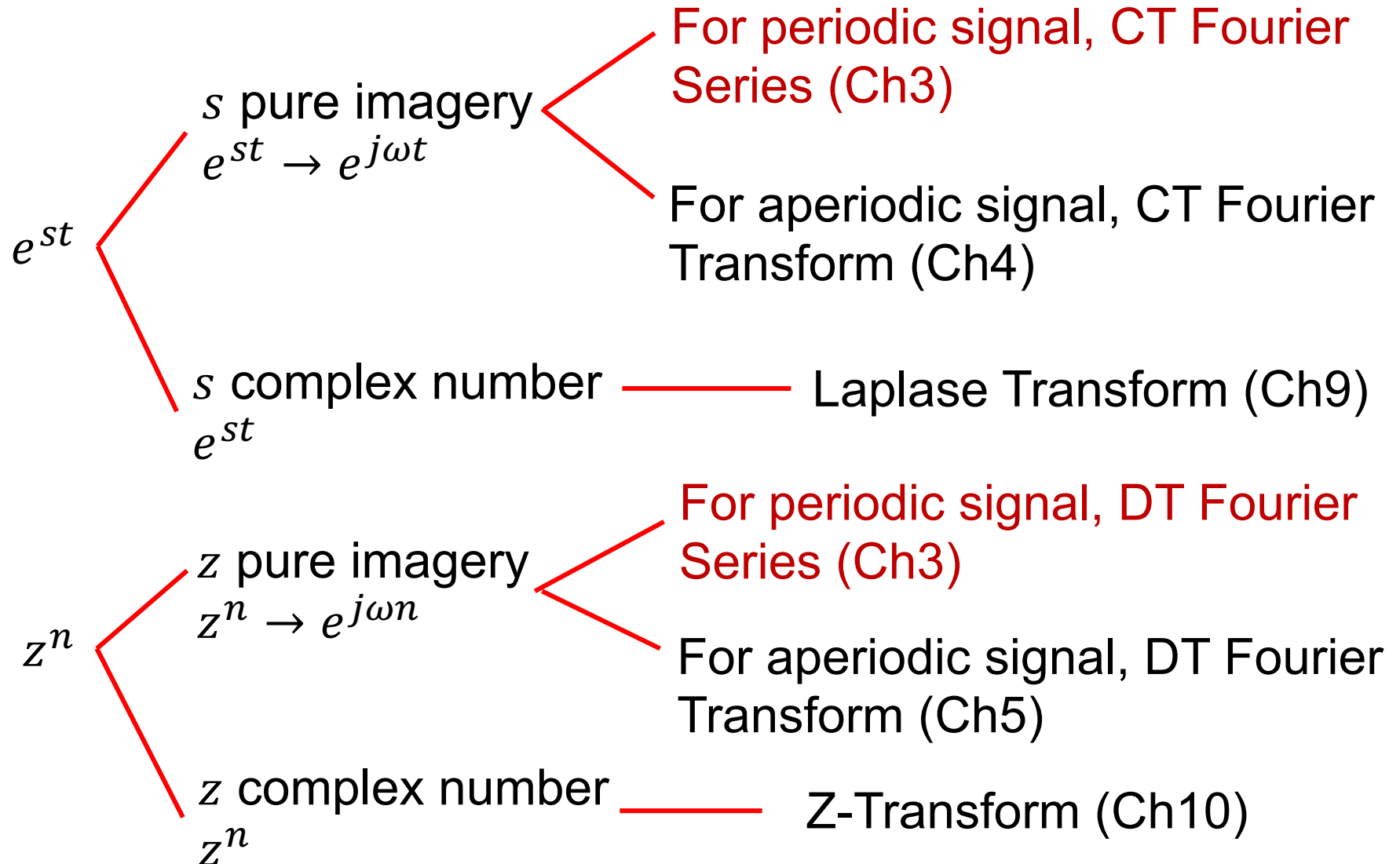
$$H[z] = \sum_{n=-\infty}^{\infty} h[n]z^{-n} \xRightarrow{z=e^{j\omega}} H(e^{j\omega}) = \sum_{n=-\infty}^{\infty} h[n]e^{-j\omega n}$$



$$x[n] = \sum_{k=\langle N \rangle} a_k e^{jk(2\pi/N)n} \xrightarrow{\text{LTI}} y[n] = \sum_{k=\langle N \rangle} a_k H(e^{jk2\pi/N}) e^{jk(2\pi/N)n}$$

$$b_k = a_k H(e^{jk2\pi/N})$$

Fourier Series and LTI Systems



Fourier Series and LTI Systems

- Example: $x(t) = \sum_{k=-3}^3 a_k e^{jk2\pi t}$ ($a_0 = 1$, $a_1 = a_{-1} = \frac{1}{4}$, $a_2 = a_{-2} = \frac{1}{2}$, $a_3 = a_{-3} = \frac{1}{3}$) is the input of a LTI system with $h(t) = e^{-t}u(t)$, determine $y(t)$.

Fourier Series and LTI Systems

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- Solution: $b_k = a_k H(j\omega) = a_k \frac{1}{1 + jk2\pi}$

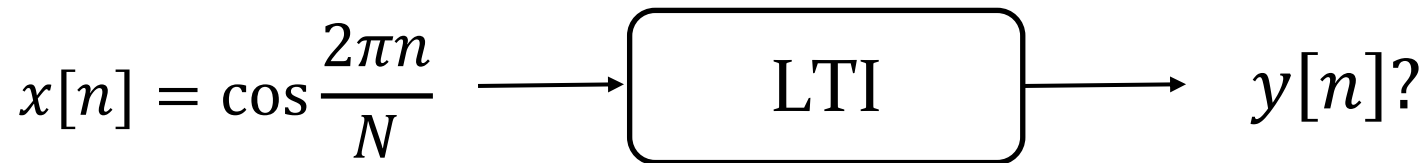
$$b_0 = 1 \cdot 1 = 1 \qquad b_1 = \frac{1}{4} \frac{1}{1 + j2\pi} \qquad b_{-1} = \frac{1}{4} \frac{1}{1 - j2\pi}$$

$$b_2 = \frac{1}{2} \frac{1}{1 + j4\pi} \qquad b_{-2} = \frac{1}{2} \frac{1}{1 - j4\pi} \qquad b_3 = \frac{1}{3} \frac{1}{1 + j6\pi} \qquad b_{-3} = \frac{1}{3} \frac{1}{1 - j6\pi}$$

Fourier Series and LTI Systems

■ Example:

$$h[n] = \alpha^n u[n], |\alpha| < 1$$



■ Solution:

$$x[n] = \frac{1}{2} e^{j(2\pi/N)n} + \frac{1}{2} e^{-j(2\pi/N)n}$$

$$H(e^{j\omega}) = \sum_{n=0}^{\infty} \alpha^n e^{-j\omega n} = \frac{1}{1 - \alpha e^{-j\omega}}, \text{ where } \omega = k \left(\frac{2\pi}{N} \right), k = \pm 1.$$

$$\begin{aligned} y[n] &= \sum_{k=\langle N \rangle} a_k H\left(e^{jk\left(\frac{2\pi}{N}\right)}\right) e^{jk\left(\frac{2\pi}{N}\right)n} \\ &= \frac{1}{2} \left(\frac{1}{1 - \alpha e^{-j2\pi/N}} \right) e^{j(2\pi/N)n} + \frac{1}{2} \left(\frac{1}{1 - \alpha e^{j2\pi/N}} \right) e^{-j(2\pi/N)n} \end{aligned}$$

Outline

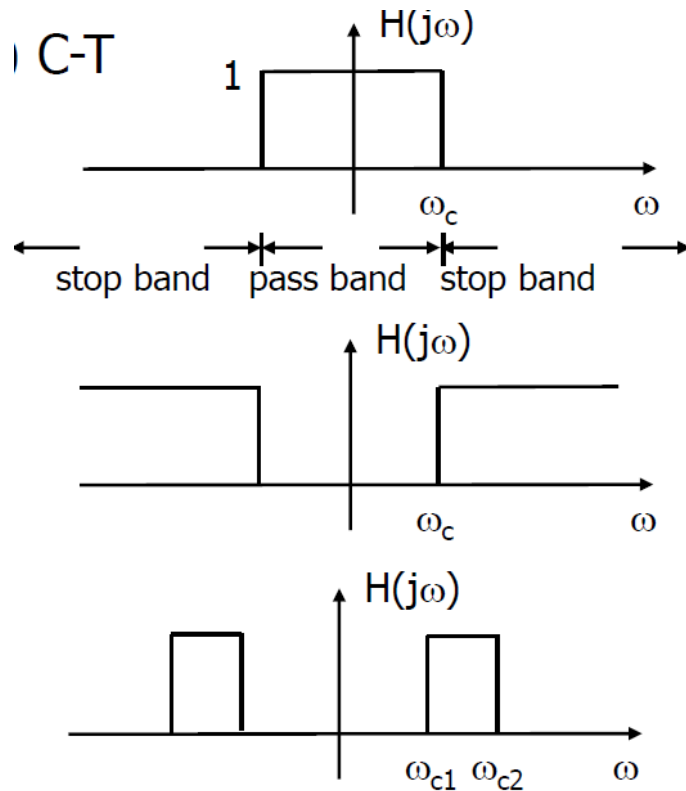
- Fourier Series and LTI Systems
- Filtering

Filtering

- Filtering: A process that changes the relative amplitude (or phase) of some frequency components.
- e.g.,
 - frequency-shaping filter
(like equalizer in a Hi-Fi system)
 - frequency-selective filter
(like low-pass, band-pass, high-pass filters)

Frequency-Selective Filter

- Select some bands of frequencies and reject others.



Ideal low-pass filter (LPF)

$$H(j\omega) = \begin{cases} 1 & \text{in pass band} \\ 0 & \text{in stop band} \end{cases}$$

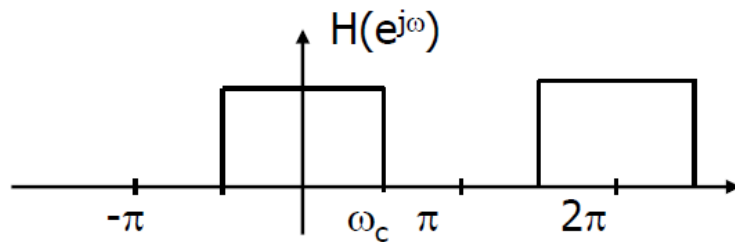
Ideal high-pass filter (HPF)

Ideal band-pass filter (BPF)

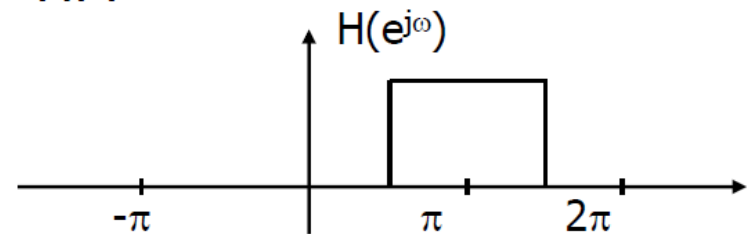
Frequency-Selective Filter

■ Discrete-time

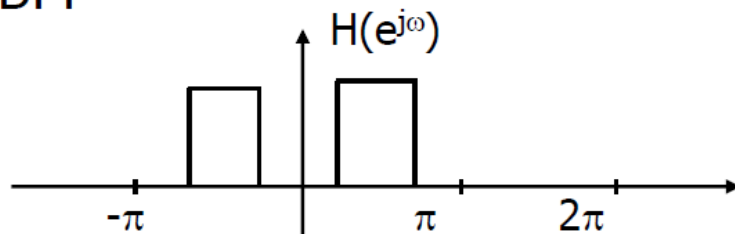
LPF



HPF



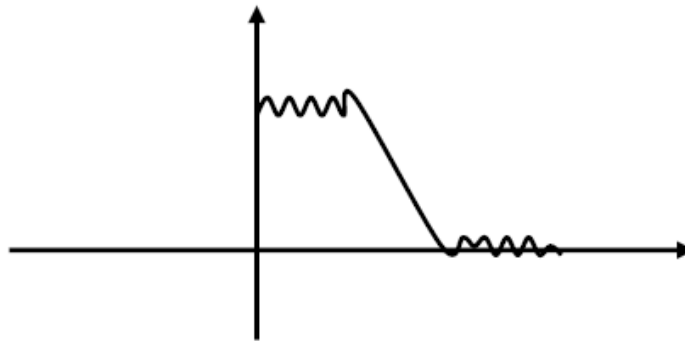
BPF



Frequency-Selective Filter

- For D-T: $H(e^{j\omega})$ is periodic with period 2π
 - Low frequencies: at around $\omega = 0, \pm 2\pi, \pm 4\pi, \dots$
 - High frequencies: at around $\omega = \pm\pi, \pm 3\pi, \dots$
- Note: ideal filters are not realizable
- Practical filters have transition band, and may have ripple in stopband and passband

e.g. LPF



Summary

- Fourier Series and LTI Systems
- Filtering
- Reference in textbook:
 - 3.8, 3.9