

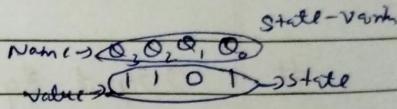
State-Space Analysis

State: It represents present operating condition of a system in terms of stored values within the system.

e.g.: L, C, sequential circuit.

State-variable: Variables that are used to represent state of a system are called as state-variables.

e.g.: i_L , v_C , Q (seq. ckt.)

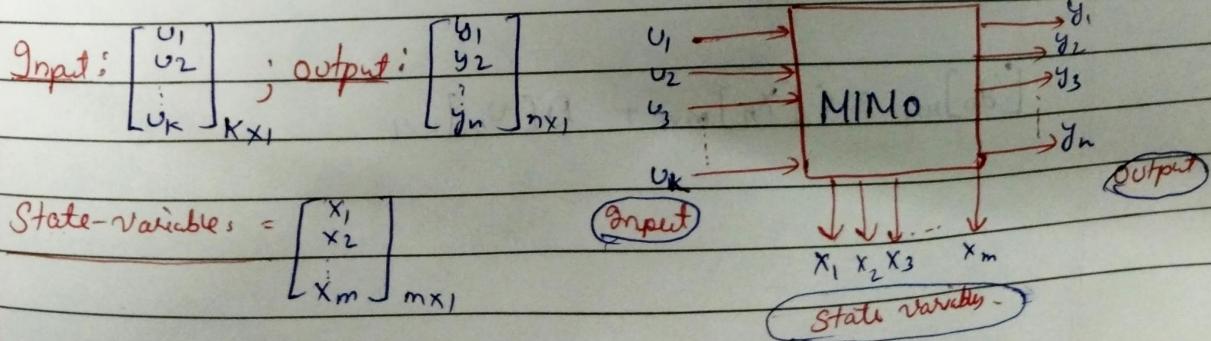


State-space: Set of all possible combination of state variable. State of a system is an element of state space.

Advantages:

- Applicable for LTI & TTV system.
- Consider initial conditions.
- All internal state variables can be considered.
- Applicable for MIMO
- Controllability & observability can be determined

Representation:



State-equation: $\dot{x} = Ax + Bu$

$$[\dot{x} = Ax + Bu] \Rightarrow [\dot{x}]_{m \times 1} = [A][x]_{m \times 1} + [B][u]_{k \times 1}$$

$$\dot{x} = \begin{bmatrix} \frac{dx_1}{dt} \\ \frac{dx_2}{dt} \\ \vdots \\ \frac{dx_m}{dt} \end{bmatrix}_{m \times 1} \quad [A]_{m \times m} \quad [B]_{m \times k}$$

Output Equation: $y = Cx + Du$

$$[y]_{n \times 1} = [C][x]_{m \times 1} + [D][u]_{k \times 1}$$

$$[C]_{n \times m} \quad [D]_{n \times k}$$

Discrete-Time-System

→ State equation:

$$[x_{n+1}]_{m \times 1} = [A][x_n]_{m \times 1} + [B][u]_{k \times 1}$$

x_n : n^{th} sample of x

x_{n+1} : $(n+1)^{\text{th}}$ sample of x

Output equation:

$$[y_n]_{n \times 1} = C[x_n]_{m \times 1} + D[u_n]_{k \times 1}$$

Differential equation to state model.

$$\frac{d^3y}{dt^3} + \frac{6}{dt^2} \frac{dy}{dt} + \frac{3}{dt} y + 4y = 7u$$

State variable : $x_1 = y$

$$x_2 = \frac{dy}{dt} = \dot{x}_1$$

$$x_3 = \frac{d^2y}{dt^2} = \ddot{x}_1$$

$$\dot{x}_3 = \frac{d^3y}{dt^3}$$

$$\dot{x}_3 = -4y - 3\frac{dy}{dt} - 6\frac{d^2y}{dt^2} + 7u$$

$$\dot{x}_3 = -4x_1 - 3\frac{dx_1}{dt} - 6\frac{d^2x_1}{dt^2} + 7u$$

$$\boxed{\dot{x}_3 = -4x_1 - 3x_2 - 6x_3 + 7u}$$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -4 & -3 & -6 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 7 \end{bmatrix} u$$

o/p -

$$y = [1 \ 0 \ 0] \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + [0] u$$

$$\frac{d^5y}{dt^5} + 0 \cdot \frac{d^4y}{dt^4} + 6 \frac{d^3y}{dt^3} + 7 \frac{d^2y}{dt^2} + 0 \cdot \frac{dy}{dt} + 9y = 110$$

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$$\frac{d^5y}{dt^5} + 6 \frac{d^3y}{dt^3} + 7 \frac{d^2y}{dt^2} + 9y = 110$$

$$\frac{d^5y}{dt^5} + 6 \frac{d^3y}{dt^3}$$

State-equation.

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \\ \dot{x}_4 \\ \dot{x}_5 \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 0 & 1 \\ -9 & -6 & -7 & -4 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \\ x_5 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \\ 11 \end{bmatrix}$$

Output eq:

$$y = [1 \ 0 \ 0 \ 0 \ 0] \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \\ x_5 \end{bmatrix} + [0]_0$$

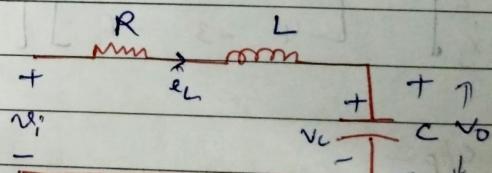
Electrical Network to state model.

$$\begin{cases} v_L = L \frac{di_L}{dt} \\ i_L = C \frac{dv_C}{dt} \end{cases}$$

$$v_L = v_s - i_L R - v_C$$

KVL

$$\frac{di_L}{dt} = -R \frac{i_L}{L} - \frac{1}{L} v_C + \frac{v_s}{L}$$



State-variable: i_L, v_C

$i_C = i_L$

KCL -

$$\frac{dv_C}{dt} = \frac{i_L}{C} ; v_0 = v_C$$

$$\begin{bmatrix} \dot{i}_L \\ \dot{v}_C \end{bmatrix} = \begin{bmatrix} -R/L & -1/L \\ 1/C & 0 \end{bmatrix} \begin{bmatrix} i_L \\ v_C \end{bmatrix} + \begin{bmatrix} 1/L \\ 0 \end{bmatrix} v_s$$

$$v_0 = [0 \ 1] \begin{bmatrix} i_L \\ v_C \end{bmatrix} + [0] v_s$$

Q1 A system is represented by $\frac{3}{dt}y + 2y = u$. What is the transfer function of the system?

$$\Rightarrow \frac{3}{dt}y + 2y = u$$

L.T of eq ①

$$3sy(s) + 2y(s) = u(s)$$

$$y(s)[3s + 2] = u(s)$$

$$\boxed{\frac{y(s)}{u(s)} = \frac{1}{3s+2}}$$

Q2 The dynamic model of a pendulum is given by

$$\frac{d^2\theta}{dt^2} + 400\theta = 100T.$$

where - θ = displacement

T = Torque

If representation in time ~~sp~~ Scale-state variable form
 $\dot{x} = \alpha x + \beta u$ can have constant? $\left\{ \alpha = ? \quad \beta = ? \right.$

$$\frac{d^2\theta}{dt^2} + 400\theta = 100T \Rightarrow \frac{d^2\theta}{dt^2} - \frac{\partial \theta}{\partial t} + 400\theta = 100T$$

$$\begin{bmatrix} \dot{\theta}_1 \\ \dot{\theta}_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -400 & 0 \end{bmatrix} \begin{bmatrix} \theta_1 \\ \theta_2 \end{bmatrix} + \begin{bmatrix} 0 \\ 100 \end{bmatrix} T$$

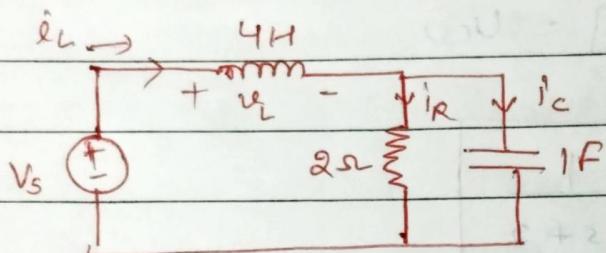
Q3

With reference to state space representation of eqⁿ
 $\frac{d^2y}{dt^2} + 2\frac{dy(t)}{dt} + 3y(t) = u(t)$, the matrix A & B are?

$$\Rightarrow \begin{bmatrix} \dot{y}_1 \\ \dot{y}_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -3 & -2 \end{bmatrix} \begin{bmatrix} y_1 \\ y_2 \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u(t)$$

Q

Consider the network shown below -



Find the state space representation for the above network -

$$KVL : \quad V_L = V_s - V_c \quad \left\{ \begin{array}{l} V_s = V_L + V_c \end{array} \right.$$

$$\frac{d\dot{i}_L}{dt} = +\frac{V_s}{4}$$

$$\frac{di_L}{dt} = \frac{V_s}{L} - \frac{V_c}{L} \Rightarrow \boxed{\frac{d\dot{i}_L}{dt} = +\frac{V_s}{4} - \frac{V_c}{4}}$$

KCL -

$$\dot{i}_c = \dot{i}_L - \dot{i}_R$$

$$\left\{ \begin{array}{l} \dot{i}_L = \dot{i}_R + \dot{i}_c \end{array} \right.$$

$$\boxed{\frac{dv_c}{dt} = \dot{i}_L - \frac{v_c}{R} = \dot{i}_L - \frac{v_c}{2}}$$

$$\boxed{\begin{bmatrix} \dot{i}_L \\ \dot{i}_c \end{bmatrix} = \begin{bmatrix} 0 & -1/4 \\ 1 & -1/2 \end{bmatrix} \begin{bmatrix} \dot{i}_L \\ v_c \end{bmatrix} + \begin{bmatrix} 1/4 \\ 0 \end{bmatrix} v_s}$$

CANONICAL FORM

Transfer function of a system is unique but there may be multiple state model associated with it & such models are called canonical form.

$$(a) \text{ Transfer function} = \frac{b(s^3 + c_2 s^2 + c_1 s + c_0)}{(s^4 + a_3 s^3 + a_2 s^2 + a_1 s + a_0)}$$

Controllable canonical form - (CCF)

$$\dot{x} = \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \\ \dot{x}_4 \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ -a_0 & -a_1 & -a_2 & -a_3 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 0 \\ b \end{bmatrix}$$

→ O/P -

$$y = [c_0 \ c_1 \ c_2 \ 1] \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix}$$

$$y = Cx + Dv$$

$$(b) T(s) = \frac{b(s^3 + c_2 s^2 + c_1 s + c_0)}{(s^4 + a_3 s^3 + a_2 s^2 + a_1 s + a_0)}$$

→ Observable canonical form (OCF) \Rightarrow ~~EEG~~

$$A = \begin{bmatrix} 0 & 0 & 0 & -a_0 \\ 0 & 0 & 0 & -a_1 \\ 0 & 1 & 0 & -a_2 \\ 0 & 0 & 1 & -a_3 \end{bmatrix}$$

$$B = \begin{bmatrix} c_0 \\ c_1 \\ c_2 \\ c_3 \end{bmatrix} \quad C = \begin{bmatrix} 0 & 0 & 0 & b \end{bmatrix}$$

$$G(s) = \frac{Y(s)}{U(s)} = \frac{20(10s+1)}{s^3 + 3s^2 + 2s + 1} \quad (\text{Transfer function})$$

Q Find the state equation & output equation.

$$\frac{Y(s)}{U(s)} = \frac{x_1(s)}{U(s)} \cdot \frac{Y(s)}{x_1(s)} = \frac{20}{s^2 + 3s^2 + 2s + 1} \cdot (10s+1)$$

$$\text{Let } \frac{x_1(s)}{U(s)} = \frac{20}{s^3 + 3s^2 + 2s + 1} \quad \text{Eq } \frac{Y(s)}{x_1(s)} = (10s+1) \quad (2)$$

eq ① can be re-written as -

$$x_1(s) [s^3 + 3s^2 + 2s + 1] = 20 U(s)$$

on taking inverse laplace -

$$\dot{x}_3 + 3x_3 + 2x_2 + x_1 = 20 u(t)$$

$$\dot{x}_3 = -x_1 - 2x_2 - 3x_3 + 20 u(t)$$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -1 & -2 & -3 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 20 \end{bmatrix} [u]$$

State equation

Now - eq ② can be re-written as -

$$\frac{Y(s)}{x_1(s)} = (10s+1)$$

$$Y(s) = (10s+1) x_1 = 10s x_1(s) + x_1(s)$$

on taking inverse Laplace -

$$y(t) = 10x_2 + x_1$$

$$y = \begin{bmatrix} 1 & 10 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix}$$

Output equation

$$\begin{cases} x_1(s) \cdot s^3 = x_3 \\ x_1(s) \cdot s^2 = \dot{x}_2 = x_3 \\ x_1(s) \cdot s^1 = \ddot{x}_1 = x_2 \\ x_1(s) = x_1 \end{cases}$$

$$\textcircled{O} \quad \frac{Y(s)}{U(s)} = \frac{K}{s^0 (s+1) (s+2) (s^2 + 1)}$$

$$\frac{Y(s)}{U(s)} = \frac{x_1(s)}{U(s)} \cdot \frac{Y(s)}{x_1(s)} = \frac{K}{(s+1)(s+2)(s^2 + 1)} \cdot 1$$

$$\textcircled{a} - \frac{x_1(s)}{U(s)} = \frac{K}{(s+1)(s+2)(s^2 + 1)} \Rightarrow \frac{K}{(s^4 + 3s^3 + 3s^2 + 3s + 2)}$$

$$x_1(s) [s^4 + 3s^3 + 3s^2 + 3s + 2] = K U(s)$$

on taking inverse-laplace -

$$\dot{x}_4 + 3x_4 + 3x_3 + 3x_2 + 2x_1 = K U(t)$$

$$\dot{x}_4 = -3x_4 - 3x_3 - 3x_2 - 2x_1 + K U(t)$$

$$\begin{cases} s x_1(s) = x_1 \\ x_1 = x_2 \\ \ddot{x}_1 = \ddot{x}_2 = x_3 \\ \dddot{x}_1 = \ddot{x}_3 = x_4 \\ \ddot{\ddot{x}}_1 = \ddot{x}_4 = x_5 \end{cases}$$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \\ \dot{x}_4 \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ -2 & -3 & -3 & -3 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 0 \\ K \end{bmatrix} [U]$$

$$\textcircled{b} - \frac{Y(s)}{X_1(s)} = 1 \Rightarrow Y(s) = X_1(s)$$

Taking inverse laplace transform -

$$y(t) = x_1$$

$$Y = [1 \ 0 \ 0 \ 0] \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix}$$

CONTROLLABILITY

A control system is said to be controllable if the initial states of the control system are changed to some other desired states by controlled input in finite duration of time.

KALAM'S TEST

Step 1 Write the matrix $\Omega_c = [B \ AB \ A^2B \ \dots \ A^{n-1}B]$

Find determinant of Ω_c , if it is not equal to 0
Then control system is controllable.

Q Verify the controllability of a control system which is represented by state equation.

$$\dot{x} = \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \begin{bmatrix} -1 & -1 \\ 1 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u$$

$$\dot{x} = Ax + Bu$$

Kalam's test: $\Omega_c = [B \ AB \ A^2B \ \dots \ A^{n-1}B]$

$$\text{Given } A = \begin{bmatrix} -1 & -1 \\ 1 & 0 \end{bmatrix}; B = \begin{bmatrix} 1 \\ 0 \end{bmatrix} \quad n=2$$

So -

$$\Omega_c = [B \ AB] = \begin{bmatrix} 1 & -1 \\ 0 & 1 \end{bmatrix}$$

$$AB = \begin{bmatrix} -1 & -1 \\ 1 & 0 \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$

$$AB = \begin{bmatrix} -1+0 & -1 \\ 1+0 & 0 \end{bmatrix} \xrightarrow{\text{minor}} \left\{ \begin{bmatrix} -1 & -1 \\ 1 & 0 \end{bmatrix} \right\} = \begin{bmatrix} 1 & -1 \\ 1 & 0 \end{bmatrix}$$

$$A \cdot B = \begin{bmatrix} -1 \\ 1 \end{bmatrix}$$

Step 2 $|Q_c| = \begin{vmatrix} 1 & -1 \\ 0 & 1 \end{vmatrix}$

$$|Q_c| = \begin{vmatrix} 1 & -1 \\ 0 & 1 \end{vmatrix}$$

$\neq 0 \rightarrow \text{controllable}$

Q The state eq. of a system is given as -

$$\dot{x}_1 = 2x_1 + x_2 + u$$

$$\dot{x}_2 = -2x_2$$

check for controllability.

$$\Rightarrow \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 2 & 1 \\ 0 & -2 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u$$

$$A = \begin{bmatrix} 2 & 1 \\ 0 & -2 \end{bmatrix} \quad B = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$

$$Q_c = [B \quad AB]$$

$$AB = \begin{bmatrix} 2 & 1 \\ 0 & -2 \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} = \begin{bmatrix} 2+0 \\ 0+0 \end{bmatrix} = \begin{bmatrix} 2 \\ 0 \end{bmatrix}$$

$$Q_c = \begin{bmatrix} 1 & 2 \\ 0 & 0 \end{bmatrix}$$

$$|Q_c| = \begin{vmatrix} 1 & 2 \\ 0 & 0 \end{vmatrix} = 0 \quad \text{Not controllable.}$$

$$A = \begin{bmatrix} -2 & 0 \\ 0 & -1 \end{bmatrix} \quad B = \begin{bmatrix} 1 \\ 1 \end{bmatrix}$$

Check the controllability of the system.

OBSERVABILITY

A control system is said to be observable if it is able to determine the initial state of the control system by observing the o/p in finite duration of time.

Kalman's test

Step 1 From the matrix $Q_o = [C^T \quad A^T C^T \quad (A^T)^2 C^T \dots \quad (A^T)^{n-1} C^T]$

Step 2 Take determinant of Q_o , If it is not equal to zero then the control system is observable.

Verify the observability of a control system which is represented in the state space model.

$$\dot{x} = \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} -2 & -2 \\ 1 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 1 \\ 1 \end{bmatrix} u$$

$$y = \begin{bmatrix} 1 & 1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

Given -
 $A = \begin{bmatrix} -2 & -2 \\ 1 & 0 \end{bmatrix} ; B = \begin{bmatrix} 1 \\ 1 \end{bmatrix} ; C = \begin{bmatrix} 1 & 1 \end{bmatrix} ; D = 0$

Step 1 $Q_o = [C^T \quad A^T C^T \quad (A^T)^2 C^T \dots \quad (A^T)^{n-1} C^T]$

$$C^T = \begin{bmatrix} 1 \\ 1 \end{bmatrix} \quad A^T = \begin{bmatrix} -2 & 1 \\ -2 & 0 \end{bmatrix} \quad \boxed{n=2}$$

$$\boxed{Q_o = [C^T \quad A^T C^T]}$$

$$A^T \cdot C^T = \begin{bmatrix} -2 & 1 \\ -2 & 0 \end{bmatrix} \begin{bmatrix} 1 \\ 1 \end{bmatrix} = \begin{bmatrix} -2+1 \\ -2+0 \end{bmatrix} = \begin{bmatrix} -1 \\ -2 \end{bmatrix}$$

$$\boxed{Q_o = \begin{bmatrix} 1 & -1 \\ 1 & -2 \end{bmatrix}}$$

Step 2 $|Q_o| = \begin{vmatrix} 1 & -1 \\ 1 & -2 \end{vmatrix} = (-2 - (-1)) = -1$

$|Q_o| \neq 0$ OBSCRVABLE

Transfer function decomposition

The process of obtaining state model from the transfer function is called as transfer function decomposition.

It is of three types -

(i) Direct decomposition -

$$f(s) = \frac{Y(s)}{U(s)} = \frac{10}{s^2 + 5s + 6}$$

$$Y(s) [s^2 + 5s + 6] = 10 U(s)$$

Let -
 $y(t) = x_1$

I.L.T.
 $\frac{d^2y}{dt^2} + 5\frac{dy}{dt} + 6y(t) = 10u(t)$

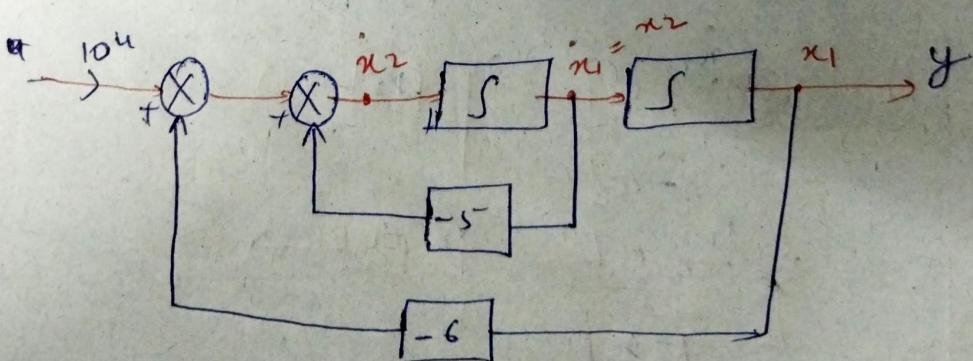
or - $\dot{x}_2 + 5x_2 + 6x_1 = 10u$

$$\dot{x}_2 = -5x_2 - 6x_1 + 10u$$

or $\boxed{\dot{x}_2 = -6x_1 - 5x_2 + 10u}$

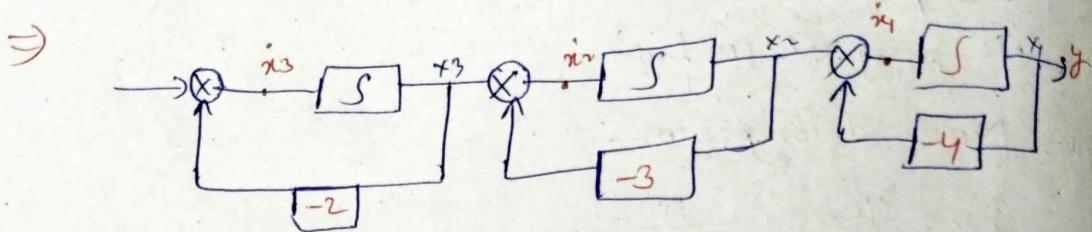
$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -6 & -5 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ 10 \end{bmatrix} u$$

$$Y = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$



② Cascade decomposition

$$\frac{Y(s)}{U(s)} = \frac{1}{(s+2)(s+3)(s+4)}$$



$$\dot{x}_1 = -4x_1 + x_2$$

$$\dot{x}_2 = -3x_2 + x_3$$

$$\dot{x}_3 = -2x_3 + 4$$

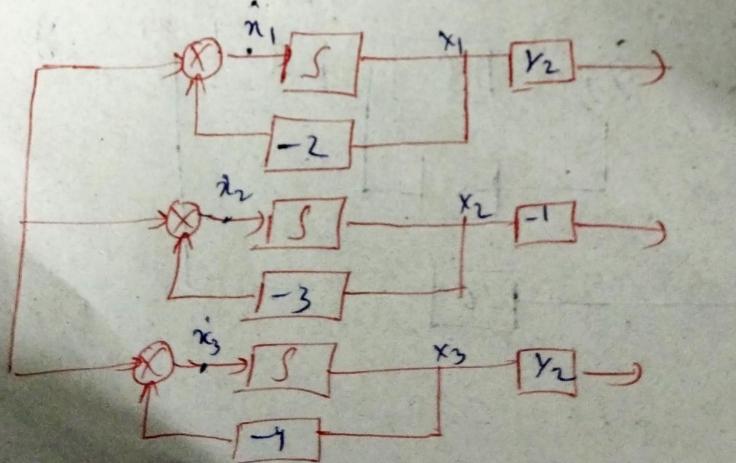
$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} -4 & 1 & 0 \\ 0 & -3 & 1 \\ 0 & 0 & -2 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} [4]$$

↑ upper triangular form

③ Parallel decomposition

$$\frac{Y(s)}{U(s)} = \frac{1}{(s+2)(s+3)(s+4)} = \frac{A}{(s+2)} + \frac{B}{(s+3)} + \frac{C}{(s+4)}$$

$$\frac{Y(s)}{U(s)} = \frac{Y_1}{(s+2)} + \frac{(-1)}{(s+3)} + \frac{(Y_2)}{(s+4)}$$



$$y = \frac{1}{2}x_1 - x_2 + \frac{1}{2}x_3$$

$$y = \begin{bmatrix} +\frac{1}{2} & -1 & \frac{1}{2} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix}$$

$$\dot{x}_1 = -2x_1 + 4$$

$$\dot{x}_2 = -3x_2 + 4$$

$$\dot{x}_3 = -4x_3 + 4$$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} -2 & 0 & 0 \\ 0 & -3 & 0 \\ 0 & 0 & -4 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 1 \\ 1 \\ 1 \end{bmatrix} e_4$$

All poles are in diagonal.