

Communication theory Project_CIE 337:

Ahmed Amgad 202200393

Abdelrahman Magdy 202200341

Mohammed Ali 202200594

Part A:

❖ Matlab code for variables:

```
%% Time space for simulation
fc = 5e3; % Carrier frequency in Hz
fs = 1000 * fc; % Sampling frequency (50 kHz) 1000 times the frequency of
the carrier for better visualization
Ts = 1 / fs; % Sampling period
T = 0.004; % Total simulation time (4 ms)
t = 0:Ts:T; % Time vector
```

1- Message Signal Generation & plotting:

• Matlab code:

```
fm1 = 1000; % Frequency of m1 in Hz

Tm2 = 2e-3; % Period of m2 in seconds

fm2 = 1 / Tm2; % Frequency of m2 in Hz

m1 = sawtooth((2*pi*fm1*t) + pi, 0); % Corrected to start from 0

m2 = zeros(size(t));

shift = mod(t, Tm2); % Repeat the step function every period

m2(shift < 0.5e-3) = 1;

m2((shift >= 0.5e-3) & (shift < 1e-3)) = 0.5;

m2((shift >= 1e-3) & (shift < 1.5e-3)) = -0.5;

m2(shift >= 1.5e-3) = -1;

figure;

subplot(2,1,1);

plot(t, m1, 'LineWidth', 1.5);
```

```

title('Message Signal m_1(t)');

xlabel('Time (ms)');

ylabel('Amplitude');

grid on;

subplot(2,1,2);

plot(t, m2, 'LineWidth', 1.5);

title('Message Signal m_2(t)');

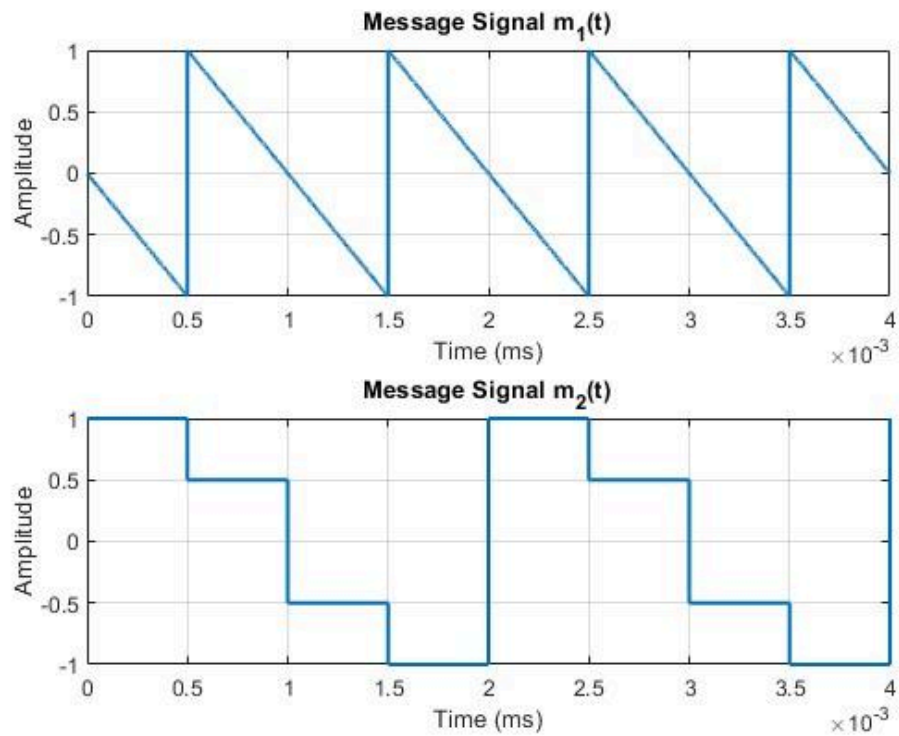
xlabel('Time (ms)');

ylabel('Amplitude');

grid on;

```

- $m_1(t)$ is a **triangular waveform** (sawtooth-type), periodic and continuous, and $m_2(t)$ is a **piecewise constant signal** (digital-style steps).
- Message signals figures:



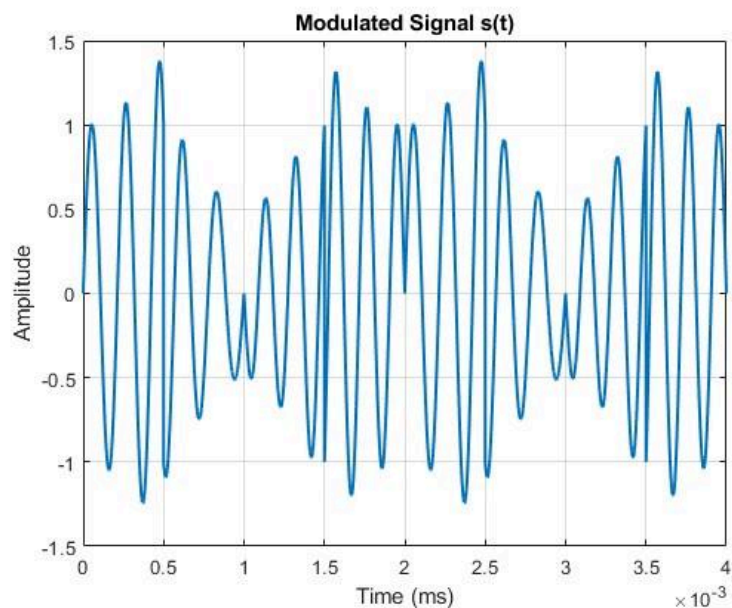
2-Generate & plot the modulated signal $s(t)$:

- $s(t) = A_{cm1}(t)\cos(2\pi f_c t) + A_{cm2}(t)\sin(2\pi f_c t)$
- Matlab code:

```
% Generating the Carrier Signals.  
c1 = cos(2*pi*fc*t);  
c2 = sin(2*pi*fc*t);  
%% Modulating the Signals using QAM Technique  
s = (m1 .* c1) + (m2 .* c2);  
figure;  
plot(t, s, 'LineWidth', 1.5);  
title('Modulated Signal s(t)');  
xlabel('Time (ms)');  
ylabel('Amplitude');  
grid on;
```

- ☐ **DSB-QAM signal** uses orthogonal carriers (sine and cosine) to load two distinct signals into the same frequency spectrum.
- ☐ The envelope or shape of the modulated signal depends on both $m_1(t)$ and $m_2(t)$
- ☐ The variations in the signal's amplitude *and* phase influence the zero-crossings and peak locations.

- Modulated Signal figure:



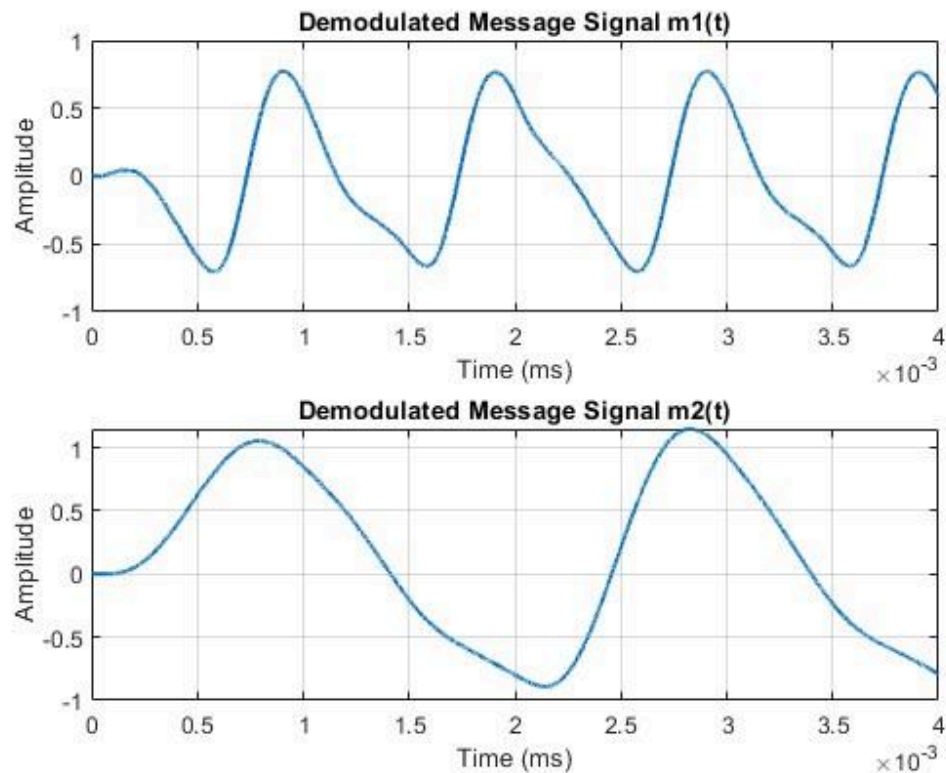
3- Implement receiver of DSB-QAM $m_1(t)$ and $m_2(t)$:

- $c_1(t) = \cos(2\pi f_c t)$
- $c_2(t) = \sin(2\pi f_c t)$
- Matlab code:

```
% Implementing the QAM Receiver
% Implementing the carrier signal (normal case)
rc1 = 2 * cos(2*pi*fc*t);
rc2 = 2 * sin(2*pi*fc*t);
x1 = s .* rc1;
x2 = s .* rc2;
% Design Butterworth low-pass filters
[b1, a1] = butter(4, 2*fm1/(fs/2)); % Filter for m1_demodulated
[b2, a2] = butter(4, 2*fm2/(fs/2)); % Filter for m2_demodulated
% Apply the Butterworth filters
m1_demodulated = filter(b1, a1, x1);
m2_demodulated = filter(b2, a2, x2);
```

- ☐ Following low-pass filtering and coherent demodulation (multiplication by $2\cos(2\pi f_c t)$, and $2\sin(2\pi f_c t)$) this plot displays the restored original signals. The apparent smoothness and loss of sharp details are caused by the low-pass filter's inability to filter part of the original message signal's higher frequency components, even while it effectively eliminates the high-frequency components (about $2f_c$) produced during demodulation. Nonetheless, the essential form and content of the material are mainly maintained.
- ☐ The used filter is a Butterworth low-pass filter with order 4, and $2*fm1/(fs/2)$ is the cut-off frequency.
- ☐ However, the receiver has successfully separated $m_1(t)$ and $m_2(t)$ due to the orthogonality of the carriers used in modulation and demodulation.

- Demodulated Signals:



4- Receiver with phase offset ($\cos(2\pi f_c t + \pi/3)$):

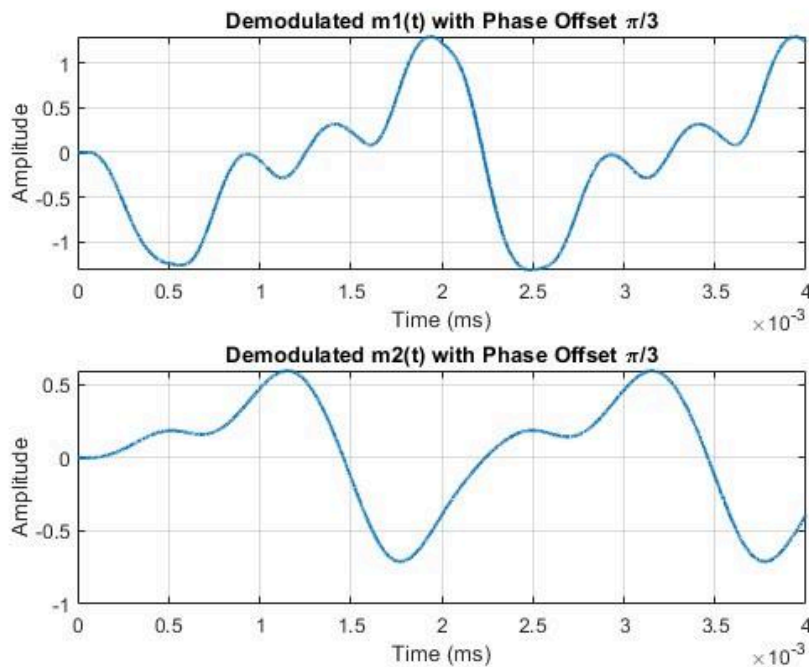
- Matlab code:

```
%% POINT 4: Phase Offset at Receiver ( $\pi/3$ )
% New receiver carrier with phase offset
rc1_offset = 2 * cos(2*pi*fc*t + pi/3);
rc2_offset = 2 * sin(2*pi*fc*t + pi/3);
% Multiply the modulated signal with the phase-shifted carrier
x1_offset = s .* rc1_offset;
x2_offset = s .* rc2_offset;
% Apply the same filters
m1_phase_offset = filter(b1, a1, x1_offset);
m2_phase_offset = filter(b2, a2, x2_offset);
```

- ☐ The used filter is the same LPF used in part 3.

- This displays the output of the demodulator branch when $2\cos(2\pi f_c t + \pi/3)$ is multiplied by the local oscillator's phase error = $\pi/3$ with respect to the transmitter's carrier. The orthogonality between the carriers is lost as a result of the phase error. Theoretically, $m_1(t)\cos(\phi) - m_2(t)\sin(\phi)$ equals the output. The output, plus filtering effects, is approximately $0.5m_1(t) - 0.866m_2(t)$, since $\cos(\pi/3)=0.5$ and $\sin(\pi/3)=0.866$. Since the output is $0.886m_1(t)+0.5m_2(t)$, this explains the significant distortion and the emergence of features from $m_2(t)$ in the $m_1(t)$ plot and the same in $m_2(t)$ plot. This is referred to as crosstalk.

- Demodulated signals figure using phase offset:

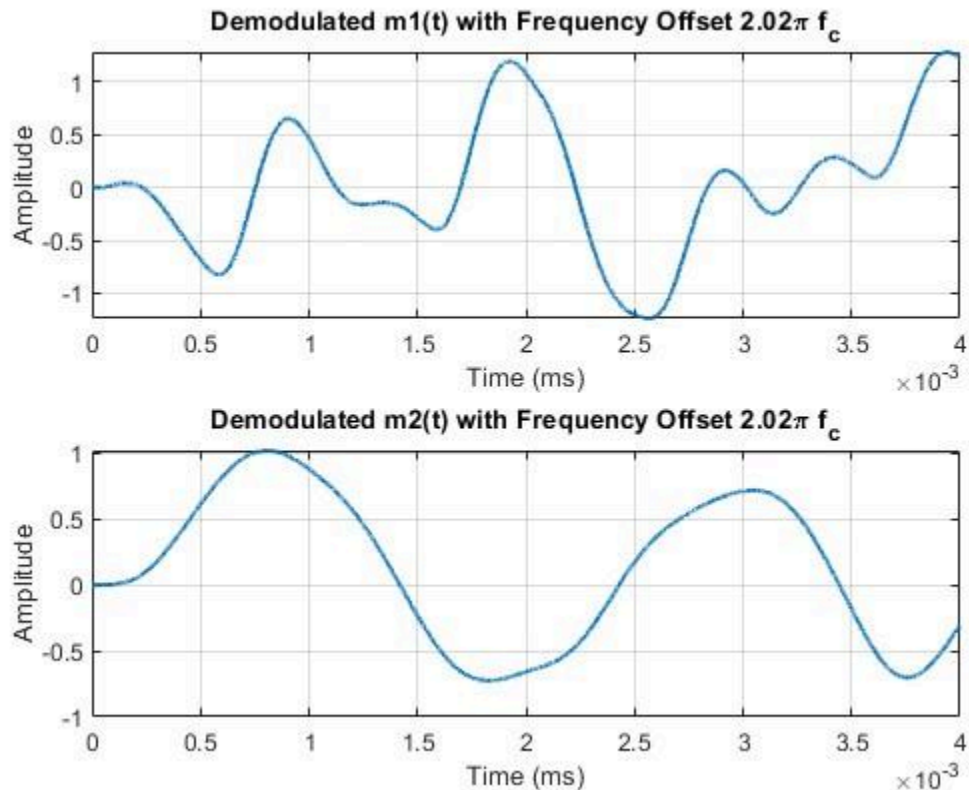


5- Receiver with frequency offset ($\cos(2.02 \pi f_c t)$):

- Matlab code:


```
%% POINT 5: Frequency Offset at Receiver (2.02πfat)
fc_offset = 1.01 * fc; % 1% frequency offset
rc1_freq_offset = 2 * cos(2*pi*fc_offset*t);
rc2_freq_offset = 2 * sin(2*pi*fc_offset*t);
% Multiply modulated signal with frequency offset carrier
x1_freq_offset = s .* rc1_freq_offset;
x2_freq_offset = s .* rc2_freq_offset;
% Filter to demodulate
m1_freq_offset = filter(b1, a1, x1_freq_offset);
m2_freq_offset = filter(b2, a2, x2_freq_offset);
```

- The outcome of demodulating using a local carrier frequency that differs somewhat from the transmitter's carrier ($f_{\text{local}} = 1.01 * f_c$) is displayed in this plot. Δf is equal to $0.01 * f_c$, which is equal to $0.01 * 5 \text{ kHz}$, or 50 Hz . Terms such as $m_1(t)\cos(2\pi\Delta f t)$ and $m_2(t)\sin(2\pi\Delta f t)$ are present in the demodulated output. The target signal and the crosstalk component are multiplied by the slow sinusoidal variation (50 Hz) represented by the $\cos(2\pi\Delta f t)$ and $\sin(2\pi\Delta f t)$ terms. The original signal is totally jumbled by this time-varying strength and crosstalk, making recovery impossible.
- Demodulated signal using frequency offset:



Part B

❖ Matlab code for variables:

```
%% Time space for simulation
fc = 10e3; % Carrier frequency in Hz
fs = 1000 * fc; % Sampling frequency (10 MHz)
Ts = 1 / fs; % Sampling period
T = 0.004; % Total simulation time (4 ms)
t = 0:Ts:T; % Time vector
Ac = 1;
```

1- Message Signal Generation & plotting:

● Matlab code:

```
%% Generating the Message Signals

fm1 = 1000; % Frequency of m1 in Hz

Tm2 = 2e-3; % Period of m2 in seconds

fm2 = 1 / Tm2; % Frequency of m2 in Hz

m1 = sawtooth((2*pi*fm1*t) + pi, 0); % Sawtooth wave for m1

m2 = zeros(size(t)); % Step function for m2

shift = mod(t, Tm2); % Repeat every Tm2

% Step levels for m2

m2(shift < 0.5e-3) = 1;

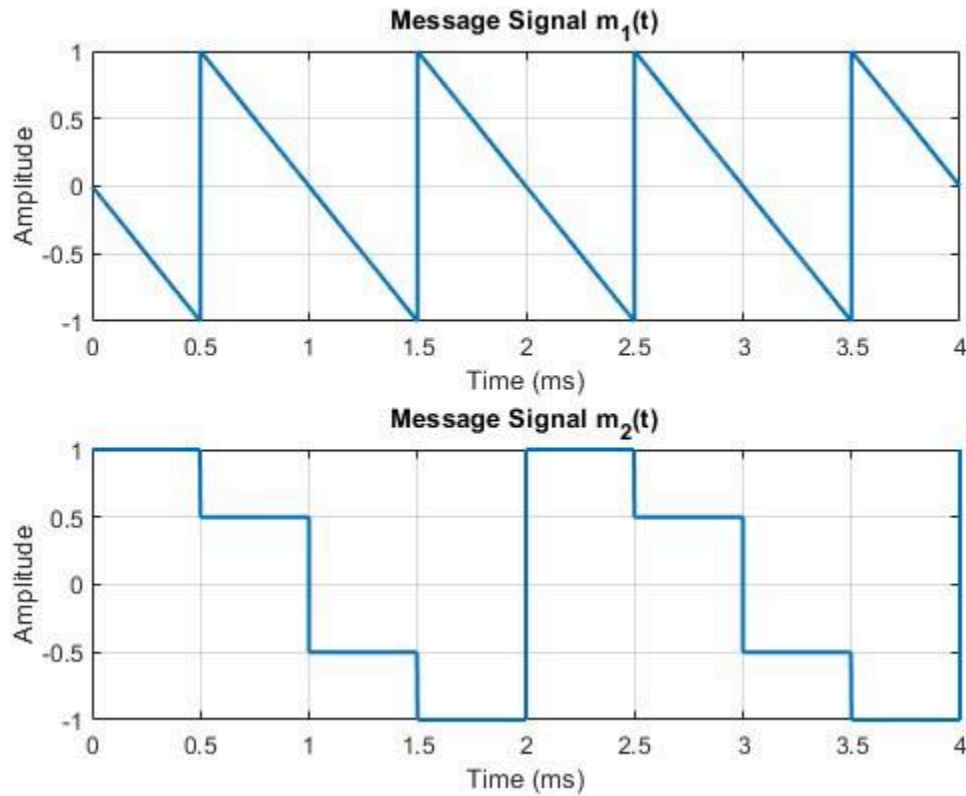
m2((shift >= 0.5e-3) & (shift < 1e-3)) = 0.5;

m2((shift >= 1e-3) & (shift < 1.5e-3)) = -0.5;

m2(shift >= 1.5e-3) = -1;
```

- $m_1(t)$ is a **triangular waveform** (sawtooth-type), periodic and continuous and $m_2(t)$ is a **piecewise constant signal** (digital-style steps).

- Generated message signals figure:



2- Phase Modulated Signal $s_1(t)$ with various k_p :

- Matlab code:

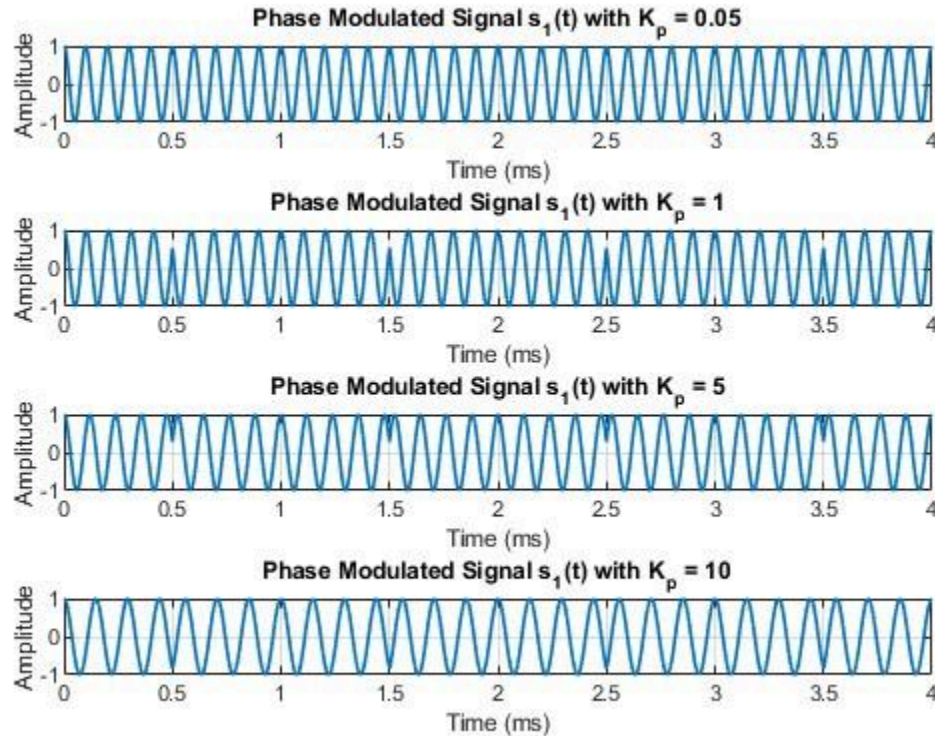
```
% Phase Modulation for different values of Kp
Kp_values = [0.05, 1, 5, 10];
figure;
for i = 1:length(Kp_values)
    Kp = Kp_values(i);
    s1 = Ac * cos(2*pi*fc*t + Kp .* m1); % Phase modulation

    subplot(length(Kp_values),1,i);
    plot(t*1000, s1, 'LineWidth', 1.2);
    title(['Phase Modulated Signal s_1(t) with K_p = ', num2str(Kp)]);
    xlabel('Time (ms)');
```

```
ylabel('Amplitude');  
grid on;  
end
```

- ☐ For $k_p=0.05$, the maximum phase deviation = $k_p * \max|m_1(t)| = 0.05 * 1 = 0.05$ radians, which is very small and characteristic of narrowband PM. This is making the signal visually similar to the carrier. The oscillations appear very regular, almost like an unmodulated carrier.
- ☐ For $k_p=1$, the maximum phase deviation is 1 radian (approx. 57.3 degrees). The phase variations are more significant. The instantaneous frequency deviations are larger, causing visible compression/expansion of cycles corresponding to the slope of $m_1(t)$.
- ☐ For $k_p=5$, the maximum phase deviation = 5 radians, or around 286 degrees,. This is referred to as Wideband PM. The instantaneous frequency changes dramatically as a result of the phase deviance. One can clearly see the optical distortion from a pure sinusoid.
- ☐ For $k_p = 10$, the maximum phase deviation = 10 radians (about 573 degrees). This PM is very wideband. More than one full cycle ($2\pi \approx 6.28$ rad) separates the phases. The sharp apparent variations in cycle density are caused by the extremely huge instantaneous frequency swings.

- Modulated signal using different Kp values figure:



3-Frequency Modulated Signal $s_2(t)$ using $m_2(t)$:

- $s_2(t) = A_c \cos(2\pi f_c t + k_f \int m_2(\tau) d\tau)$.
- Matlab code:**

```

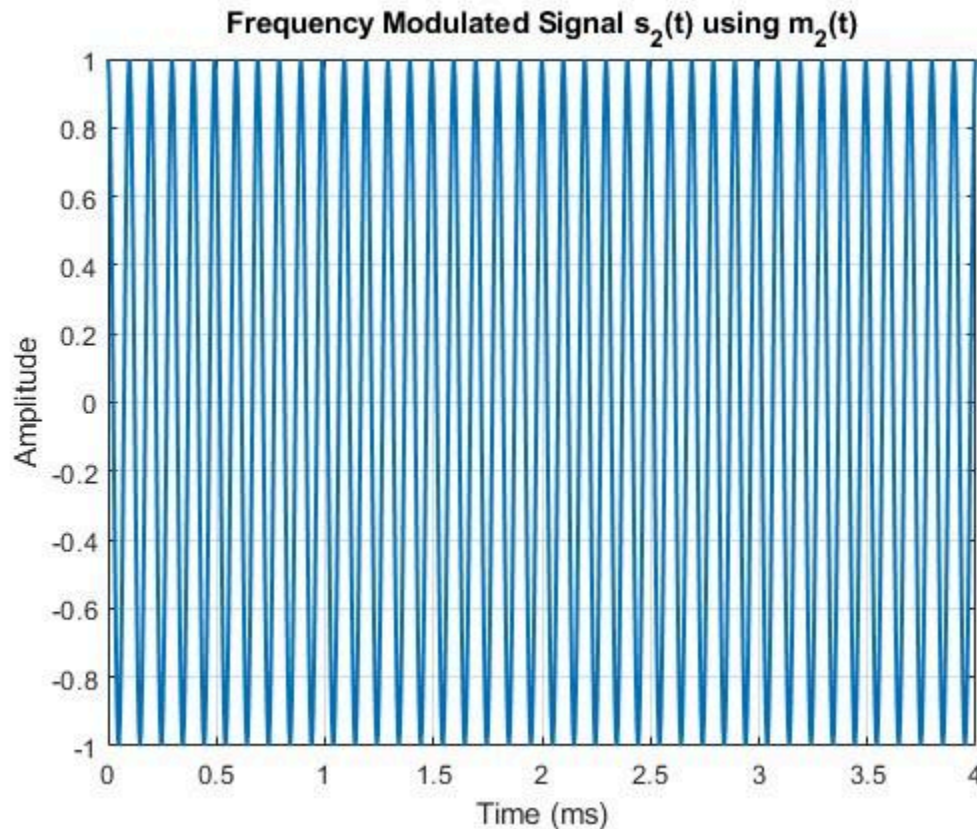
%% Frequency Modulation using m2(t), m1(t)
Kf = 1000; % Frequency sensitivity
Kf3 = 2000;
% Integrate m2(t) over time
int_m1 = cumsum(m1) * Ts;
int_m2 = cumsum(m2) * Ts; % Numerical integration
% Frequency modulated signal
s2 = Ac * cos(2*pi*fc*t + Kf * int_m2);
s3 = Ac * cos(2*pi*fc*t + Kf3 * int_m1);

```

- ☐ The instantaneous frequency is $f_i(t) = f_c + (k_f/2\pi) * m_2(t)$. The frequency deviation constant Δf_{unit} is equal to $k_f/2\pi \approx 159 \text{ Hz/Volt}$ in this case, where $f_c=10\text{kHz}$ and $k_f=1000$ (assuming units of rad/s/Volt).

- ☐ $F_i = 10000 + 159 * 1 \approx 10159$ Hz for $m_2(t) = +1$ (0-0.5ms).
- ☐ $F_i = 10000 + 159 * 0.5 \approx 10080$ Hz for $m_2(t) = +0.5$ (0.5-1ms).
- ☐ $F_i = 10000 + 159 * (-0.5) \approx 9920$ Hz when $m_2(t) = -0.5$ (1-1.5 ms).
- ☐ F_i is equal to $10000 + 159 * (-1) \approx 9841$ Hz for $m_2(t) = -1$ (1.5-2ms).
- ☐ Alternatively, frequencies would be 11 kHz, 10.5 kHz, 9.5 kHz, and 9 kHz if $k_f=1000$ Hz/Volt was the intended value.
- ☐ As the instantaneous frequency is precisely proportional to the amplitude of the message signal, the graphic effectively depicts FM via a staircase signal. The frequency changes to match constant values when $m_2(t)$ moves between constant levels; this is seen in the figure as varying cycle density.

- Frequency Modulated Signal $s_2(t)$ using $m_2(t)$ figure:



4- Frequency Modulated Signal $s_3(t)$ using $m_1(t)$:

- $s_3(t) = A_c \cos(2\pi f_c t + k_f \int m_1(\tau) d\tau)$.
- Matlab code:

```
%% Frequency Modulation using m2(t), m1(t)
Kf = 1000; % Frequency sensitivity
Kf3 = 2000;
% Integrate m2(t) over time
int_m1 = cumsum(m1) * Ts;
int_m2 = cumsum(m2) * Ts; % Numerical integration
% Frequency modulated signal
s2 = Ac * cos(2*pi*f_c*t + Kf * int_m2);
s3 = Ac * cos(2*pi*f_c*t + Kf3 * int_m1);
```

- ☐ The shape of the wave $m_1(t)$, which varies linearly throughout linear ramps of $m_1(t)$, determines the instantaneous frequency, $f_i(t)$. The lowest frequency is 9682 Hz, and the highest frequency is $10000 + 318 * 1$. With continuous changes in $m_1(t)$ translating into continuous changes in the instantaneous frequency, the graphic effectively illustrates FM with the sawtooth wave $m_1(t)$. At the peak of $m_1(t)$, the frequency is at its highest, and at the minimum of $m_1(t)$, it is at its lowest.

- Frequency Modulated Signal $s_3(t)$ using $m_1(t)$ figure:

