Ch10: Centralized Control - Motion Control - Part 1

Amin Fakhari, Spring 2024 P2

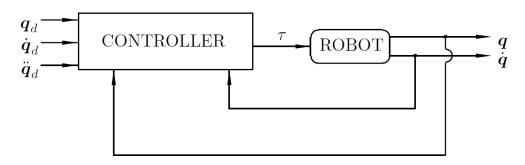
Motion Control:PD/PIDPD with FeedforwardPD with Gravity CompensationPD+Inverse Dynamic ControlT-Space ControlOOOOOOOOOOOOOOOOOOOOOStore Unit

Motion Control Objective

Given desired joint position $q_d(t) \in \mathbb{R}^n$, velocity $\dot{q}_d(t) \in \mathbb{R}^n$, and acceleration $\ddot{q}_d(t) \in \mathbb{R}^n$, we wish to find joint torques/forces $\tau \in \mathbb{R}^n$ such that the joint position $q(t) \in \mathbb{R}^n$ follow (asymptotically) $q_d(t)$ accurately:

$$\lim_{t \to \infty} \mathbf{q}(t) = \mathbf{q}_d(t) \quad \Rightarrow \quad \lim_{t \to \infty} \mathbf{e}(t) = \mathbf{0}$$

$$e(t) = q_d(t) - q(t) \in \mathbb{R}^n$$
 position error



$$\dot{\boldsymbol{e}}(t) = \dot{\boldsymbol{q}}_d(t) - \dot{\boldsymbol{q}}(t) \in \mathbb{R}^n$$
 velocity error

The most common motion controllers:

- PD/PID Control
- PD with Feedforward Control
- PD Control with Gravity Compensation
- PD+ Control
- Inverse Dynamic Control

PD/PIDPD with Feedforward
OOPD with Gravity Compensation
OPD+Inverse Dynamic Control
OOT-Space Control
OO



Motion Control:

Stony Brook University

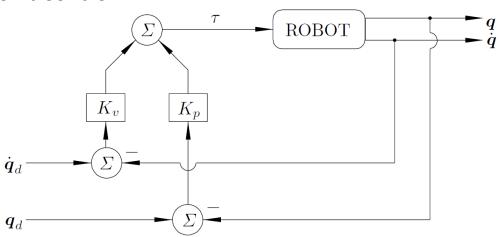
PD Control

The PD control law is given by $au = K_p e + K_v \dot{e}$

$$e = q_d - q$$

 K_p , $K_v \in \mathbb{R}^{n \times n}$ are symmetric positive definite matrices. If $K_p = \text{diag}\{K_{p,i}\}$, $K_v = \text{diag}\{K_{v,i}\}$, the controller is called PD Independent Joint Control.

This is the simplest (linear) controller that may be used to control robot manipulators.



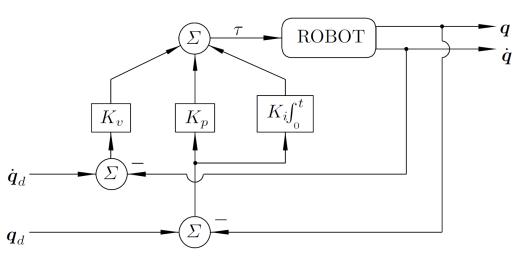
- For any $K_p = K_p^T > 0$, $K_v = K_v^T > 0$, it is guaranteed that e(t) and $\dot{e}(t)$ are bounded for all initial conditions.
- The error bound decreases, as $K_{v,i}$ become larger (in case $K_v = \text{diag}\{K_{v,i}\}$), however, large $K_{v,i}$ can saturate the robot actuators.

PID Control

The residual error at steady state <u>due to gravity</u> with PD control can be removed to some extend using the PID control law which is given by

$$\boldsymbol{\tau} = \boldsymbol{K}_{p}\boldsymbol{e} + \boldsymbol{K}_{v}\dot{\boldsymbol{e}} + \boldsymbol{K}_{i}\int_{0}^{t}\boldsymbol{e}(\tau)d\tau \qquad \boldsymbol{e} = \boldsymbol{q}_{d} - \boldsymbol{q}$$

 $K_p, K_v, K_i \in \mathbb{R}^{n \times n}$ (position, velocity, and integral gains) are symmetric positive definite matrices. If $K_p = \text{diag}\{K_{p,i}\}$, $K_v = \text{diag}\{K_{v,i}\}$, $K_i = \text{diag}\{K_{i,i}\}$, the controller is called PID Independent Joint Control.



PD/PID PD with Feedforward PD with Gravity Compensation PD+ Inverse Dynamic Control T-Space Control OO OOOOOOO

PD with Feedforward Control

PD with Feedforward Control

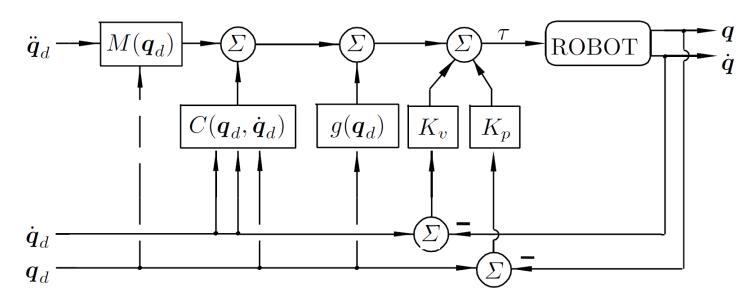
The PD with Feedforward control law is given by

$$\tau = K_p e + K_v \dot{e} + M(q_d) \ddot{q}_d + C(q_d, \dot{q}_d) \dot{q}_d + g(q_d)$$

 $e = q_d - q$

For reducing the number of computations in real time implementation.

 K_p , $K_v \in \mathbb{R}^{n \times n}$ are symmetric positive definite matrices.



Inverse Dynamic Control

PD with Feedforward Control (cont.)

The closed-loop dynamic equation is derived as

$$\begin{split} M(q)\ddot{q} + C(q,\dot{q})\dot{q} + g(q) &= K_{p}e + K_{v}\dot{e} + M(q_{d})\ddot{q}_{d} + C(q_{d},\dot{q}_{d})\dot{q}_{d} + g(q_{d}) \\ \frac{d}{dt} \begin{bmatrix} e \\ \dot{e} \end{bmatrix} &= \begin{bmatrix} \dot{e} \\ M(q)^{-1} [-K_{p}e - K_{v}\dot{e} - C(q,\dot{q})\dot{e} - h(t,e,\dot{e})] \end{bmatrix} \end{split}$$

$$h(t,e,\dot{e}) = [M(q_{d}) - M(q)]\ddot{q}_{d} + [C(q_{d},\dot{q}_{d}) - C(q,\dot{q})]\dot{q}_{d} + [g(q_{d}) - g(q)]$$

We can show that

- Origin $(e, \dot{e}) = \mathbf{0} \in \mathbb{R}^{2n}$ is an equilibrium, independently of the gain matrices K_p , K_v .
- The number of equilibria of the system depends on the proportional gain \pmb{K}_p .
- By choosing K_p sufficiently large, then the system has a unique equilibrium at origin.
- By choosing K_p , K_v sufficiently large, the origin is globally uniformly asymptoticly stable.



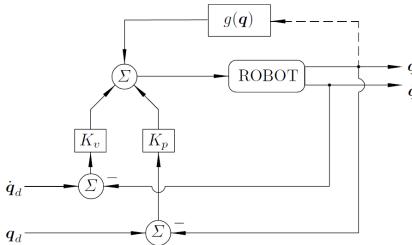
PD Control with Gravity Compensation

The PD control law with gravity compensation is given by

$$au = K_p e + K_v \dot{e} + g(q)$$

$$e = q_d - q$$

 K_p , $K_v \in \mathbb{R}^{n \times n}$ are symmetric positive definite matrices.



- In general, if q_d is not constant (i.e., motion control), then the controller guarantees bounded tracking errors e(t) about zero, but the error never goes exactly to zero.
- The error bound decreases, as the PD gains become larger (in case $K_p = \text{diag}\{K_{p,i}\}$, $K_v = \text{diag}\{K_{v,i}\}$).
- PD Controller with Gravity Compensation may behave batter than a PID controller for the same PD gains. However, the integral term can reject other terms besides gravity (e.g., friction of some sorts).

PD+ Control

Motion Control:

Stony Brook University

PD+

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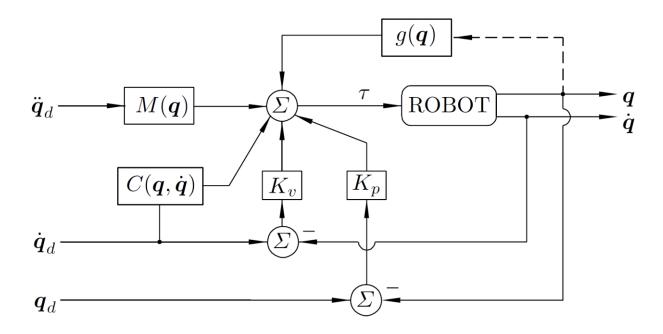
PD+ Control

The PD+ control law is given by

$$\tau = K_p e + K_v \dot{e} + M(q) \ddot{q}_d + C(q, \dot{q}) \dot{q}_d + g(q)$$

$$e = q_d - q$$

 K_p , $K_v \in \mathbb{R}^{n \times n}$ are symmetric positive definite matrices.



PD+ Control (cont.)

The closed-loop dynamic equation is derived as $M(q)\ddot{e} + C(q,\dot{q})\dot{e} = -K_pe - K_v\dot{e}$

$$\frac{d}{dt} \begin{bmatrix} \mathbf{e} \\ \dot{\mathbf{e}} \end{bmatrix} = \begin{bmatrix} \dot{\mathbf{e}} \\ \mathbf{M}(\mathbf{q})^{-1} [-\mathbf{K}_p \mathbf{e} - \mathbf{K}_v \dot{\mathbf{e}} - \mathbf{C}(\mathbf{q}, \dot{\mathbf{q}}) \dot{\mathbf{e}}] \end{bmatrix} \qquad q = q_d - \mathbf{e}$$

PD+

The system is **nonautonomous**, and origin $(e, \dot{e}) = 0 \in \mathbb{R}^{2n}$ is the only equilibrium point.

Consider a Lyapunov function candidate as $V(t, \boldsymbol{e}, \dot{\boldsymbol{e}}) = \frac{1}{2} \dot{\boldsymbol{e}}^T \boldsymbol{M}(\boldsymbol{q}) \dot{\boldsymbol{e}} + \frac{1}{2} \boldsymbol{e}^T \boldsymbol{K}_p \boldsymbol{e} > 0$.

$$\dot{V}(t, \boldsymbol{e}, \dot{\boldsymbol{e}}) = \dot{\boldsymbol{e}}^T \boldsymbol{M}(\boldsymbol{q}) \ddot{\boldsymbol{e}} + \frac{1}{2} \dot{\boldsymbol{e}}^T \dot{\boldsymbol{M}}(\boldsymbol{q}) \dot{\boldsymbol{e}} + \boldsymbol{e}^T \boldsymbol{K}_p \dot{\boldsymbol{e}}$$

Using closed-loop dynamic equation, and
$$\dot{m{e}}^T \left[rac{1}{2} \dot{m{M}}(m{q}) - m{C}(m{q}, \dot{m{q}})
ight] \dot{m{e}} = m{0}$$

$$\dot{V}(t, \boldsymbol{e}, \dot{\boldsymbol{e}}) = -\dot{\boldsymbol{e}}^T \boldsymbol{K}_v \dot{\boldsymbol{e}} \leq 0$$

Thus, the origin $(e, \dot{e}) = 0$ is stable.

- Using more advance theorems (e.g., Matrosov's theorem) or a different Lyapunov function, we can show that the origin is globally uniformly asymptoticly stable.

Inverse Dynamic Control

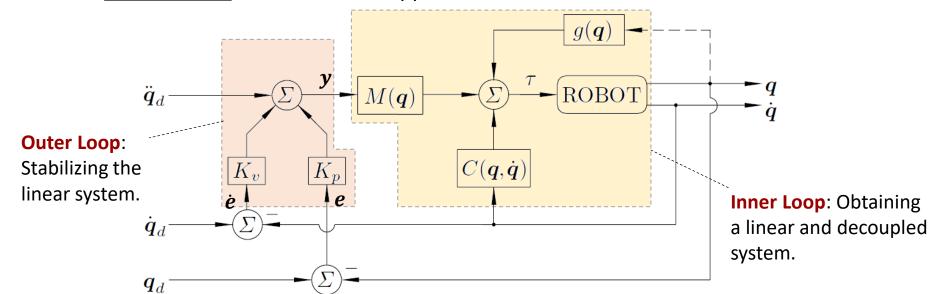
Inverse Dynamic Control (or Computed Torque Control)

The idea of Inverse Dynamics control is to seek a nonlinear feedback control law which results in a linear closed-loop system. The control law is given by

$$au=M(q)y+C(q,\dot{q})\dot{q}+g(q)$$
 $e=q_d-q$ $y=\ddot{q}_d+K_v\dot{e}+K_pe$ ($ightarrow$ PD control with feedforward acceleration)

 K_p , $K_v \in \mathbb{R}^{n \times n}$ are symmetric positive definite matrices.

This is a model-based motion control approach.



Inverse Dynamic Control

The closed-loop dynamic equation is derived as

$$M(q)\ddot{q} + C(q,\dot{q})\dot{q} + g(q) = M(q)y + C(q,\dot{q})\dot{q} + g(q) \longrightarrow M(q)\ddot{q} = M(q)y$$

$$\stackrel{(n \text{ uncoupled linear } \ddot{q} = y \longrightarrow \ddot{q} = \ddot{q}_d + K_v\dot{e} + K_pe$$

$$\ddot{e} + K_v\dot{e} + K_pe = 0 \longrightarrow \frac{d}{dt}\begin{bmatrix} e \\ \dot{e} \end{bmatrix} = \begin{bmatrix} \dot{e} \\ -K_ve - K_v\dot{e} \end{bmatrix} = \begin{bmatrix} 0 & I_n \\ -K_v & -K_v \end{bmatrix}\begin{bmatrix} e \\ \dot{e} \end{bmatrix}$$

The system is **linear** and **autonomous**, and the origin $(e, \dot{e}) = \mathbf{0} \in \mathbb{R}^{2n}$ is the unique equilibrium point.

Let's introduce the constant $\varepsilon \in \mathbb{R}_{++}$ satisfying $\lambda_{\min}(K_v) > \varepsilon > 0$, thus, $\lambda_{\min}(K_v)x^Tx > \varepsilon x^Tx$, and since $\lambda_{\min}(K_v)x^Tx \leq x^TK_vx$, then $x^T(K_v - \varepsilon I_n)x > 0 \ \forall x \neq 0 \in \mathbb{R}^n$. This means $K_v - \varepsilon I_n > 0$ and $K_p + \varepsilon K_v - \varepsilon^2 I_n > 0$. Now, a Lyapunov function candidate is

$$V(\boldsymbol{e}, \dot{\boldsymbol{e}}) = \frac{1}{2} \begin{bmatrix} \boldsymbol{e} \\ \dot{\boldsymbol{e}} \end{bmatrix}^T \begin{bmatrix} \boldsymbol{K}_p + \varepsilon \boldsymbol{K}_v & \varepsilon \boldsymbol{I}_n \\ \varepsilon \boldsymbol{I}_n & \boldsymbol{I}_n \end{bmatrix} \begin{bmatrix} \boldsymbol{e} \\ \dot{\boldsymbol{e}} \end{bmatrix} = \frac{1}{2} [\dot{\boldsymbol{e}} + \varepsilon \boldsymbol{e}]^T [\dot{\boldsymbol{e}} + \varepsilon \boldsymbol{e}] + \frac{1}{2} \boldsymbol{e}^T [\boldsymbol{K}_p + \varepsilon \boldsymbol{K}_v - \varepsilon^2 \boldsymbol{I}_n] \boldsymbol{e} > 0$$

$$\Rightarrow V(\boldsymbol{e}, \dot{\boldsymbol{e}}) = \frac{1}{2} \dot{\boldsymbol{e}}^T \dot{\boldsymbol{e}} + \frac{1}{2} \boldsymbol{e}^T [\boldsymbol{K}_p + \varepsilon \boldsymbol{K}_v] \boldsymbol{e} + \varepsilon \boldsymbol{e}^T \dot{\boldsymbol{e}} > 0$$

Inverse Dynamic Control

$$\dot{V}(\boldsymbol{e}, \dot{\boldsymbol{e}}) = \ddot{\boldsymbol{e}}^T \dot{\boldsymbol{e}} + \boldsymbol{e}^T [\boldsymbol{K}_p + \varepsilon \boldsymbol{K}_v] \dot{\boldsymbol{e}} + \varepsilon \dot{\boldsymbol{e}}^T \dot{\boldsymbol{e}} + \varepsilon \boldsymbol{e}^T \ddot{\boldsymbol{e}} \qquad \overset{\ddot{\boldsymbol{e}} + \boldsymbol{K}_v \dot{\boldsymbol{e}} + \boldsymbol{K}_p \boldsymbol{e} = \boldsymbol{0}}{\Longrightarrow}$$

$$\Rightarrow \dot{V}(\boldsymbol{e}, \dot{\boldsymbol{e}}) = -\dot{\boldsymbol{e}}^T [\boldsymbol{K}_v - \varepsilon \boldsymbol{I}_n] \dot{\boldsymbol{e}} - \varepsilon \boldsymbol{e}^T \boldsymbol{K}_p \boldsymbol{e} < 0$$

Thus, the origin $(e, \dot{e}) = \mathbf{0}$ is globally asymptotically stable for any initial condition $q(0), \dot{q}(0) \in \mathbb{R}^n$:

$$\lim_{t\to\infty} \mathbf{e}(t) = \mathbf{0} \qquad \lim_{t\to\infty} \dot{\mathbf{e}}(t) = \mathbf{0}$$

 \Rightarrow Thus, the motion control objective is achieved.

Note: Since the closed-loop equation is linear and autonomous, the origin is globally exponentially stable.

Note: Friction at the joints may also affect the position error. Moreover, in the presence of bounded disturbance $\tau_{\text{dist}}(t)$, error e(t) remains bounded.

Note: This controller is a special case of the method of <u>feedback linearization</u> for nonlinear systems.

Inverse Dynamic Control: Parameter Selection

$${\it K}_p$$
 and ${\it K}_v$ may be chosen diagonal as:

$$K_p = \text{diag}\{K_{p,i}\} = \text{diag}\{\omega_{n,i}^2\}$$

 $K_v = \text{diag}\{K_{v,i}\} = \text{diag}\{2\zeta_i\omega_{n,i}\}$

With this choice, the closed-loop equation is n decoupled 2nd-order linear ODEs. The natural frequency $\omega_{n,i} \in \mathbb{R}$ determines the <u>speed of response</u> (the larger, the faster) and the damping ratio $\zeta_i \in \mathbb{R}$ determines the <u>existence of overshoot</u> in joint error e(t).

Note 1: It may be useful to select the desired responses at the end of the arm faster than near the base, where the masses that must be moved are heavier.

Note 2: It is undesirable for the robot to exhibit overshoot (e.g., since this could cause impact for paths near the workpiece surface). Therefore, the damping ratios are usually selected $\zeta_i=1$ to have a critically damped responses.

Note 3: If the gains $K_{p,i}$, $K_{v,i}$ are too large, the control torque may reach its upper limits and saturate some or all of the actuators.

Note 4: The implementation of this control scheme requires the real-time computation of M(q) and $C(q, \dot{q})$, and g(q), which may be computationally expensive.

Inverse Dynamic Control: Example

Consider the equation of a pendulum of length l and mass m concentrated at its tip, subject to the action of gravity g and to which is applied a torque τ at the axis of rotation. Drive the inverse dynamic control law.

$$ml^2\ddot{\theta} + mgl\sin\theta = \tau$$



Approximate Inverse Dynamic Control

In some cases, M(q) is not known exactly (e.g., unknown payload mass), or $h(q,\dot{q})=C(q,\dot{q})\dot{q}+g(q)$ is not known exactly (e.g., unknown friction terms). Then $\widehat{M}(q)$ and $\widehat{h}(q,\dot{q})$ could be the best estimate we have for these terms. On the other hand, we might simply wish to avoid computing M(q) and $h(q,\dot{q})$ at each short sample time and instead compute more simpler $\widehat{M}(q)$ and $\widehat{h}(q,\dot{q})$. The approximate inverse dynamic control law is given by

$$\tau = \widehat{M}(q)y + \widehat{h}(q, \dot{q})$$

$$y = \ddot{q}_d + K_v \dot{e} + K_p e$$

$$e = q_d - q$$

It can be shown that even if $\widehat{M} \neq M$ and $\widehat{h} \neq h$, the performance of the controller can be <u>acceptable</u> if the symmetric positive definite matrices $K_p, K_v \in \mathbb{R}^{n \times n}$ are selected large enough.

PID Inverse Dynamic Control

In the presence of <u>unknown constant disturbances</u> ($\tau_{\rm dist}$), PD control gives a nonzero steady-state error. Thus, we by including an integrator (I) in the outer loop, we can achieve a PID inverse dynamic controller as

$$\tau = M(q)y + C(q, \dot{q})\dot{q} + g(q)$$

$$y = \ddot{q}_d + K_v\dot{e} + K_pe + K_i \int_0^t e(\tau)d\tau$$

$$e = q_d - q$$

 K_p , K_v , $K_i \in \mathbb{R}^{n \times n}$ are symmetric positive definite matrices.

The closed-loop dynamic equation is derived as

$$\ddot{e} + K_{v}\dot{e} + K_{p}\dot{e} + K_{i}e = 0$$

Note: If control gains are diagonal ($K_p = \text{diag}\{K_{p,i}\}$, $K_v = \text{diag}\{K_{v,i}\}$, $K_i = \text{diag}\{K_{i,i}\}$), for closed-loop stability, based on Routh-Hurwitz criterion, we require that

$$K_{i,i} < K_{p,i}K_{v,i}$$

Task Space Control

Task Space Control

Since the robot interacts with the external environment and objects in it, it may be more convenient to express the motion as a trajectory of the end-effector in task space. If the end-effector trajectory be specified by $x_d(t) \in \mathbb{R}^m$ or $T_d(t) \in SE(3)$, $V_d(t) \in \mathbb{R}^6$:

Method 1: Converting a desired trajectory in task space to joint-space and proceed with joint-space control.

$$\begin{cases} \boldsymbol{q}_{d}(t) = \boldsymbol{f}^{-1}\big(\boldsymbol{x}_{d}(t)\big) \\ \dot{\boldsymbol{q}}_{d}(t) = \bar{\boldsymbol{f}}^{-1}\big(\dot{\boldsymbol{x}}_{d}(t)\big) \\ \ddot{\boldsymbol{q}}_{d}(t) = \bar{\boldsymbol{f}}^{-1}\big(\ddot{\boldsymbol{x}}_{d}(t)\big) \end{cases} \quad \text{or} \quad \begin{cases} \boldsymbol{q}_{d}(t) = \boldsymbol{T}^{-1}\big(\boldsymbol{T}_{d}(t)\big) \\ \dot{\boldsymbol{q}}_{d}(t) = \boldsymbol{J}^{\dagger}\big(\boldsymbol{q}_{d}(t)\big)\boldsymbol{\mathcal{V}}_{d}(t) \\ \ddot{\boldsymbol{q}}_{d}(t) = \boldsymbol{J}^{\dagger}\big(\boldsymbol{q}_{d}(t)\big)\big(\dot{\boldsymbol{\mathcal{V}}}_{d}(t) - \dot{\boldsymbol{J}}\big(\boldsymbol{q}_{d}(t)\big)\dot{\boldsymbol{q}}_{d}(t)\big) \end{cases}$$

Drawback: This requires significant computing power. To reduce the computational load, we can first compute $q_d(t)$, then perform a numerical differentiation to compute $\dot{q}_d(t)$ and $\ddot{q}_d(t)$.

Task Space Control (cont.)

Method 2: Developing a control law in the task space using the robot dynamics expressed either in <u>joint space</u> or <u>task space</u>.

$$\tau = M(q)\ddot{q} + C(q, \dot{q})\dot{q} + g(q)$$

$$\tau = M_{C}(q)\dot{V} + C_{C}(\theta, V)V + g_{C}(q)$$

$$\tau = J^{T}(q)\mathcal{F}$$

$$(or)$$

$$F = M_{C}(q)\ddot{x} + C_{C}(q, \dot{x})\dot{x} + g_{C}(q)$$

$$\tau = J_{a}^{T}(q)F$$

Difficulties:

- Task space controllers always require computation of manipulator Jacobian. Thus, the presence of **singularities** and/or **redundancy** influences the Jacobian, and the induced effects are somewhat difficult to handle with a task space controller (e.g., we must use Jacobian pseudoinverse or other redundancy handling techniques).
- Expressing the joint limits in easier in joint space than task space.
- Here, let's consider a nonredundant manipulator avoiding singularities to develop the control laws.

Position Control: PD Control with Gravity Compensation

(Based on Robot Dynamics in J-Space & Min. Coord. Rep. of EE Frame)

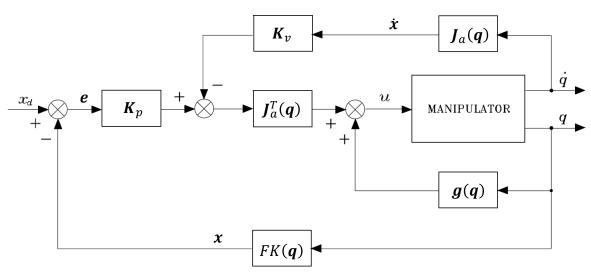
Given a constant end-effector pose x_d , the PD control law with gravity compensation is given by

$$\tau = J_a^T(q) (K_p e - K_v \dot{x}) + g(q)$$

$$e = x_d - x$$

$$\dot{x} = J_a(q) \dot{q}$$

 K_p , $K_v \in \mathbb{R}^{n \times n}$ are symmetric positive definite matrices.



Note: If measurements of x and \dot{x} are made directly in the task space, FK(q) and $J_a(q)$ are not required; however, it is necessary to measure q to update both $J_a^T(q)$ and g(q) on-line.

Position Control: PD Control with Gravity Compensation

(Based on Robot Dynamics in J-Space & Min. Coord. Rep. of EE Frame) (cont.)

The closed-loop dynamic equation is derived as

$$M(q)\ddot{q} + C(q,\dot{q})\dot{q} + g(q) = J_a^T(q)K_pe - J_a^T(q)K_vJ_a(q)\dot{q} + g(q)$$

$$\frac{d}{dt} \begin{bmatrix} \mathbf{e} \\ \dot{\mathbf{q}} \end{bmatrix} = \begin{bmatrix} -\dot{\mathbf{x}} \\ \mathbf{M}(\mathbf{q})^{-1} (J_a^T(\mathbf{q}) K_p \mathbf{e} - J_a^T(\mathbf{q}) K_v J_a(\mathbf{q}) \dot{\mathbf{q}} - C(\mathbf{q}, \dot{\mathbf{q}}) \dot{\mathbf{q}} \end{bmatrix} \qquad \mathbf{x} = \mathbf{x}_d - \mathbf{e}$$

The system is **autonomous** (since x_d is constant), and it has a unique equilibrium point at origin $(e, \dot{q}) = 0 \in \mathbb{R}^{2n}$.

Consider a Lyapunov function candidate as

$$V(\dot{\boldsymbol{q}}, \boldsymbol{e}) = \frac{1}{2} \dot{\boldsymbol{q}}^T \boldsymbol{M}(\boldsymbol{q}) \dot{\boldsymbol{q}} + \frac{1}{2} \boldsymbol{e}^T \boldsymbol{K}_P \boldsymbol{e} > 0 \quad (PD)$$

Kinetic energy of the arm

$$\dot{V}(\dot{q}, e) = \dot{q}^T M(q) \ddot{q} + \frac{1}{2} \dot{q}^T \dot{M}(q) \dot{q} + \dot{e}^T K_P e$$

$$\dot{\boldsymbol{q}}^T \left[\frac{1}{2} \dot{\boldsymbol{M}} - \boldsymbol{C} \right] \dot{\boldsymbol{q}} = \mathbf{0}$$

$$\dot{V}(\dot{q}, e) = -\dot{x}^T K_v \dot{x} \le 0 \qquad \Rightarrow$$

Using LaSalle (invariant set) theorem, the origin $(e, \dot{q}) = 0$ is globally asymptotically stable.

$$\lim_{t\to\infty} \boldsymbol{e}(t) = \mathbf{0}$$

Motion Control: Inverse Dynamic Control

(Based on Robot Dynamics in J-Space & Min. Coord. Rep. of EE Frame)

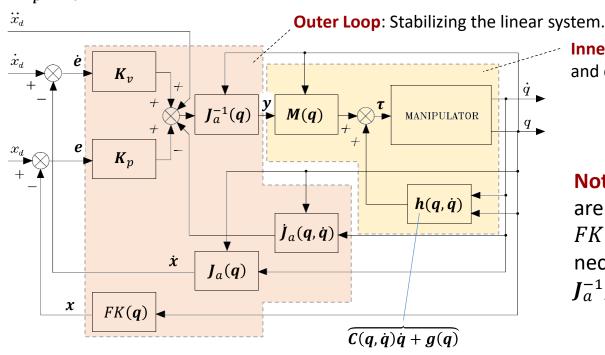
The inverse dynamic control law is given by

$$\tau = M(q)y + C(q, \dot{q})\dot{q} + g(q)$$

$$y = J_a^{-1}(q)(\ddot{x}_d + K_v\dot{e} + K_pe - \dot{J}_a(q, \dot{q})\dot{q})$$

$$e = x_d - x$$

 K_p , $K_v \in \mathbb{R}^{n \times n}$ are symmetric positive definite matrices.



Inner Loop: Obtaining a linear and decoupled system.

Note: If measurements of x and \dot{x} are made directly in the task space, FK(q) is not required; however, it is necessary to measure q, \dot{q} to update J_a^{-1} , \dot{J}_a , M, C, and g on-line.

Motion Control: Inverse Dynamic Control

(Based on Robot Dynamics in J-Space & Min. Coord. Rep. of EE Frame) (cont.)

The closed-loop dynamic equation is derived as

$$M(q)\ddot{q} + C(q,\dot{q})\dot{q} + g(q) = M(q) \left[J_a^{-1}(q) \left(\ddot{x}_d + K_v \dot{e} + K_p e - \dot{J}_a(q,\dot{q})\dot{q} \right) \right] + C(q,\dot{q})\dot{q} + g(q) \right]$$

$$\ddot{q} = J_a^{-1}(q) \left(\ddot{x}_d + K_v \dot{e} + K_p e - \dot{J}_a(q,\dot{q})\dot{q} \right) = y$$

$$\ddot{x} = J_a(q)\ddot{q} + \dot{J}_a(q,\dot{q})\dot{q}$$

$$\ddot{e} + K_v \dot{e} + K_p e = 0 \longrightarrow \frac{d}{dt} \begin{bmatrix} e \\ \dot{e} \end{bmatrix} = \begin{bmatrix} \dot{e} \\ -K_p e - K_v \dot{e} \end{bmatrix} = \begin{bmatrix} 0 & I_n \\ -K_p & -K_v \end{bmatrix} \begin{bmatrix} e \\ \dot{e} \end{bmatrix}$$

The system is **decoupled**, **linear** and **autonomous**, and the origin $(e, \dot{e}) = 0 \in \mathbb{R}^{2n}$ is the unique equilibrium point.

Similar to Inverse Dynamic Control in joint space, the origin $(e, \dot{e}) = 0$ is globally asymptotically (exponentially) stable for any initial condition $q(0), \dot{q}(0) \in \mathbb{R}^n$:

$$\lim_{t\to\infty} \boldsymbol{e}(t) = \mathbf{0}$$

A Remark on Computation of Error

$$e = \begin{bmatrix} e_R \\ e_p \end{bmatrix} = \begin{bmatrix} e_R \\ p_d - p \end{bmatrix}, \qquad \dot{e} = \begin{bmatrix} \dot{e}_R \\ \dot{e}_n \end{bmatrix} = \begin{bmatrix} \dot{e}_R \\ \dot{p}_d - \dot{p} \end{bmatrix}$$

$$\dot{m{e}} = egin{bmatrix} \dot{m{e}}_R \ \dot{m{e}}_p \end{bmatrix} = egin{bmatrix} \dot{m{e}}_R \ \dot{m{p}}_d - \dot{m{p}} \end{bmatrix}$$

Computation of e_R , \dot{e}_R depends on the orientation representation of end-effector frame:

(1) Euler Angles: Method 1:
$$\boldsymbol{e}_R = \boldsymbol{\phi}_d - \boldsymbol{\phi}$$

$$e_R = \phi_d - \phi$$

 $\dot{\boldsymbol{e}}_{R}=\dot{\boldsymbol{\phi}}_{d}-\dot{\boldsymbol{\phi}}$

Method 2:
$$\mathbf{e}_R = \text{EulerAngles}(\mathbf{R}^T \mathbf{R}_d)$$

 $\boldsymbol{\phi} \in \mathbb{R}^3$

 $\mathbf{R} \in SO(3)$

Assumption: There is no kinematic or representation singularities.

(2) Angle and Axis (Exponential Coordinates):

Method 1:
$$\mathbf{R}_e = \mathbf{R}^T \mathbf{R}_d$$
 , $\log(\mathbf{R}_e) = [\widehat{\boldsymbol{\omega}}] \theta$, $\mathbf{e}_R \coloneqq \widehat{\boldsymbol{\omega}} \theta$ (in EE frame) $\mathbf{e}_R \coloneqq \mathbf{R} \widehat{\boldsymbol{\omega}} \theta$ (in base frame)

$$\log(\pmb{R}_e) = [\widehat{\pmb{\omega}}] \theta$$
 , (in EE frame)

$$oldsymbol{e}_R\coloneqq\widehat{oldsymbol{\omega}} heta$$
 (in EE frame)

$$oldsymbol{e}_R\coloneqq oldsymbol{R}\widehat{oldsymbol{\omega}} heta$$
 (in base frame)

A Remark on Computation of Error

Method 2:
$$\mathbf{R}_e = \mathbf{R}^T \mathbf{R}_d$$
, UnitQuat(\mathbf{R}_e) =
$$\begin{bmatrix} \cos \theta / 2 \\ \sin \theta / 2 \widehat{\boldsymbol{\omega}} \end{bmatrix} = \begin{bmatrix} q_0 \\ q_1 \\ q_2 \\ q_3 \end{bmatrix}$$
 $\mathbf{e}_R \coloneqq \begin{bmatrix} q_1 \\ q_2 \\ q_3 \end{bmatrix}$ (in EE frame)
$$\mathbf{e}_R \coloneqq \mathbf{R} \begin{bmatrix} q_1 \\ q_2 \\ q_3 \end{bmatrix}$$
 (in base frame)

• If ω and ω_d are measured/given in the base frame: $\dot{\boldsymbol{e}}_R\coloneqq \boldsymbol{\omega}_d-\boldsymbol{\omega}$

$$\dot{\boldsymbol{e}} = \begin{bmatrix} \dot{\boldsymbol{e}}_R \\ \dot{\boldsymbol{e}}_p \end{bmatrix} = \begin{bmatrix} \boldsymbol{\omega}_d - \boldsymbol{\omega} \\ \dot{\boldsymbol{p}}_d - \dot{\boldsymbol{p}} \end{bmatrix} = \begin{bmatrix} \boldsymbol{\omega}_d \\ \dot{\boldsymbol{p}}_d \end{bmatrix} - \begin{bmatrix} \boldsymbol{\omega} \\ \dot{\boldsymbol{p}} \end{bmatrix} = \begin{bmatrix} \boldsymbol{\omega}_d \\ \dot{\boldsymbol{p}}_d \end{bmatrix} - \boldsymbol{J}_g \dot{\boldsymbol{q}}$$

Motion Control: Inverse Dynamic Control

(Based on Robot Dynamics in T-Space)

Inverse dynamic control law when end-effector trajectory is specified by $\mathbf{T}_d(t) \in SE(3)$, $\mathbf{V}_d(t) \in \mathbb{R}^6$:

$$\tau = J^{\mathrm{T}}(q) \big(M_{C}(q) y + C_{C}(\theta, \mathcal{V}) \mathcal{V} + g_{C}(q) \big)$$

$$p = \frac{d}{dt} \big(\big[\mathrm{Ad}_{T^{-1}T_{d}} \big] \mathcal{V}_{d} \big) + K_{p}T_{e} + K_{d}\mathcal{V}_{e}$$

$$\downarrow \text{Configuration Error}$$

$$\mathsf{Feedforward\ acceleration}$$

$$\mathsf{Expressed\ in\ the\ actual\ EE}$$

$$\mathsf{frame}\ T.$$

$$\mathsf{Twist\ Error}$$

$$\mathsf{Expressed\ in\ the\ actual\ EE}$$

$$\mathsf{frame}\ T.$$

$$\mathsf{Expressed\ in\ the\ actual\ EE\ frame\ } T.$$

PID:
$$\mathbf{y} = \frac{d}{dt} ([\mathrm{Ad}_{\mathbf{T}^{-1}\mathbf{T}_d}] \mathbf{v}_d) + \mathbf{K}_p \mathbf{T}_e + \mathbf{K}_d \mathbf{v}_e + \mathbf{K}_i \int \mathbf{T}_e(t) dt$$