

Note: all the simulations are under $TT\ 25^{\circ}C$

1. Use Composer and Hspice to simulate the differential amplifier as shown at Fig. 1 with **$V_{dd} = 1.5V$ and output loading (CL) = $1pF$** . Please design the bias voltage (V_b), input common mode voltage (V_{in_CM}), device size (W/L) of M1~M3 and resistor (R_d) to make differential gain **$|A_{DM}| > 20$, $CMRR > 25\text{ dB}$, and $-3dB\text{ bandwidth} > 1.5MHz$** . Please make sure all the MOSFETs operate in saturation region. (75%)

Hint: You can use the Hspice command showing below in your .sp file

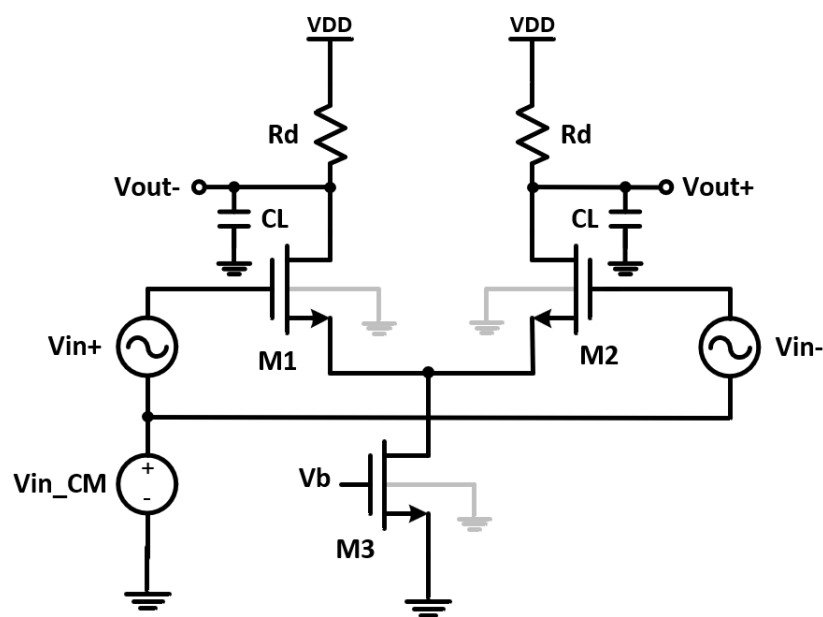


Fig. 1

- (a) Describe your design consideration. (How you choose the size of M1~M3 , R_d , the bias voltage V_b , and V_{in_CM}) (10%)
- (b) Please use “.op” command to print out the small signal parameters of active devices. Please make sure all the MOSFETs operate in saturation region. (5%)
- (c) Please use “.tf” command to print out the small signal parameters. (5%)
- (d) Use the parameters in (b) to calculate the $|A_{DM}|$, check your calculation with the simulation results. (5%)
- (e) Please use “.ac” command to find **common mode gain** at 10kHz and use the A_{DM} from (c) to find **CMRR**. ($CMRR = 20\log(\frac{A_{DM}}{A_{CM-CM}})$) (5%)

(f) Use the parameters in (b) to calculate the **CMRR**, check your calculation with the simulation results. (5%)

(g) Please simulate and plot the frequency response of your design. Use “.pz” to simulate and mark the **dominant pole** on this curve. (5%)

(h) Use the parameters in (a) to calculate the **dominant pole**, and check your calculation with the simulation results. (5%)

(i) Please use the command below to find the input range which fits the requirement $|A_{DM}| > 5$. Please print out the small signal parameters at the maximum and minimum to prove the mos is not in subthreshold region. (10%)
Hint: Input range should be continuous.

(j) Please discuss your design flow and calculate the figure of merit (FoM) value

$$\frac{\text{Total current (uA)}}{\text{Input range (mV)} \times -3\text{dB bandwidth (MHz)}}$$
 Comment how to improve FoM of your design. (5%+15%)

Hint: Please use the Vdd current.

Example:

```
**** voltage sources

subckt
element 0:vip      0:vb      0:vdd      0:vin      0:vss
volts    500.0000m  500.0000m  1.5000  500.0000m  0.
current  0.         0.         -25.8755u  0.         25.8755u
power    0.         0.         38.8132u  0.         0.

total voltage source power dissipation= 38.8132u watts
```

2. Design a 1:6 wide-swing cascade current source as shown in Fig. 2(a) with Vdd=1.8V. (25%)

(a) With Iref = 20uA (Iout = 120uA), design the W/L sizes of M1~M4, and the dc bias Vb to get Rout > 700kΩ when Vout = 300mV. (10%)

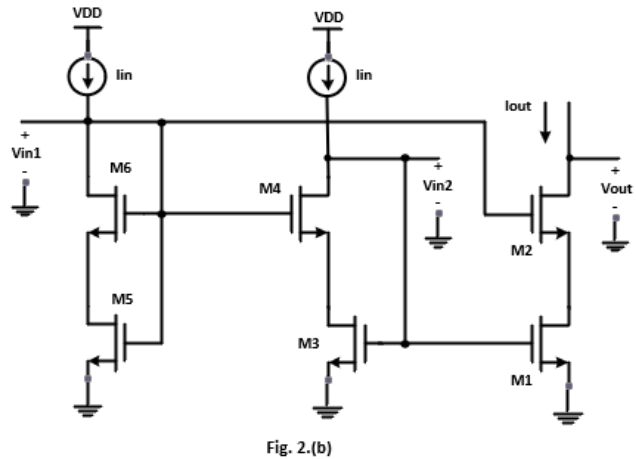
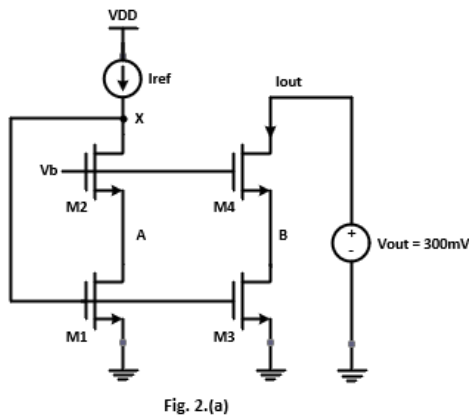
Hint: You can use the following Hspice command in your .sp file to find Rout.

```
.DC Vout 0 1.8 0.01
.meas dc deriv find deriv('I(M3)') at 300mV
.meas dc Rout PARAM='1/deriv'
```

Example:

```
***** dc transfer curves tnom= 25.000 temp= 25.000 *****
deriv= 1.3196u
rout= 757.8167k
```

- (b) Use the circuit structure as shown in Fig. 2(b) as a reference to design a bias generation circuit of V_b with $I_{in} = 20\mu A$ ($I_{out} = 120\mu A$). State the M5's and M6's (Fig. 2. (b)) design strategy and show in hand calculation. And express V_{in1} , V_{in2} , and V_{out} in terms of V_{ov} and V_{th} . (15%)



Reference code

Differential mode:

```
*Input
Vip vip 0 dc Vin_CM ac 0.5 0
Vin vin 0 dc Vin_CM ac 0.5 180

*Gain/Bandwidth
.op
.tf V(voutp, voutn) vip
.pz V(voutp, voutn) vip
.ac dec 10 1 100G
.print ac v(voutp) v(voutn) v(voutp, voutn)
.probe Vdb(voutp, voutn) Vp(voutp, voutn)
.meas ac dcgain_in_db max Vdb(voutp, voutn)
.meas ac BW when Vdb(voutp, voutn) = 'dcgain_in_db - 3'
```

Result examples:

Adm:

```
****      small-signal transfer characteristics

v(voutp,voutn)/vip          = 20.5256 |Adm|
input resistance at vip     = 1.000e+20
output resistance at v(voutp,voutn) = 153.2831k
```

Pole/Zero:

```
***** pole/zero analysis |

input = 0:vip      output = v(voutp,voutn)

      poles (rad/sec)      poles (hertz) 注意單位
real      imag      real      imag
-8.51523x  0.      -1.35524x  0.
-10.0208x  0.      -1.59485x  0.  dominant pole 請看第 2 個
-57.8283x  0.      -9.20365x  0.
      zero 最近的關係相消

      zeros (rad/sec)      zeros (hertz)
real      imag      real      imag
-8.43947x  0.      -1.34318x  0.
-57.6666x  0.      -9.17793x  0.
5.46304g   0.      869.470x   0.

**** constant factor = 18.0480m
```

BW:

```
dcgain_in_db= 26.2459      at= 1.0000
      from= 1.0000      to= 100.0000g
bw= 1.5912x -3dB bandwidth
```

Common mode:

```
*Input
Vip vip 0 dc Vin_CM ac 1 0
Vin vin 0 dc Vin_CM ac 1 0

*Gain
.op
.ac dec 10 1 100G
.meas ac acm_in_db find vdb(voutp) at=10k
```

Result example:

```
***** ac analysis tnom= 25.000 temp= 25.000 *****
acm_in_db= 641.0592m Common mode gain in dB
```

$$\text{CMRR} = 20 \log(20.5256) - 641.0592\text{m} = 25.605 \text{ dB}$$

Input range test:

```
*Input
.param sweepv=Vin_CM
Vip vip 0 dc 'sweepv' ac 0.5 0
Vin vin 0 dc '2*Vin_CM-sweepv' ac 0.5 180

*Gain
.op
.dc sweepv 0 '2*Vin_CM' 1m
.tf V(voutp, voutn) vip
```

Result example:

Please find the input range around your V_{in_CM} (0.5V for example) which fits the requirement. (Make sure the range is **the continuous range**)

481.0000m	1.000e+20	28.1476k	
482.0000m	1.000e+20	29.0007k	
483.0000m	1.000e+20	30.0169k	5.0859
484.0000m	1.000e+20	31.2398k	5.2313
485.0000m	1.000e+20	32.7294k	5.4098
486.0000m	1.000e+20	34.5706k	5.6321
487.0000m	1.000e+20	36.8869k	5.9134
488.0000m	1.000e+20	39.8615k	6.2767
489.0000m	1.000e+20	43.7735k	6.7568
490.0000m	1.000e+20	49.0494k	7.4073
491.0000m	1.000e+20	56.3372k	8.3096
492.0000m	1.000e+20	66.5372k	9.5773
493.0000m	1.000e+20	80.5628k	11.3276
494.0000m	1.000e+20	98.3243k	13.5540
495.0000m	1.000e+20	117.1996k	15.9332
496.0000m	1.000e+20	132.9484k	17.9329
497.0000m	1.000e+20	143.4520k	19.2787
498.0000m	1.000e+20	149.4481k	20.0525
499.0000m	1.000e+20	152.4020k	20.4301
500.0000m	1.000e+20	153.2831k	20.5256
501.0000m	1.000e+20	152.4020k	20.3679
502.0000m	1.000e+20	149.4481k	19.9018
503.0000m	1.000e+20	143.4520k	18.9785
504.0000m	1.000e+20	132.9484k	17.3796
505.0000m	1.000e+20	117.1996k	14.9995
506.0000m	1.000e+20	98.3243k	12.1603
507.0000m	1.000e+20	80.5628k	9.4964
508.0000m	1.000e+20	66.5372k	7.3959
509.0000m	1.000e+20	56.3372k	5.8684
510.0000m	1.000e+20	49.0494k	4.7757
511.0000m	1.000e+20	43.7735k	3.9824

For example, my input
common mode is 0.5V

Please find the input range
around your input common mode

Then print out the small signal parameters at $V_{ip} = 0.483V$; $V_{in} = 2*V_{in_CM}-0.483$ and $V_{ip} = 0.509V$; $V_{in} = 2*V_{in_CM}-0.509$ to prove the mos region.

For more detail, you can check Lecture 4, Page 12 !!!

天梯分數計算方式

1. 得到天梯分數資格為需要滿足 $FoM < 1$

- ## 2. 天梯分數級距

1~2 名: 15 分

3~5 名: 13 分

6~10 名: 10 分

11~20 名: 8 分

21 名以後: 5 分

- ### 3. 天梯表格連結:

<https://docs.google.com/spreadsheets/d/1Cb1Ykvu3HXtgSUoKY7dYbgchWF>

[2wmQXm/edit#gid=1062528342](#)

[illegible]

排名會顯示左邊

學號

名字(可隨便取)

填入你的結果