

COEG 304: Instrumentation and Control

Chapter 5: Stability Analysis

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Stability Analysis

5.1 Definition of stability and stable system

5.1.1 Stability depending on pole and zero location

5.2 Routh Hurwitz criterion

5.3 Examples to apply RH criteria





- Stability is the most important system specification.
- If a system is unstable, transient response and steady-state error are moot points.
- An unstable system cannot be designed for a specific transient response or steady-state error requirements.
- If so, What is stability?
- If we may recall, the total response of the system is the sum of the forced and natural response, i.e.

$$c(t) = c_{forced}(t) + c_{natural}(t)$$

• A linear, time-invariant system is *stable* if the natural response approaches zero as time approaches infinity.

5.1 Definition of stability and stable system

Forced response

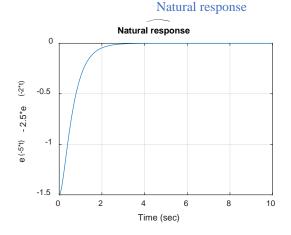


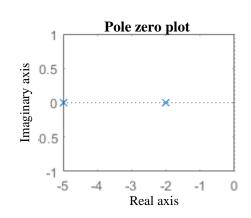
- A linear, time-invariant system is *stable* if the natural response approaches zero as time approaches infinity.
- Consider the following second-order systems:

$$G_1(s) = \frac{15}{(s+2)(s+5)}$$

- $G_1(s)$ has two real poles, i.e. $s_1 = -2$ and $s_2 = -5$.
- The step response of $G_1(s)$ is:

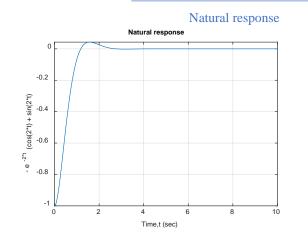
•
$$c_1(t) = e^{-5t} - 2.5e^{-2t} + 1.5$$

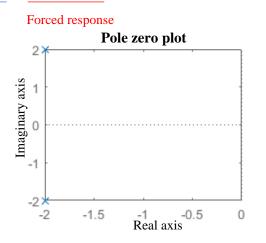




$$G_2(s) = \frac{8}{(s^2 + 4s + 8)}$$

- $G_2(s)$ has a complex-conjugate pair of poles, i.e. $s_{1,2} = -2 \pm j2$
- The step response of $G_2(s)$ is:
- $c_2(t) = -e^{-2t}[\cos(2t) + \sin(2t)] + 1$









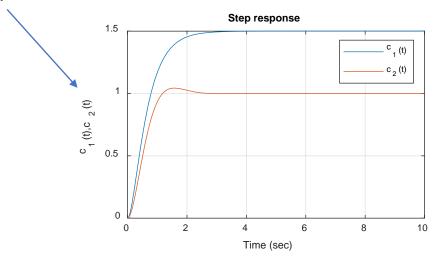
$$G_1(s) = \frac{15}{(s+2)(s+5)}$$

$$G_2(s) = \frac{8}{(s^2+4s+8)}$$

$$c_1(t) = e^{-5t} - 2.5e^{-2t} + 1.5$$

$$c_2(t) = -e^{-2t}[\cos(2t) + \sin(2t)] + 1$$

- Step response in both cases are superposition of natural (transient) and forced (steady-state) response.
- In both cases, the natural response decays to zero as $t \to \infty$.
- Step responses characteristic of stable system.







• Now, consider the following similar-looking systems:

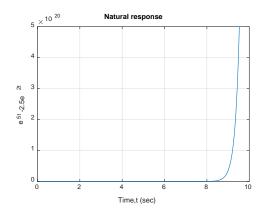
$$G_3(s) = \frac{15}{(s-2)(s-5)}$$

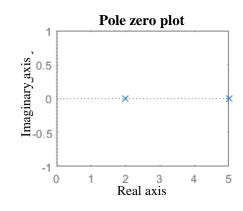
- $G_3(s)$ has two real poles, i.e. $s_1 = 2$ and $s_2 = 5$.
- The step response of $G_1(s)$ is:

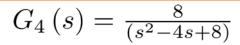
•
$$c_3(t) = e^{5t} - 2.5e^{2t} + 1.5$$

Natural response

Forced response





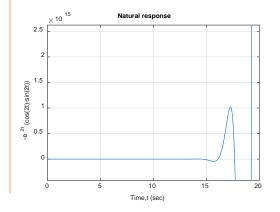


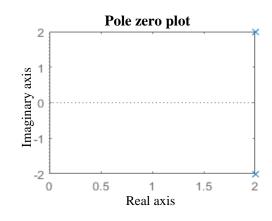
- $G_4(s)$ has a complex-conjugate pair of poles, i.e. $s_{1,2}=2\pm j2$
- The step response of $G_2(s)$ is:

•
$$c_4(t) = -e^{2t}[\cos(2t) - \sin(2t)] + 1$$



Forced response









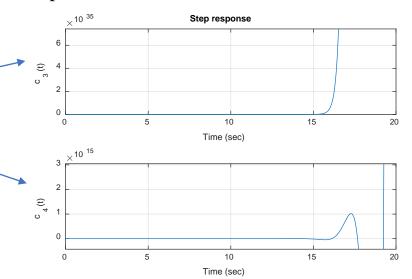
$$G_3(s) = \frac{15}{(s-2)(s-5)}$$

$$G_4(s) = \frac{8}{(s^2-4s+8)}$$

$$c_3(t) = e^{5t} - 2.5e^{2t} + 1.5$$

$$c_4(t) = -e^{2t}[\cos(2t) - \sin(2t)] + 1$$

- Again, step response consists of a natural response and a forced response component.
- However, as $t \to \infty$, the natural response do not decay to zero
 - Such system blow up-Why?
 - Since, exponential terms are positive.
- Step response characteristic of unstable systems



• A linear, time-invariant system is *unstable* if the natural response grows without bound as time approaches infinity

5.1 Definition of stability and stable system



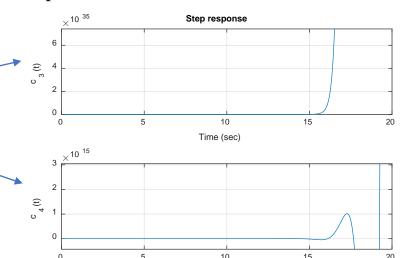
$$G_3(s) = \frac{15}{(s-2)(s-5)}$$

$$G_4\left(s\right) = \frac{8}{\left(s^2 - 4s + 8\right)}$$

•
$$c_3(t) = e^{5t} - 2.5e^{2t} + 1.5$$

•
$$c_4(t) = -e^{2t}[\cos(2t) - \sin(2t)] + 1$$

- Again, step response consists of a natural response and a forced response component.
- However, as $t \to \infty$, the natural response do not decay to zero
 - Such system blow up-Why?
 - Since, exponential terms are positive.
- Step response characteristic of unstable systems



Time (sec)

• A linear, time-invariant system is *unstable* if the natural response grows without bound as time approaches infinity



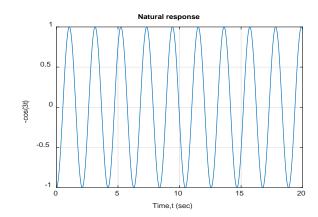


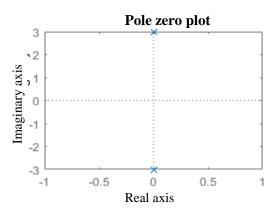
- So, poles in LHP corresponds to stable system while poles in RHP correspond to unstable system.
- It seems that the imaginary axis is the boundary for stability.
- What if poles are on the imaginary axis?
- Consider the following system:

$$G_5\left(s\right) = \frac{9}{\left(s^2 + 9\right)}$$

- Two purely imaginary poles associated with above system are: $s_{1,2} = \pm j3$
- Step response of $G_5(s)$ is $c_5(t) = -\cos(3t) + 1$

Natural response Forced response

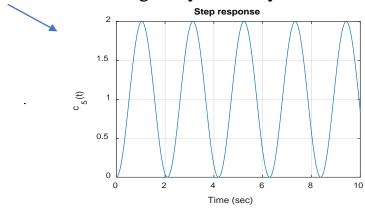








- We can observe that natural response neither decays to zero, nor grows without bound.
 - It oscillates indefinitely
 - Such system is marginally stable
- Step response characteristics of marginally stable system.



• A linear, time-invariant system is *marginally stable* if the natural response neither decays nor grows but remains constant or oscillate as time approaches infinity.





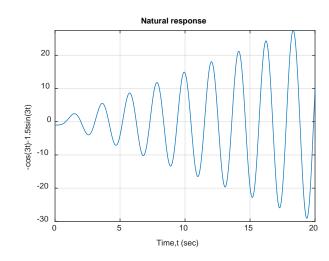
- We'll look at one more interesting case.
- Consider the following system [with repeated imaginary poles]

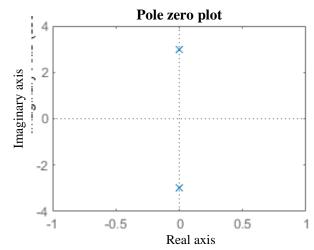
$$G_6(s) = \frac{81}{(s^2+9)^2}$$

- Repeated poles on the imaginary axis are: $s_{1,2} = \pm j3$ and $s_{3,4} = \pm j3$
- The step response of the system is : $c_6(t) = -\cos(3t) \frac{3t\sin(3t)}{2} + 1$

Natural response

Forced response



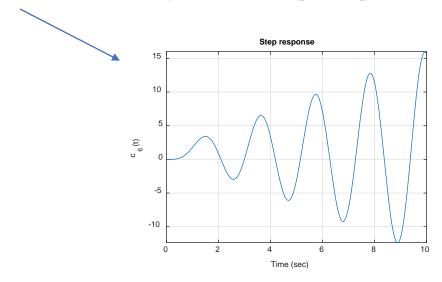






•
$$c_6(t) = -\cos(3t) - \frac{3t\sin(3t)}{2} + 1$$

- As observed, multiplying time factor causes the natural response to grow without bound.
 - An unstable system
 - Multiple identical poles on the imaginary axis implies an unstable system
- Step response characteristics of unstable system due to repeated poles in imaginary axis:

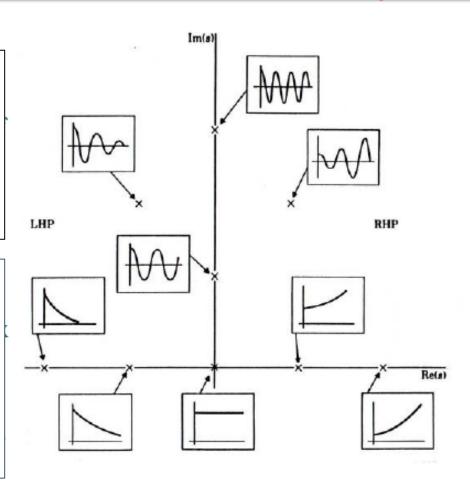


5.1 Definition of stability and stable system



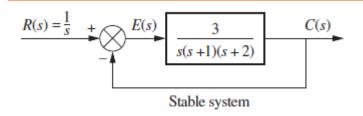
Summary

- Let us summarize our definitions of stability for linear, time-invariant systems Using the natural response:
 - A system is stable if the natural response approaches zero as time approaches infinity.
 - A system is unstable if the natural response approaches infinity as time approaches infinity.
 - A system is marginally stable if the natural response neither decays nor grows but remains constant or oscillates.
- Let us summarize our definitions of stability for linear, time-invariant systems Using the total response:
 - A system is stable if *every* bounded input yields a bounded output.
 - A system is unstable is *any* bounded input yields an unbounded output.



5.1 Definition of stability and stable system [based on closed loop transfer function]



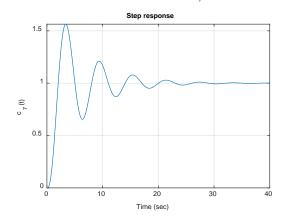


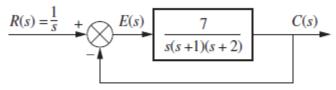
Closed loop system transfer function will be:

$$G_7(s) = \frac{C(s)}{R(s)} = \frac{3}{(s(s+1)(s+2)+3)}$$

• The poles of the above system are:

$$s_1 = -2.6717, s_{2,3} = -0.1642 \pm 1.0469i$$





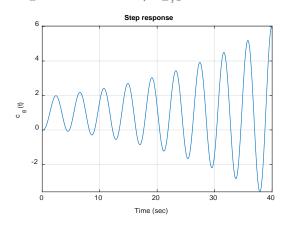
Unstable system

• Closed loop system transfer function will be:

$$G_8(s) = \frac{C(s)}{R(s)} = \frac{7}{(s(s+1)(s+2)+7)}$$

• The poles of the above system are:

$$s_1 = -3.0867, s_{2,3} = 0.0434 \pm 1.5053i$$



5.1 Definition of stability and stable system [based on closed loop transfer function]



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- Stable systems has closed-loop transfer functions with poles only in the left half-plane.
- Unstable system have closed-loop transfer functions with at least one pole in the right half-plane and/or poles of multiplicity greater than 1 on the imaginary axis.
- Marginally stable systems have closed-loop transfer functions with only imaginary axis poles of multiplicity 1 and poles in the left half-plane.

References



- [1] Nise, Norman S., Control Systems Engineering, 7th ed, Hoboken, NJ: Wiley, 2004.
- [2] K. Webb, Class Lecture, Topic: "Section 6: Stability" ESE499, College of Engineering, OSU-Cascades, Oregon, *accessed on*: June 2
- [3] Reymond T. Stefani, Design of Feedback Control Systems, 4th ed, Oxford, 2004

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Thank You!

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