

Final

1 a)

$$y_{ss} = A |G(s)| \sin(\omega t + \phi)$$

$$u(t) = .5 \sin(20t)$$

$$y_{ss} = .5 |G(s)| \sin(20t + \phi)$$

$$G(j\omega) = \frac{100j\omega + 100}{j\omega(j\omega^2 + 16j\omega + 100)} = \frac{100(j\omega + 1)}{j\omega \left( \frac{j\omega^2}{100} + \frac{16}{100}j\omega + 1 \right)} \cdot \frac{1}{100} = \frac{j\omega + 1}{j\omega \left( \frac{j\omega^2}{100} + \frac{16}{100}j\omega + 1 \right)}$$

$$\left| \frac{100j\omega + 100}{j\omega(j\omega^2 + 16j\omega + 100)} \right| \quad \omega = 20$$

$$= \left| \frac{100(20j) + 100}{20j((20j)^2 + 16(20j) + 100)} \right| \quad 20^2 = -400$$

$$= \left| \frac{2000j + 100}{20j(320j - 300)} \right|$$

$$= \left| \frac{6400j^2 - 6000j}{2000j + 100} \right|$$

$$= \left| \frac{64j^2 - 60}{20j + 1} \right|$$

$$= \frac{|-124|}{|20j + 1|}$$

$$|G(j\omega)| = \frac{\sqrt{\omega^2 + 1}}{\omega \sqrt{\left(1 - \left(\frac{\omega}{10}\right)^2\right)^2 + \left(\frac{16}{10}\right)^2}}$$

$$\begin{aligned} \text{gain} &= 20 \log(\sqrt{\omega^2 + 1}) - 20 \log(\omega) - 20 \log\left(\sqrt{\left(1 - \left(\frac{\omega}{10}\right)^2\right)^2 + \left(\frac{16\omega}{10}\right)^2}\right) \\ &= 20 \log(\sqrt{20^2 + 1}) - 20 \log(20) - 20 \log\sqrt{\left(1 - \left(\frac{20}{10}\right)^2\right)^2 + \left(\frac{16(20)}{10}\right)^2} \\ &= 21.911 \end{aligned}$$

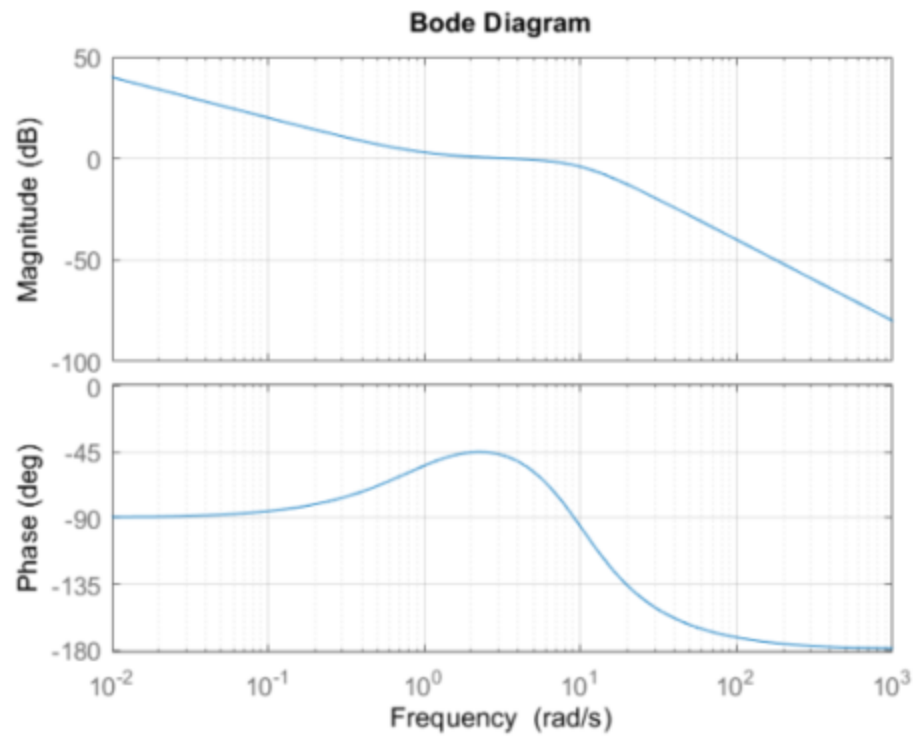
$$\text{phase} = \angle G(j\omega) = \angle(j\omega + 1) - \angle(j\omega(j\omega^2 + 16j\omega + 100))$$

$$= \tan^{-1}(\omega) - 90^\circ - \tan^{-1}\left(\frac{16\omega/10}{1 - (\omega/10)^2}\right) = \tan^{-1}(20) - 90^\circ - \tan^{-1}\left(\frac{16(20)/10}{1 - (20/10)^2}\right) = 81.7818^\circ$$

$$y_{ss} = .5(21.911) \sin(20t + 81.7818^\circ)$$

1b)

```
sys = tf([100,100],[1,16,100,0])  
bode(sys)  
grid on
```





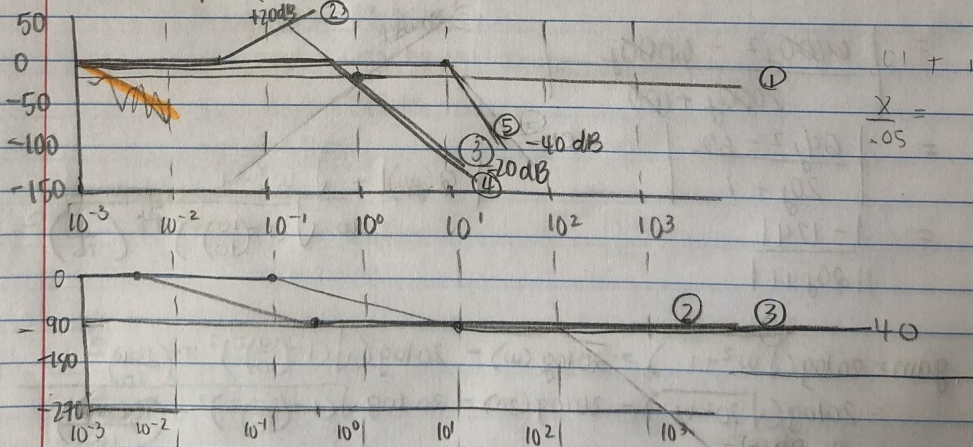
Final

$$2. \quad G(s) = \frac{200s + 10}{s(s+1)(s^2 + 16s + 100)} = \frac{10(20s + 1)}{s(s+1)\left(\left(\frac{s}{10}\right)^2 + \frac{16s}{100} + 1\right)} \cdot \frac{1}{10^0} = \frac{.1(20s + 1)}{s(s+1)\left(\left(\frac{s}{10}\right)^2 + \frac{16s}{100} + 1\right)}$$

factor	gain	phase
① $.1$	$20 \log(.1) = -20$	0 [0 until]
② $\frac{s}{.05} + 1$ [ $\omega_0 = .05$ ]	0 dB until .05 then +20 dB	$.1(.05) = .005$ low $0^\circ$ then high $-90^\circ$
③ $\frac{1}{s}$	-20 dB goes thru 0 @ $\omega = 1$	$10(.05) = .5$ [-90° after] $-90^\circ$
④ $\frac{1}{s+1}$ $\omega_0 = 1$	0 until $\omega = 1$ then -20	[0 until] $.1(1) = .1 \times 10^{-1}$ $10(1) = 10$ [-90° after]
⑤ $\frac{s^2 + 16s + 100}{10}$ $\omega_0 = 10$	0 dB until 10 then -40	$\frac{10}{10^8} = 1 \times 10^{-7}$ $10(10^8) = 10^9$ $0^\circ$ $-180^\circ$

$$\left(\frac{s}{10}\right)^2 + \frac{16s}{100} + 1$$

$$2\zeta \frac{s}{100} = \frac{16s}{100} \Rightarrow \zeta = 8$$

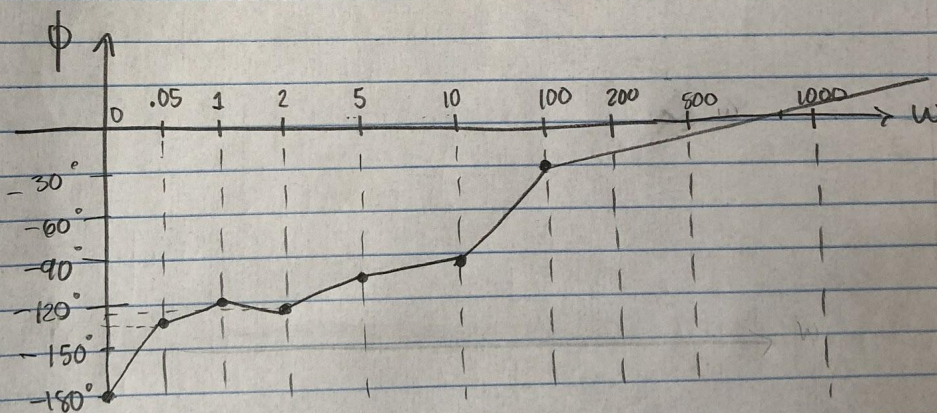
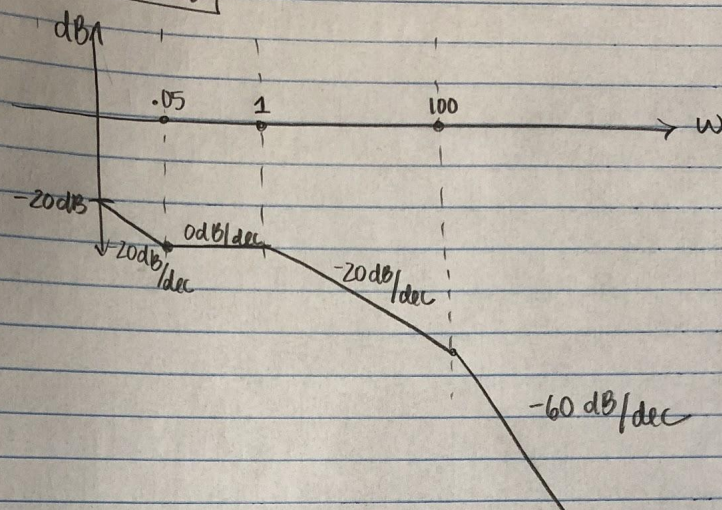




2 cont.

Final

Final plots





Final

3. a) gain crossover frequency: Magnitude (in dB) = 0  
 $\Rightarrow \boxed{2 \text{ rad/sec}}$

Phase crossover frequency: phase  $\angle$  (in deg) =  $180^\circ$

$\Rightarrow$  phase  $\angle$  is  $-180^\circ$  between  $3 \text{ rad/s}$  &  $4 \text{ rad/sec}$

$\Rightarrow$  so phase crossover frequency =  $\sqrt{3 \cdot 4} = \sqrt{12} = \boxed{3.46 \text{ rad/s}}$

gain margin = - Mag @ cross over frequency

Slope from frequency  $2 \text{ rad/sec}$  onward is  $-40 \text{ dB/dec}$

Mag @ phase crossover frequency = Mag @  $2 \text{ rad/s}$  - Slope  $\cdot (\log(\text{phase crossover freq}) - \log(2))$

$$= 0 - 40 \log(3.46/2)$$

$$= \boxed{-9.54 \text{ rad/sec}}$$

Phase margin =  $180 + \text{phase } \angle \text{ @ cross over}$

$$= 180 + (-135^\circ)$$

$$= \boxed{45^\circ}$$

- b) both the gain and phase margins are positive so the system is stable

c) CLTF for bode plot is  $\frac{k}{s(s+2)}$

@  $2 \text{ rad/s}$  mag =  $20 \text{ dB}$

$$20 \log \left( \frac{k}{2(2)} \right) = 20$$

$$k = 2(2)(10) = 4$$

$$\text{CLTF} = \frac{4}{s^2 + 2s + 4}$$

damping factor =  $.5$

$$\text{max overshoot} = e^{-\left(\frac{\xi \pi}{\sqrt{1-\xi^2}}\right)} = e^{-\left(\frac{.5\pi}{\sqrt{1-.5^2}}\right)} = \boxed{.1630}$$



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3. cont.

d) Initial value of slope is equal to  $-20\text{dB/dec}$ , because of the pole @ the origin. From the OLTF we can see that initial (before  $2\text{ rad/s}$ ) only effects the pole @ origin. So initial slope is due to pole @ origin.

4.

phase margin =  $30^\circ$

gain crossover freq =  $15\text{ rad/sec}$

Safety factor  $\angle = 15^\circ$

$$\phi = 30 + 15 = 45^\circ$$

lead compensator:  $G_c(s) = \frac{(sT+1)}{(1+\alpha Ts)}$

phase  $\angle$ :  $\tan^{-1}(wT) - \tan^{-1}(\alpha wT) = \phi$

$$\phi = \tan^{-1} \left( \frac{wT - \alpha wT}{1 + w^2 \alpha T^2} \right)$$

$$\sin \phi = \frac{1 - \alpha}{1 + \alpha}$$

$$\alpha = \frac{1 - \sin \phi}{1 + \sin \phi}$$

$$\alpha = \frac{1 - \sin(45^\circ)}{1 + \sin(45^\circ)} = \frac{1 - 0.707}{1 + 0.707} = \underline{0.1716}$$

$$T = \frac{1}{w_c \sqrt{\alpha}} = \frac{1}{15 \sqrt{0.1716}} = 0.1610$$

gain:

$$\left| \frac{k_c \sqrt{(w_c T)^2 + 1}}{\sqrt{(w_c \alpha T)^2 + 1}} \right| = 1$$

$$\left| \frac{k_c \sqrt{(15(0.1610))^2 + 1}}{\sqrt{(15(0.1716)(0.1610))^2 + 1}} \right| = 1$$

$$\underline{k = 0.5}$$



Final

4 cont.

System design:

$$PM = 45^\circ$$

$$\text{gain} = 15 \text{ rad/sec}$$

$$\phi = 60^\circ$$

$$\frac{1-\alpha}{1+\alpha} = \sin \phi$$

$$\alpha = \frac{1 - \sin \phi}{1 + \sin \phi} = \frac{1 - \frac{\sqrt{3}}{2}}{1 + \frac{\sqrt{3}}{2}} = \frac{.134}{1.866} = \underline{.071}$$

$$T = \frac{1}{\omega_c \sqrt{\alpha}} = \frac{1}{15 \sqrt{.071}} = .25$$

final lead Compensator function:

$$G_c(s) = \frac{k_c (sT + 1)}{(sT\alpha + 1)}$$

$$G_c(s) = \frac{.5(.25s + 1)}{1 + .0175s}$$