

Computer Exercise IV: Digital Modulation

I- Prelab Assignment:

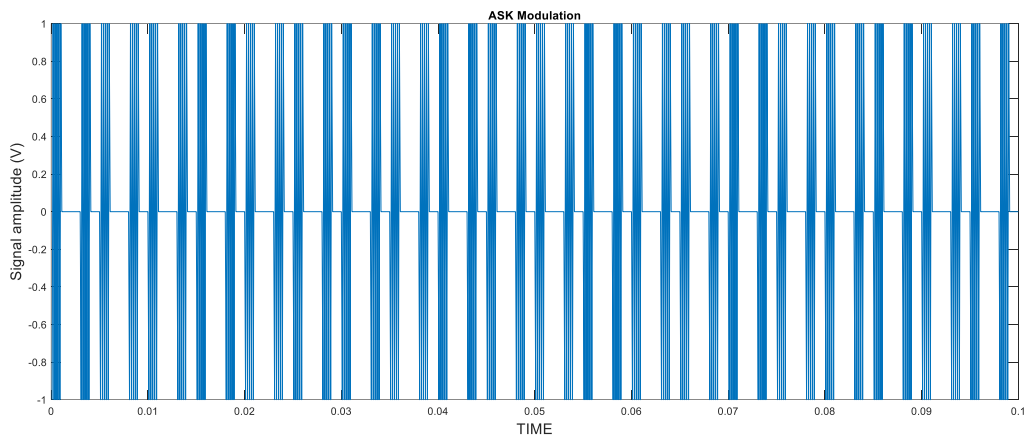
1.

$$b_i = 1, 0, 0, 1, 0 \quad \forall i = 1, \dots, 5$$

$$b(t) = b_i \quad 0 \leq t \leq 1/R_b$$

a)

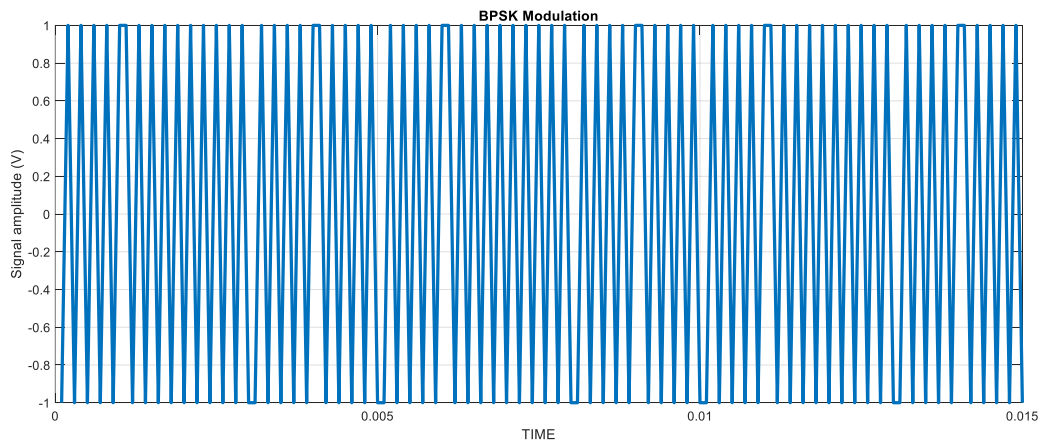
```
clear all
clc
b = [1 0 0 1 0]; Rb = 1000;
simulation = 100; e = 1;
time = 0.0001 : 0.0001 : simulation*(1/Rb);
for i = 1:simulation
    if e > length(b)
        e = 1;
    end
    Ask(1+(i-1)*10:i*10) = b(e)*cos(2*pi*(5e+3)*time(1+(i-1)*10:i*10));
    e = e + 1; % index of binary sequence
end
plot(time, Ask)
```



b)

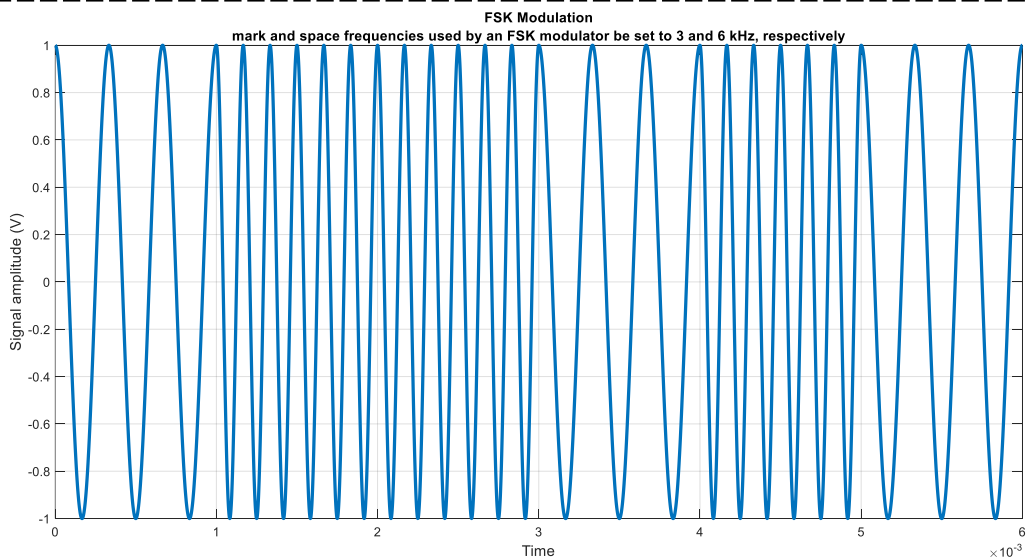
```
clear all
clc
b = [1 0 0 1 0]; Rb = 1000;
simulation = 15; e = 1;
time = 0.0001 : 0.0001 : simulation*(1/Rb);
for i = 1:simulation
    if e > length(b)
        e = 1;
    end
    Psk(1+(i-1)*10:i*10) = cos(2*pi*(5e+3)*time(1+(i-1)*10:i*10)+pi+pi*b(e));
end
```

```
e = e + 1;% index of binary sequence
end
plot(time,Psk)
```



c)

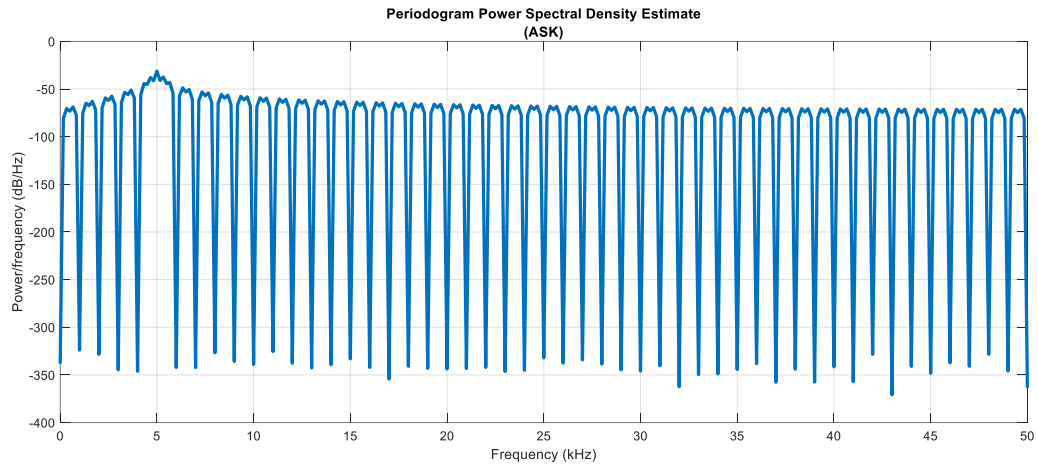
```
clear all
clc
b = [1 0 0 1 0];Rb = 1000;
simulation = 20;e = 1;
time = 0.000001 : 0.000001 : simulation*(1/Rb);
for i = 1:simulation
    if e > length(b)
        e = 1;
    end
    Fsk(1+(i-1)*1000:i*1000) = cos(2*pi*((6e+3)-b(e)*(3e+3))*...
    time(1+(i-1)*1000:i*1000));
    e = e + 1;% index of binary sequence
end
plot(time,Fsk)
```



2.

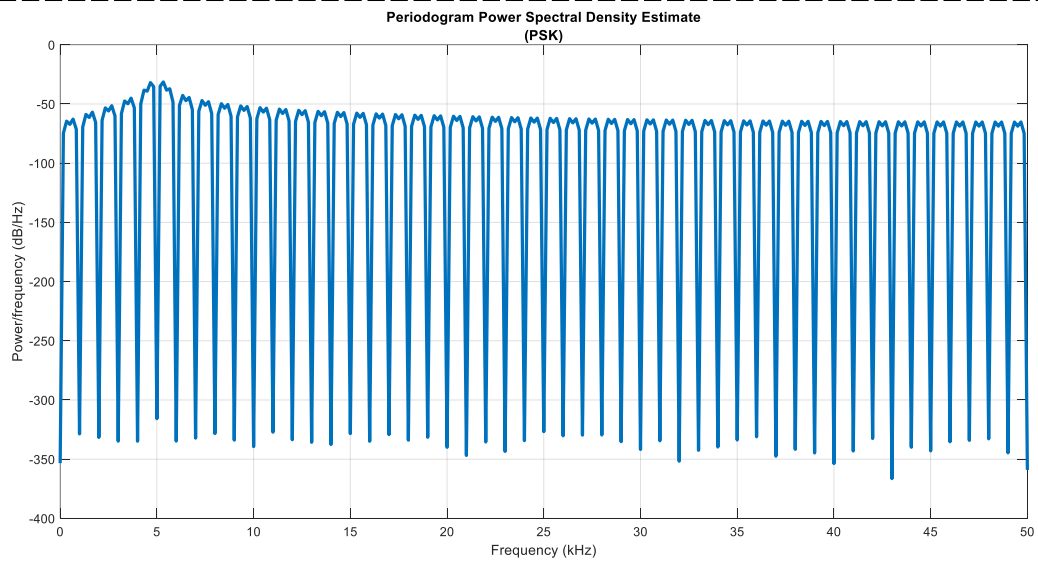
For ASK :

```
plot(psd(spectrum.periodogram,Ask,'Fs',100000,'NFFT',length(Ask)));
```



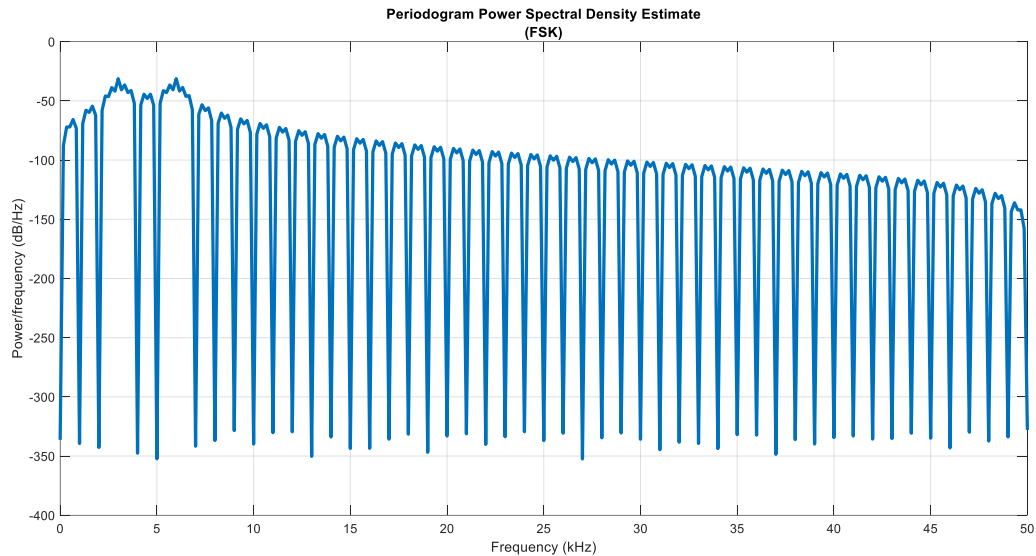
For psk :

```
plot(psd(spectrum.periodogram,Psk,'Fs',100000,'NFFT',length(Psk)));
```



For FSK :

```
plot(psd(spectrum.periodogram,Fsk,'Fs',100000,'NFFT',length(Fsk)));
```



3.

```
v = Ask .* cos(2*pi*(5e+3)*time);
Matchfilter = ones(1,10);
W = conv(Matchfilter,v);
% sampling at 10th member because it's at the center of convolved signal
S = [W(10),W(20),W(30),W(40),W(50)];
bb = S > 0.5
bb =
```

1 0 0 1 0

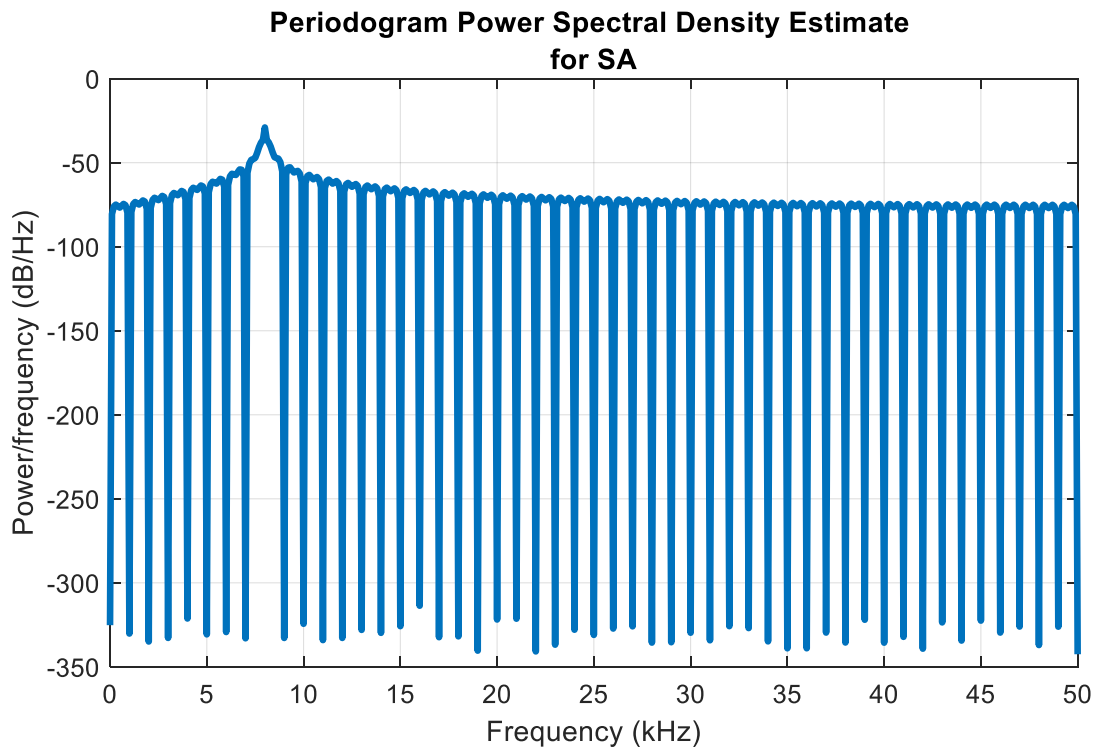
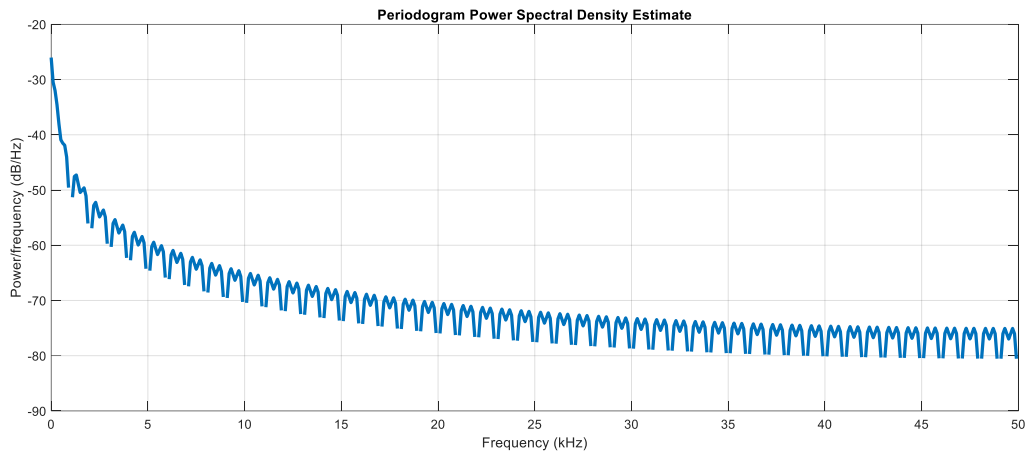
A. Generation of Modulated Signals

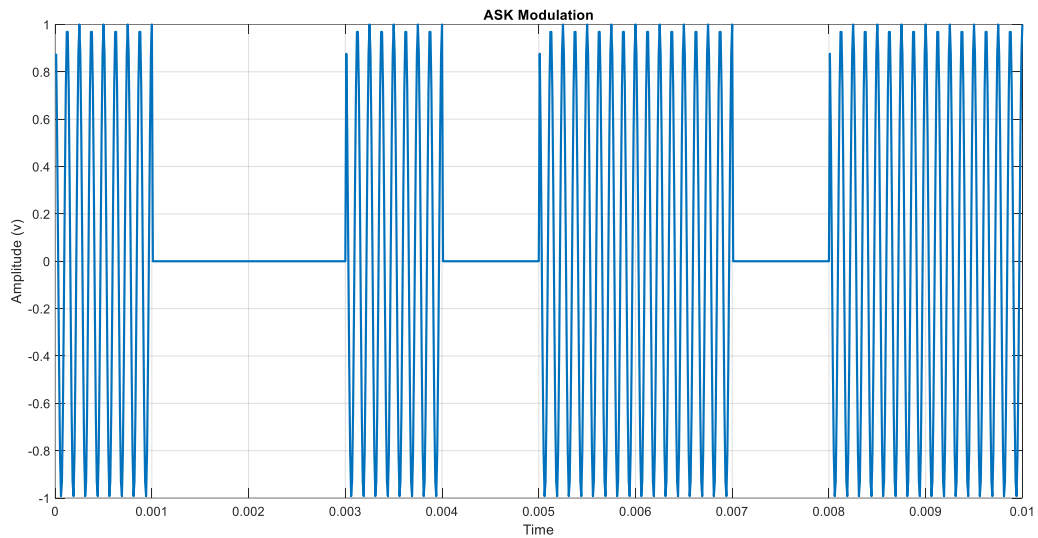
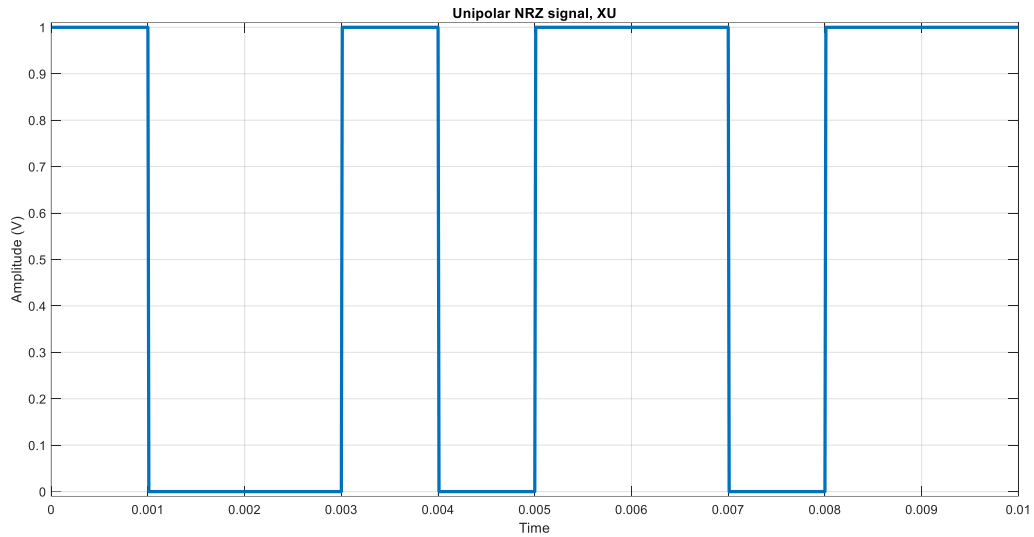
*Amplitude-Shift Keying (ASK)

A.1, 2, 3

```
clear all
clc
b = [1 0 0 1 0]; Rb = 1000;
b(6:10) = randi([0 1],[1 5]);
simulation = 10;
time = 0.00001 : 0.00001 : simulation*(1/Rb);
XU = zeros(1,length(time));
for i = 1:simulation
    XU(1+(i-1)*100:i*100) = b(i);
end
SA = XU.*cos(2*pi*((8e+3))*time);
plot(time,XU)
```

```
axis([0 max(time) -0.01 1.01])
figure
plot(time,SA)
figure
plot(psd(spectrum.periodogram,XU,'Fs',100000,'NFFT',length(XU)));
figure
plot(psd(spectrum.periodogram,SA,'Fs',100000,'NFFT',length(SA)));
```





*Phase-Shift Keying (PSK)

A.4 , 5 , 6

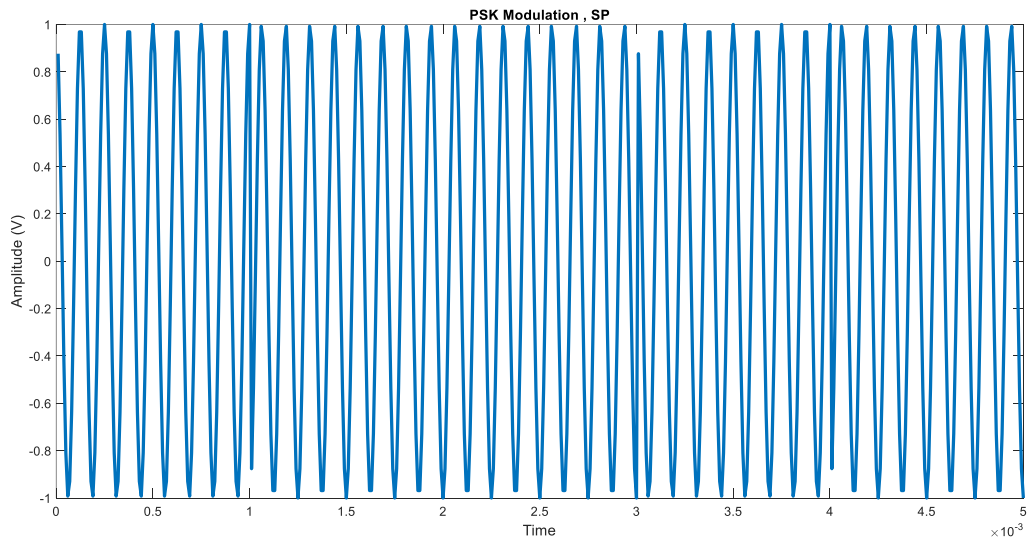
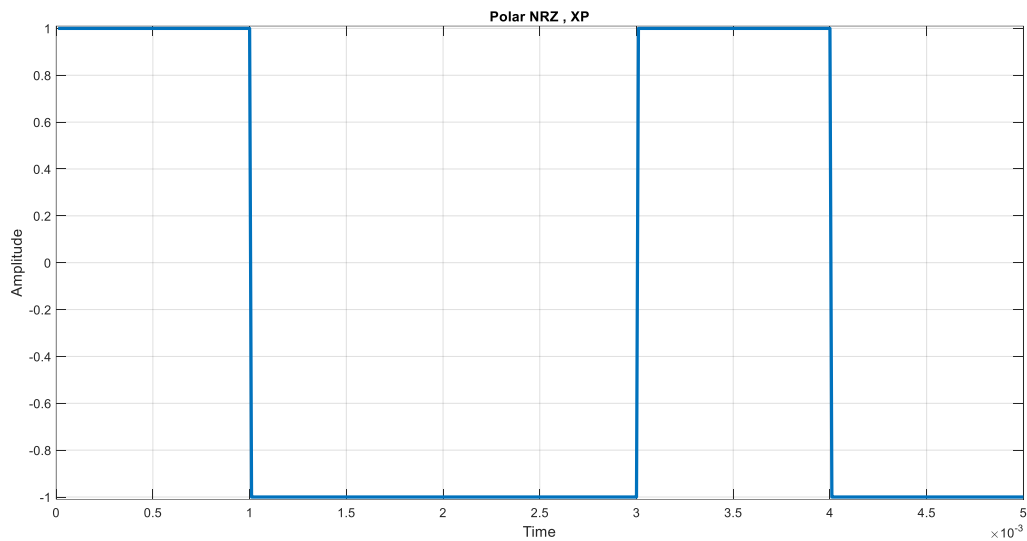
phase difference between sp and the carrier $\cos(2\pi fct)$ during the first = 0° and second = 180°

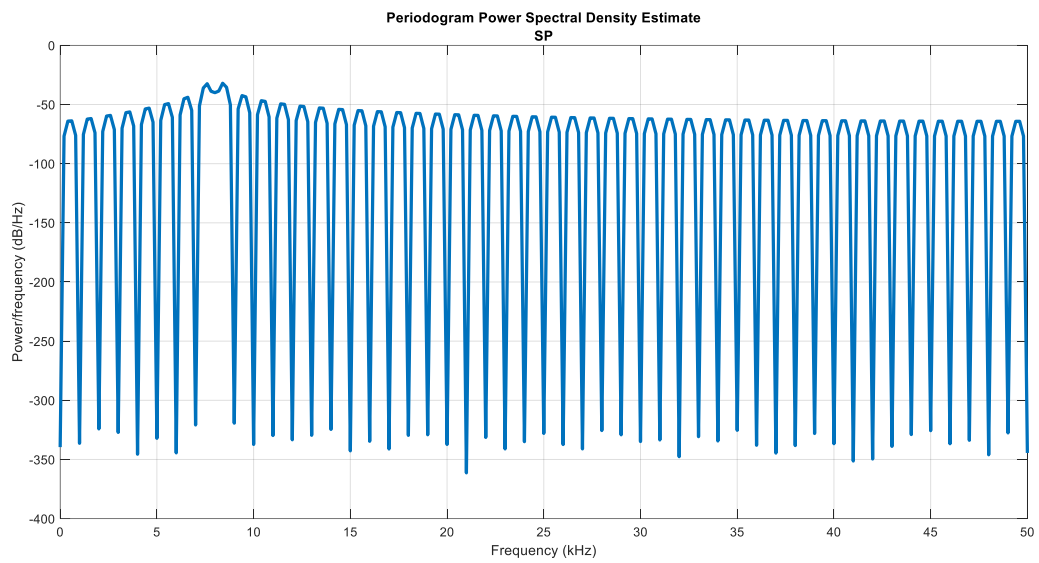
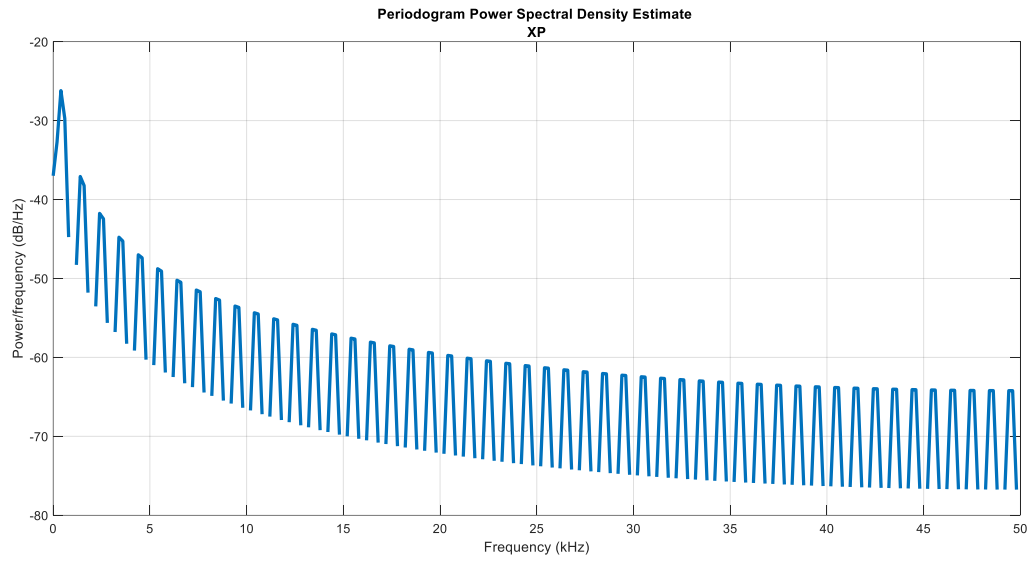
```
clear all
clc
b = [1 0 0 1 0]; Rb = 1000;
b(6:10) = randi([0 1], [1 5]);
simulation = 5;
time = 0.00001 : 0.00001 : simulation*(1/Rb);
XP = zeros(1, length(time));
for i = 1:simulation
    XP(1+(i-1)*100:i*100) = 2*b(i)-1;
end
```

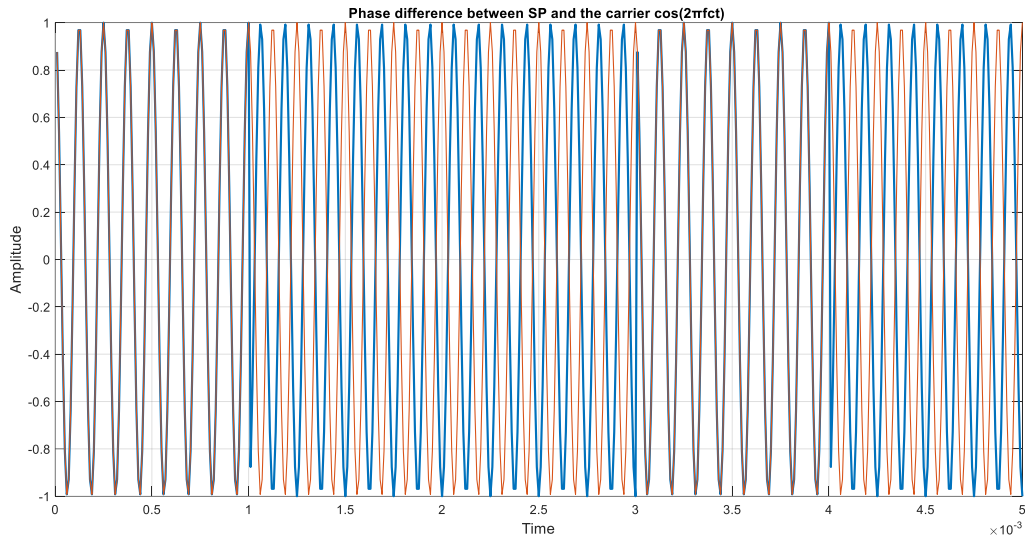
```

SP = XP.*cos(2*pi*((8e+3))*time);
plot(time,XP)
axis([0 max(time) -1.01 1.01])
figure
plot(time,SP)
figure
plot(time,SP)
hold on
plot(time,cos(2*pi*((8e+3))*time))
figure
plot(psd(spectrum.periodogram,XP,'Fs',100000,'NFFT',length(XP)));
figure
plot(psd(spectrum.periodogram,SP,'Fs',100000,'NFFT',length(SP)));

```



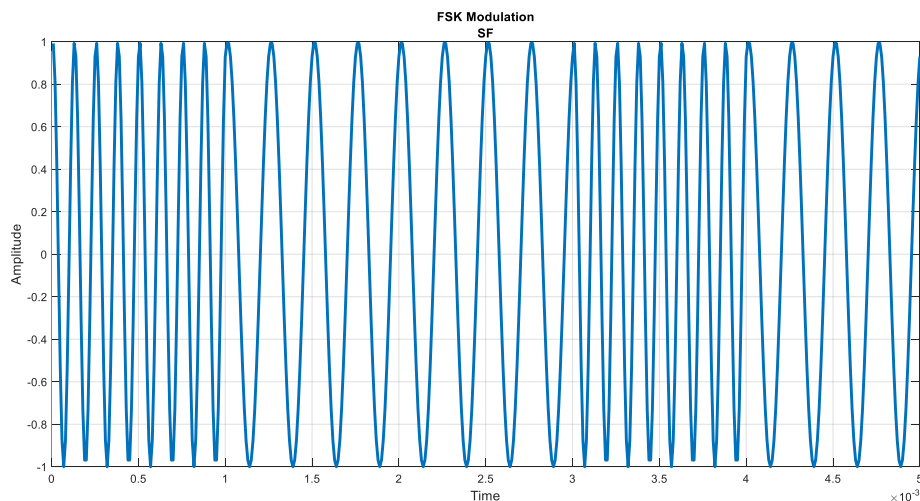


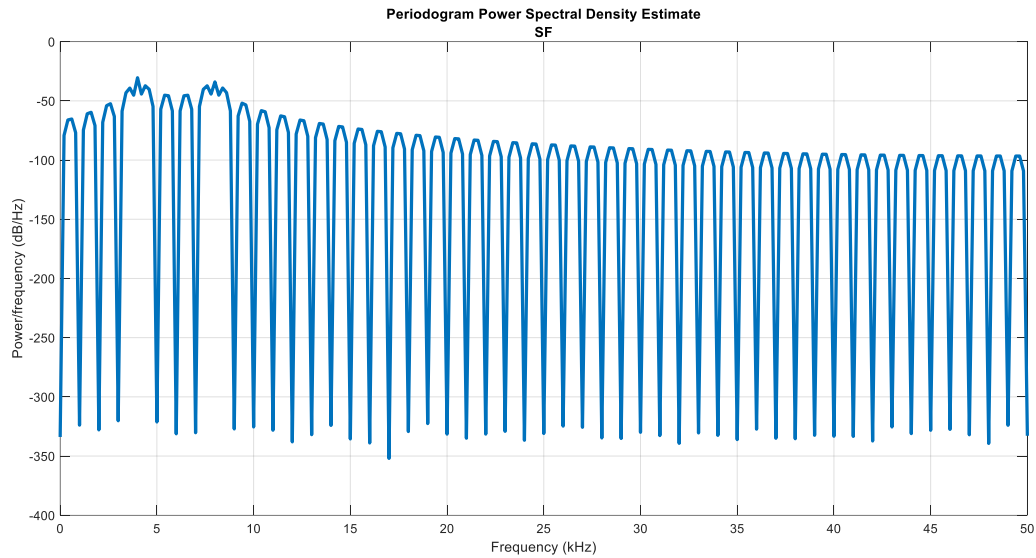


*Frequency-Shift Keying (FSK)

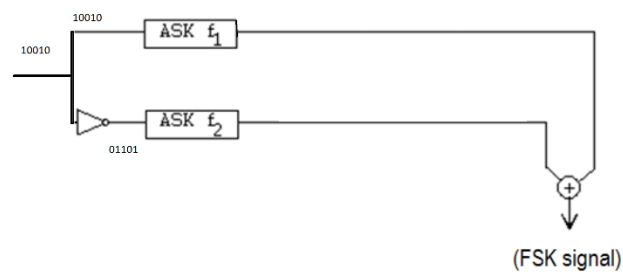
A.6 , 7

```
clear all
clc
b = [1 0 0 1 0]; Rb = 1000;
b(6:10) = randi([0 1], [1 5]);
simulation = 5;
time = 0.00001 : 0.00001 : simulation*(1/Rb);
XP = zeros(1, length(time));
for i = 1:simulation
    XP(1+(i-1)*100:i*100) = 2*b(i)-1;
end
SF = vco(XP, [4e+3 8e+3], 1e+5);
plot(time, SF)
figure
plot(psd(spectrum.periodogram, SF, 'Fs', 100000, 'NFFT', length(SF)));
```





One way to implement an FSK signal is by using two ASK signals with different carrier frequencies:



Where : f_1 = mark frequency(4 kHz) and f_2 = space frequency(8 kHz)

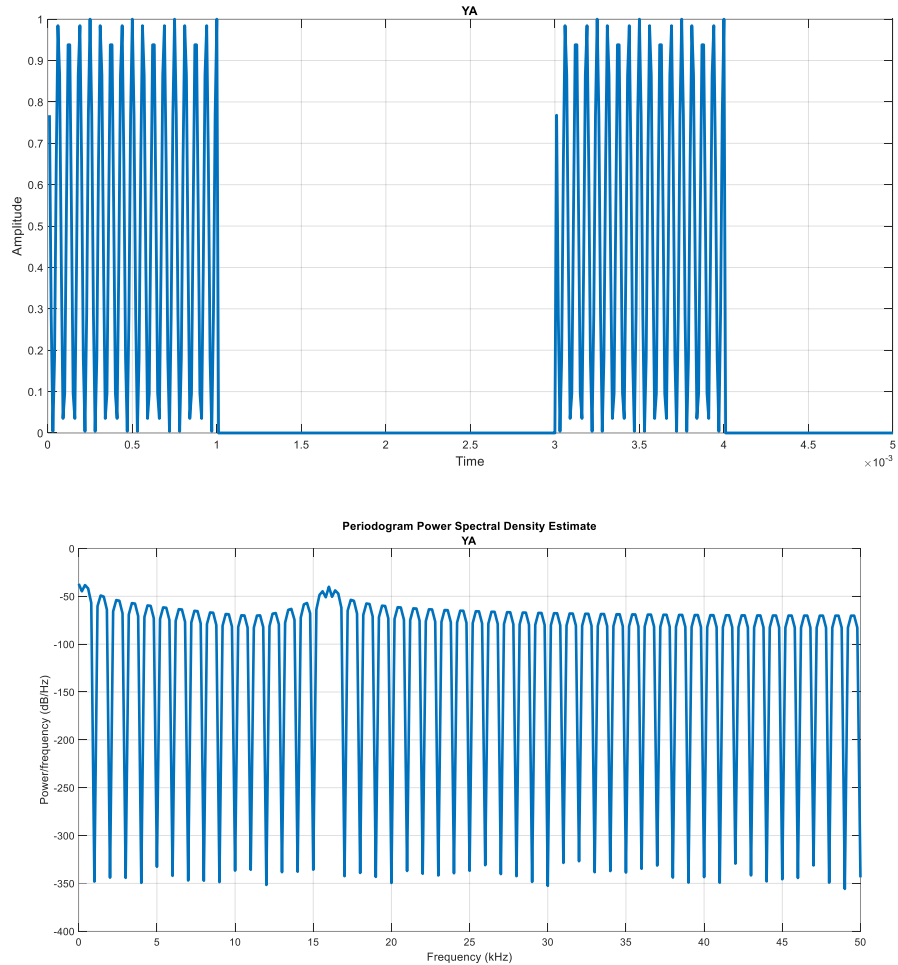
Surely FSK is not efficient bandwidth . ASK is desire as this point.

B. Coherent and Noncoherent Detection

*Coherent Detection

B.1

```
YA = SA.*cos(2*pi*((8e+3))*time);
plot(time,YA)
figure
plot(psd(spectrum.periodogram,YA,'Fs',100000,'NFFT',length(YA)));
```

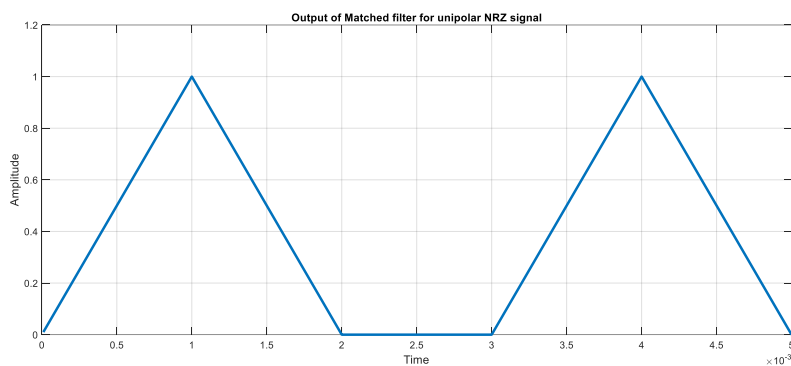
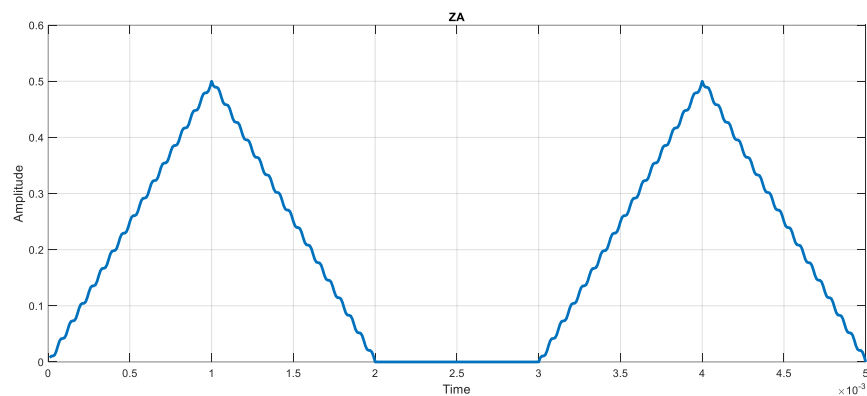


B.2

By assuming matched filter is normalized to it's energy.

```
YA = SA.*cos(2*pi*((8e+3))*time);
ZA = conv(0.01*ones(1,100),YA);
Output_of_unipolar_NRZ_signal = conv(0.01*ones(1,100),XA);
plot(time(1:500),ZA(1:500))
figure
plot(time(1:500),Output_of_unipolar_NRZ_signal(1:500))
```

Because two mentioned output are equivalent to each other.



B.3

By assuming matched filter is normalized to it's energy.

| Phase Error | Peak Amplitude [V] |
|-------------|--------------------|
| 0° | 0.5 |
| 20° | 0.47 |
| 60° | 0.25 |
| 80° | 0.09 |
| 120° | 0 |

Surely Phase Error = 0° is desirable for BER .

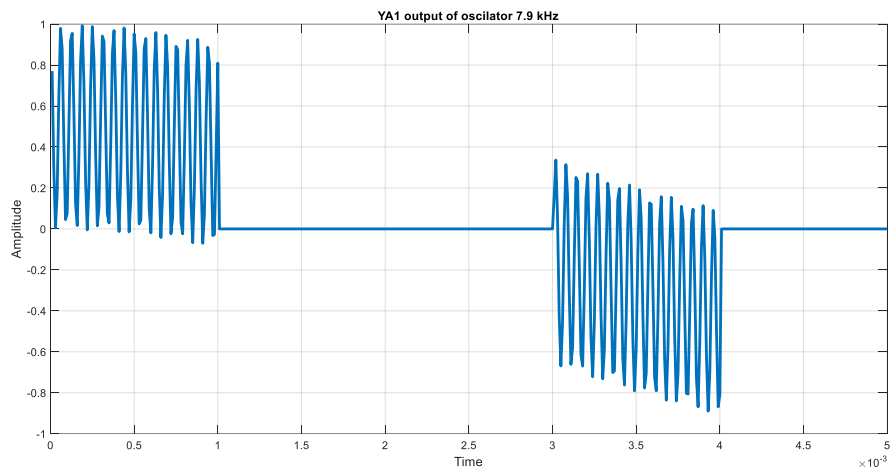
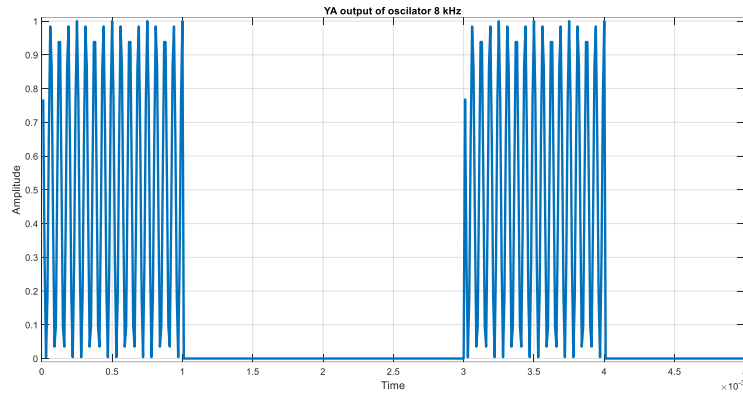
B.4

Assuming sampler samples at $1/R_b$ and $V_{th} = 0.5$:

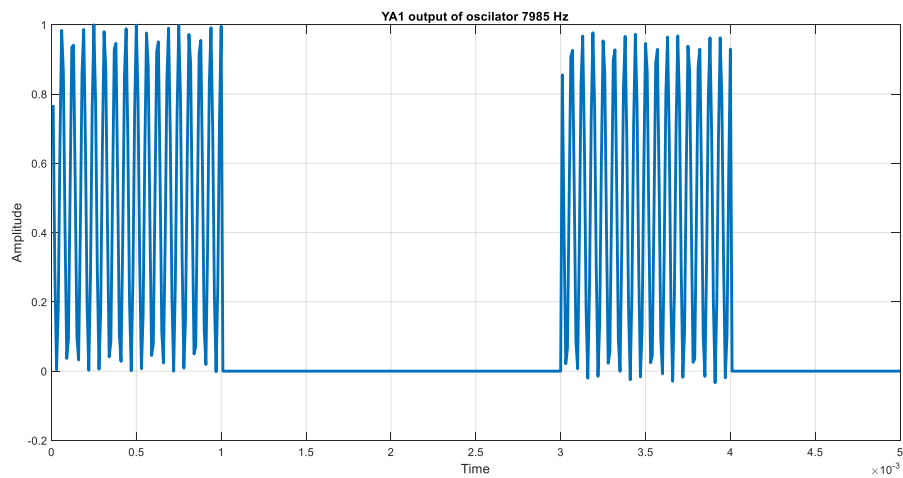
At Phase Error = $60^\circ \Rightarrow \hat{b} = 00000$

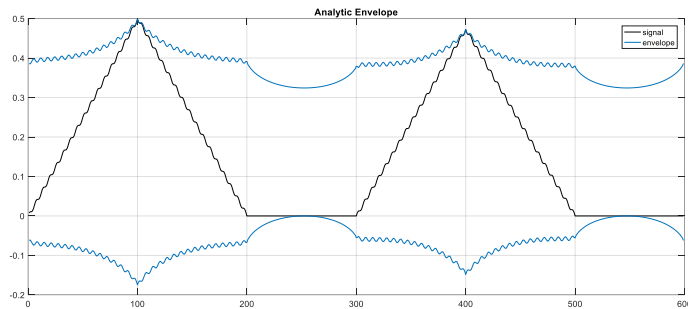
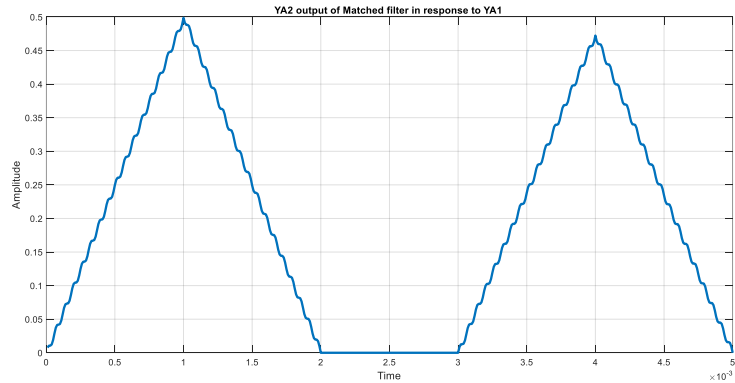
At Phase Error = $120^\circ \Rightarrow \hat{b} = 00000$

B.5



So if we consider $V_{th} = 0.5$ all the sequence cannot be decoded correctly.





frequency of the envelope $= f_c - f_0 = 8 \text{ kHz} - 7985 \text{ Hz} = 15 \text{ Hz}$

C. System Performance Under Noise

Coherent Detection

C.1,2,3

```
clear all
clc
b = [1 0 0 1 0]; Rb = 1000;
b(6:500) = randi([0 1], [1 495]);
simulation = 500;
time = 0.00001 : 0.00001 : simulation*(1/Rb);
XA = zeros(1, length(time));
for i = 1:simulation
    XA(1+(i-1)*100:i*100) = b(i);
end
SA = XA.*cos(2*pi*((8e+3))*time);
Y = SA + sqrt(0.5)*randn(1, length(SA));
YA = Y.*cos(2*pi*((8e+3))*time);
ZA = conv(0.01*ones(1, 100), YA);
eyediagram(ZA, 200)
figure
plot(time(1:500), SA(1:500), time(1:500), Y(1:500))
legend('SA, modulated signal', 'Y, the output of channel')
% sampling at 1/Rb or at center of eyediagram and Vth = 1/2*(top to down of
open eye) = 0.25
```

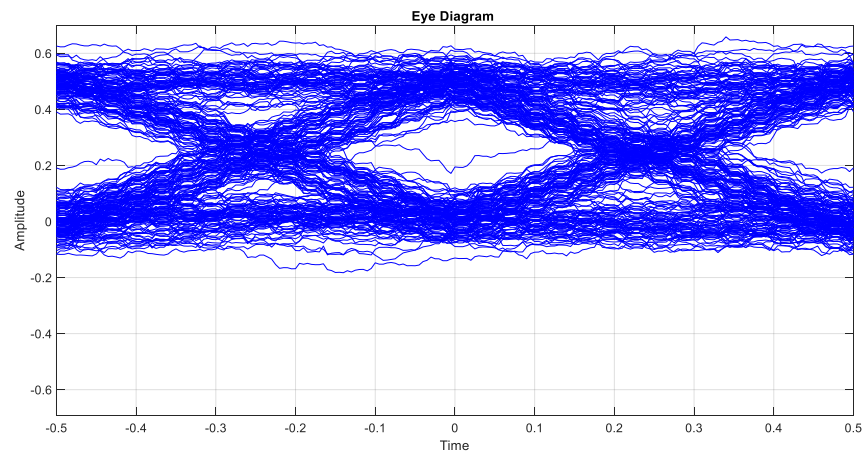
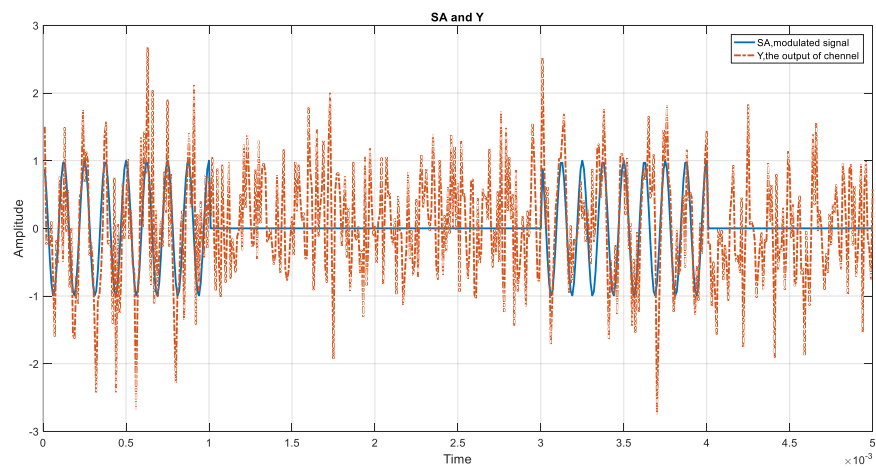
```

Z = ZA(100:100:end); % every bit interval includes 100 samples
bb = Z > 0.25;
noe = sum(abs(bb-b)); % number of error
BER = noe/simulation
% BER theory
N0 = 2e-5;
EBN0 = (sum(SA.*SA)/(Rb*length(SA)))/N0;
BER_theory = qfunc(sqrt(EBN0))

```

BER = 0

BER_theory = 2.2049e-04



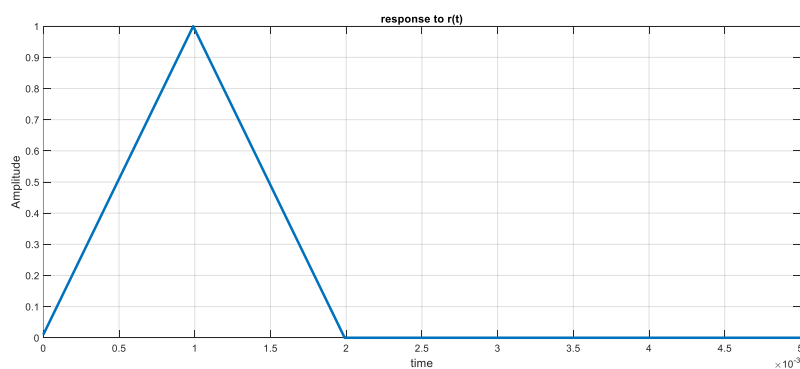
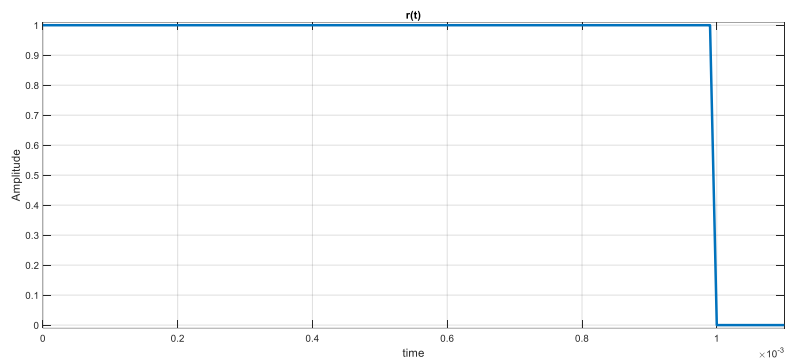
Computer Exercise V: Matched Filter and Bit Error Rate (BER)

I- Prelab Assignment:

A:

a,b)

```
clear all
clc
simulation = 5; T = 1e-3;
rt = zeros(1, simulation*100);
t = 0.00001:0.00001:simulation*T;
x = 1; % impulse
nsamp = 100;
rt(1:100) = rectpulse(x, nsamp);
plot(t, rt(1:length(t)))
axis([0 0.0011 -0.01 1.01])
figure
out = 0.01*conv(rt, rt);
plot(t, out(1:length(t)))
```



c)

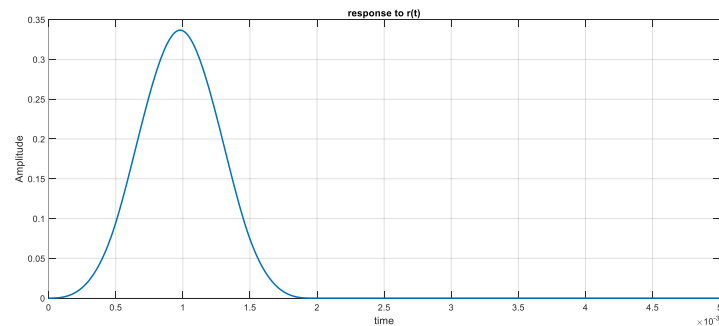
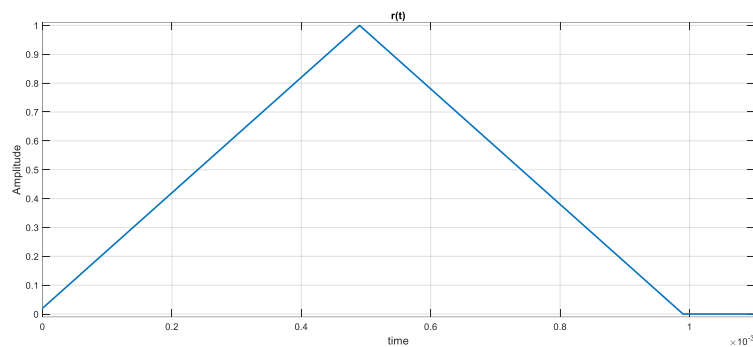
```
clear all
clc
```



```

simulation = 5; T = 1e-3;
rt = zeros(1,1+simulation*100);
t = 0:0.00001:simulation*T;
x = 1;% impulse
nsamp = 100;
rt1 = rectpulse(x,nsamp/2);
plot(t(1:50),rt1)
figure
rt(1:99) = 0.02*conv(rt1,rt1);
plot(t,rt(1:length(t)))
axis([0 0.0011 -0.01 1.01])
figure
out = 1/99*conv(rt,rt);
plot(t,out(1:length(t)))

```



B:

a,b,c)

Assuming ; $W = 10$ kHz is channel bandwidth

```

clear all
clc
b = [1 0 0 1 0]; Rb = 1000; W = 1e+4;
b(6:500) = randi([0 1],[1 495]);
simulation = 500;
time = 0.00001 : 0.00001 : simulation*(1/Rb);
X = zeros(1,length(time));
for i = 1:simulation

```

```

X(1+(i-1)*100:i*100)= 2*b(i)-1;
end
T = 1e-3;
Y = X + sqrt(0.5)*randn(1,length(X));
rt = zeros(1,simulation*100);
x = 1;% impulse
nsamp = 100;
rt(1:100) = rectpulse(x,nsamp);
out = 0.01*conv(Y,rt);
plot(time,out(1:length(time)))
rms_noise = sqrt(var(out))
Peak_out_matched_filter = max(abs(out))
the_average_energy_X = sum(X.*X)/(Rb*length(X))
N0 = 1e-4;
EBN0 = the_average_energy_X/N0;
Pe = qfunc(sqrt(2*EBN0))

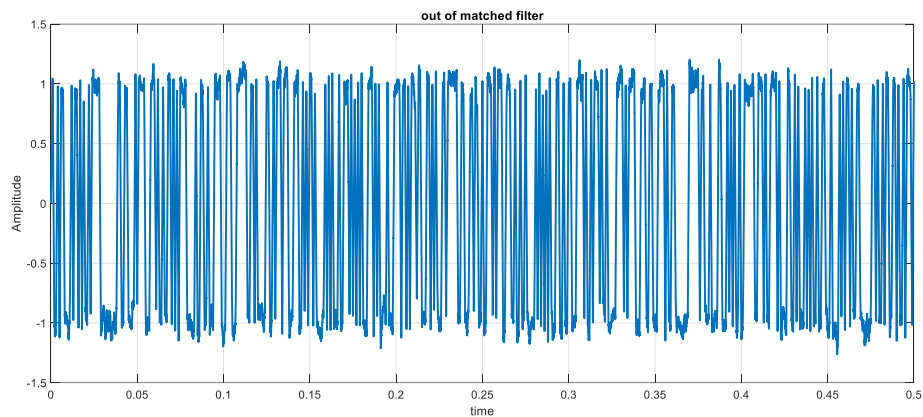
```

rms_noise = 0.5777

Peak_out_matched_filter = 1.2629

the_average_energy_X = 1.0000e-03

Pe = 3.8721e-06



C:

a,b)

$$\frac{1}{1 + j2\pi fRC} \leftrightarrow \frac{1}{RC} e^{-\frac{t}{RC}} u(t) \quad , \quad \Delta f = 1 \text{ kHz}$$

```

clear all
clc
% assuming sampel per symbol = 1
simulation = 1000000;
b = [1 0 0 1 0];Rb = 1000;W = 1e+3;
b(6:simulation)= randi([0 1],[1 simulation-5]);
time = 0.001 : 0.001 : simulation*(1/Rb);
X = zeros(1,length(time));

```

```

for i = 1:simulation
X(1+(i-1)*1:i*1)= 2*b(i)-1;
end
T = 1e-3;
Y = X + sqrt(0.05)*randn(1,length(X));
channel = (2000*pi)*exp(-(2000*pi)*time);
out_channel = (1/length(time))*conv(Y,channel);
receive_signal = (1/length(time))*conv(out_channel,channel);
detected_sig = receive_signal(1:1:simulation*1);
bb = detected_sig > 0 ;
noe = sum(abs(bb-b));
Pe = noe/simulation
plot(time,out_channel(1:length(time)))
rms_noise = sqrt(var(out_channel))
Peak_out_channel = max(abs(out_channel))
Pe =

```

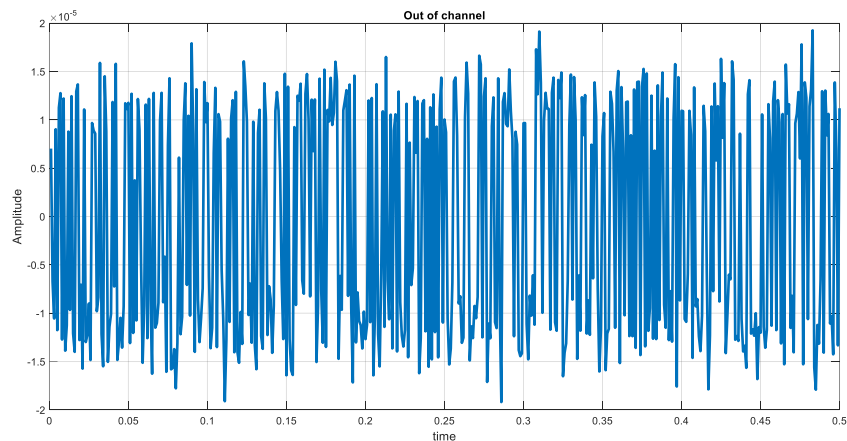
6.0000e-06

rms_noise =

8.5023e-06

Peak_out_channel =

2.4881e-05

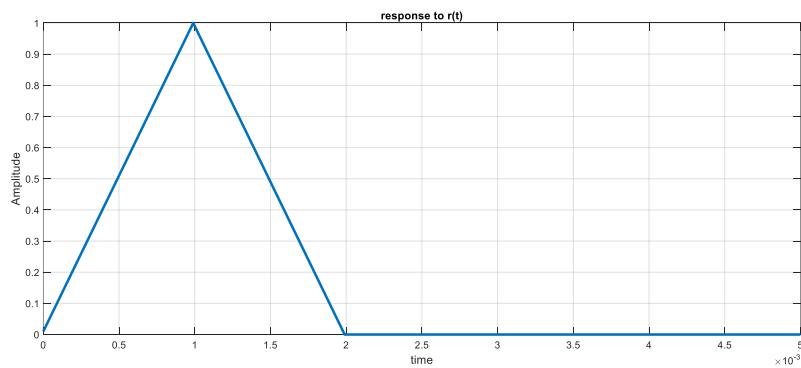
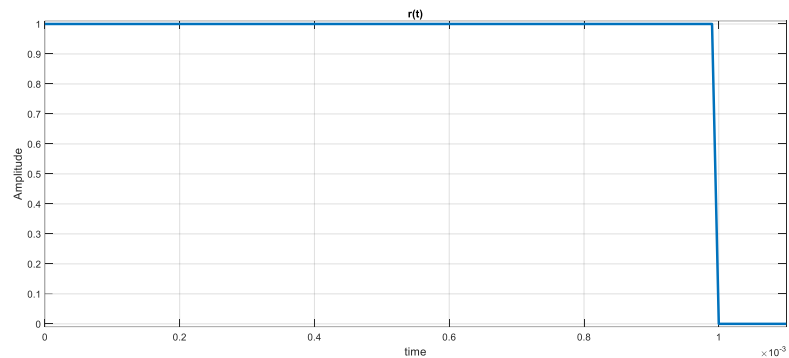


II- Procedure:

A. Characteristics of Matched Filters

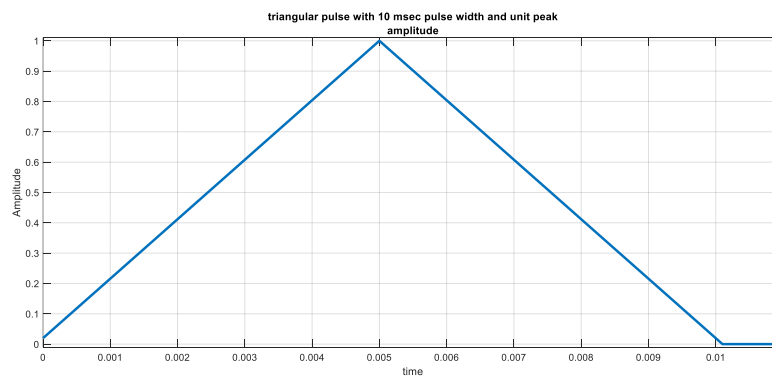
A.1 ,2 ,3

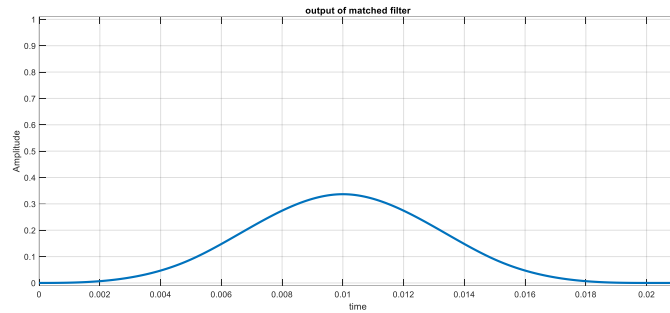
$$r = \text{matched filter based on } r = \text{rect}\left(\frac{t - \frac{T}{2}}{T}\right), T = 1\text{ms}$$



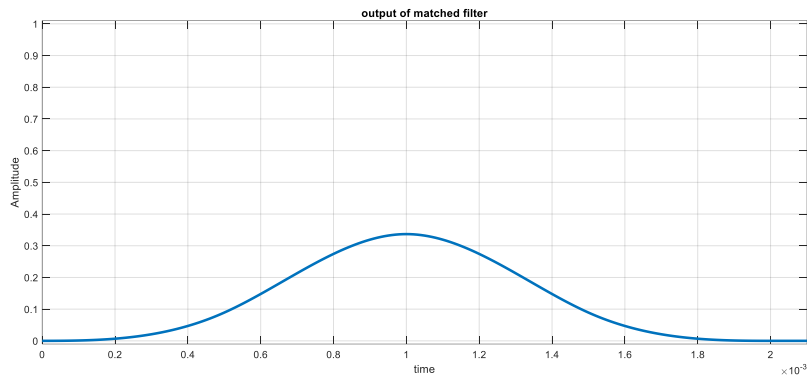
The time when the filter output reaches its maximum value = $T = 1\text{ms}$

A.4



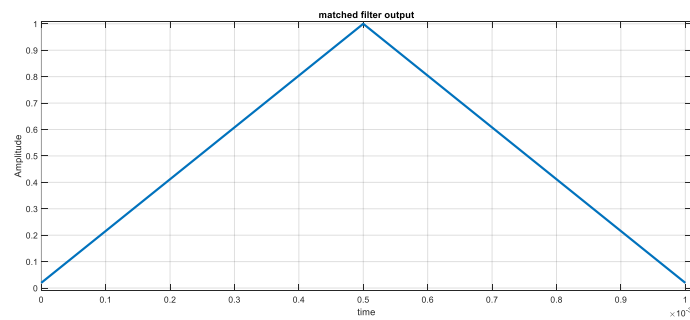
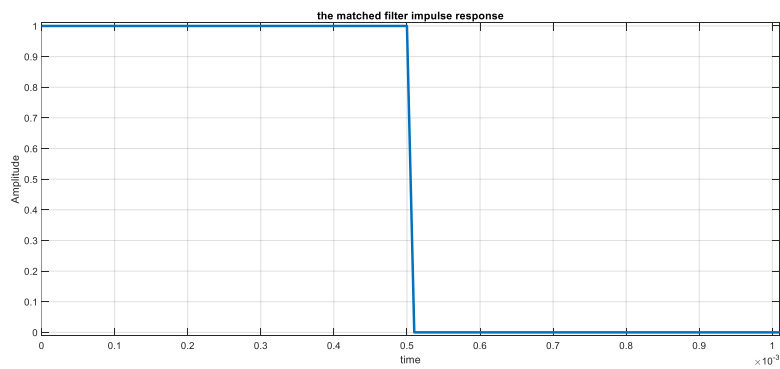


The time when the filter output reaches its maximum value = $T = 10\text{ms}$

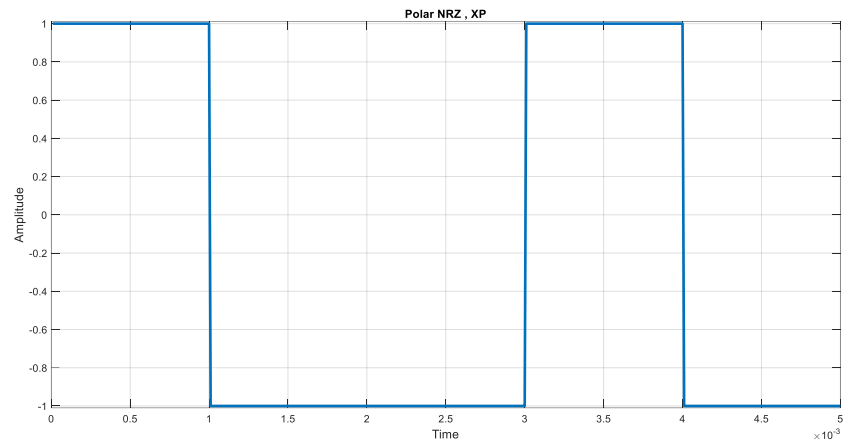


The time when the filter output reaches its maximum value = $T = 1\text{ms}$

A.5

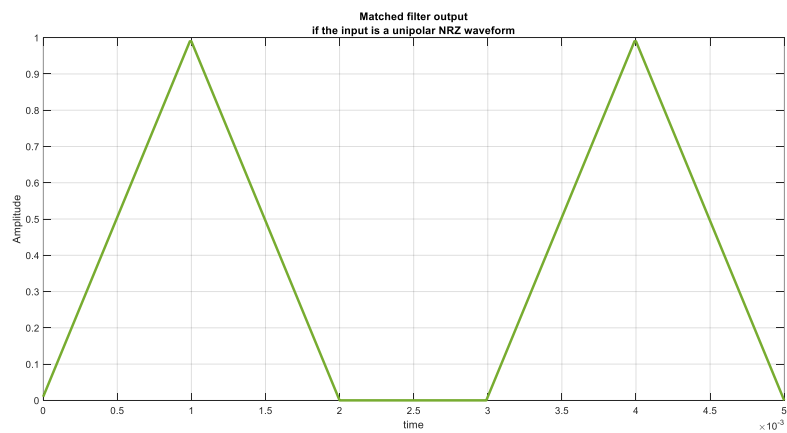
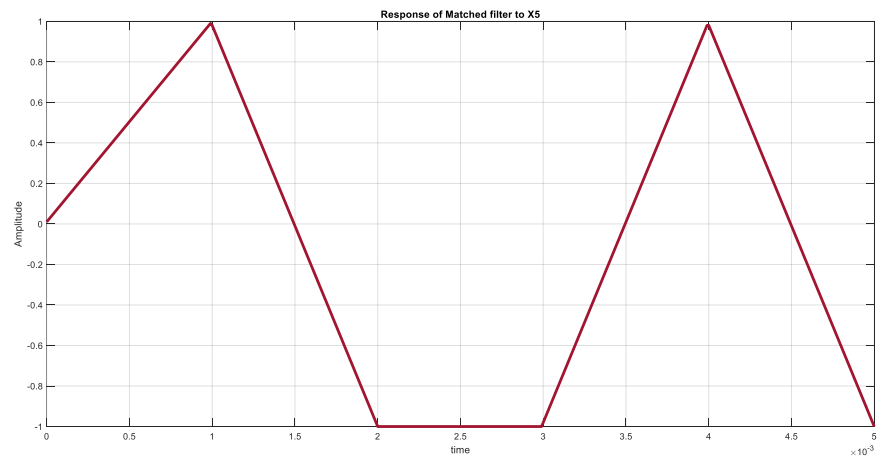


A.6



A.7

Assuming selected matched filter is rect.



B. Signal Detection

B.1 , 2 , 3, 4

$$\frac{1}{1+j2\pi fRC} \leftrightarrow \frac{1}{RC} e^{-\frac{t}{RC}} u(t) \quad , \quad RC = \frac{1}{2\pi(4900)}, \Delta f = 4.9 \text{ kHz}$$

```
clear all
clc
simulation = 10;
b10 = [1 0 0 1 0]; Rb = 1000; W = 1e+3;
b10(6:simulation) = randi([0 1], [1 simulation-5]);
time = 0 : 0.00001 : simulation*(1/Rb);
X10 = zeros(1, length(time));
for i = 1:simulation
    X10(1+(i-1)*100:i*101) = 2*b10(i)-1;
end
channel = (2*pi*(4900))*exp(-(2*pi*(4900))*time);
out_channel = (1/length(time))*conv(X10, channel);
Y10 = out_channel + sqrt(2/2)*randn(1, length(out_channel));
y10 = Y10(1+100:100:1+simulation*100);
bar_y10 = y10 > 0 ;
noe1 = sum(abs(bar_y10-b10));
Pe1 = noe1/simulation
rt1 = zeros(1, 1+simulation*100);
x = 1; % impulse
nsamp = 101;
rt1(1:101) = rectpulse(x, 101);
Z10 = (1/length(time))*conv(Y10, rt1);
z10 = Z10(1+100:100:1+simulation*100);
bb = z10 > 0 ;
noe2 = sum(abs(bb-b10));
Pe2 = noe2/simulation
% if sampling instants other than KT
zz10 = Z10(1+100+90:100:1+simulation*100+90);
bb1 = zz10 > 0 ;
noe3 = sum(abs(bb1-b10));
Pe3 = noe3/simulation
plot(time, Y10(1:length(time)))
figure
plot(time, Z10(1:length(time)))
```

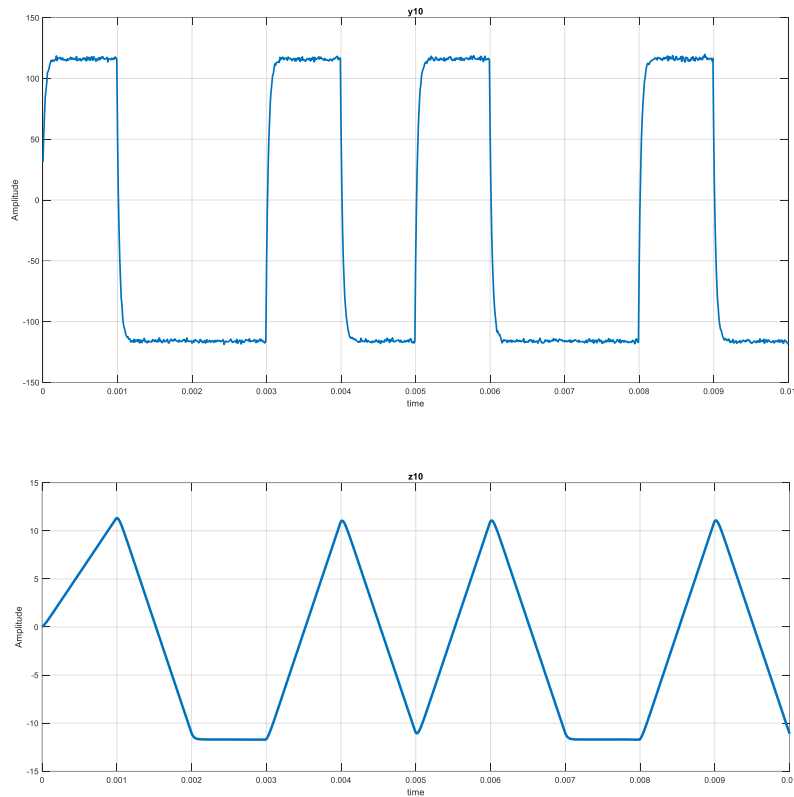
Pe1 = 0

Pe2 = 0

Pe3 = 0.6000

→ Decoding after Matched filter is easier because Matched filter can degrade the noise effect.

→ If sampling instants other than those specified (KT) the probability of error will be larger (Pe3) because we are sampling wrong sampels(next sampels) or correct sampels with amplitude lower than noise.



C. Matched-Filter Receiver

C.1,2

An example of code for this part just for the case $\sigma^2 = 0.5 W$ and assumes the sample number = 10.

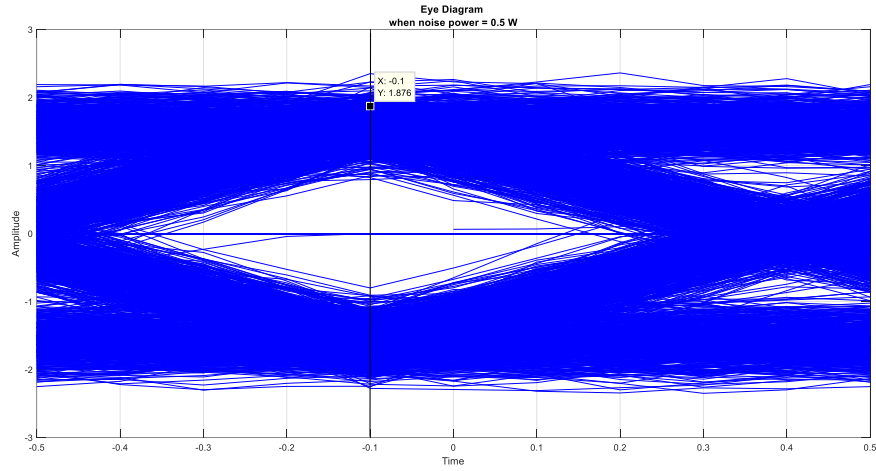
```
clear all
clc
simulation = 2000;
nsamp = 10;
b = [1 0 0 1 0]; Rb = 1000; W = 1e+3;
b(6:simulation) = randi([0 1], [1 simulation-5]);
time = 0.000001 : 0.0001 : simulation*(1/Rb);
X = zeros(1, length(time));
for i = 1:simulation
    X(1+(i-1)*nsamp:i*nsamp) = 2*b(i)-1;
end
channel = (2*pi*(4900))*exp(-(2*pi*(4900))*time);
y = X + sqrt(0.5/2)*randn(1, length(X));
out_channel = (1/length(time))*conv(y, channel);
rt1 = zeros(1, simulation*nsamp);
x = 1; % impulse
rt1(1:nsamp) = rectpulse(x, nsamp);
z = (1/nsamp)*conv(out_channel, rt1);
eyediagram(z, nsamp)
% with help of eyediagram : sampling instant = -0.1*Tb + KTb , V_th = 0
z10 = z(nsamp-1:nsamp:simulation*nsamp-1);
```



```

bb = z10 > 0 ;
noe = sum(abs(bb-b));
Pe_empirical = noe/simulation
% Theory Pe
N0 = 0.5/(4900);
Eb = sum(X.*X)/(Rb*length(X));
EbN0 = Eb/N0;
Pe_theoretical = qfunc(sqrt(2*EbN0))

```



| $\sigma^2 [W]$ | Pe empirical | Pe theoretical |
|----------------|---|------------------|
| 0.5 | 0 (or $O(10^{-6})$ when at least 10^6 bits generated) | 4.7735e-06 |
| 1 | 0 (or $O(10^{-4})$ when at least 10^4 bits generated) | 8.7256e-04 |
| 1.5 | 0.0015 | 0.0053 |
| 2 | 0.004 | 0.0134 |

C.3

Sampling at KTb :

| $\sigma^2 [W]$ | Pe empirical |
|----------------|---|
| 1 | 0 (or $O(10^{-4})$ when at least 10^4 bits generated) |
| 1.5 | 0.0015 |
| 2 | 0.0045 |

C.4

$$\alpha = \frac{\text{Out. Matched filter}(t = Tb - 0.5(0.1)Tb \text{ or } -0.9(0.1)Tb)}{\text{Out. Matched filter}(t = Tb)}$$

$$P_{theoretical} = Q\left(\sqrt{\frac{2\alpha Eb}{N_0}}\right)$$

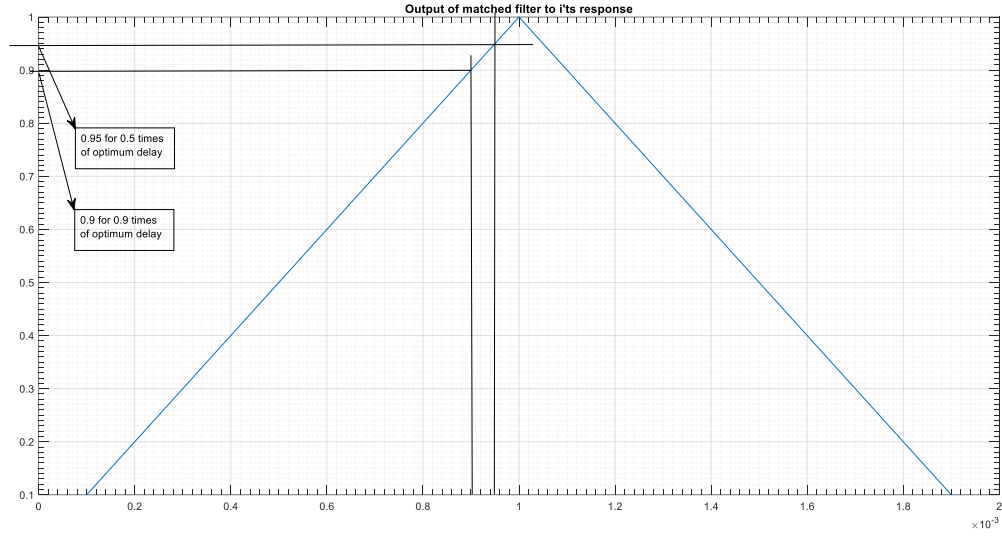
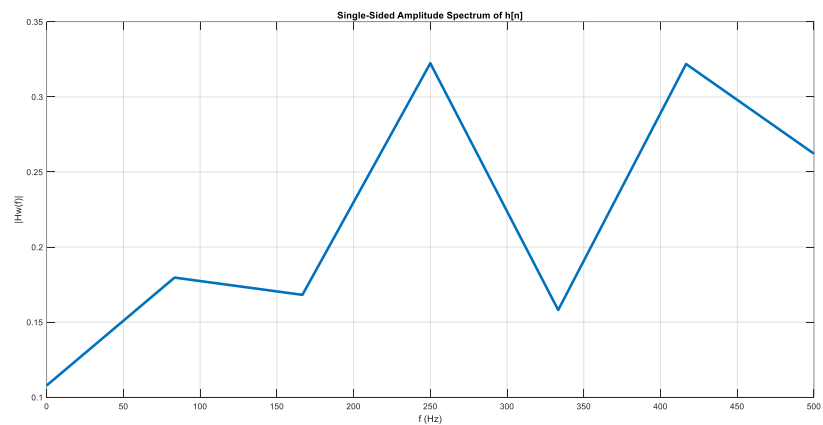
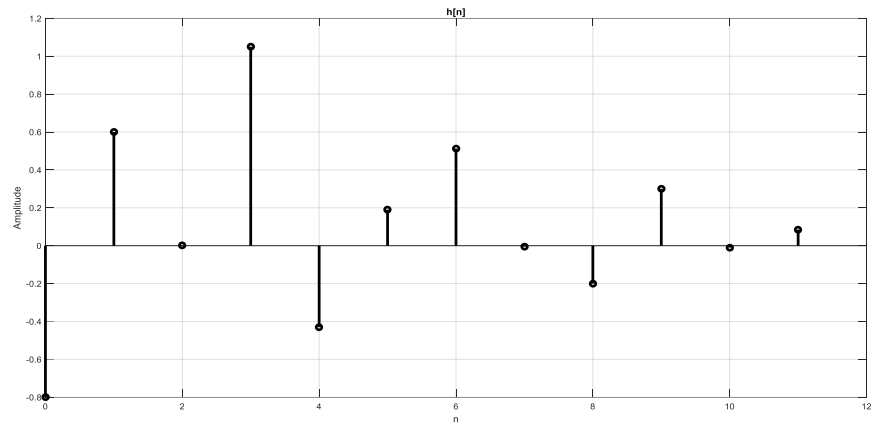


Table 1 : sample instant: $KTb - 0.5(0.1)Tb$

| σ^2 [W] | Pe empirical | Pe theoretical |
|----------------|---|------------------|
| 0.5 | 0 (or $O(10^{-6})$ when at least 10^6 bits generated) | 7.9769e-06 |
| 1 | 0 (or $O(10^{-4})$ when at least 10^4 bits generated) | 0.0011 |
| 1.5 | 0.002 | 0.0064 |
| 2 | 0.045 | 0.0155 |

Computer Exercise VI : Equalization

a)



```
clear all
clc
Fs = 1000; L = 11;
h = [-0.8 0.6 0.002 1.05 -0.43 0.19 0.512 -0.005 -0.2 0.3 -0.01 0.085];
n = 0:L;
H = fft(h);
f = Fs*(0:(length(h)/2))/length(h);
P2 = abs(H/(L+1));
Hw = P2(1:(L+1)/2+1);
Hw(2:end-1) = 2*Hw(2:end-1);
plot(n,h)
title('h[n]')
xlabel('n')
ylabel('Amplitude')
figure
```

```

plot(f,Hw)
title('Single-Sided Amplitude Spectrum of h[n]')
xlabel('f (Hz)')
ylabel('|Hw(f)|')

```

b)

Assuming : $W = Fs$, roll-of-factor = 0 :

$$SNR = \frac{2Eav}{N0} \left(\int_{-W}^W \frac{1}{|H(f)|^2} df \right)^{-1}, \quad N0 = 0.2 W/Hz$$

$$SNR = 0.0720$$

```

clear all
clc
Fs = 1000;L =11;
h = [-0.8 0.6 0.002 1.05 -0.43 0.19 0.512 -0.005 -0.2 0.3 -0.01 0.085];
n = 0:L;
H = fft(h);
f = Fs*(0:0.01:(length(h)/2))/length(h);
P2 = abs(H/(L+1));
Hw = P2(1:0.01:(L+1)/2+1);
Hw(2:end-1) = 2*Hw(2:end-1);
plot(n,h)
title('h[n]')
xlabel('n')
ylabel('Amplitude')
figure
plot(f,Hw)
title('Single-Sided Amplitude Spectrum of h[n]')
xlabel('f (Hz)')
ylabel('|Hw(f)|')
SNR = 2/0.2*(sum(1./(Hw).^2)*0.01)^-1

```

c)

i)

$$Nf + 1 = 2K + 1, C_{opt} = \Gamma^{-1}\xi:$$

```

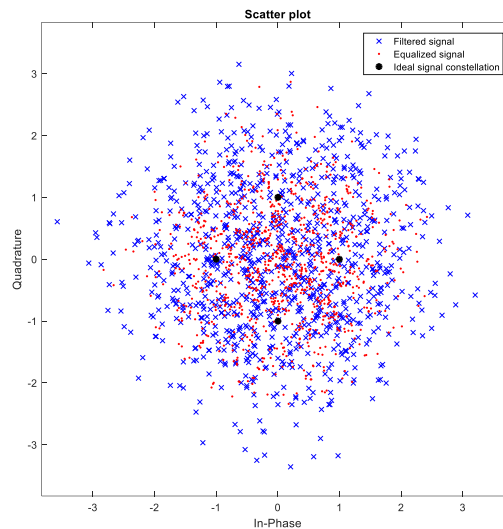
clear all
clc
Fs = 1000;L =11;ntab = 25;
h = [-0.8 0.6 0.002 1.05 -0.43 0.19 0.512 -0.005 -0.2 0.3 -0.01 0.085];
n = 0:L;
H = fft(h);
f = Fs*(0:0.01:(length(h)/2))/length(h);
P2 = abs(H/(L+1));
Hw = P2(1:0.01:(L+1)/2+1);

```

```

Hw(2:end-1) = 2*Hw(2:end-1);
% peak distortion criterion for ZFE EQ.
c = [ h zeros(1,13)];
g = [h(1) zeros(1,24)];
Gama = toeplitz(c,g);
zeta = [zeros(1,12) 1 zeros(1,12)]';
Copt = inv(Gama)*zeta;% ZFE EQ. coefficients
% Set up parameters and signals.
M = 4; % Alphabet size for modulation
msg = randint(1500,1,M); % Random message
modmsg = pskmod(msg,M); % Modulate using QPSK.
trainlen = 500; % Length of training sequence
filtmsg = filter(h,1,modmsg); % Introduce channel distortion
constellation = pskmod([0:M-1],M); % Set signal constellation.
symboltest = 1/norm(Copt)*conv(Copt,filtmsg);% Equalize.
% Plot signals.
hh = scatterplot(filtmsg,1,trainlen,'bx'); hold on;
scatterplot(symboltest,1,trainlen,'r.',hh);
scatterplot(constellation,1,0,'k*',hh);
legend('Filtered signal','Equalized signal',...
'Ideal signal constellation');
hold off;

```



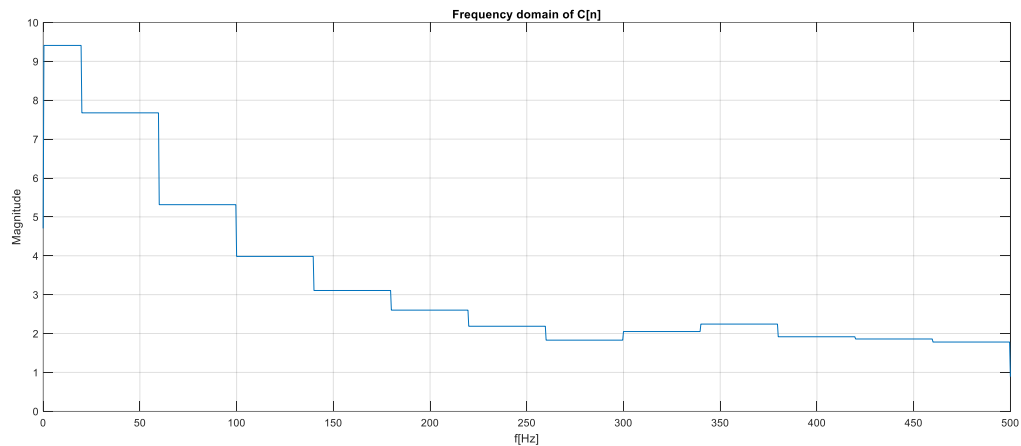
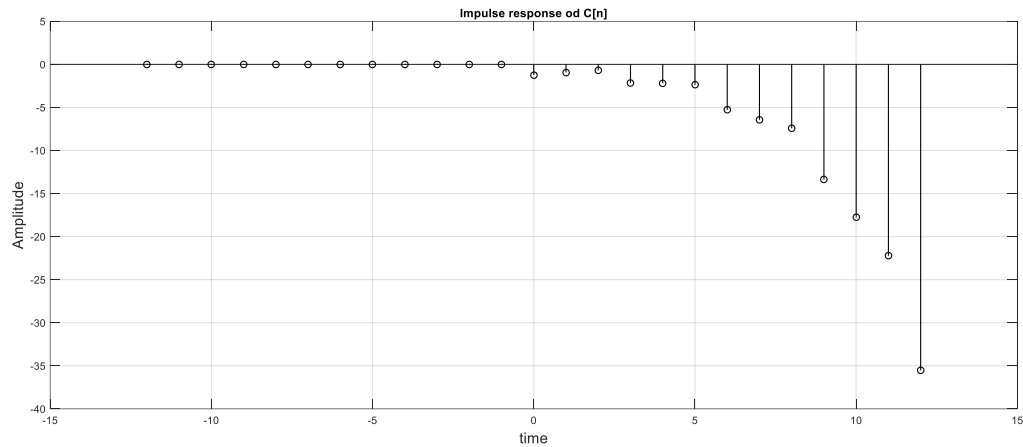
```

nn = -(ntab-1)/2:(ntab-1)/2;
plot(nn,Copt)
figure
FCopt = fft(Copt);
ff = Fs*(0:0.01:(length(Copt)/2))/length(Copt);
PP2 = abs(FCopt/(ntab));
HwCopt = PP2(1:0.01:(ntab)/2+1);
HwCopt(2:end-1) = 2*HwCopt(2:end-1);
plot(ff,HwCopt)

```

By attending to $q = \text{conv}(\text{Copt}, h)$

$$SNR = \frac{2Eb}{N_0(\sum q[m \neq 0])} (1 - \sum q[m \neq 0])^2 = 10 \times 153$$



ii)

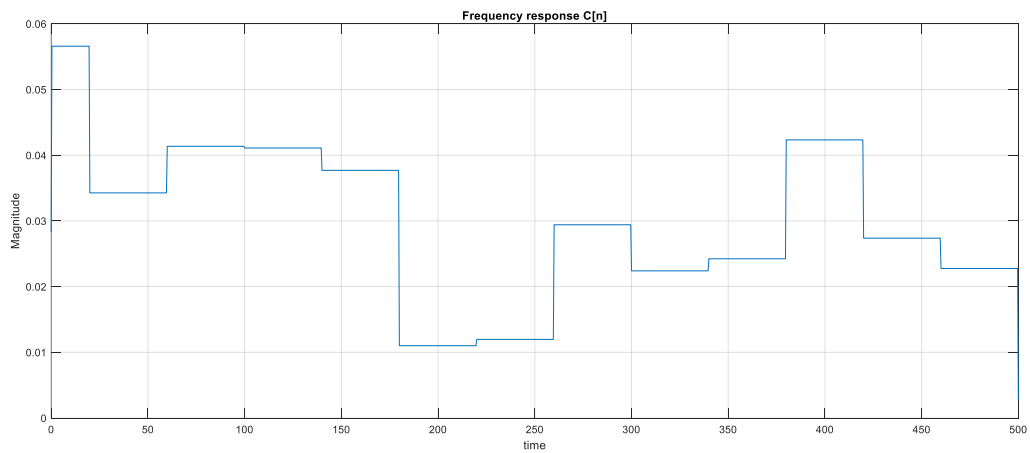
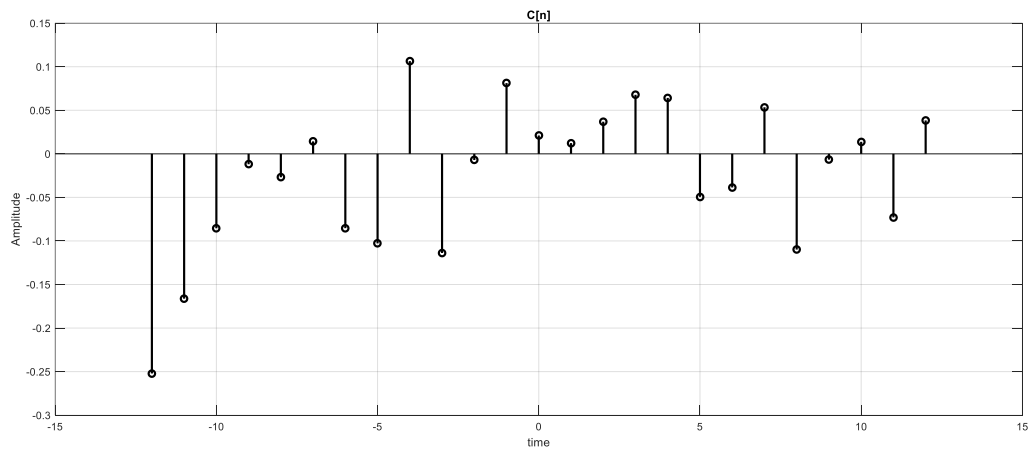
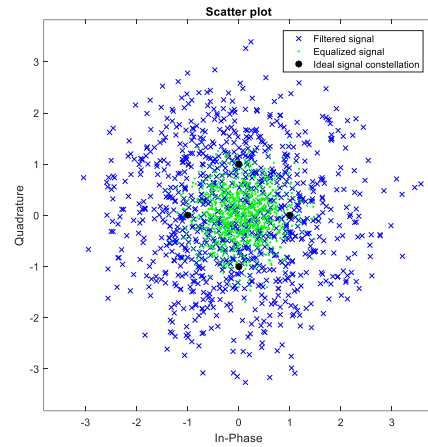
```
clear all
clc
Fs = 1000; L = 11; ntab = 25;
h = [-0.8 0.6 0.002 1.05 -0.43 0.19 0.512 -0.005 -0.2 0.3 -0.01 0.085];
% Set up parameters and signals.
M = 4; % Alphabet size for modulation
msg = randint(1500,1,M); % Random message
modmsg = pskmod(msg,M); % Modulate using QPSK.
trainlen = 500; % Length of training sequence
filtmsg = filter(h,1,modmsg); % Introduce channel distortion.
% Equalize the received signal.
[e,W] = lms1(0.01,ntab,filtmsg,modmsg); % Create an equalizer object.
constellation = pskmod([0:M-1],M); % Set signal constellation.
[symbolest] = conv(W,filtmsg); % Equalize.
```

```

% Plot signals.
h = scatterplot(filtmsg,1,trainlen,'bx'); hold on;
scatterplot(symboltest,1,trainlen,'g.',h);
scatterplot(constellation,1,0,'k*',h);
legend('Filtered signal','Equalized signal',...
'Ideal signal constellation');
hold off
nn = -(ntab-1)/2:(ntab-1)/2;
figure
plot(nn,W)
figure
Feq1 = fft(W);
ff = Fs*(0:0.01:(length(W)/2))/length(W);
PP2 = abs(Feq1/(ntab));
Hweq1 = PP2(1:0.01:(ntab)/2+1);
Hweq1(2:end-1) = 2*Hweq1(2:end-1);
plot(ff,Hweq1)

function [e,w]=lms1(mu,M,u,d)
% Call:
% [e,w]=lms(mu,M,u,d);
%
% Input arguments:
% mu = step size, dim 1x1
% M = filter length, dim 1x1
% u = input signal, dim Nx1
% d = desired signal, dim Nx1
%
% Output arguments:
% e = estimation error, dim Nx1
% w = final filter coefficients, dim Mx1
%initial values: 0
w=zeros(M,1);
%number of samples of the input signal
N=length(u);
%Make sure that u and d are column vectors
u=u(:);
d=d(:);
%LMS
for n=M:N
uvec=u(n:-1:n-M+1);
e(n)=d(n)-w'*uvec;
w=w+mu*uvec*conj(e(n));
end
e=e(:);

```



iii)

```
clear all
clc
```



```

Fs = 1000;L =11;ntabfor = 25;ntabback = 3;
h = [-0.8 0.6 0.002 1.05 -0.43 0.19 0.512 -0.005 -0.2 0.3 -0.01 0.085];
% Set up parameters and signals.
M = 4; % Alphabet size for modulation
msg = randint(1500,1,M); % Random message
modmsg = pskmod(msg,M); % Modulate using QPSK.
trainlen = 500; % Length of training sequence
filtmsg = filter(h,1,modmsg); % Introduce channel distortion.
% Equalize the received signal.
eq1 = dfe(ntabfor,ntabback,rls(0.3)); % DFE
eq1.SigConst = pskmod([0:M-1],M); % Set signal constellation.
[symbolest,yd] = equalize(eq1,filtmsg,modmsg(1:trainlen)); % Equalize.
% Plot signals.
h = scatterplot(filtmsg,1,trainlen,'bx'); hold on;
scatterplot(symbolest,1,trainlen,'g.',h);
scatterplot(eq1.SigConst,1,0,'k*',h);
legend('Filtered signal','Equalized signal',...
'Ideal signal constellation');
hold off;

```

