# The Size of Optimal Sequence Sets for Synchronous CDMA Systems

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#### **Abstract**

The sum capacity on a symbol-synchronous CDMA system having processing gain N and supporting K power constrained users is achieved by employing at most 2N-1 sequences. Analogously, the minimum received power (energy-per-chip) on the symbol-synchronous CDMA system supporting K users that demand specified data rates is attained by employing at most 2N-1 sequences. If there are L oversized users in the system, at most 2N-L-1 sequences are needed. 2N-1 is the minimum number of sequences needed to guarantee optimal allocation for single dimensional signaling. N orthogonal sequences are sufficient if a few users (at most N-1) are allowed to signal in multiple dimensions. If there are no oversized users, these split users need to signal only in two dimensions each. The above results are shown by proving a converse to a well-known result of Weyl on the interlacing eigenvalues of the sum of two Hermitian matrices, one of which is of rank 1. The converse is analogous to Mirsky's converse to the interlacing eigenvalues theorem for bordering matrices.

#### **Index Terms**

Code division multiple access (CDMA), eigenvalues, interlacing eigenvalues, inverse eigenvalue problems, sequences, tight frames

#### I. INTRODUCTION

Consider a symbol-synchronous code-division multiple access (CDMA) system. The kth user is assigned an N-sequence  $s_k \in \mathbb{R}^N$  of unit energy, i.e.,  $s_k^t s_k = 1$ . The processing gain is N chips, and the number of users is K, with K > N. User k modulates the vector  $s_k$  by its data symbol  $X_k \in \mathbb{R}$  and transmits  $X_k s_k$  over N chips. This transmission interferes with other users' transmissions and is corrupted by noise. The received signal is modeled by

$$Y = \sum_{k=1}^{K} s_k X_k + Z,$$

where Z is a zero-mean Gaussian random vector with covariance  $I_N$ , the  $N \times N$  identity matrix.

We consider two problems already studied in the literature. **Problem I**: User k has a power contraint  $p_k$  units per chip, *i.e.*,  $E[X_k^2] \leq Np_k$ . The goal then is to assign sequences and data rates to users so that the sum of the individual rates at which the users can transmit data reliably (in an asymptotic sense) is maximized. The maximum value  $C_{sum}$  is called the sum capacity. Viswanath and Anantharam [1] show the following. Let  $p_{tot} = \sum_{k=1}^{K} p_k$ .

- Oversized users, i.e., those capable of transmitting at large powers relative to other users' power contraints, are best allocated non-interfering sequences;
- others are allocated generalized Welch-bound equality (GWBE) sequences [1];
- $C_{sum} \leq (1/2) \log(1 + p_{tot})$ , with equality if and only if no user is oversized;
- no user is oversized if  $Np_k \leq p_{tot}$  for every user k.

**Problem II**, a dual to Problem I, is one where user k demands reliable transmission at a minimum rate  $r_k$  bits/chip. The goal is to assign sequences and powers to users so that despite their mutual interference and noise, each of the users can transmit reliably at or greater than their required rates, and the sum of the received powers (energy/chip) at the base-station is minimized. Guess [2] shows results analogous to those of Problem I.

- Oversized users, i.e., those that demand large rates relative to other users' requirements, are best allocated non-interfering sequences;
- others are allocated GWBE sequences;
- the received sum power is lower bounded by  $\exp\{2r_{tot}\}-1$ , where  $r_{tot}=\sum_{k=1}^{K}r_k$ , with equality if and only if no user is oversized;
- no user is oversized if  $Nr_k \leq r_{tot}$  for every user k.

Once the optimal sequences are identified and the power or rate allocated to a user determined, the quantities have to be signaled to the typically geographically separated users. This "control-plane" signaling eats up some bandwidth on the downlink. This can be a significant fraction of the available resources when the system is dynamic: users may enter and leave

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the system, or the channel may vary as is typical in wireless channels. The sequences and allocations may need periodic updates. Every such update requires the transmission to each user of an N-vector representing the sequence and a real number representing the power or rate allocated. It is therefore of interest to identify schemes that result in reduced signaling overhead.

In this paper, for both Problems I and II, we come up with a generalized WBE sequence allocation for the non-oversized users. The allocation employs at most 2N-L-1 sequences, where L is the number of oversized users. To prove this, we only need to restrict our attention to the case when no user is oversized and show that we need at most 2N-1 sequences. The 2N-L-1 requirement in the presence of oversized users follows immediately. This upper bound on the number of sequences enables reduced signaling on the downlink when there are a large number of users relative to the processing gain.

Following Guess [2], we show achievability with such an allocation via a successively canceling decoder. Indeed, if there are fewer sequences than users, two or more users will be assigned the same sequence and they will overlap with each other. Nonlinear receivers, among which the successively canceling decoder is an example, are therefore necessary. Our algorithm to identify the sequences and powers or rates allocated requires O(KN) floating point operations and is numerically stable.

The solution to the above problem draws from an interesting result in matrix theory. If A is a matrix, let  $A^*$  denote its Hermitian conjugate. If A is Hermitian, let  $\sigma(A)$  denote its eigenvalues in decreasing order. The following is a well-known result due to Weyl (see for e.g., [3, Section 4.3]). If A is an  $N \times N$  Hermitian matrix with  $\sigma(A) = \lambda = (\lambda_1, \dots, \lambda_N)$ , and  $c \in \mathbb{C}^N$ , then  $\sigma(A + cc^*) = (\hat{\lambda}_1, \dots, \hat{\lambda}_N) = \hat{\lambda}$  satisfies the interlacing property

$$\hat{\lambda}_1 \ge \lambda_1 \ge \hat{\lambda}_2 \ge \lambda_2 \ge \dots \ge \hat{\lambda}_N \ge \lambda_N. \tag{1}$$

In this paper, we demonstrate a converse to Weyl's result. We show that for any Hermitian A with  $\sigma(A) = \lambda$ , and any real  $\hat{\lambda}$  such that  $\lambda$  and  $\hat{\lambda}$  satisfy (1), there exists  $c \in \mathbb{C}^N$  such that  $\sigma(A + cc^*) = \hat{\lambda}$ . This converse facilitates a sequence allocation algorithm that, for both Problems I and II, satisfies the declared properties on the number of sequences. The above problem falls within the class of structured inverse eigenvalue problems. Chu [4], Chu and Golub [5] provide excellent reviews of similar inverse eigenvalue problems. The problem we have solved appears to be new.

We then investigate the need for 2N-1 sequences. While there are rate requirements (respectively, power constraints) for which fewer sequences suffice, in general, 2N-1 is the minimum number required to meet any set of power (respectively, rate) constraints.

Finally, we show that if some users can be split, N sequences suffice. This is useful because any set of N orthogonal sequences will then work. For example, we may employ the standard basis as in a TDMA system, or a set of Walsh sequences when N is a power of 2 as in a CDMA system. Moreover, this set can be fixed up front. As the system evolves, it is sufficient to send an index to this set, thereby making the sequence signaling on the downlink rather easy. Some users may be signaled more than one index. If there are no oversized users, at most N-1 users are split into exactly two each, and therefore will need two indices. Because their energy is now concentrated in a two-dimensional subspace instead of one, the benefits of spreading (such as robustness to jamming) is obtained to a lesser degree. The allocation is however optimal; it will either maximise sum capacity or minimise sum power.

It is perhaps obvious that if multi-dimensional signaling is allowed, then N orthogonal sequences suffice. Indeed, the goal of an optimal allocation is to ensure that energy is spread equally in all dimensions leading to a GWBE allocation. A sufficiently fine splitting of the users into virtual users with smaller requirements, coupled with multiple dimensional signaling per user, will achieve this. The interesting aspects of the above result, however, are an identification of the number of users that need be split and the resulting dimensionality of their signaling.

The problem of sequence detection has attracted much attention since the work of Rupf and Massey [6] who consider the scenario where users have identical power constraints and the goal is to maximize the sum capacity. Viswanath and Anantharam [1] extend the results for users with differing power constraints. Guess [2] studies the dual problem (Problem II) and obtains sequences that minimize sum power subject to users' rate constraints. Waldron [7] shows that GWBE sequences are tight frames, a generalization of bases useful in analyzing wavelet decompositions. Tropp and others [8] show that sequence design for Problem I is a structured inverse singular value problem [5] and provide a numerically stable algorithm whose complexity is O(KN) floating point operations. Our proposed algorithm (first presented in [9]) has the same O(KN) complexity and numerical stability properties. Moreover, our algorithm guarantees optimality with at most 2N-1 sequences and will work for any ordering of users.

Several iterative algorithms for finding optimal sequences have been proposed. As our focus is on finite-step algorithms, we refer the interested reader to a work of Tropp and others [10] and references therein. Another line of research has been that of sequence design for suboptimal linear receivers. Viswanath and others in [11] focus on linear MMSE receivers and identify sequences and powers to meet a per-user signal to interference ratio (SIR) requirement. Guess [12] extends the results to decision-feedback receivers. In our work, we place no restriction on the receivers.

The paper is organized as follows. Section II provides the preliminaries and states the problems. Section III discusses the converse to Weyl's result. Section IV provides the new algorithms, shows their optimality, and verifies that at most 2N-1 sequences are utilized. It also discusses the complexity and numerical stability of the algorithms. Section V deals with the need for 2N-1 sequences and Section VI discusses the rate-splitting approach to minimize the number of sequences. Section VII provides some concluding remarks.

#### II. PRELIMINARIES AND PROBLEM STATEMENTS

Suppose user k is assigned sequence  $s_k$  and is received at power  $p_k$ . Let S be the  $N \times K$  matrix  $[s_1 \ s_2 \ \cdots \ s_K]$ . Then, the capacity region [13] can be written as

$$C(S,p) = \bigcap_{J \subset \{1,\dots,K\}} \left\{ (r_1,\dots,r_K) \in \mathbb{R}_+^K : \sum_{k \in J} r_k \le \frac{1}{2N} \log \left| I_N + \sum_{k \in J} N p_k \cdot s_k s_k^t \right| \right\},$$

$$(2)$$

where |A| denotes the determinant of the matrix A, and  $r_k$  is user k's data rate in bits/chip.

The rate vector  $r = (r_1, \dots, r_K)$  is a *vertex* of this capacity region if

$$r_k \stackrel{\Delta}{=} \frac{1}{2N} \log |A_k| - \frac{1}{2N} \log |A_{k-1}|, \ 1 \le k \le K,$$
 (3)

where  $A_0 \stackrel{\Delta}{=} I_N$ , and  $A_k \stackrel{\Delta}{=} (A_{k-1} + Np_k \cdot s_k s_k^t)$  for  $k = 1, \dots, K$ . The matrix  $A_k$  is real, symmetric, and positive-definite. The vertex (cf. (3)) satisfies  $r_k \ge 0$  and  $r \in C(S, p)$ . These are deduced as follows. Let  $\sigma(A_k) = \left(\lambda_1^{(k)}, \dots, \lambda_N^{(k)}\right)$ . Clearly  $\lambda_1^{(0)} = \cdots = \lambda_N^{(0)} = 1$ . Weyl's result (1) indicates

$$\lambda_1^{(k)} \geq \lambda_1^{(k-1)} \geq \lambda_2^{(k)} \geq \lambda_2^{(k-1)} \geq \dots \geq \lambda_N^{(k)} \geq \lambda_N^{(k-1)} \geq 1,$$

which implies that  $|A_k| \geq |A_{k-1}|$ , and hence  $r_k \geq 0$ . Moreover,  $r \in C(S,p)$  because this rate point can be achieved via successive decoding. (Alternatively, C(S, p) is a polymatroid [14], and therefore contains all its vertices in  $\mathbb{R}^N_+$ ).

With the above ideas fixed, let us now re-state Problems I and II.

- **Problem I**: Given a per user power contraint of  $p = (p_1, \dots, p_K)$  where no user is oversized (i.e.,  $Np_k \le p_{tot}$  for every user k), find S and r that satisfy  $r \in C(S, p)$  and  $r_{tot} = (1/2) \log(1 + p_{tot})$ , the maximum sum-rate among all sequence and rate allocations.
- **Problem II**: Given a per user rate requirement of  $r=(r_1,\cdots,r_K)$  bits/chip where no user is oversized (i.e.,  $Nr_k \leq r_{tot}$ for every user k), find S and p that satisfy  $r \in C(S, p)$  and  $p_{tot} = \exp\{2r_{tot}\} - 1$ , the minimum value among all power and sequence allocations.

Section IV identifies the matrices  $A_k$  and the sequence matrix S that solve Problems I and II. Vertices will play an important role in the solution.

### III. CONVERSE TO WEYL'S RESULT

Proposition 1: (Converse to Weyl's result) Let A be an  $N \times N$  complex Hermitian matrix with  $\sigma(A) = (\lambda_1, \dots, \lambda_N)$ . Let  $(\hat{\lambda}_1, \cdots, \hat{\lambda}_N)$  be N real numbers satisfying the interlacing inequality (1). Then, there exists a  $c \in \mathbb{C}^N$  such that  $\sigma(A + cc^*) = 0$  $(\hat{\lambda}_1, \dots, \hat{\lambda}_N)$ . If, in addition, A is a real, there exists  $c \in \mathbb{R}^N$  with the same properties. 

This result is a pleasing dual to an existence result shown by Mirsky [15] for bordered matrices. The problem falls under the category of additive inverse eigenvalue problems [4], [5]. Proposition 1 however appears to be new. A proof can be found in Appendix I.

Proposition 1 is crucial to this paper. It guarantees the existence of a program that takes in the matrix A, a set of prescribed eigenvalues  $\hat{\lambda}$  that interlace the eigenvalues of A, and outputs  $c = c(A, \hat{\lambda})$  such that  $A + cc^*$  is the desired perturbation of A. Let us now focus on the nature of c in a special circumstance that will be of relevance to the sequence design problem.

Proposition 2: Let A be an  $N \times N$  complex Hermitian matrix with  $\sigma(A) = \lambda = (\lambda_1, \dots, \lambda_N)$ . Let A be unitarily diagonalized by the unitary matrix U, i.e.,  $A = U\Lambda U^*$ , where  $\Lambda = \text{diag}\{\lambda\}$ . Fix l satisfying  $1 \le l \le N$ . Let  $\hat{\lambda}$  be N real numbers such that  $\hat{\lambda}_j = \lambda_j$  for all j except when j = l. Furthermore, let  $\hat{\lambda}$  and  $\lambda$  satisfy the interlacing property (1). Then

- 1) with  $c = \left(\sqrt{\hat{\lambda}_l \lambda_l}\right) u_l$ , where  $u_l$  is the lth column of U, we have  $\sigma(A + cc^*) = \hat{\lambda}$ ; 2) A and  $(A + cc^*)$  are simultaneously diagonalized by U.

Proof: This is easily verified by looking at the assignment in the proof of Proposition 1. Instead of doing this, we give a direct proof. Let  $U = [u_1 \cdots u_N]$ , where  $u_i \in \mathbb{C}^N$ . Then  $Au_l = \lambda_l u_l$ , and  $u_i^* u_i = 0$  if  $i \neq l$ . Hence for every  $i \neq l$ ,

$$(A + cc^*) u_i = Au_i + c (c^*u_i) = Au_i + 0 = \lambda_i u_i,$$

and thus  $\lambda_i$ ,  $i \neq l$ , are N-1 eigenvalues of  $A+cc^*$ . Furthermore,

$$(A + cc^*) u_l = \lambda_l u_l + (\hat{\lambda}_l - \lambda_l) u_l (u_l^* u_l) = \hat{\lambda}_l u_l,$$

and thus  $\hat{\lambda}_l$  is an eigenvalue of  $A + cc^*$ . This proves the first statement.

The eigenvectors for the two Hermitian matrices are the same and therefore the same unitary matrix (of eigenvectors) diagonalizes them.

#### IV. SEQUENCE ASSIGNMENTS

Let the rate vector r, the sequence matrix S, and the power vector p be such  $r \in C(S, p)$ . Then

$$p_{tot} \ge \exp\left\{2r_{tot}\right\} - 1,\tag{4}$$

where  $p_{tot} = \sum_{k=1}^{K} p_k$ , and  $r_{tot} = \sum_{k=1}^{K} r_k$ . Indeed, this can be easily seen through the following sequence of inequalities. Let  $P = \text{diag}\{p\}$ , a  $K \times K$  diagonal matrix with diagonal entries given by the elements of p.

$$p_{tot} = \operatorname{trace} P$$

$$= \operatorname{trace} PS^{t}S \qquad (5)$$

$$= \operatorname{trace} SPS^{t} \qquad (6)$$

$$= \frac{1}{N} \cdot \operatorname{trace} (I_{N} + NSPS^{t}) - 1$$

$$= \frac{1}{N} \cdot \operatorname{trace} A_{K} - 1$$

$$\geq |A_{K}|^{1/N} - 1 \qquad (7)$$

$$\geq \exp\{2r_{tot}\} - 1, \qquad (8)$$

where (5) follows because the diagonal entries of  $S^tS$  are all 1; (6) follows because trace AB = trace BA when the products AB and BA are well-defined; (7) follows because arithmetic mean of the eigenvalues exceeds the geometric mean of the eigenvalues; and (8) follows because  $r \in C(S,p)$  which implies that  $r_{tot} \leq \frac{1}{2N} \log |A_K|$ .

For Problem I, given a set of power constraints, (4) indicates that the maximum achievable sum capacity is upperbounded by  $\frac{1}{2}\log\left(1+p_{tot}\right)$ , as observed by Viswanath and Anantharam in [1]. Analogously, for Problem II, given a set of rate constraints, (4) says that the minimum sum power is lowerbounded by  $\exp\left\{2r_{tot}\right\}-1$ . If these bounds are achieved, then equality in (7) and (8) indicate that all the eigenvalues of  $A_K$  are the same, and in particular,

$$A_K = (1 + p_{tot}) I_N = \exp\{2r_{tot}\}I_N,$$

a well-known property of GWBE sequences ([1, Section IV]).

## A. Algorithm for Problem II

The signal and interference matrix in the absence of any users is  $A_0$ . The matrix with all users taken into account is  $A_K$ . Let us now assign sequences in a *sequential* fashion, *i.e.*, one user after another.

Loosely speaking, we begin with  $A_0$  and its eigenvalues  $\lambda^{(0)} = (1, \dots, 1)$ . We then fill up dimensions with energies from users, one user after another, jumping to the next dimension when a dimension reaches  $\exp\{2r_{tot}\}$  upon addition of a user. Each time a user is added, say user k, we ensure that  $\lambda^{(k-1)}$  and  $\lambda^{(k)}$  satisfy the interlacing property. Indeed, this is guaranteed precisely because there are no oversized users. Moreover, they differ in at most two eigenvalues. Upon each addition, we then simply appeal to Proposition 1 to get  $c_k$  for user k. The sequence  $s_k$  is then  $s_k = c_k/||c_k||$  and power  $p_k = (c_k^t c_k)/N$ .

The following algorithm makes this notion precise. For simplicity, we assume that users are assigned in the increasing order of their indices. The order of the users is immaterial as will become apparent in the proof of correctness of the procedure. Our assignment will make the given set of rates the *vertex* of some capacity region.

#### Algorithm 3: Problem II

• Initialization: Set the following:

$$\begin{array}{lcl} \lambda_n^{(k)} & \leftarrow & 1, & \text{for } 0 \leq k \leq K, & 1 \leq n \leq N; \\ k & \leftarrow & 1; \\ n & \leftarrow & 1; \\ \lambda_{\max} & \leftarrow & \exp\left\{2r_{tot}\right\}; \\ A_0 & \leftarrow & I_N. \end{array}$$

- Step 1: (All users are done). If k > K, stop.
- Step 2 Case (a): (Only one eigenvalue changes). If

$$\lambda_n^{(k-1)} \cdot \exp\left\{2Nr_k\right\} < \lambda_{\max},\tag{9}$$

then, set

$$\lambda_n^{(k)} \leftarrow \lambda_n^{(k-1)} \cdot \exp\left\{2Nr_k\right\}. \tag{10}$$

Go to Step 3.

• Step 2 - Case (b): (Only one eigenvalue changes; dimension n is to be filled up). If

$$\lambda_n^{(k-1)} \cdot \exp\left\{2Nr_k\right\} = \lambda_{\max},\tag{11}$$

then, set

$$\lambda_n^{(j)} \leftarrow \lambda_{\max}, \quad \text{for } j = k, \cdots, K.$$
 (12)

Now set

$$n \leftarrow n + 1$$
.

Go to Step 3.

• Step 2 - Case (c): (Two eigenvalues change; dimension n is to be filled up). If we are in this step, then

$$\lambda_n^{(k-1)} \cdot \exp\left\{2Nr_k\right\} > \lambda_{\max},\tag{13}$$

Set the following:

$$\lambda_n^{(j)} \leftarrow \lambda_{\max}, \quad \text{for } j = k, \cdots, K,$$
 (14)

$$\lambda_{n+1}^{(k)} \leftarrow \frac{\lambda_{n+1}^{(k-1)} \cdot \lambda_n^{(k-1)} \cdot \exp\left\{2Nr_k\right\}}{\lambda_{\max}}.$$
(15)

Observe that  $\lambda_{n+1}^{(k-1)}=1$ . Now set

$$n \leftarrow n + 1$$

• Step 3: (Assign power and sequence for user k). Identify the vector

$$c_k = c\left(A_{k-1}, \lambda^{(k)}\right)$$

from Proposition 1. Then set

$$s_k \leftarrow c_k/||c_k||,$$
  
 $p_k \leftarrow (c_k^t c_k)/N.$ 

A sequence and power are now allocated to user k. Finally,

$$A_k \leftarrow A_{k-1} + c_k c_k^t, k \leftarrow k+1.$$

Go to Step 1. 
$$\Box$$

We now make some remarks on the computations and the numerical stability of the algorithm. The matrices  $A_k$  need not be explicitly computed. It is sufficient to identify and store the unitary matrices  $U_k = \begin{bmatrix} u_1^{(k)} & \cdots & u_N^{(k)} \end{bmatrix}$  that diagonalize  $A_k$ . Computation of  $c\left(A_{k-1},\lambda^{(k)}\right)$  utilizes only  $U_{k-1}$ ,  $\lambda^{(k-1)}$ , and the new  $\lambda^{(k)}$ , as seen in the proof of Proposition 1. For Step 2(c) at most two eigenvalues change in going from  $A_{k-1}$  to  $A_k$ . So the two matrices share N-2 eigenvectors and exactly two eigenvectors change. These are computed via a rotation in the (n,n+1)st plane as follows:

$$\begin{bmatrix} u_n^{(k)} \ u_{n+1}^{(k)} \end{bmatrix} = \begin{bmatrix} u_n^{(k-1)} \ u_{n+1}^{(k-1)} \end{bmatrix} \begin{bmatrix} \alpha & -\beta \\ \beta & \alpha \end{bmatrix},$$

where

$$\alpha = \sqrt{\frac{\left(\hat{\lambda}_{n} - \lambda_{n+1}\right)\left(\lambda_{n} - \hat{\lambda}_{n+1}\right)}{\left(\hat{\lambda}_{n} - \hat{\lambda}_{n+1}\right)\left(\lambda_{n} - \lambda_{n+1}\right)}},$$

$$\beta = \sqrt{1 - \alpha^{2}} = \sqrt{\frac{\left(\hat{\lambda}_{n+1} - \lambda_{n+1}\right)\left(\hat{\lambda}_{n} - \lambda_{n}\right)}{\left(\hat{\lambda}_{n} - \hat{\lambda}_{n+1}\right)\left(\lambda_{n} - \lambda_{n+1}\right)}},$$

with  $\lambda \stackrel{\Delta}{=} \lambda^{(k-1)}$  and  $\hat{\lambda} \stackrel{\Delta}{=} \lambda^{(k)}$  for simplicity. Moreover,  $p_k = \hat{\lambda}_{n+1} + \hat{\lambda}_n - (\lambda_{n+1} + \lambda_n)$ , and

$$c_k = y_n \ u_n^{(k-1)} + y_{n+1} \ u_{n+1}^{(k-1)}$$

where

$$y_n = \sqrt{\frac{\left|\lambda_n - \hat{\lambda}_n\right| \left|\lambda_n - \hat{\lambda}_{n+1}\right|}{\left|\lambda_n - \lambda_{n+1}\right|}},$$

$$y_{n+1} = \sqrt{\frac{\left|\lambda_{n+1} - \hat{\lambda}_n\right| \left|\lambda_{n+1} - \hat{\lambda}_{n+1}\right|}{\left|\lambda_n - \lambda_{n+1}\right|}}.$$

These facts can be verified by direct substitution. Furthermore several assignments in (14) and in the initialization step can be implicitly assumed. The upshot is that computation of sequences as well as the new eigenvectors requires only O(N) floating point operations. Steps 2(a) and 2(b) do not result in a change in eigenvectors. The sequence in these simpler cases is a scaled version of  $u_n^{(k-1)}$  and requires only N floating point operations for computation. Hence the algorithm's complexity is O(KN) floating point operations.

The rotation requires identification of  $\alpha$  and  $\beta$ ; these are between 0 and 1 and hence their computation is numerically stable. Furthermore, it is easy to verify from the interlacing inequality that

$$\max\left\{\left(\lambda_n - \hat{\lambda}_{n+1}\right), \left(\hat{\lambda}_{n+1} - \lambda_{n+1}\right)\right\} \le \lambda_n - \lambda_{n+1},$$

and therefore computations of  $y_n$  and  $y_{n+1}$  are also numerically stable. This verifies that the algorithm is numerically stable. Before we give the proof of correctness, we illustrate the algorithm with an example.

Example 4: Let N=2. Consider four users with rate requirements  $r_1, r_2, r_3$ , and  $r_4$ , such that  $r_k \leq r_{tot}/N$ , for  $k=1, \dots, 4$ , i.e., there are no oversized users.

Fig. 1 represents the eigenvalue space of the matrices  $A_k$ ,  $k=0,1,\cdots,4$ . Our interest is in the upper sector in the positive quadrant which represents  $\lambda_1 \geq \lambda_2 \geq 1$ . The algorithm results in a sequence of  $\left(\lambda_1^{(k)}, \lambda_2^{(k)}\right)$ , the eigenvalues of  $A_k$ .

The given rates should be supportable. This places a condition on the product of the eigenvalues

$$\lambda_1^{(k)} \lambda_2^{(k)} \ge \exp\left\{2N \sum_{j=1}^k r_j\right\}, \ k = 1, 2, 3, 4,$$

i.e., after user k is added, the eigenvalue pair should lie beyond the kth hyperbola. Observe that  $N+Np_{tot}=$  trace  $A_K=\lambda_1^{(K)}+\lambda_2^{(K)}$  is least, and therefore  $p_{tot}$  is least, if  $\lambda_1^{(K)}=\lambda_2^{(K)}=\lambda_{\max}=\exp\{2r_{tot}\}$  as is indicated by the tangent to the outermost hyperbola in the figure. An allocation that reaches this point is a GWBE allocation. The proof below shows that this is possible if none of the users are oversized.

As the users are added sequentially, the interlacing condition says that the eigenvalues should lie in infinite half-strips shown in Fig. 1. The condition for k=1 is degenerate; it is the half line represented by  $\lambda_2=0,\ \lambda_1\geq 1$ .

The first user is added with minimum power to meet the interlacing and rate conditions. The other users are added to minimize the number of non-unit eigenvalues (as is the first user). A new eigenvalue is enlarged only if the previous eigenvalue has reached  $\lambda_{\max}$ . The choices of eigenvalue pairs that follow this rule are indicated in the figure as solid circles. The steps executed for the four users are 2(a), 2(c), 2(a), 2(b), respectively, for the four users.

Note that the choice is not greedy in the powers that are allocated. Such a "power-hungry" allocation is not guaranteed to reach the minimum power point, whereas the above algorithm guarantees such an allocation.

*Proof:* We now give the proof of correctness of Algorithm 3. When a new user, say user k, is added, two conditions on the set of eigenvalues of the signal and interference matrix should be satisfied. One, the added user's rate should be supportable and hence,

$$\frac{1}{2N}\log|A_k| = \frac{1}{2N}\log\prod_{l=1}^{N}\lambda_l^{(k)} \ge \sum_{j=1}^{k}r_j;$$

this imposes a lower bound on the product of the eigenvalues. It is easy to verify that the algorithm satisfies the above inequality with equality (see (16)). Two, the interlacing eigenvalues condition should hold because the signal and interference matrix  $A_k$  is an as yet unknown rank-1 perturbation of the matrix  $A_{k-1}$ .

Interlacing eigenvalues: We first ensure that an execution of Step 2 results in a set of eigenvalues  $\lambda^{(k)} = \left(\lambda_1^{(k)}, \cdots, \lambda_N^{(k)}\right)$  that interlace the eigenvalues  $\lambda^{(k-1)} = \left(\lambda_1^{(k-1)}, \cdots, \lambda_N^{(k-1)}\right)$ . For a particular k, exactly one of the cases (a), (b), or (c) holds.

Consider case (a) or (b). Then exactly one eigenvalue changes between  $\lambda^{(k-1)}$  and  $\lambda^{(k)}$ . Indeed, either (9) or (11) holds, and

$$\lambda^{(k-1)} = \left(\underbrace{\lambda_{\max}, \cdots, \lambda_{\max}}_{n-1}, \lambda_n^{(k-1)}, \underbrace{1, \cdots, 1}_{N-n}\right),$$

$$\lambda^{(k)} = \left(\underbrace{\lambda_{\max}, \cdots, \lambda_{\max}}_{n-1}, \lambda_n^{(k)}, \underbrace{1, \cdots, 1}_{N-n}\right).$$

To check that the interlacing inequality holds, it is sufficient to check that

$$\lambda_{\max} \overset{(\alpha)}{\geq} \lambda_n^{(k)} \overset{(\beta)}{>} \lambda_n^{(k-1)} \overset{(\gamma)}{\geq} 1.$$

But this is easily verified. Inequality  $(\alpha)$  holds because of (9) and (10) in case (a) and because of (11) and (12) in case (b). Next, inequality  $(\beta)$  holds because  $r_k > 0$  in both (10) and (12). Furthermore, inequality  $(\gamma)$  is preserved at all steps.

To check that the interlacing inequality holds under Step 2 case (c), note that if this step is executed, (13) holds, and the eigenvalues are given by

$$\lambda^{(k-1)} = \left(\underbrace{\lambda_{\max}, \cdots, \lambda_{\max}}_{n-1}, \lambda_n^{(k-1)}, 1, \underbrace{1, \cdots, 1}_{N-n-1}\right),$$

$$\lambda^{(k)} = \left(\underbrace{\lambda_{\max}, \cdots, \lambda_{\max}}_{n-1}, \lambda_{\max}, \lambda_{n+1}^{(k)}, \underbrace{1, \cdots, 1}_{N-n-1}\right).$$

To see that the interlacing inequality holds, it is sufficient to check that

$$\lambda_{\max} \stackrel{(\alpha)}{=} \lambda_n^{(k)} \stackrel{(\beta)}{\geq} \lambda_n^{(k-1)} \stackrel{(\gamma)}{\geq} \lambda_{n+1}^{(k)} \stackrel{(\delta)}{>} 1.$$

This is can be verified as follows.  $(\alpha)$  holds because of (14). At all times, our choice of eigenvalues is such that  $1 \le \lambda_n^{(k-1)} \le \lambda_{\max}$  and therefore  $(\beta)$  holds. To check  $(\gamma)$  observe that (15) gives

$$\lambda_{n+1}^{(k)} = \frac{\lambda_{n+1}^{(k-1)} \cdot \lambda_n^{(k-1)} \cdot \exp\{2Nr_k\}}{\lambda_{\max}} \le \lambda_n^{(k-1)}$$

because  $\lambda_{n+1}^{(k-1)} = 1$  and  $\exp\{2Nr_k\} \le \lambda_{\max} = \exp\{2r_{tot}\}$  for there are no oversized users. Lastly  $(\delta)$  follows because of (13) and (15).

*Termination*: After an execution of Step 3, k is the index of the next user to be allocated, and it is easy to check via induction and (10), or (11) and (12), or (14) and (15), that

$$|A_{k-1}| = \exp\left\{2N\sum_{j=1}^{k-1} r_j\right\}. \tag{16}$$

This ensures that every one of the K users can be added without the eigenvalues exceeding  $\lambda_{max}$  at any stage and in any of the dimensions because

$$|A_{k-1}| = \exp\left\{2N\sum_{j=1}^{k-1} r_j\right\} \le \exp\left\{2Nr_{tot}\right\}$$

for every k. In particular, if at the end of some Step 3 execution we have n=N for some  $k \leq K$ , then users  $k, k+1, \cdots, K$  can be added without the need for an additional dimension, i.e., Step 2(c) will no longer be executed. Moreover, because of (16), we have  $|A_K| = \exp\{2Nr_{tot}\}$  and therefore the last Step 2 to be executed will be Step 2(c) thereby correctly filling up the last eigenvalue at termination.

Validity of the sequential procedure: The previous steps indicate that the interlacing inequality holds at every step of the algorithm, and therefore a sequence matrix S and a power allocation p will be put out by the algorithm. The given r, by design, is a vertex of the capacity region C(S,p). As C(S,p) is a polymatroid, it contains all its vertices, and therefore  $r \in C(S,p)$ . (Alternatively, a successive cancellation receiver indicates that r is achievable). This implies that the allocations S and p support the given rate vector r, and the proof of correctness is complete.

#### B. Algorithm for Problem I

We now look at the dual problem of sequence and rate allocation for sum capacity maximization given users' power constraints  $p = (p_1, \dots, p_K)$ . Guess [2] explores the duality between these two problems.

### Algorithm 5: Problem I

• Initialization: Set the following:

- Step 1: (All users are done). If k > K, stop.
- Step 2 Case (a): (Only one eigenvalue changes). If

$$\lambda_n^{(k-1)} + Np_k < \lambda_{\max},\tag{17}$$

then, set

$$\lambda_n^{(k)} \leftarrow \lambda_n^{(k-1)} + Np_k. \tag{18}$$

Go to Step 3.

• Step 2 - Case (b): (Only one eigenvalue changes; dimension n is to be filled up). If

$$\lambda_n^{(k-1)} + Np_k = \lambda_{\max},\tag{19}$$

then, set

$$\lambda_n^{(j)} \leftarrow \lambda_{\max} \text{ for } j = k, \cdots, K.$$
 (20)

Now set

$$n \leftarrow n + 1$$
.

Go to Step 3.

• Step 2 - Case (c): (Two eigenvalues change; dimension n is to be filled up). If we are in this step, then

$$\lambda_n^{(k-1)} + Np_k > \lambda_{\text{max}}. (21)$$

Set the following:

$$\lambda_n^{(j)} \leftarrow \lambda_{\max} \text{ for } j = k, \cdots, K;$$
 (22)

$$\lambda_{n}^{(j)} \leftarrow \lambda_{\max} \text{ for } j = k, \cdots, K;$$

$$\lambda_{n+1}^{(k)} \leftarrow \lambda_{n+1}^{(k-1)} + \lambda_{n}^{(k-1)} + Np_k - \lambda_{\max}.$$
(22)

Observe that  $\lambda_{n+1}^{(k-1)}=1$ . Set

$$n \leftarrow n + 1$$
.

• Step 3: (Assign rate and sequence for user k). Identify the vector

$$c_k = c\left(A_{k-1}, \lambda^{(k)}\right)$$

from Proposition 1. Then set

$$A_k \leftarrow A_{k-1} + c_k c_k^t,$$

$$s_k \leftarrow c_k / \|c_k\|,$$

$$r_k \leftarrow \frac{1}{2N} \log |A_k| - \frac{1}{2N} \log |A_{k-1}|.$$

A sequence and rate are now allocated to user k. Finally,

$$k \leftarrow k + 1$$
.

Go to Step 1. 
$$\Box$$

We now illustrate the algorithm with an example.

Example 6: Let N=2. Consider a numerical example with four users having power limits  $p=(p_1,p_2,p_3,p_4)=(2,2,3,1)$ .

It is easy to check that no user is oversized. Refer to Fig. 2, the eigenvalue space of the matrices  $A_k$  for  $k = 0, 1, \dots, 4$ . Because trace  $A_k = N + N \sum_{j=1}^k p_j$ , the eigenvalues are constrained to lie within the confines of the four triangular regions bounded by

$$\lambda_1 \ge 1$$
,  $\lambda_2 \ge 1$ ,  $\lambda_1 + \lambda_2 \le N + N \sum_{j=1}^k p_j$ 

as given in the figure. The last bounding line  $\lambda_1 + \lambda_2 = N + Np_{tot}$  is a tangent to the region

$$\lambda_1 \lambda_2 \ge (1 + p_{tot})^N$$

at the point  $\lambda_1 = \lambda_2 = 1 + p_{tot}$ . If this point can be reached, then we have a GWBE allocation and the sum rate is

$$\frac{1}{2}\log(1+p_{tot}) = \frac{1}{2}\log 9,$$

the maximum among all allocations meeting the power constraints given by the aforementioned tangent line.

The algorithm ensures that at each step the eigenvalues interlace and the power constraints met. As before the number of non-unit eigenvalues are kept at a minimum at each step. The sequence of eigenvalue pairs are

$$\begin{array}{lll} (\lambda_1^{(1)},\lambda_2^{(1)}) & = & (5,1), \\ (\lambda_1^{(2)},\lambda_2^{(2)}) & = & (9,1), \\ (\lambda_1^{(3)},\lambda_2^{(3)}) & = & (9,7), \text{ and} \\ (\lambda_1^{(4)},\lambda_2^{(4)}) & = & (9,9) \end{array}$$

The interesting aspect of this numerical example is that users 1 and 2 share a sequence, say  $s_1$ , and users 3 and 4 share another sequence, say  $s_3$ , and moreover,  $s_1^t s_3 = 0$ . This completely separates users 1 and 2 from users 3 and 4, attains the minimum  $\frac{1}{2} \log (1 + p_{tot})$  bound, and yet gives the benefits of spreading. Any pair of orthogonal vectors suffice. We explore this possibility further in Section VI and show how to obtain orthogonal sequences under mild rate splitting.

*Proof:* We now give the proof of correctness of Algorithm 5. The proof parallels the proof of correctness of Algorithm 3. We provide it here only to highlight the steps that rely on the absence of oversized users.

When a new user, say user k, is added, the power constraint

trace 
$$A_k = \sum_{l=1}^N \lambda_l^{(k)} \le \sum_{j=1}^k p_k$$
,

and the interlacing eigenvalue condition are met. The power constraint is in fact met with equality. The rate assigned to this user is given by (3). We now verify these statements.

Interlacing eigenvalues: An execution of Step 2 results in  $\lambda^{(k)}$  that interlace  $\lambda^{(k-1)}$ . Indeed, for a particular k, exactly one of cases (a), (b), or (c) is executed. Consider case (a) or (b). Then exactly one eigenvalue changes between  $\lambda^{(k-1)}$  and  $\lambda^{(k)}$ , and

$$\lambda^{(k-1)} = \left(\underbrace{\lambda_{\max}, \cdots, \lambda_{\max}}_{n-1}, \lambda_n^{(k-1)}, \underbrace{1, \cdots, 1}_{N-n}\right),$$

$$\lambda^{(k)} = \left(\underbrace{\lambda_{\max}, \cdots, \lambda_{\max}}_{n-1}, \lambda_n^{(k)}, \underbrace{1, \cdots, 1}_{N-n}\right).$$

To check that the interlacing inequality holds, it is sufficient to check that

$$\lambda_{\max} \overset{(\alpha)}{\geq} \lambda_n^{(k)} \overset{(\beta)}{>} \lambda_n^{(k-1)} \overset{(\gamma)}{\geq} 1.$$

But this is easily verified. Inequality  $(\alpha)$  holds because of (17) and (18), or (19) and (20), as the case may be. Inequality  $(\beta)$  holds because in (18) or in (19) we have  $p_k > 0$ . The algorithm is such that  $(\gamma)$  is preserved at all steps.

We now verify the interlacing inequality holds under Step 2(c). As a consequence of (21), we have

$$\lambda^{(k-1)} = \left(\underbrace{\lambda_{\max}, \cdots, \lambda_{\max}}_{n-1}, \lambda_n^{(k-1)}, 1, \underbrace{1, \cdots, 1}_{N-n-1}\right),$$

$$\lambda^{(k)} = \left(\underbrace{\lambda_{\max}, \cdots, \lambda_{\max}}_{n-1}, \lambda_{\max}, \lambda_{n+1}^{(k)}, \underbrace{1, \cdots, 1}_{N-n-1}\right).$$

To see that the interlacing inequality holds, it is sufficient to check that

$$\lambda_{\max} \stackrel{(\alpha)}{=} \lambda_n^{(k)} \stackrel{\beta}{\geq} \lambda_n^{(k-1)} \stackrel{(\gamma)}{\geq} \lambda_{n+1}^{(k)} \stackrel{(\delta)}{>} 1.$$

This can be verified as follows. Equality  $(\alpha)$  holds because of the assignment in (22) for j=k. At all times, our choice of eigenvalues is such that  $1 \le \lambda_n^{(k-1)} \le \lambda_{\max}$  and therefore  $(\beta)$  holds. To check  $(\gamma)$  observe that (23) gives

$$\lambda_{n+1}^{(k)} = \lambda_{n+1}^{(k-1)} + \lambda_n^{(k-1)} + Np_k - \lambda_{\max} \le \lambda_n^{(k-1)}$$

because  $\lambda_{n+1}^{(k-1)}=1$  and  $1+Np_k\leq \lambda_{\max}=1+p_{tot}$  for there are no oversized users. Lastly  $(\delta)$  follows because of (21), the assignment in (23), and the equality  $\lambda_{n+1}^{(k-1)}=1$ .

*Termination*: After an execution of Step 3, user k is the next user to be allocated. As before, it is easy to check via induction, and (18), or (19) and (20), or (22) and (23), as the case may be, that

trace 
$$A_{k-1} = N + N \sum_{j=1}^{k-1} p_j$$
,

i.e., the power constraints of the just added user is respected. This also ensures that every one of the K users can be added because

trace 
$$A_{k-1} = N + N \sum_{j=1}^{k-1} p_j \le N + N p_{tot} = N \lambda_{\max}$$

for every k. In particular, if at the end of some Step 3 we have n = N for some  $k \le K$ , then users  $k, k + 1, \dots, K$  can be added without the need for an additional dimension, i.e., Step 2(c) will no longer be executed.

Validity of the sequential procedure: The previous steps indicate that the interlacing inequality holds at every step of the algorithm, and therefore a sequence matrix S and a rate allocation r will be put out by the algorithm. The obtained r, by design, is a vertex of the capacity region C(S,p). C(S,p) is a polymatroid; it contains all its vertices, and therefore  $r \in C(S,p)$ .

#### V. NUMBER OF SEQUENCES

In this section, we study the number of sequences needed to design systems that solve Problems I and II.

#### A. 2N-1 sequences suffice

We first show that 2N-1 sequences suffice regardless of the number of users K. If K>2N-1, then there are fewer sequences assigned than users and therefore some users share the same sequence; they will completely overlap with one other. But a successive cancelation receiver enables us to receive data from all such users if powers or rates are suitably assigned.

Theorem 7: There is an optimal sequence allocation with at most 2N-1 distinct sequences for both Problems I and II. Furthermore, when there are L oversized users, there is an optimal sequence allocation with at most 2N-L-1 distinct sequences.

*Proof:* Suppose that there are no oversized users. We will show that Algorithms 3 and 5 result in a design with the desired property. In both algorithms, a new sequence is put out if either

- Step 2(c) is executed, i.e., a "break-out" in a new dimension n+1 occurs, or
- if one of Step 2(a) or Step 2(b) is executed for the first time in a new dimension n, i.e., n was updated when the previous user was added.

The former of these conditions happens at most N-1 times because it results in a break-out in a new dimension and there are at most N dimensions. The latter happens at most N times because in any dimension there is only one first step. All subsequent steps in the same dimension do not result in a new sequence because of Proposition 2. This ensures that the algorithm puts out a sequence matrix with at most 2N-1 sequences.

Suppose now that there are L oversized users. These users should be assigned L sequences, orthogonal to each other, and orthogonal to the space spanned by the sequences assigned to the remaining users. There are N-L dimensions available for the remaining K-L nonoversized users. An appeal to the case with no oversized users results in at most 2(N-L)-1 for these K-L users. In total, therefore, we need at most L+2(N-L)-1=2N-L-1 sequences.

B. 2N-1 sequences are necessary for one-dimensional signaling

We now provide a converse to the results in the previous subsection.

Theorem 8: For Problem I, given N, there exist K > N and power constraints  $p = (p_1, \dots, p_K)$  with no user oversized, such that any rate and sequence allocation with 2N - 2 or fewer sequences results in a sum capacity

$$r_{tot} < \frac{1}{2}\log\left(1 + p_{tot}\right).$$

Similarly, for Problem II, given N, there exist K > N and rate requirements  $r = (r_1, \dots, r_k)$  with no user oversized, such that any power and sequence allocation with 2N - 2 or fewer sequences results in a minimum power

$$p_{tot} > \exp\left\{2r_{tot}\right\} - 1.$$

*Proof:* Consider Problem I. Consider K = 2N - 1 users with a symmetric power constraint

$$p = \left(\underbrace{\frac{p_{tot}}{K}, \frac{p_{tot}}{K}, \cdots, \frac{p_{tot}}{K}}_{2N-1}\right).$$

Let  $[s_1, s_2, \dots, s_K]$  be any sequence assignment with 2N-2 or fewer sequences, and  $(r_1, \dots, r_K)$  a rate assignment. At least two users share a sequence. Without loss of generality we may assume that these are users 1 and 2, *i.e.*,  $s_1 = s_2$ . The basic idea is to show that users 1 and 2 put together make an oversized *compound* user with power constraint  $p_1 + p_2$  in a 2N-2 user system. Indeed,

$$p_1 + p_2 = \frac{2p_{tot}}{K} = \frac{2p_{tot}}{2N - 1} > \frac{p_{tot}}{N},$$

which makes the compound user oversized. The sum capacity of the system is therefore strictly smaller than  $\frac{1}{2}\log(1+p_{tot})$ . We now formalize this idea.

 $A_2$  is given by

$$A_2 = I_N + Np_1 s_1 s_1^t + Np_2 s_2 s_2^t = I_N + \frac{2Np_{tot}}{K} s_1 s_1^t,$$

with eigenvalues

$$\lambda^{(2)} = \left(1 + \frac{2Np_{tot}}{K}, \underbrace{1, \cdots, 1}_{N-1}\right),\,$$

and therefore

$$\lambda_1^{(K)} \ge \lambda_1^{(2)} = 1 + \frac{2Np_{tot}}{K} = 1 + \frac{2Np_{tot}}{2N - 1} > 1 + p_{tot},$$

where the first inequality is a consequence of Weyl's result applied to  $A_K = A_2 + \sum_{k=3}^K Np_k s_k s_k^t$ . Consequently, we have the following sequence of inequalities:

$$\frac{1}{2}\log\left(1 + p_{tot}\right) = \frac{1}{2}\log\left(\frac{\operatorname{trace}\ A_K}{N}\right) \tag{24}$$

$$> \frac{1}{2N}\log|A_K| \tag{25}$$

$$\geq r_{tot},$$
 (26)

where (24) holds because the power constraint implies trace  $A_K = N + Np_{tot}$ , (25) holds because the arithmetic mean of the eigenvalues is *strictly* larger than their geometric mean, a consequence of the fact that one of the eigenvalues  $\lambda_1^{(K)}$  is strictly larger than the arithmetic mean of the eigenvalues, and (26) is a necessary condition to support the rate vector r. This proves the first part of the theorem.

The proof for Problem II is very similar. Once again, we look at K=2N-1 users with symmetric rate requirements  $r=\left(\frac{r_{tot}}{K},\cdots,\frac{r_{tot}}{K}\right)$ , and make users 1 and 2 share a sequence. Then users 1 and 2 put together form an oversized compound user because

$$r_1 + r_2 = \frac{2r_{tot}}{K} = \frac{2r_{tot}}{2N - 1} > \frac{r_{tot}}{N}.$$

Now observe that

$$\sigma(A_2) = \left(\lambda_1^{(2)}, \underbrace{1, \cdots, 1}_{N-1}\right)$$

and therefore to support the rate of the two users we need

$$\frac{1}{2N}\log \lambda_1^{(2)} = \frac{1}{2N}\log|A_2| \ge r_1 + r_2 > \frac{r_{tot}}{N}$$

which leads to

$$\lambda_1^{(K)} \ge \lambda_1^{(2)} > \exp\{2r_{tot}\}.$$

The same sequence of inequalities, (24), (25), and (26), once again hold. The strict inequality in (25) holds because the largest eigenvalue is now strictly bigger than the geometric mean of the eigenvalues. Thus  $p_{tot} > \exp\{2r_{tot}\} - 1$ , which completes the proof of the second part of the theorem.

# VI. Two Dimensional signaling and sufficiency of N orthogonal sequences

Our goal in this section is to show how to achieve sum capacity (respectively minimum sum power) by using at most N orthogonal sequences. Thus far, each user has been confined to signal along a single dimension. In some cases, as in Example 6, it is possible to get an optimal assignment with N orthogonal sequences. In general we need 2N-1 sequences to achieve sum capacity or minimum power. However, by letting a few users' signals span two dimensions instead of one, it is possible to achieve optimality with N orthogonal sequences. We start with the following definition.

Definition 9: A vector  $x = (x_1, x_2, \dots, x_K)$  has a symmetric sum partition of size N, if there is a partition of the users  $\{1, 2, \dots, K\}$  into N subsets  $S_1, S_2, \dots, S_N$ , such that

$$\sum_{k \in S_n} x_k = \frac{1}{N} \sum_{k=1}^K x_k = \frac{x_{tot}}{N}, \tag{27}$$

for  $n=1,2,\cdots,N$ . The subsets  $S_1,S_2,\cdots,S_N$  will be referred to as the symmetric sum partition.

Remark: If x has a symmetric sum partition of size N, no user is oversized. This is because, any user k' belongs to  $S_n$ , for some n, and (27) implies  $x_{k'} \leq \sum_{k \in S_n} x_k = x_{tot}/N$ . Note that the power constraint vector p of Example 6 has a symmetric sum partition of size 2.

Proposition 10: If the rate vector r has a symmetric sum partition of size N, then N orthogonal sequences are sufficient to attain the minimum sum power  $p_{tot} = \exp\{2r_{tot}\} - 1$ . Analogously, if the power constraint vector p has a symmetric sum partition of size N, then N orthogonal sequences are sufficient to attain the sum capacity  $r_{tot} = \frac{1}{2} \log(1 + p_{tot})$ .

*Proof:* We will prove the proposition for Problem II. A similar argument holds for Problem I. Let  $S_1, S_2, \dots, S_N$  be the symmetric sum partition of the users. An execution of the Algorithm 3 that assigns sequences and powers to all users in a subset  $S_n$  before assigning to users in another subset will result in an orthogonal allocation. This is because

$$\exp\left\{2N\sum_{k\in S_n}r_k\right\} = \exp\left\{2r_{tot}\right\} = \lambda_{\max},$$

which implies that when all users in  $S_n$  are assigned, exactly one dimension is completely filled to  $\lambda_{max}$ . Users in one subset will therefore be assigned the same sequence, and different subsets are assigned orthogonal sequences. We will however prove the proposition directly.

Let  $S_1, S_2, \dots, S_N$  be the symmetric sum partition of the users. Assign sequences and powers as follows: if  $k \in S_n$ , then

$$s_k = e_n, (28)$$

$$p_k = \frac{\Im_k(n)}{N} \left[ \exp\{2Nr_k\} - 1 \right],$$
 (29)

where

$$\mathfrak{I}_{k}\left(n\right) \stackrel{\Delta}{=} \exp \left\{2N \sum_{j:j \in S_{n}, j < k} r_{j}\right\} \tag{30}$$

is the interference suffered by user k due to presence of other users in the same subset  $S_n$  with a smaller index.

It is easy to see that

$$r_k = \frac{1}{2N} \log \left( 1 + \frac{Np_k}{\Im_k(n)} \right) \tag{31}$$

is achievable via successive interference cancelation, where the highest index user in this subset is decoded first. Users in other subsets do not cause interference to users in this subset. Observe that the total power assigned to users in any subset  $S_n$  is given by

$$\sum_{k \in S_n} p_k = \frac{1}{N} \left( \exp\left\{ 2r_{tot} \right\} - 1 \right), \tag{32}$$

where (32) follows by substitution of (29) and (30) in the left side of (32) and by observing that the resulting sum over  $S_n$  has only two terms that survive.

From (32) we see that total power allocated to all users is

$$p_{tot} = \sum_{n=1}^{N} \sum_{k \in S_n} p_k = \exp\{2r_{tot}\} - 1,$$

thus showing that N orthogonal sequences are optimal.

Having derived a sufficient condition for optimality of N orthogonal sequences, let us see how to manufacture this condition from a given set of power constraints or rate requirements.

Proposition 11: Every strictly positive vector x representing power constraints or rate requirements for K non-oversized users can be cast into a vector x' for K' virtual users, where  $K \leq K' \leq K + N - 1$ , and x' is such that it has a symmetric sum partition of size N. Moreover x' is obtained by splitting K' - K users into exactly two virtual users each.  $\square$ 

Proof: Consider the cumulative requirement

$$X_k = \sum_{i \le k} x_i.$$

 $X_k$  is a strictly increasing function of k and  $X_K = x_{tot}$ . Hence there exist N-1 distinct users with indices  $k_j$ ,  $j=1,2,\cdots,N-1$ , such that

$$X_{k_{j}-1} = \sum_{i=1}^{k_{j}-1} x_{i} < \frac{jx_{tot}}{N},$$

$$X_{k_{j}} = \sum_{i=1}^{k_{j}} x_{i} \ge \frac{jx_{tot}}{N}.$$
(33)

If strict inequality holds in (33), split user  $k_j$ 's rate as

$$x_{k_j} = \underbrace{\left(\frac{jx_{tot}}{N} - X_{k_j-1}\right)}_{x'_{k_j}} + \underbrace{\left(X_{k_j} - \frac{jx_{tot}}{N}\right)}_{x''_{k_j}}.$$
(34)

If equality holds in (33) leave the user as is. For users that will be split, obtain x' from x by replacing  $x_{k_j}$ , the requirement for user  $k_j$ , by requirements  $x'_{k_j}$  and  $x''_{k_j}$  for two virtual users, where  $x'_{k_j}$  and  $x''_{k_j}$  are as in (34).

It follows that x' is a vector of size K', where  $K \leq K' \leq K + N - 1$ , and x' has a symmetric sum partition of size N.

Users whose rates or power constraints are split are assigned two orthogonal sequences and will span a two dimensional subspace. The design subsequently results in N orthogonal sequences.

It is immediate that even for the case with oversized users, the oversized users will be split into at most N virtual users. Non-oversized users will be split into at most two virtual users. The resulting vector r' has at most K+N-1 elements and has a symmetric sum partition of size N. Thus N orthogonal sequences are sufficient if rate or power splitting and multi-dimensional signaling is allowed for some users. We make this precise in the following Proposition. The proof is identical to the proof of Proposition 11 and therefore omitted.

Proposition 12: Every strictly positive vector x representing power constraints or rate requirements for K users, regardless of the presence of oversized users, can be cast into a vector x' of size K', where x' has a symmetric sum partition of size N, and  $K \le K' \le K + N - 1$ .

### VII. CONCLUDING REMARKS

The work in this paper was primarily motivated by a desire to reduce the amount of signaling necessary to communicate the sequences to the geographically separated users. This signaling usually eats up precious bandwidth in the downlink. This is especially a problem when the channel changes, or when users enter or leave the system. These significant events trigger a communication of a new set of sequences and rates/powers in the downlink.

We showed that 2N-1 sequences are sufficient to attain maximum sum capacity or minimum sum power. This results in some savings on the downlink when the number of users are large. The base station may broadcast the 2N-1 sequences as common information and indicate to each user only the index from the set and the rate or power allocated. If there are oversized users, fewer sequences are needed. The allocations are optimal and the algorithms have a simple geometric interpretation. The algorithms take only O(KN) steps, have the same complexity as that of [8, Algorithm 4], and the computations are numerically stable. We also argued that there exist rate or power constraints that necessitate 2N-1 sequences.

We then saw that with a small penalty in the spreading factor for a few users, N orthogonal sequences are sufficient. The users and the base station can then agree on a fixed orthogonal set of N sequences. The base station only needs to signal the powers/rates and the indices corresponding to the sequences allocated to a user.

The proposed algorithms identify a tight frame that meets the supplied constraints. They may therefore find application in other areas where tight frames arise.

The assumed multiple access channel model, however, has severe limitations. The uplink wireless channel typically suffers from multipath effects, fading, and asynchronism. Moreover, the users are not active all the time. Yet, if the users can all be synchronized via, for example, a Global Positioning System (GPS) receiver, our results give some interesting design insights. For frequency-flat slow fading channels where all the users with a tight delay constraint have to be served simultaneously, multi-dimensional signaling (more commonly referred to multi-code where the code is a spreading sequence) can allow communication at sum capacity or minimum power. Orthogonal sequences are sufficient and the signaling in the downlink is significantly reduced. A successively canceling decoder is necessary, but the complexity of this receiver is reduced to a great extent because optimal decoding for the K users decouples into decoding for N separate non-interfering groups.

Fairness of the allocation has not been considered in this paper. However, with the rate splitting approach that separates virtual users into groups, the decoding order of virtual users within a group can be cycled to get a fairer allocation. The first user to be decoded in a group treats all others in the group as interference and suffers the most. Cycling ensures that this and other such disadvantageous positions are shared in time by all users.

# APPENDIX I PROOF OF PROPOSITION 1

*Proof:* Observe that the complex Hermitian matrix A can be unitarily diagonalized as follows

$$A = U\Lambda U^*$$
.

where  $U = [u_1 \ u_2 \ \cdots \ u_N]$  is the matrix of eigenvectors of A and  $\Lambda = \text{diag}\{\lambda_1, \cdots, \lambda_N\}$ , a diagonal matrix with real entries. Assuming that the Proposition holds for real and diagonal matrices, we can find a real vector y such that  $\sigma(\Lambda + yy^*) = (\hat{\lambda}_1, \cdots, \hat{\lambda}_N)$ . Clearly,

$$\sigma\left(U(\Lambda + yy^*)U^*\right) = \sigma\left(\Lambda + yy^*\right).$$

and therefore the desired rank-1 perturbation of A with the prescribed eigenvalues is  $A + (Uy)(Uy)^*$ , i.e., c = Uy. If A is a real and symmetric matrix, U may be taken to be a real orthogonal matrix, and therefore c is a real vector.

It is therefore sufficient to prove the Proposition for diagonal matrices with real entries. Let A now be such a matrix. Let  $y \in \mathbb{R}^N$  be a nonzero vector. Let  $y_N \neq 0$ . The characteristic polynomial  $p_{\hat{A}}(t)$  of  $\hat{A} = A + yy^T$  is computed as

$$\frac{(c)}{\underbrace{\left(\frac{t-\lambda_{N}}{y_{N}}v^{T}(tI_{N-1}-\Lambda_{N-1})^{-1}}\right)}^{I_{N-1}} 1 \\
 \vdots \\
 \frac{tI_{N-1}-\Lambda_{N-1}}{\underbrace{\left(\frac{t-\lambda_{N}}{y_{N}}v^{T}\right)}^{T}} \frac{t-\lambda_{N}-y_{N}^{2}}{t-\lambda_{N}-y_{N}^{2}} \\
 \vdots \\
 \frac{I_{N-1}}{\underbrace{\left(\frac{t-\lambda_{N}}{y_{N}}v^{T}\right)}^{T}} \frac{t-\lambda_{N}-y_{N}^{2}}{t-\lambda_{N-1}} \\
 \vdots \\
 \frac{I_{N-1}}{\underbrace{\left(\frac{t-\lambda_{N}}{y_{N}}v^{T}\right)}^{T}} \frac{t-\lambda_{N}-y_{N}^{2}}{t-\lambda_{N-1}} \\
 \vdots \\
 \frac{tI_{N-1}-\Lambda_{N-1}}{\underbrace{\left(\frac{t-\lambda_{N}}{y_{N}}\right)}^{T}} \frac{1}{t-\lambda_{N}} \\
 \vdots \\
 \frac{tI_{N-1}-\Lambda_{N-1}}{t-\lambda_{N}-1} \frac{0}{t-\lambda_{N}} \\
 \vdots \\
 \frac{tI_{N-1}-\lambda_{N-1}}{t-\lambda_{N}-1} \frac{1}{t-\lambda_{N}} \\
 \vdots \\
 \frac{tI_{N-1}-\lambda_{N-1}}{t-\lambda_{N}-1} \frac{1}{t-\lambda_{N}-1} \frac{1}{t-\lambda_{N}-1} \\
 \vdots \\
 \frac{tI_{N-1}-\lambda_{N-1}}{t-\lambda_{N}-1} \frac{1}{t-\lambda_{N}-1} \frac{1}{t-\lambda_{N}-1} \\
 \vdots \\
 \frac{tI_{N-1}-\lambda_{N-1}}{t-\lambda_{N}-1} \frac{1}{t-\lambda_{N}-1} \frac{1}{t-$$

where (b) follows by setting the columns of the matrix in (a) to be  $h_i, i=1,\dots,N$ , then by replacing column  $h_i$  by  $h_i-h_Ny_i/y_N$ , for  $i=1,\dots,N-1$ , and then by setting  $v=(y_1,\dots,y_{N-1})^T\in\mathbb{R}^{N-1}$  and  $\Lambda_{N-1}=\operatorname{diag}(\lambda_1,\dots,\lambda_{N-1});$  (c) follows by multiplying the matrix in (b) on either side by triangular matrices that have determinant 1; (d) follows by multiplying out the matrices in (c). Observe that the conclusion remains unchanged if  $y_N=0$ , but instead some  $y_k\neq 0$  for some other k; such a k exists because y is nonzero.

Define the Nth degree polynomials

$$f(t) \stackrel{\triangle}{=} \prod_{i=1}^{N} (t - \hat{\lambda_i}),$$
  
$$g(t) \stackrel{\triangle}{=} \prod_{i=1}^{N} (t - \lambda_i).$$

Suppose now that the eigenvalues in  $\lambda$  are all distinct. By the Euclidean algorithm, we have

$$f(t) = g(t) + r(t),$$

where r(t) is a polynomial of degree at most N-1. We also have  $f(\lambda_i) = r(\lambda_i)$  for  $i = 1, 2, \dots, N$  because  $g(\lambda_i) = 0$ . The polynomial r(t) being known at N different points can be written explicitly by using the Lagrange interpolation formula

$$r(t) = \sum_{i=1}^{N} f(\lambda_i) \frac{g(t)}{g'(\lambda_i)(t - \lambda_i)}.$$

Thus,

$$\frac{f(t)}{g(t)} = 1 + \frac{r(t)}{g(t)} = 1 + \sum_{i=1}^{N} \frac{f(\lambda_i)}{g'(\lambda_i)} \frac{1}{t - \lambda_i}.$$
 (36)

Comparing (35) and (36), it is clear that we are done if we can set  $y_i^2 \stackrel{\triangle}{=} -f(\lambda_i)/g'(\lambda_i)$ , for  $i=1,\dots,N$ . To do this, however, we must show that  $f(\lambda_i)/g'(\lambda_i) \leq 0$ . Interlacing property ensures that this is indeed the case. Indeed, we can easily see that

$$f(\lambda_i) = (-1)^{N-i+1} \prod_{j=1}^{N} |\lambda_i - \hat{\lambda}_j|$$
  
$$g'(\lambda_i) = (-1)^{N-i} \prod_{j=1, j \neq i}^{N} |\lambda_i - \lambda_j|,$$

and therefore  $f(\lambda_i)$  and  $g'(\lambda_i)$  are always of opposite signs. Thus we may take

$$y_i = \sqrt{\frac{\prod_{j=1}^N |\lambda_i - \hat{\lambda}_j|}{\prod_{j=1, j \neq i}^N |\lambda_i - \lambda_j|}}.$$
(37)

Next suppose that  $\lambda$  has multiplicities. Observe that  $\lambda$  and  $\hat{\lambda}$  interlace. Consequently, if the spectrum  $\lambda$  has multiplicity k for a particular value  $\theta$ , i.e.,  $\lambda_{l+j} = \theta$  for some l and for  $j = 1, \dots, k$ , then the spectrum  $\hat{\lambda}$  has multiplicity k-1, or k, or k+1 for that  $\theta$ . This is because  $\hat{\lambda}_{l+j}, j = 1, \dots, k-1$ , are pegged at  $\theta$  due to the interlacing property. So the multiplicity of  $\theta$  is at least k-1. It may further happen that  $\hat{\lambda}_l = \theta$  or  $\hat{\lambda}_{l+k} = \theta$ , or both, leading to the other possibilities.

We can therefore set

$$\frac{f(t)}{g(t)} = \frac{\tilde{f}(t)}{\tilde{g}(t)},$$

where the polynomials  $\tilde{f}$  and  $\tilde{g}$  have the same degree L, with  $1 \le L \le N$ , and have no common factors. Moreover,  $\tilde{g}$  has no multiplicities and the zeros of  $\tilde{f}$  and  $\tilde{g}$  interlace. The same arguments that lead to (37) hold with  $y_{l_i}$  set as in (37) for those indices that survive and  $y_j = 0$  when  $j \ne l_i, i = 1, \dots, L$ . The details are easy to fill and therefore omitted.

#### REFERENCES

- [1] P. Viswanath and V. Anantharam, "Optimal sequences and sum capacity of synchronous CDMA systems," *IEEE Trans. Inform. Theory*, vol. 45, no. 6, pp. 1984 1991, Sep. 1999.
- [2] T. Guess, "CDMA with power control and sequence design: The capacity region with and without multidimensional signaling," *IEEE Trans. Inform. Theory*, vol. 50, no. 11, pp. 2604 2619, Nov. 2004.
- [3] R. A. Horn and C. R. Johnson, Matrix Analysis. Cambridge University Press, 1999.
- [4] M. T. Chu, "Inverse eigenvalue problems," SIAM Rev., vol. 40, pp. 1 39 (electronic), 1998.
- [5] M. T. Chu and G. H. Golub, "Structured inverse eigenvalue problems," Acta Numer., vol. 11, pp. 1 71, Jan. 2002.
- [6] M. Rupf and J. L. Massey, "Optimum sequence multisets for synchronous code-division multi-access channels," IEEE Trans. Inform. Theory, vol. 40, no. 4, pp. 1261 1266, Jul. 1994.
- [7] S. Waldron, "Generalized Welch bound equality sequences are tight frames," IEEE Trans. Inform. Theory, vol. 49, no. 9, pp. 2307 2309, Sep. 2002.
- [8] J. A. Tropp, I. S. Dhillon, and J. R. W. Heath, "Finite-step algorithms for constructing optimal CDMA signature sequences," *IEEE Trans. Inform. Theory*, vol. 50, no. 11, pp. 2916–2921, Nov. 2004.
- [9] R. Sundaresan, "On sequence design for a synchronous CDMA channel," in Proc. 2004 IEEE Int. Symp. on Information Theory, 2004.
- [10] J. A. Tropp, I. S. Dhillon, J. R. W. Heath, and T. Strohmer, "Designing structured tight frames via an alternating projection method," *IEEE Trans. Inform. Theory*, vol. 51, no. 1, pp. 188 208, Jan. 2005.
- [11] P. Viswanath, V. Anantharam, and D. Tse, "Optimal sequences, power control, and user capacity of synchronous CDMA systems with linear MMSE multiuser receivers," *IEEE Trans. Inform. Theory*, vol. 45, no. 6, pp. 1968 1983, Sep. 1999.
- [12] T. Guess, "Optimal sequences for CDMA with decision-feedback receivers," IEEE Trans. Inform. Theory, vol. 49, no. 4, pp. 886 900, Apr. 2003.
- [13] S. Verdú, "Capacity region of gaussian CDMA channels: the symbol-synchronous case," in *Proceedings of the 24th Allerton Conference*, Oct. 1986, pp. 1025 1034.
- [14] D. N. C. Tse and S. Hanly, "Multiaccess fading channels Part I: Polymatroid structure, optimal resource allocation and throughput capacities," *IEEE Trans. Inform. Theory*, vol. IT-44, pp. 2796–2815, Nov. 1998.
- [15] L. Mirsky, "Matrices with prescribed characteristic roots and diagonal elements," J. London Math. Soc., vol. 33, pp. 14 21, 1958.

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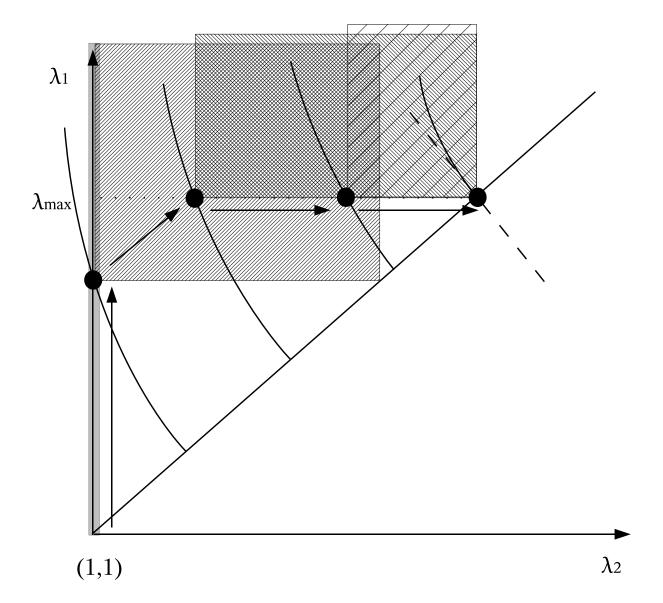


Fig. 1. Rate constrained sequence allocation algorithm

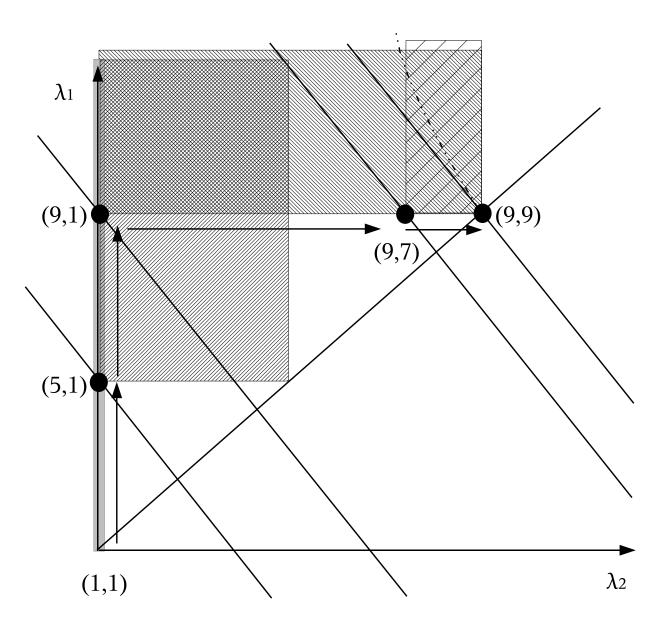


Fig. 2. Power constrained sequence allocation algorithm