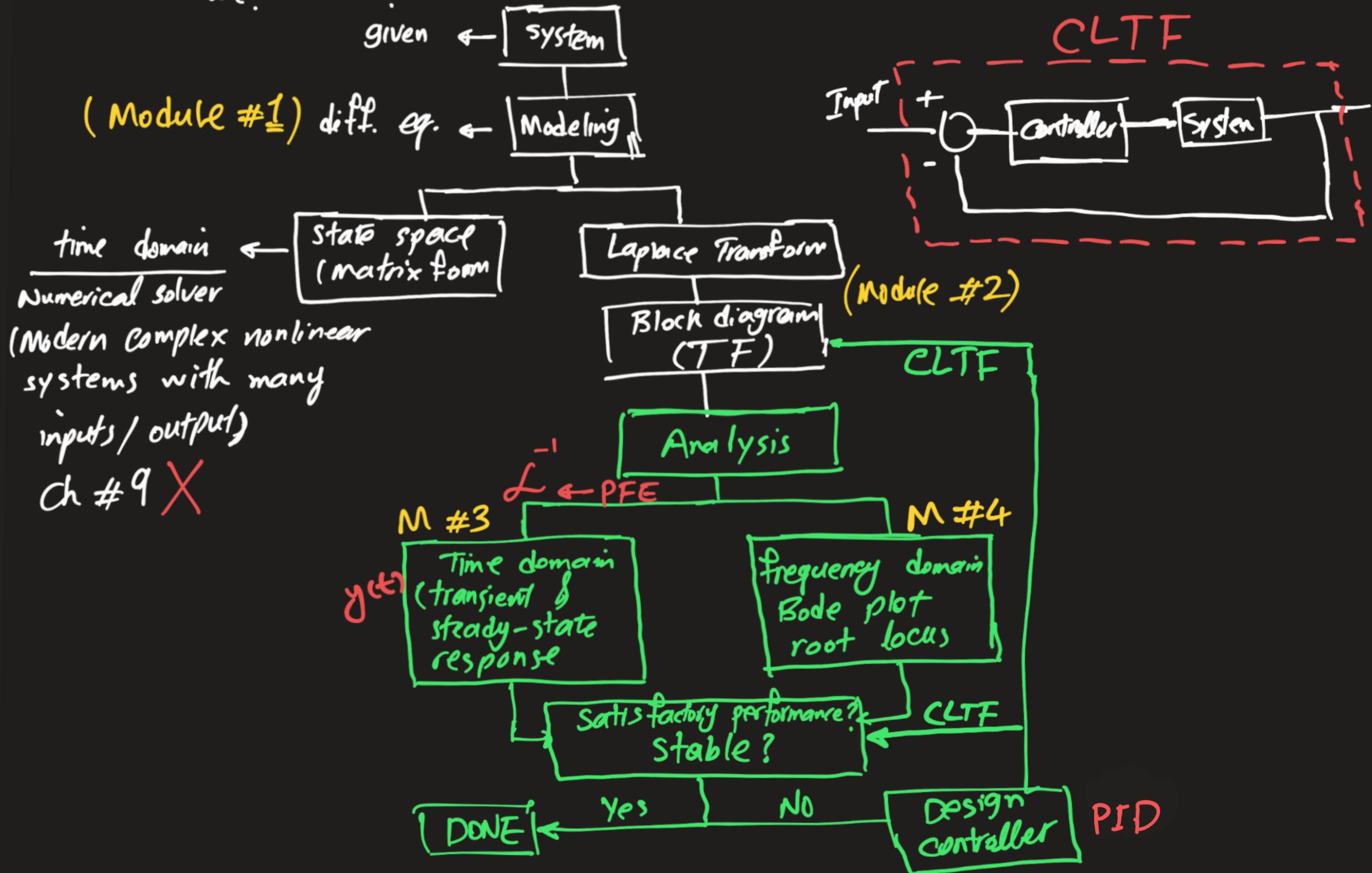


Where are we?



Frequency Response of LTI systems : Chapter 7

LTI: Linear Time Invariant: systems are described by linear equations w/ coefficients that do not change w/ time

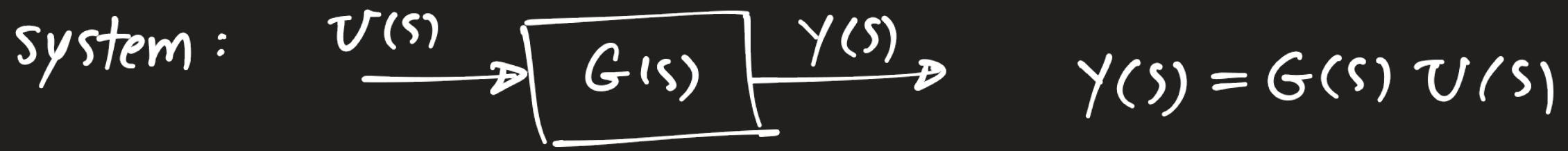
Transient Response: How the output evolves w/ time.

- Adv. {
- useful for non-periodic inputs (impulse, step, ramp)
 - Explicitly shows closed-loop poles, which indicate trade-offs in transient specs. (t_r , M_p , etc.)
- Disadv. {
- stability is inferred separately using Routh criterion.
 - Difficult to infer performance for higher-order systems
 - Requires explicit knowledge of TF
 - Does not work for periodic inputs (sinusoidal)

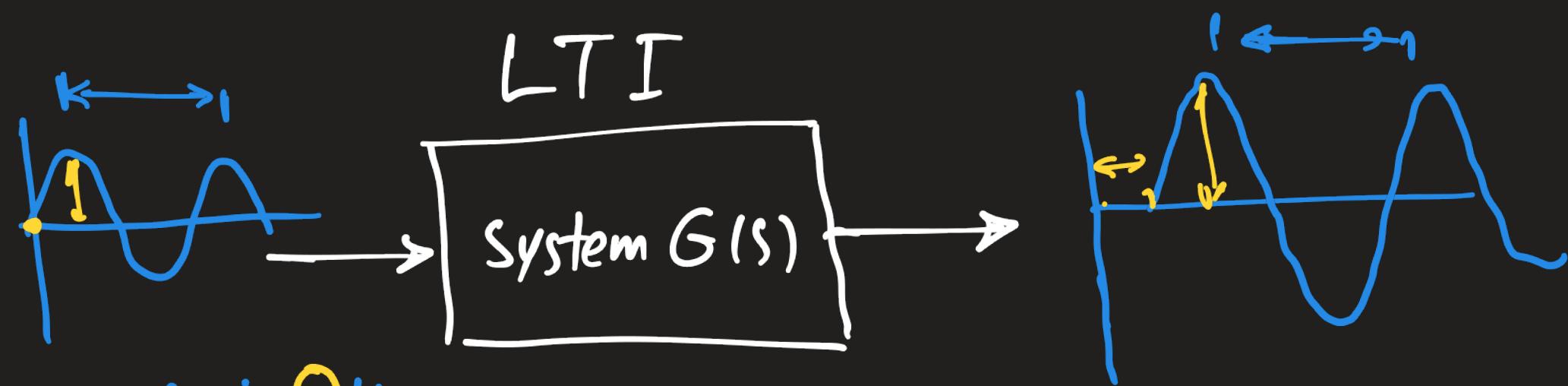
Frequency Response : How a system responds to a sinusoidal input in steady state.

- Advantages {
- » supplements transient response analysis.
 - » performance and stability inferred from same analysis.
 - » graphical \rightarrow easier
 - » can use measured data instead of a model (TF)
 - » independent of system order.
 - » any periodic input can be analyzed.

- Disadvantages {
- \rightarrow less explicit about poles and time behavior.
 - \rightarrow does not work for non-periodic inputs



Objective: Show that a periodic input to a LTI system will result in a periodic steady-state output with the same frequency while amplitude & phase can change.



$$u(t) = A \sin(\omega t)$$

↑
amplitude

↓
frequency

$$\underline{y_{ss}} = B \sin(\omega t + \phi)$$

↑
Amplitude

↓
phase

$$Y(s) = G(s) U(s)$$

$$u(t) = \underline{A} \sin(\underline{\omega}t)$$

$$\int \sin(\omega t) = U(s) = \frac{A\omega}{s^2 + \omega^2}$$

$$\rightarrow s^2 + \omega^2 = (s + j\omega)(s - j\omega) \leftarrow$$

$$PFE: Y(s) = \frac{C_1}{s - \alpha_1} + \frac{C_2}{s - \alpha_2} + \dots + \frac{C_n}{s - \alpha_n} + \frac{\textcircled{K}_1}{s + j\omega} + \frac{\textcircled{K}_2}{s - j\omega}$$

$$\stackrel{-1}{\mathcal{L}} \Rightarrow y(t) = C_1 e^{-\alpha_1 t} + C_2 e^{-\alpha_2 t} + \dots + C_n e^{-\alpha_n t} + K_1 e^{-j\omega t} + K_2 e^{j\omega t}$$

they all decay to zero at $s = s$ ($t \rightarrow \infty$)

Steady-state :
$$Y_{ss} = \lim_{t \rightarrow \infty} y(t) = \underline{K}_1 e^{-j\omega t} + \underline{K}_2 e^{j\omega t}$$

$$K_1 = [Y(s) \cdot (s + j\omega)]_{s=-j\omega} = \left[G(s) \frac{A\omega}{(s + j\omega)(s - j\omega)} \cdot (s + j\omega) \right]_{s=-j\omega} = -\frac{AG(-j\omega)}{2j}$$

$$K_2 = [Y(s) \cdot (s - j\omega)]_{s=j\omega} = \frac{AG(j\omega)}{2j}$$

Reminder: Any complex quantity like $\vec{Z} = a + jb$ can be written in the exp. form: $\vec{Z} = |\vec{Z}| e^{j\phi}$

$$|\vec{Z}| = |a+jb| = \sqrt{a^2+b^2}$$

$$\phi = \angle \vec{Z} = \angle(a+jb) = \tan^{-1} \frac{b}{a} = \phi$$

$G(j\omega)$ is a complex quantity: $\underline{\underline{G(j\omega)}} = |\underline{\underline{G(j\omega)}}| e^{j\phi}$

$$\boxed{\phi = \angle G(j\omega)}$$

$$\Rightarrow K_1 = -\frac{A |G(j\omega)| e^{-j\phi}}{2j}, \quad K_2 = \frac{A |G(j\omega)| e^{j\phi}}{2j}$$

$$\Rightarrow y_{ss} = K_1 e^{-j\omega t} + K_2 e^{j\omega t} = A |G(j\omega)| \left[e^{j(\omega t + \phi)} - e^{-j(\omega t + \phi)} \right]$$

$$\boxed{y_{ss} = \overbrace{A |G(j\omega)|}^{\checkmark} \sin(\omega t + \phi)}$$

Euler formula:
 $\sin(\omega t + \phi)$

$$u(t) = A \sin(\omega t + \phi)$$

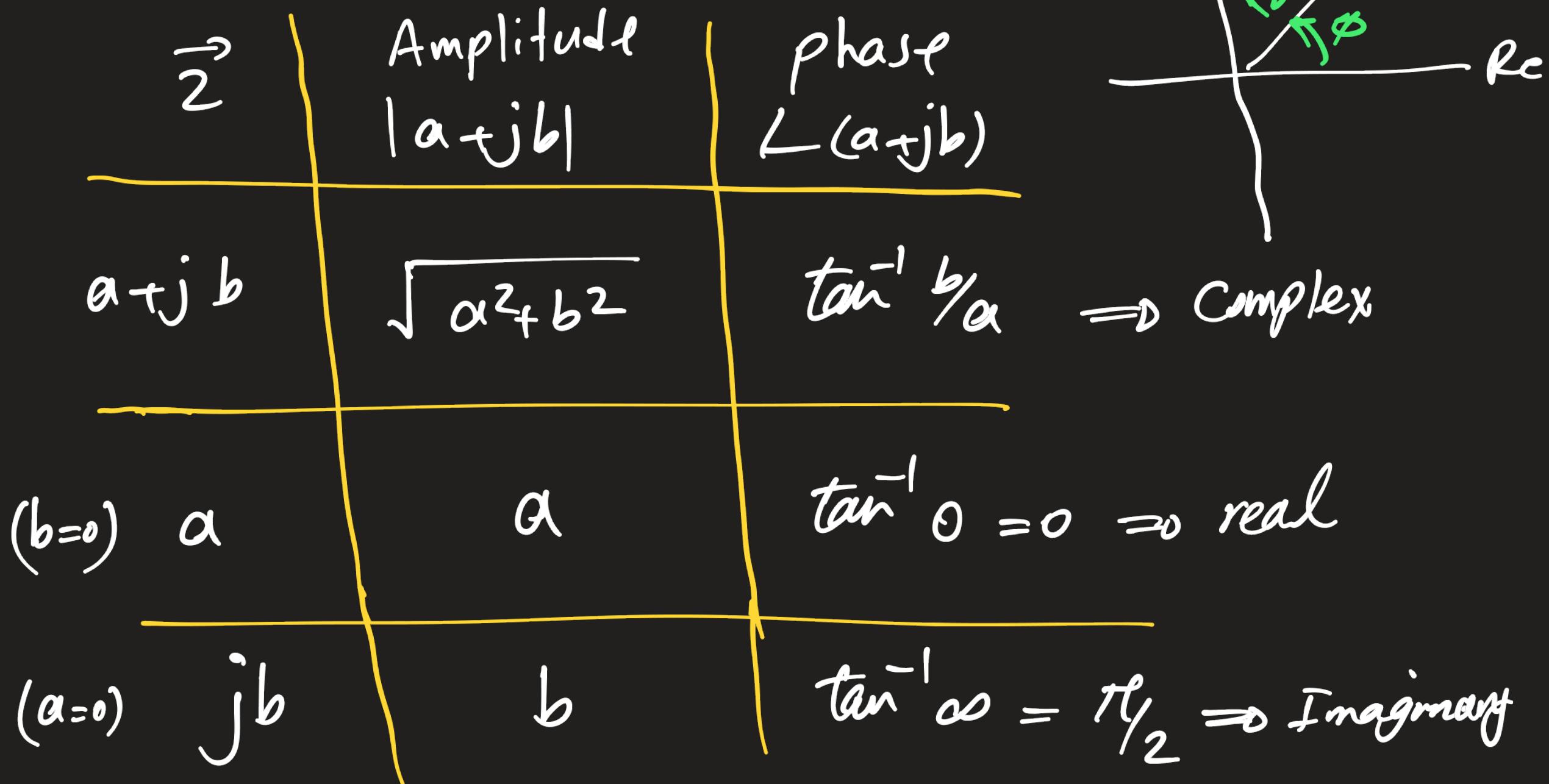
$$y_{ss} = \underbrace{A |G(j\omega)|}_B \sin(\omega t + \underbrace{\angle G(j\omega)}_{\phi})$$

$$\text{Gain} = \frac{B}{A} = \frac{A |G(j\omega)|}{A} = \boxed{|G(j\omega)| = \text{Gain}} *$$

$$\text{Phase} = \phi = \underbrace{\angle G(j\omega)}_{\text{Phase}} *$$

$$y_{ss} = A \underbrace{|G(j\omega)|}_{\text{Gain}} \sin(\omega t + \underbrace{\angle G(j\omega)}_{\text{Phase}})$$

Reminder : Amplitude & phase of a complex, real, & imaginary number.



Reminder (Cont.) : Amplitude & phase of a fraction

$$\text{Amplitude} = \left| \frac{a+jb}{c+jd} \right| = \frac{|a+jb|}{|c+jd|} = \frac{\sqrt{a^2+b^2}}{\sqrt{c^2+d^2}}$$

$$\phi = \angle \left(\frac{a+jb}{c+jd} \right) = \angle(a+jb) - \angle(c+jd)$$

$$\phi = \tan^{-1} \frac{b}{a} - \tan^{-1} \frac{d}{c}$$