Signals and Systems

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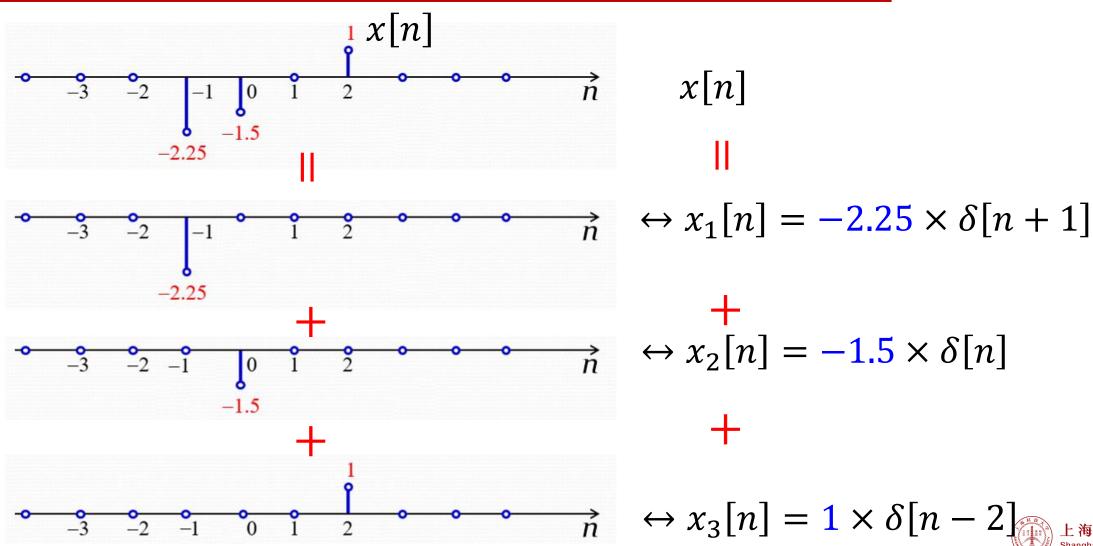


Chapter 2: Linear Time-Invariant Systems

- **□** Discrete-Time LTI Systems
- **□** Continuous-Time LTI Systems
- **□** Properties of LTI Systems
- **□** Differential or Difference Equations

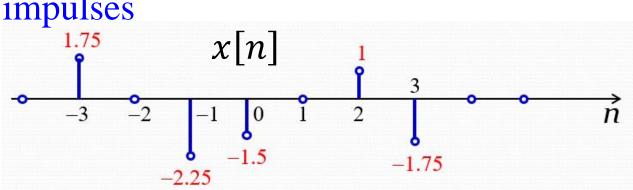


Representation of Discrete-Time Signals in Terms of Impulse



Representation of Discrete-Time Signals in Terms of Impulse

☐ An arbitrary sequence can be represented as the weighted sum of shifted unit impulses



$$x[n] = 1.75\delta[n+3] - 2.25\delta[n+1] - 1.5\delta[n] + \delta[n-2] - 1.75\delta[n-3]$$

☐ A general form

$$x[n] = \sum_{k=0}^{\infty} x[k]\delta[n-k]$$
 Sifting property of $\delta[n]$



Discrete-Time Unit Impulse Response and the Convolution-Sum

The response of a system to a unit impulse sequence $\delta[n]$ is called impulse response, denoted by h[n]





Discrete-Time Unit Impulse Response and the Convolution-Sum

- ☐ How to calculate the impulse response of a system
- For any system whose input-output relationship is defined by

$$y[n] = f\{x[n]\}$$

the impulse response h[n] is calculated as

$$h[n] = f\{\delta[n]\}$$
 replace $x[n]$ by $\delta[n]$



Discrete-Time Unit Impulse Response and the Convolution-Sum

- ☐ How to calculate the impulse response of a system
- Examples: a system is defined as

$$y[n] = a_1 x[n] + a_2 x[n-1] + a_3 x[n-2] + a_4 x[n-3]$$

its impulse response h[n] is

$$h[n] = a_1 \delta[n] + a_2 \delta[n-1] + a_3 \delta[n-2] + a_4 \delta[n-3]$$



Discrete-Time Unit Impulse Response and the Convolution-Sum

- ☐ How to calculate the impulse response of a system
- Examples: a system is defined as

$$y[n] = \sum_{k=-\infty}^{n} x[k]$$

its impulse response h[n] is

$$h[n] = \sum_{k=-\infty}^{n} \delta[k]$$



Discrete-Time Unit Impulse Response and the Convolution-Sum

- ☐ How to calculate the impulse response of a system
- Examples: a system is defined as

$$y[n] = x[n-1] + \frac{1}{2}(x[n-2] + x[n])$$

its impulse response h[n] is

$$h[n] = \frac{\delta}{\delta}[n-1] + \frac{1}{2}(\frac{\delta}{\delta}[n-2] + \frac{\delta}{\delta}[n])$$



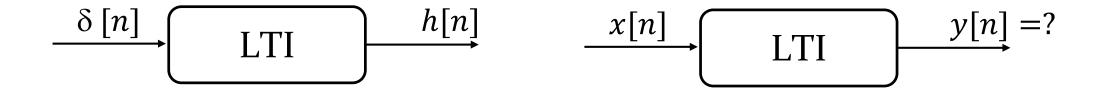
Discrete-Time Unit Impulse Response and the Convolution-Sum

- ☐ An LTI discrete system is completely characterized by its impulse response
- ☐ In other words, knowing the impulse response one can compute the output of the LTI system for an arbitrary input



Discrete-Time Unit Impulse Response and the Convolution-Sum

☐ The impulse response completely characterizes an LTI system



 \square Recall, an arbitrary input x[n] can be expressed as a linear combination of shifted unit impulses

$$x[n] = \sum_{k=-\infty}^{\infty} x[k]\delta[n-k]$$



Discrete-Time Unit Impulse Response and the Convolution-Sum

For any
$$k = k_0$$
 $\delta[n]$ \longrightarrow LTI \longrightarrow $h[n]$

$$\delta[n - k_0] \longrightarrow \text{LTI} \longrightarrow h[n - k_0]$$

$$x[k_0]\delta[n - k_0] \longrightarrow x[k_0]h[n - k_0]$$

$$x[n] = \sum_{k=-\infty}^{\infty} x[k]\delta[n - k] \longrightarrow x[k]h[n - k]$$

$$\downarrow x[n] = \sum_{k=-\infty}^{\infty} x[k]\delta[n - k]$$

$$\downarrow x[n] = \sum_{k=-\infty}^{\infty} x[k]\delta[n - k]$$



Discrete-Time Unit Impulse Response and the Convolution-Sum

$$x[n] = \sum_{k=-\infty}^{\infty} x[k]\delta[n-k]$$

$$LTI \qquad y[n] = \sum_{k=-\infty}^{\infty} x[k]h[n-k]$$

 \Box $\sum x[k]h[n-k]$ is referred as to the **convolution-sum**

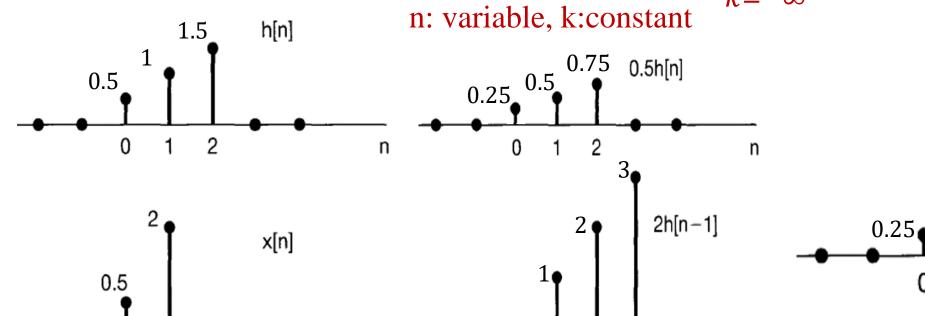
$$x[n] \longrightarrow \boxed{\qquad \qquad} y[n] = x[n] * h[n]$$



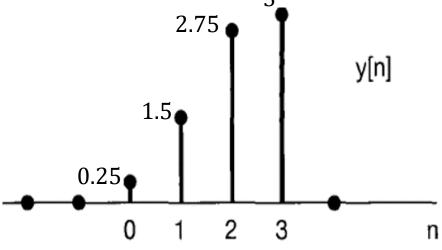
Discrete-Time Unit Impulse Response and the Convolution-Sum

 \square Convolution-Sum calculation – Method 1: sum of k shifted and scaled h[n]

$$x[n] \longrightarrow \boxed{\text{LTI}} \quad y[n] = \sum_{k=-\infty} x[k]h[n-k] = x[n] * h[n]$$



n



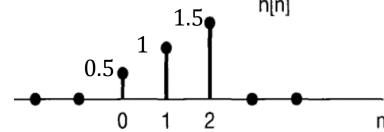


Discrete-Time Unit Impulse Response and the Convolution-Sum

 \square Convolution-Sum calculation – Method 2: calculate y[n] for each n

$$y[n] = \sum_{k=-\infty}^{\infty} x[k]h[n-k] = x[n] * h[n]$$

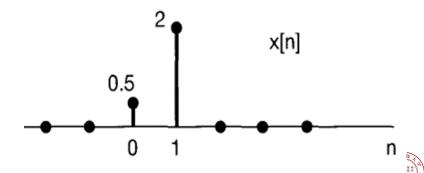
Step 1: determine the range of k $k \in \{0,1\}$



 \triangleright Step 2: determine the range of n

$$[n-k] \in \{0,1,2\} \leftrightarrow n \in \{0,1,2,3\},$$

For other n, $y[n]=0$



Discrete-Time Unit Impulse Response and the Convolution-Sum

 \square Convolution-Sum calculation – Method 2: calculate y[n] for each n

$$y[\mathbf{n}] = \sum_{k=-\infty} x[k]h[\mathbf{n} - k] = x[n] * h[n]$$

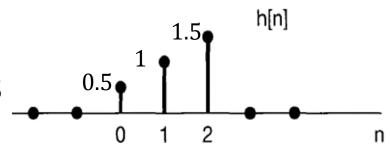
> Step 3: calculate y[n] for each n

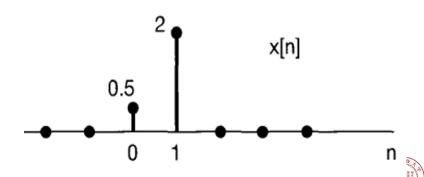
$$y[0] = \sum_{k=0}^{1} x[k]h[0-k] = x[0]h[0] + x[1]h[-1] = 0.25$$

$$y[1] = \sum_{k=0}^{1} x[k]h[1-k] = x[0]h[1] + x[1]h[0] = 1.5$$

$$y[2] = \sum_{k=0}^{1} x[k]h[2-k] = x[0]h[2] + x[1]h[1] = 2.75$$

$$y[3] = \sum_{k=0}^{1} x[k]h[3-k] = x[0]h[3] + x[1]h[2] = 3$$





Discrete-Time Unit Impulse Response and the Convolution-Sum

☐ Convolution-Sum calculation – Method 3

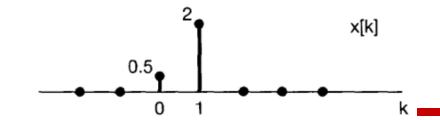
$$y[n] = x[n] * h[n] = \sum_{k=-\infty} x[k]h[n-k]$$

For each n:

- Step 1: change time variables $x[n] \to x[k]$, $h[n] \to h[k]$, and reverse $h[k] \to h[-k]$
- ightharpoonup Step 2: Shift $h[-k] \rightarrow h[n-k]$, n is considered as a constant
- \triangleright Step 3: multiply $x[k] \cdot h[n-k]$
- \triangleright Step 4: Summation $\sum_{k=-\infty}^{\infty} x[k] \cdot h[n-k]$

Change n, repeat step 1 to 4, calculate another y[n]

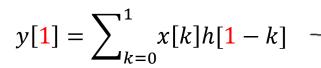




The Convolution-Sum

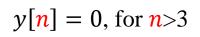
- Convolution-Sum calculation y[n] = 0, for n < 0
 - Method 3
 - If the lengths of the two sequences are *M* and *N*, then the sequence generated by the convolution is of length *M*+*N*-1

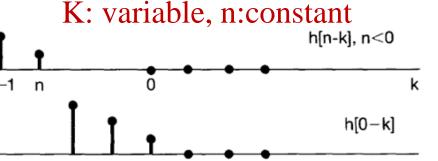
$$y[0] = \sum_{k=0}^{1} x[k]h[0-k]$$

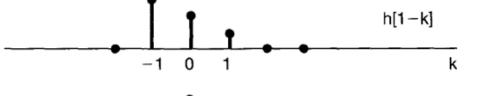


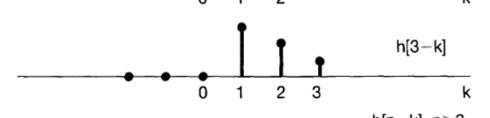
$$y[2] = \sum_{k=0}^{1} x[k]h[2-k]$$

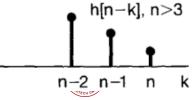
$$y[3] = \sum_{k=0}^{1} x[k]h[3-k]$$











h[2-k]

The Convolution-Sum

Examples

$$y[n] = x[n] * \delta[n] = \sum_{k=-\infty}^{\infty} x[k]\delta[n-k] = x[n]$$

$$y[n] = x[n] * \delta[n - d] = \sum_{k = -\infty}^{\infty} x[k] \delta[n - k - d] \quad \text{Let } k + d = k'$$

$$= \sum_{k' = -\infty}^{\infty} x[k' - d] \delta[n - k']$$

$$= x[n - d] * \delta[n] = x[n - d]$$



The Convolution-Sum

Examples

$$\xrightarrow{x[n]} \qquad \qquad f[n] \qquad \xrightarrow{y[n]}$$

$$x[n] \xrightarrow{h[n-m]} y_1[n] = ?$$

$$y_{1}[n] = x[n] * h[n - m] = \sum_{k = -\infty}^{\infty} x[k]h[n - k - m]$$
 Let $k + m = k'$

$$= \sum_{k' = -\infty}^{\infty} x[k' - m]h[n - k']$$

$$= x[n - m] * h[n] = y[n - m]$$



Chapter 2: Linear Time-Invariant Systems

- **□** Discrete-Time LTI Systems
- **□** Continuous-Time LTI Systems
- **□** Properties of LTI Systems
- **□** Differential or Difference Equations



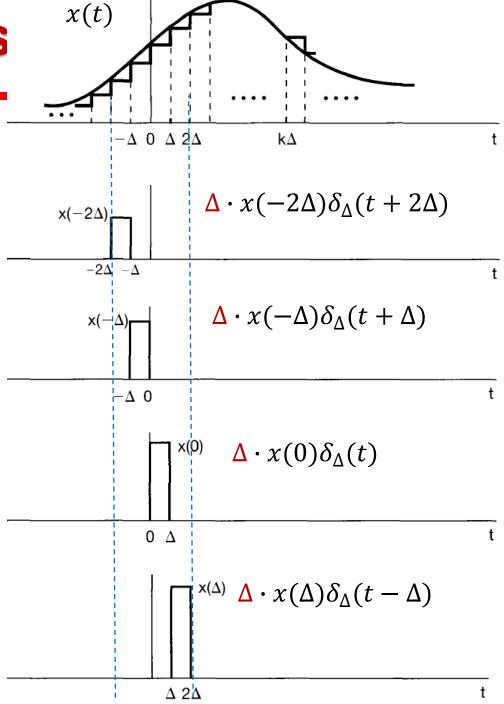
Continuous-Time Signals in Terms of Impulse



$$\delta_{\Delta}(t) = \begin{cases} \frac{1}{\Delta}, & 0 \leq t \leq \Delta \\ 0, other wise \end{cases}$$

$$\hat{x}(t) = \sum_{k=-\infty}^{\infty} x(k\Delta) \delta_{\Delta}(t - k\Delta) \cdot \Delta$$

$$x(t) = \lim_{\Lambda \to 0} \hat{x}(t)$$



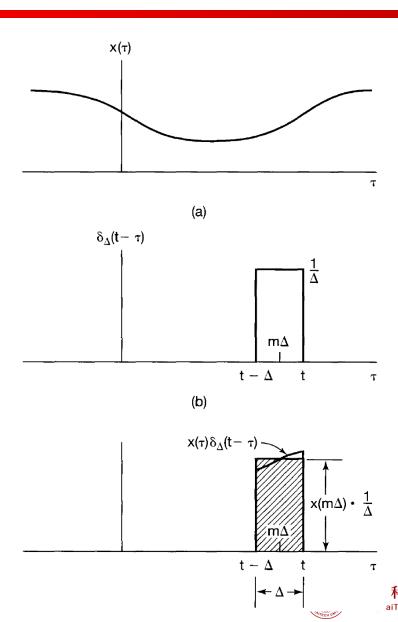
Continuous-Time Signals in Terms of Impulse

 \Box "staircase" approximation of x(t)

$$\widehat{x}(t) = \sum_{k=-\infty}^{\infty} x(k\Delta) \delta_{\Delta}(t - k\Delta) \cdot \Delta$$

$$x(t) = \lim_{\Delta \to 0} \hat{x}(t) = \int_{-\infty}^{\infty} x(\tau) \delta(t - \tau) d\tau$$

Sifting property of $\delta(t)$



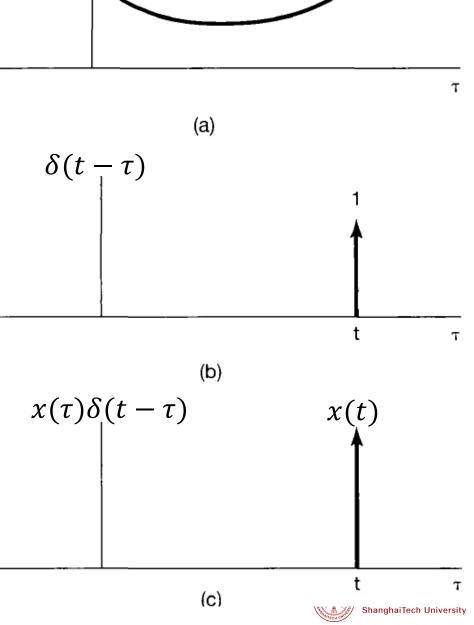
Continuous-Time Signals in Terms of Impul

 \square Using sampling property of $\delta(t)$

$$\int_{-\infty}^{\infty} x(\tau)\delta(t-\tau) d\tau = ?$$

$$x(t)\delta(t-t_0) = x(t_0)\delta(t-t_0)$$
 sampling property

$$x(\tau)\delta(t-\tau) = x(t)\delta(t-\tau)$$
 t: constant



 $x(\tau)$



Continuous-Time Signals in Terms of Impulse

☐ An example

$$x(t) = \int_{-\infty}^{\infty} x(\tau)\delta(t-\tau) d\tau$$

$$u(t) = \int_{-\infty}^{\infty} u(\tau)\delta(t-\tau) d\tau = \int_{0}^{\infty} \delta(t-\tau) d\tau$$



Continuous-Time Unit Impulse Response and Convolution Integral

☐ Continuous-Time Unit Impulse Response

$$\xrightarrow{\delta(t)} \boxed{\text{LTI}} \xrightarrow{h(t)}$$

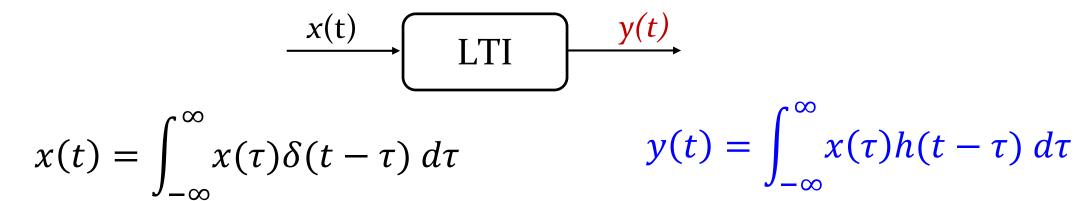
What about
$$x(t)$$
 TTI $y(t) = ? \delta_{\Delta}(t)$ TTI $h_{\Delta}(t)$

$$\hat{x}(t) = \sum_{k=-\infty}^{\infty} x(k\Delta) \delta_{\Delta}(t - k\Delta) \cdot \Delta \qquad \hat{y}(t) = \sum_{k=-\infty}^{\infty} x(k\Delta) h_{\Delta}(t - k\Delta) \cdot \Delta$$

$$x(t) = \lim_{\Delta \to 0} \hat{x}(t) = \int_{-\infty}^{\infty} x(\tau)\delta(t-\tau) d\tau \qquad y(t) = \lim_{\Delta \to 0} \hat{y}(t) = \int_{-\infty}^{\infty} x(\tau)h(t-\tau) d\tau$$

Continuous-Time Unit Impulse Response and Convolution Integral

☐ Continuous-Time Unit Impulse Response



Integral of weighted and shift impulses

Convolution integral

$$\int_{-\infty}^{\infty} x(\tau)h(t-\tau) d\tau = x(t) * h(t)$$



Continuous-Time Unit Impulse Response and Convolution Integral

☐ Compute convolution integral

$$x(t) * h(t) = \int_{-\infty}^{\infty} x(\tau)h(t - \tau) d\tau$$

- Change time variables $x(t) \to x(\tau)$, $h(t) \to h(\tau)$, and reverse $h(\tau) \to h(-\tau)$
- ightharpoonup Shift $h(-\tau) \to h(t-\tau)$
- ightharpoonup Multiply $x(\tau) \cdot h(t-\tau)$



Continuous-Time Unit Impulse Response and Convolution Integral

☐ Compute convolution integral: examples

$$x(t) * h(t) = \int_{-\infty}^{\infty} x(\tau)h(t - \tau) d\tau$$
$$x(t) * \delta(t) = \int_{-\infty}^{\infty} x(\tau)\delta(t - \tau) d\tau = x(t)$$

$$x(t) * \delta(t - t_0) = \int_{-\infty}^{\infty} x(\tau)\delta(t - \tau - t_0) d\tau = \int_{-\infty}^{\infty} x(\tau)\delta(t - (\tau + t_0)) d\tau$$

$$= \int_{-\infty}^{\infty} x(\tau' - t_0)\delta(t - \tau') d\tau' = x(t - t_0) * \delta(t)$$

$$= x(t - t_0)$$



Continuous-Time Unit Impulse Response and Convolution Integral

☐ Compute convolution integral: examples

$$x(t) = e^{-at}u(t), h(t) = u(t), a > 0 x(t)*h(t) = ?$$

$$y(t) = \int_{-\infty}^{\infty} e^{-a\tau} u(\tau) \cdot u(t-\tau) d\tau$$

For
$$t < 0$$
 $x(\tau) \cdot h(t - \tau) = 0$ $\Rightarrow y(t) = 0$

For
$$t \ge 0$$
 $y(t) = \int_0^t e^{-a\tau} d\tau = \frac{-1}{a} e^{-at} \Big|_0^t = \frac{1}{a} (1 - e^{-at})$

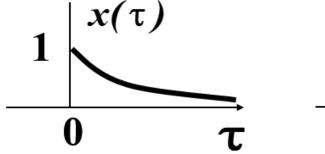


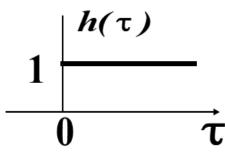
Continuous-Time Unit Impulse Response and Convolution Integral

☐ Compute convolution integral: Graphical Solution

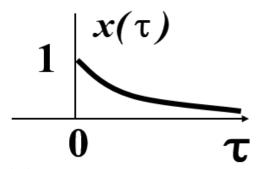
$$x(t) = e^{-at}u(t), h(t) = u(t), a > 0$$

$$x(t) * h(t) = ?$$

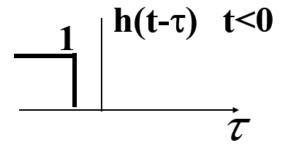


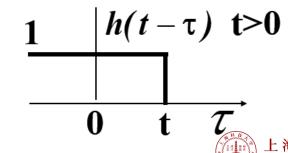


$$y(t) = \int_0^t e^{-a\tau} d\tau = \frac{-1}{a} e^{-at} \Big|_0^t = \frac{1}{a} (1 - e^{-at})$$



 τ : variable, t: constant

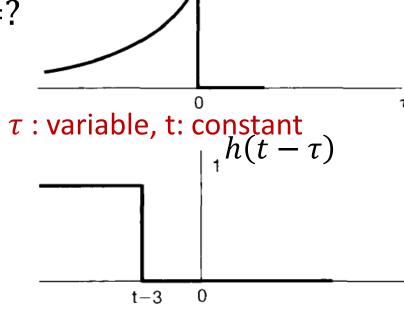


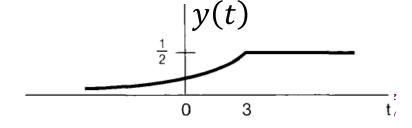


Continuous-Time Unit Impulse Response and Convolution Integral

☐ Computation convolution integral: examples

$$x(t) = e^{2t}u(-t)$$
 $h(t) = u(t-3)$ $x(t) * h(t) = ?$

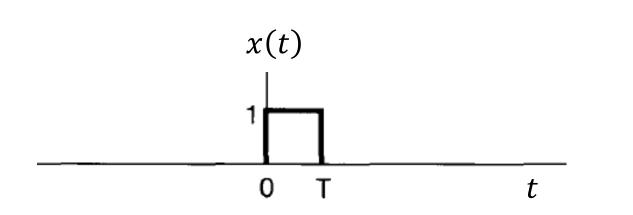


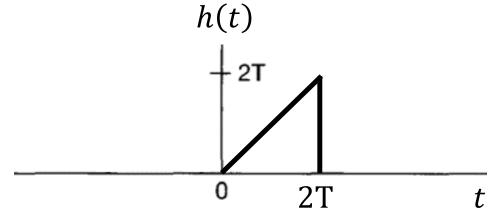


Continuous-Time Unit Impulse Response and Convolution Integral

☐ Compute convolution integral: examples

$$x(t) = \begin{cases} 1, 0 < t < T \\ 0, \text{ otherwise} \end{cases} \quad h(t) = \begin{cases} t, 0 < t < 2T \\ 0, \text{ otherwise} \end{cases}$$





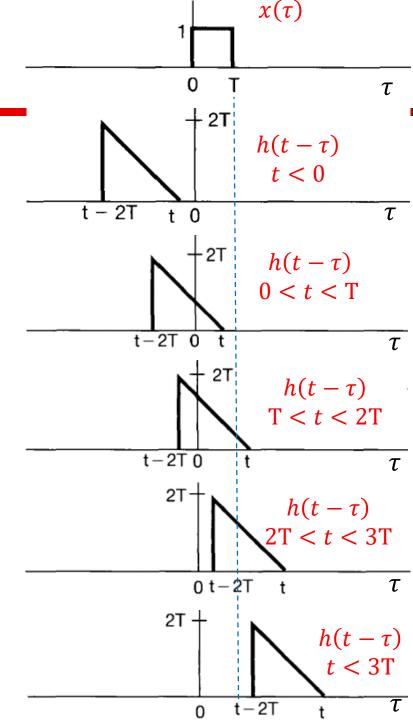
$$x(t) * h(t) = ?$$



Convolution Integral

☐ Computation: examples

$$y(t) = \begin{cases} 0, t < 0 \\ \int_0^t (t - \tau) d\tau = \frac{1}{2}t^2, 0 < t < T \\ \int_0^T (t - \tau) d\tau = Tt - \frac{1}{2}T^2, T < t < 2T \\ \int_{t-2T}^T (t - \tau) d\tau = -\frac{1}{2}t^2 + Tt + \frac{3}{2}T^2, 2T < t < 3T \\ 0, t > 3T \end{cases}$$



Chapter 2: Linear Time-Invariant Systems

- **□** Discrete-Time LTI Systems
- **□** Continuous-Time LTI Systems
- **☐** Properties of LTI Systems
- **□** Differential or Difference Equations



Properties of LTI Systems

The commutative property

$$x[n] * h[n] = \sum_{k=-\infty}^{\infty} x[k]h[n-k] = \sum_{m=-\infty}^{\infty} h[m]x[n-m] = h[n] * x[n]$$

□ Continuous-time x(t) * h(t) = h(t) * x(t)

$$x(t) * h(t) = \int_{-\infty}^{\infty} x(\tau)h(t-\tau) d\tau = \int_{-\infty}^{\infty} h(\tau')x(t-\tau')d\tau' = h(t) * x(t)$$



The distribute property

☐ Discrete-time

$$x[n] * (h_1[n] + h_2[n]) = x[n] * h_1[n] + x[n] * h_2[n]$$

Proof

$$x[n] * (h_1[n] + h_2[n]) = \sum_{k=-\infty}^{\infty} x[k] (h_1[n-k] + h_2[n-k])$$

$$= \sum_{k=-\infty}^{\infty} x[k] h_1[n-k] + \sum_{k=-\infty}^{\infty} x[k] h_2[n-k])$$

$$= x[n] * h_1[n] + x[n] * h_2[n]$$



The distribute property

☐ Continuous-time

$$x(t) * (h_1(t) + h_2(t)) = x(t) * h_1(t) + x(t) * h_2(t)$$

Proof

$$x(t) * (h_1(t) + h_2(t)) = \int_{-\infty}^{\infty} x(\tau) (h_1(t - \tau) + h_2(t - \tau)) d\tau$$
$$= \int_{-\infty}^{\infty} x(\tau) h_1(t - \tau) d\tau + \int_{-\infty}^{\infty} x(\tau) h_2(t - \tau) d\tau$$

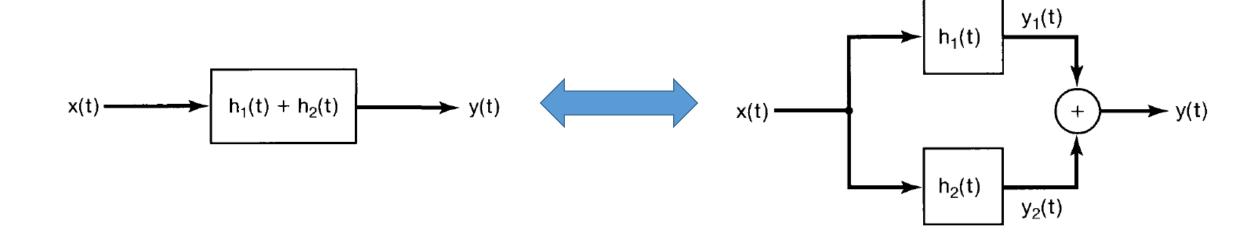
$$= x(t) * h_1(t) + x(t) * h_2(t)$$



The distribute property

☐ Continuous-time

$$x(t) * (h_1(t) + h_2(t)) = x(t) * h_1(t) + x(t) * h_2(t)$$





The associative property

Discrete-time
$$x[n] * (h_1[n] * h_2[n]) = (x[n] * h_1[n]) * h_2[n]$$

 $x[n] * (h_1[n] * h_2[n]) = x[n] * y[n], y[n] = \sum_{m=-\infty}^{\infty} h_1[m]h_2[n-m]$

$$x[n] * (h_1[n] * h_2[n]) = x[n] * y[n],$$

$$y[n] = \sum_{m=-\infty}^{\infty} h_1[m]h_2[n-m]$$

$$=\sum_{k=-\infty}^{\infty}$$

$$]y[n-k] = \sum_{k}^{n}$$

$$=\sum_{k=-\infty}^{\infty}x[k]y[n-k]=\sum_{k=-\infty}^{\infty}x[k]\sum_{m=-\infty}^{\infty}h_1[m]h_2[n-k-m]$$

Let
$$k + m = l$$

$$=\sum_{k=-\infty}^{\infty}x[k]\sum_{l=-\infty}^{\infty}h_1[l-k]h_2[n-l]$$

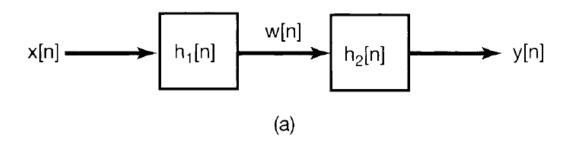
$$=\sum_{l=-\infty}^{\infty}\sum_{m=-\infty}^{\infty}x[k]h_1[l-k]h_2[n-l]$$

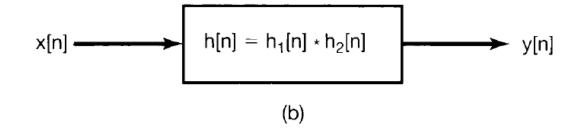
$$= \sum_{l=-\infty}^{\infty} (x[l] * h_1[l]) h_2[n-l] = (x[n] * h_1[n]) * h_2[n]$$

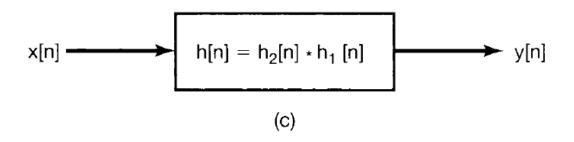


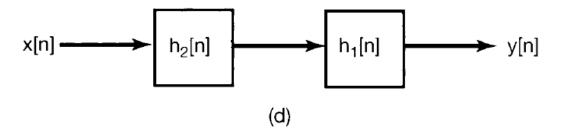
The associative property

☐ Discrete-time











The associative property

Continuous-time
$$x(t) * (h_1(t) * h_2(t)) = (x(t) * h_1(t)) * h_2(t)$$

 $x(t) * (h_1(t) * h_2(t)) = x(t) * \int_{-\infty}^{\infty} h_1(\tau)h_2(t - \tau)d\tau$
 $= \int_{-\infty}^{\infty} x(\tau') \int_{-\infty}^{\infty} h_1(\tau)h_2(t - \tau' - \tau)d\tau d\tau'$
Let $\tau' + \tau = \tau''$ $= \int_{-\infty}^{\infty} x(\tau') \int_{-\infty}^{\infty} h_1(\tau'' - \tau')h_2(t - \tau'')d\tau'' d\tau''$
 $= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} x(\tau')h_1(\tau'' - \tau')d\tau' h_2(t - \tau'')d\tau''$
 $= \int_{-\infty}^{\infty} x(\tau'') * h_1(\tau'') h_2(t - \tau'')d\tau'' = (x(t) * h_1(t)) * h_2(t)$

LTI systems with and without memory

 \square Discrete-time system without memory only if h[n] = 0 for all $n \neq 0$

$$h[n] = h[0]\delta[n] = k\delta[n]$$
 $y[n] = kx[n]$ Why?

 \Box Continuous-time system without memory only if h(t) = 0 for all $t \neq 0$

$$h(t) = h(0)\delta(t) = k\delta(t) \qquad y(t) = kx(t)$$



Invertibility for LTI systems

If $h_0(t) * h_1(t) = \delta(t)$, the system with impulse response $h_1(t)$ is the inverse of the system with impulse response $h_0(t)$

$$x(t)$$
 $h_0(t)$ $h_1(t)$ $w(t)=x(t)$

 \square Similarly, if $h_0[n] * h_1[n] = \delta[n]$, the system with impulse response $h_1[n]$ is the inverse of the system with impulse response $h_0[n]$



Invertibility for LTI systems

Examples

Consider $h_0[n] = u[n]$, determine the inverse system $h_1[n]$

$$hold$$
 $hold$
 $hold$

$$\delta[n] = u[n] - u[n-1] = u[n] * (\delta[n] - \delta[n-1])$$

$$\therefore h_1[n] = \delta[n] - \delta[n-1]$$



Invertibility for LTI systems

Examples

Consider the LTI system consisting of a pure time shift

$$y(t) = x(t - t_0),$$

determine the inverse system.



Causality for LTI systems

- \square If h[n] = 0 for n < 0, or h(t) = 0 for t < 0, the system is causal
- \square Equivalent to the condition of initial rest: if $t \le t_0$, x(t) = 0, then $y(t_0) = 0$

$$y[n] = \sum_{k=-\infty}^{n} x[k]h[n-k] \quad \text{or} \quad y[n] = \sum_{k=0}^{\infty} h[k]x[n-k]$$

$$y(t) = \int_{-\infty}^{t} x(\tau)h(t-\tau) d\tau \quad \text{or} \quad y(t) = \int_{0}^{\infty} h(\tau)x(t-\tau)d\tau$$



Causality for LTI systems

- Examples
 - Accumulator: $y[n] = \sum_{l=-\infty}^{n} x[l]$ Causal LTI system

$$h[n] = \sum_{l=-\infty}^{n} \delta[l] = u[n] \qquad h[n] = 0 \text{ for } n < 0$$

• Factor 2 interpolator: $y[n] = x[n] + \frac{1}{2}(x[n-1] + x[n+1])$

Non-Causal LTI system

$$h[n] = \delta[n] + rac{1}{2}(\delta[n-1] + \delta[n+1])$$
 $h[n]
eq 0 ext{ for } n = -1 < 0$ 上海科技大学 Shanghai Tech University

Stability for LTI systems

- \square A discrete LTI system is stable if h[n] is absolutely summable
- \square A continuous LTI system is stable if h(t) is absolutely integrable

$$\sum_{k=-\infty}^{\infty} |h[k]| < \infty$$
 absolutely summable

$$\int_{-\infty}^{\infty} |h(\tau)| d\tau < \infty \qquad \text{absolutely integrable}$$



Stability for LTI systems

☐ Proof: "if and only if" (Sufficient and necessary condition)

$$|y[n]| = \left| \sum_{k=-\infty}^{\infty} h[k]x[n-k] \right| \le \sum_{k=-\infty}^{\infty} |h[k]x[n-k]| \le \sum_{k=-\infty}^{\infty} |h[k]| |x[n-k]|$$

$$\therefore |y[n]| \le \sum_{k=-\infty} |h[k]| |x[n-k]|$$

If
$$|x[n-k]| \le B_x$$
 $|y[n]| \le B_x \sum_{k=-\infty}^{\infty} |h[k]|$

If and only if
$$\sum_{k=-\infty}^{\infty} |h[k]| < \infty$$
 $|y[n]| < \infty$



Stability for LTI systems

☐ Proof: continuous case

$$|f|x(t-\tau)| \le B_x$$

$$|y(t)| = \left| \int_{-\infty}^{\infty} h(\tau)x(t-\tau)d\tau \right| \le \int_{-\infty}^{\infty} |h(\tau)| \cdot |x(t-\tau)|d\tau \le B_x \int_{-\infty}^{\infty} |h(\tau)|d\tau$$

If and only if
$$\int_{-\infty}^{\infty} |h(\tau)| d\tau < \infty$$
 $|y(t)| < \infty$



Stability for LTI systems

□ Examples

$$y[n] = x[n - n_0]$$

$$h[n] = \delta[n - n_0]$$

$$\sum_{n=-\infty}^{\infty} |h[n]| = \sum_{n=-\infty}^{\infty} |\delta[n-n_0]| = 1$$



Stability for LTI systems

■ Examples

$$h[n] = \alpha^n u[n] \qquad |\alpha| \le 1$$

$$\sum_{n=-\infty}^{\infty} |h[n]| = \sum_{n=-\infty}^{\infty} |\alpha^n| u[n] = \sum_{n=0}^{\infty} |\alpha^n| = \frac{1}{1-|\alpha|} \quad \text{If } |\alpha| < 1 \text{ , the system is stable}$$

If $|\alpha| = 1$, the system is unstable



The unit step response of LTI systems

The unit step response, s[n], corresponding to the output with input x[n] = u[n]

$$s[n] = u[n] * h[n] = \sum_{-\infty}^{\infty} h[k]u[n-k] = \sum_{k=-\infty}^{n} h[k]$$

$$u[n] = \sum_{k=-\infty}^{n} \delta[k] \qquad \qquad s[n] = \sum_{k=-\infty}^{n} h[k]$$

$$h[n] = s[n] - s[n-1]$$



The unit step response of LTI systems

The unit step response, s(t), corresponding to the output with input x(t)=u(t)

$$s(t) = u(t) * h(t) = \int_{-\infty}^{\infty} h(\tau)u(t - \tau)d\tau = \int_{-\infty}^{t} h(\tau)d\tau$$

$$u(t) = \int_{-\infty}^{t} \delta(\tau) d\tau \qquad s(t) = \int_{-\infty}^{t} h(\tau) d\tau$$

$$h(t) = \frac{ds(t)}{dt} = s'(t)$$



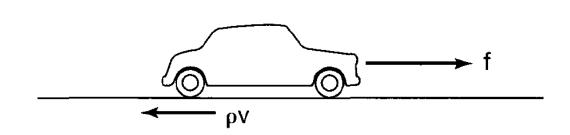
Chapter 2: Linear Time-Invariant Systems

- **□** Discrete-Time LTI Systems
- **□** Continuous-Time LTI Systems
- **□** Properties of LTI Systems
- **□** Differential or Difference Equations

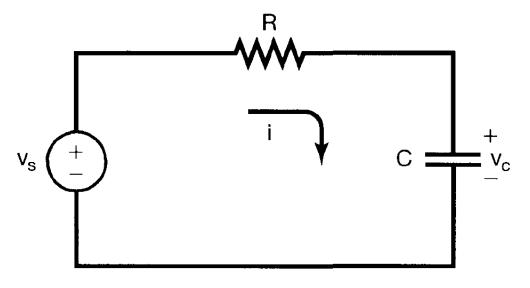


Differential equation

☐ First order system



$$\frac{dv(t)}{dt} + \frac{\rho}{m}v(t) = \frac{1}{m}f(t)$$



$$\frac{dv_c(t)}{dt} + \frac{1}{RC}v_c(t) = \frac{1}{RC}v_s(t)$$



Differential equation

- ☐ Linear constant-coefficient DE
- $\frac{dy(t)}{dt} + ay(t) = bx(t)$
- ☐ Describes an implicit relationship between the input and the output

☐ Can not completely characterize a LTI system

☐ Auxiliary conditions are required to solve the DE: causal (initial rest condition)



Differential equation

☐ First order system: example

$$\frac{dy(t)}{dt} + 2y(t) = x(t)$$

If
$$x(t) = Ke^{3t}u(t)$$
 $y(t) = ?$

□ Solution:

$$y(t) = y_p(t) + y_h(t)$$

 $y_p(t)$: particular solution, forced response (same form as input)

$$y_h(t)$$
: Homogenous solution
$$\frac{dy(t)}{dt} + 2y(t) = 0$$



Differential equation

☐ First order system: example

$$\frac{dy(t)}{dt} + 2y(t) = x(t)$$

If
$$x(t) = Ke^{3t}u(t)$$
 $y(t) = ?$

☐ Particular solution:

$$3Ye^{3t} + 2Ye^{3t} = Ke^{3t} \longrightarrow Y = K/5 \longrightarrow y_p(t) = \frac{K}{5}e^{3t} \text{ for } t > 0$$

□ Homogenous solution: Let $y_h(t) = Ae^{st}$, for t>0

$$Ase^{st} + 2Ae^{st} = 0$$
 \Longrightarrow $s = -2$ \Longrightarrow $y_h(t) = Ae^{-2t}$

$$y(t) = Ae^{-2t} + \frac{K}{5}e^{3t}$$
, for $t > 0$



Differential equation

$$y(t) = Ae^{-2t} + \frac{K}{5}e^{3t}$$
, for $t > 0$

- ☐ Auxiliary condition is required to determine *A*
- \Box Initial rest as auxiliary condition for causal LTI systems: y(0) = 0

$$A + \frac{K}{5} = 0 \implies A = -\frac{K}{5} \implies y(t) = \frac{K}{5} (e^{3t} - e^{-2t}), \text{ for } t > 0$$
$$= \frac{K}{5} (e^{3t} - e^{-2t}) u(t)$$



Differential equation

☐ General case: Nth-order linear constant-coefficient differential equation

$$\sum_{k=0}^{N} a_k \frac{d^k y(t)}{dt^k} = \sum_{k=0}^{M} b_k \frac{d^k x(t)}{dt^k}$$

- □ Particular solution + Homogenous solution: $y(t) = y_p(t) + y_h(t)$
 - $y_p(t)$: forced response (same form as input)
 - $y_h(t)$: Natural response, $\sum_{k=0}^{N} a_k \frac{d^k y(t)}{dt^k} = 0$
- \square Initial rest as auxiliary condition, that is if x(t) = 0 for $t \le t_0$,

$$y(t_0) = \frac{dy(t_0)}{dt} = \dots = \frac{d^{N-1}y(t_0)}{dt^{N-1}} = 0$$



Difference equation

☐ General case: Nth-order linear constant-coefficient difference equation

$$\sum_{k=0}^{N} a_k y[n-k] = \sum_{k=0}^{M} b_k x[n-k]$$

- □ Particular solution + Homogenous solution: $y[n] = y_p[n] + y_h[n]$
 - $y_p[n]$: forced response (same form as input)
 - $y_h[n]$: Natural response, $\sum_{k=0}^{N} a_k y[n-k] = 0$
- Initial rest as auxiliary condition, that is if x[n] = 0 for $n \le n_0$, $y[n_0] = y[n_0-1] = \dots = y[n_0-(N-1)] = 0$



Difference equation

☐ Recursive solution:

$$y[n] = \frac{1}{a_0} \left\{ \sum_{k=0}^{M} b_k x[n-k] - \sum_{k=1}^{N} a_k y[n-k] \right\}$$

Particular case N=0

$$y[n] = \frac{1}{a_0} \sum_{k=0}^{M} b_k x[n-k]$$
 Non-recursive equation

$$h[n] = \frac{1}{a_0} \sum_{k=0}^{M} b_k \delta[n-k]$$
 Finite impulse response

(FIR) system



Difference equation

- Recursive solution: example $y[n] \frac{1}{2}y[n-1] = x[n]$
 - Consider $x[n] = K\delta[n]$ and take initial rest: y[-1] = 0

$$y[0] = x[0] + \frac{1}{2}y[-1] = K$$
 $y[1] = x[1] + \frac{1}{2}y[0] = \frac{1}{2}K$

$$y[2] = x[2] + \frac{1}{2}y[1] = \left(\frac{1}{2}\right)^2 K$$
 ... $y[n] = x[n] + \frac{1}{2}y[n-1] = \left(\frac{1}{2}\right)^n K$

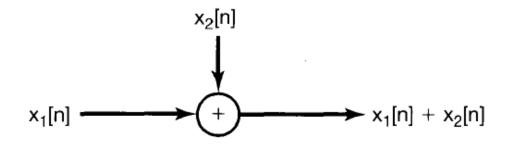
$$\therefore h[n] = \left(\frac{1}{2}\right)^n u[n] \qquad Infinite impulse response (IIR) system$$

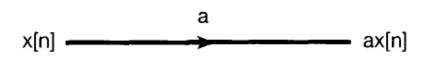
Generally
$$\sum_{k=0}^{N} a_k y[n-k] = \sum_{k=0}^{M} b_k x[n-k] \begin{cases} N = 0, \text{ FIR system} \\ N > 0, \text{ usually IIR system} \end{cases}$$

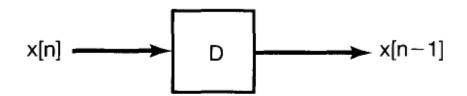


Block Diagram Representations

☐ Basic elements: discrete-time

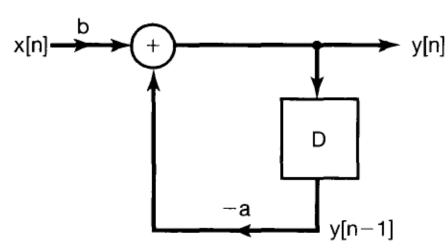






$$y[n] + ay[n-1] = bx[n]$$

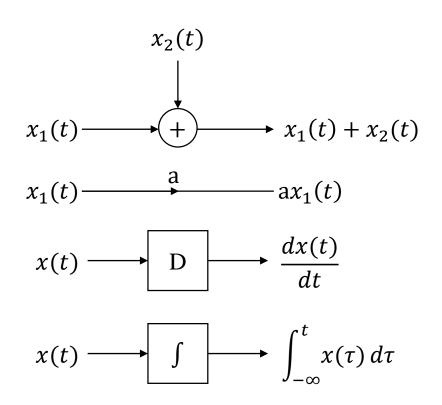
$$y[n] = -ay[n-1] + bx[n]$$





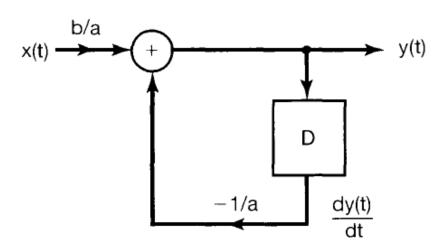
Block Diagram Representations

☐ Basic elements: continuous-time



$$\frac{dy(t)}{dt} + ay(t) = bx(t)$$

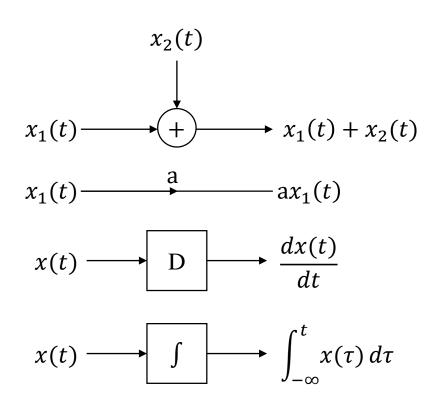
$$y(t) = -\frac{1}{a}\frac{dy(t)}{dt} + \frac{b}{a}x(t)$$





Block Diagram Representations

☐ Basic elements: continuous-time



$$\frac{dy(t)}{dt} + ay(t) = bx(t)$$
$$\frac{dy(t)}{dt} = -ay(t) + bx(t)$$

$$y(t) = \int_{-\infty}^{t} [bx(\tau) - ay(\tau)]d\tau$$

