

Resistive Operation

0.02 / 948

Threshold Voltage

- The 'V_{gs}' voltage at which 'strong inversion' occurs is called threshold vge (V_t)

Resistive Operation

- Since, V_{gs} = V_t, let us observe what happens at different voltages of 'V_{gs} > V_t'
- Hence, in the channel (shown in yellow circle), Induced charge (Q_i) \propto (V_{gs} - V_t)
- Let's analyze at V_{gs} = 1V and small V_{ds} (say 0.05V). Assume V_t (NMOS) = 0.45V
- Let 'x' axis be along the channel and 'y' axis be perpendicular to channel, as shown

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- Let's analyze at V_{gs} = 1V and small V_{ds} (say 0.05V). Assume V_t (NMOS) = 0.45V
- Let 'x' axis be along the channel and 'y' axis be perpendicular to channel, as shown
- Now, in the channel (shown in yellow circle), Induced charge at any point 'x' $Q_i(x) \propto -([V_{gs} - V(x)] - V_t)$ i.e.

First order analysis:

$$Q_i(x) \propto -([V_{gs} - V(x)] - V_t)$$

i.e.

$$Q_i(x) = -C_{ox} ([V_{gs} - V(x)] - V_t)$$

From device point of view, we have 2 kinds of currents

- 1) Drift current = Current due to potential difference
- 2) Diffusion current = Current due to difference in carrier concentration

Here we focus on drift current

First order analysis:

$$Q_i(x) \propto -([V_{gs} - V(x)] - V_t)$$

i.e.

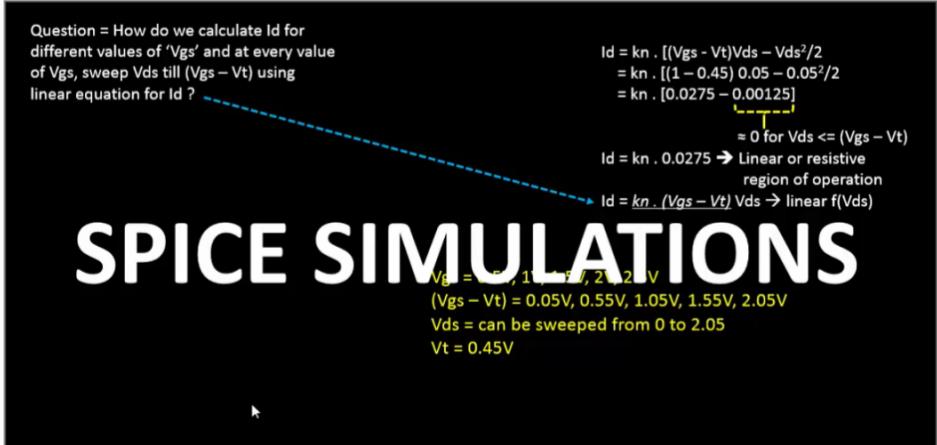
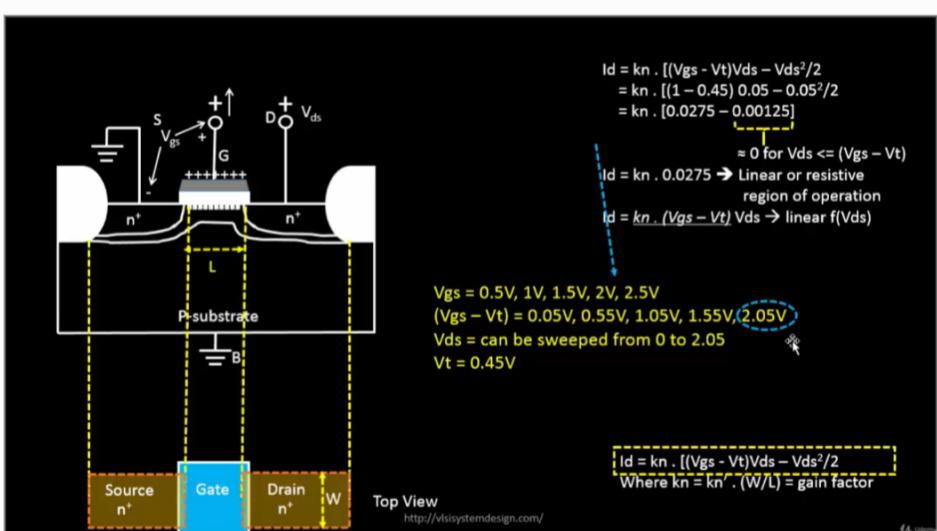
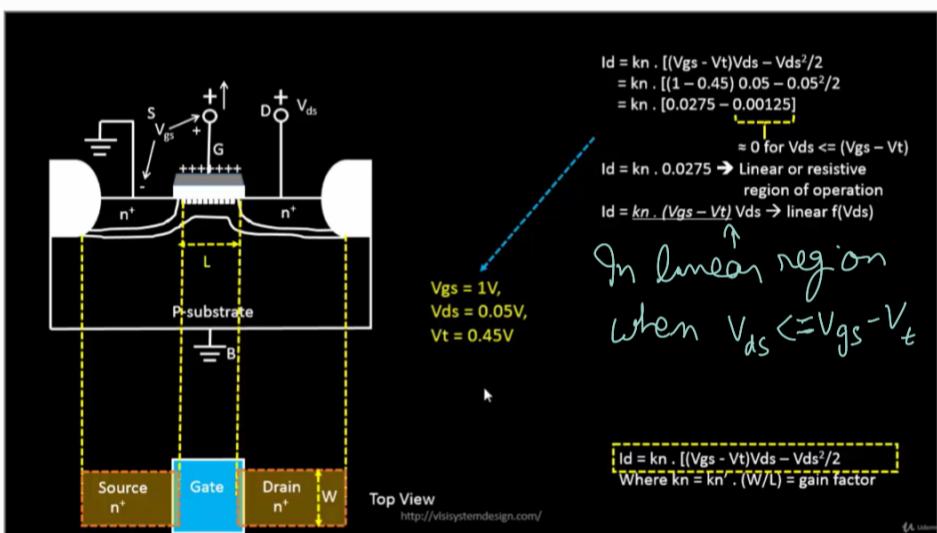
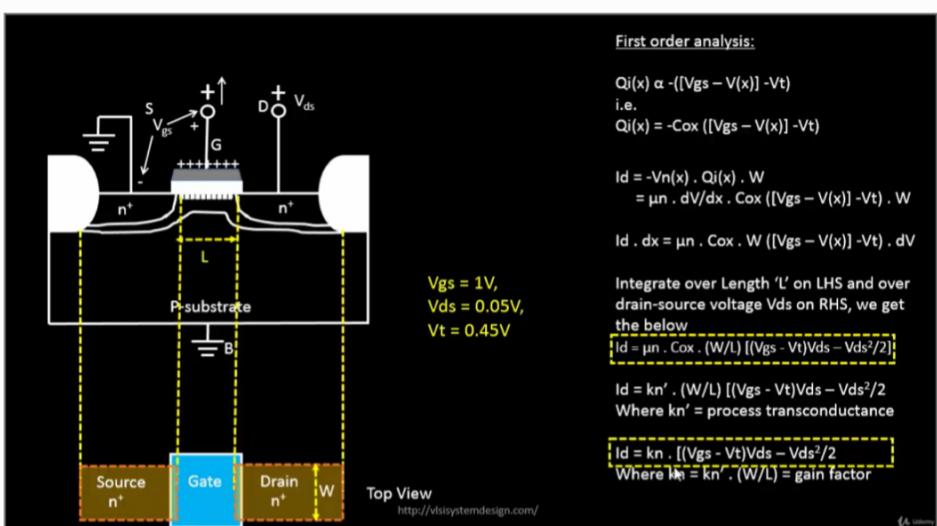
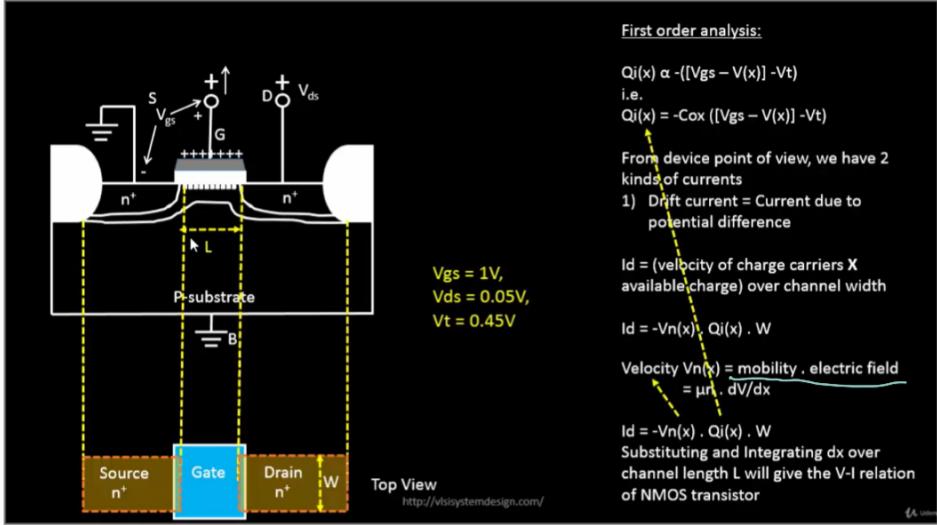
$$Q_i(x) = -C_{ox} ([V_{gs} - V(x)] - V_t)$$

From device point of view, we have 2 kinds of currents

- 1) Drift current = Current due to potential difference

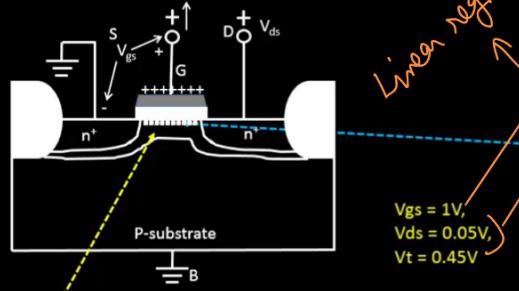
$I_d = (\text{velocity of charge carriers } \times \text{available charge}) \text{ over channel width}$

Source n+, Gate, Drain n+, W, Top View http://systemdesign.com/



Saturation Region

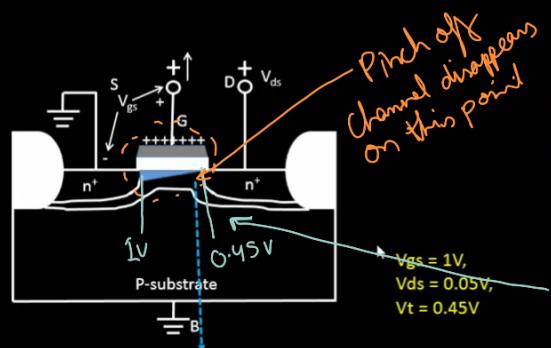
Channel voltage = $V_{gs} - V_{ds}$



$V(x)$ is voltage at a point 'x' along channel
 $V_{gs} - V(x)$ is the gate-to-channel voltage at that point.

V_{gs}	V_{ds}	$V_{gs} - V_{ds}$		V_t
1	0.05	0.95	>	0.45
1	0.15	0.85	>	0.45
1	0.25	0.75	>	0.45
1	0.35	0.65	>	0.45
1	0.45	0.55	>	0.45
1	0.55	0.45	=	0.45
1	0.65	0.35	<	0.45
1	0.75	0.25	<	0.45
1	0.85	0.15	<	0.45
1	0.95	0.05	<	0.45

Linear region

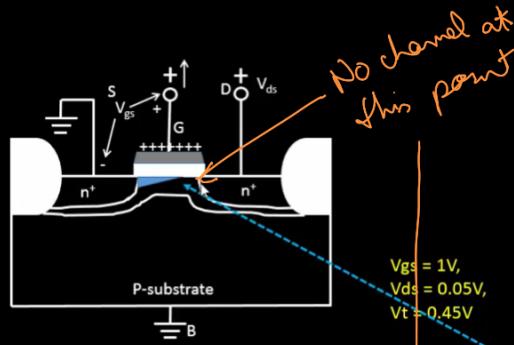


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1	0.65	0.35	<	0.45
1	0.75	0.25	<	0.45
1	0.85	0.15	<	0.45
1	0.95	0.05	<	0.45

Pinch-off
Channel disappears
on this point

$V(x)$ is voltage at a point 'x' along channel
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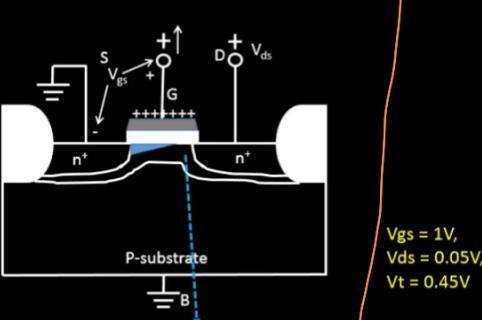


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1	0.85	0.15	<	0.45
1	0.95	0.05	<	0.45

No channel at this point

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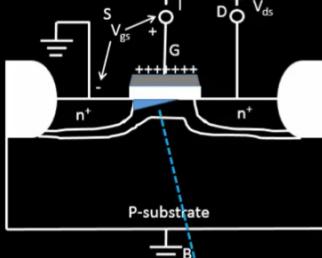
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Pinch-Off region – no channel near drain region
 Pinch-off condition: $V_{gs} - V_{ds} \leq V_t$

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Channel voltage = $V_{gs} - V_{ds}$



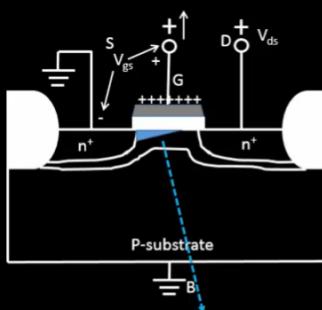
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Voltage over the channel remains constant = $V_{gs} - Vt$

$$V_{gs} = 1V, V_{ds} = 0.05V, V_t = 0.45V$$

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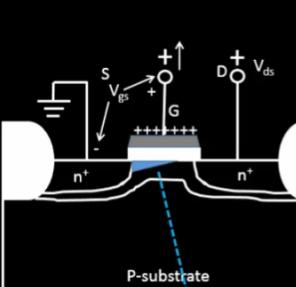
Voltage over the channel remains constant = $V_{gs} - Vt$

$$V_{gs} = 1V, V_{ds} = 0.05V, V_t = 0.45V$$

Channel voltage = $V_{gs} - V_{ds}$

$$Id = kn \cdot [(V_{gs} - V_t)V_{ds} - V_{ds}^2/2] \\ \text{Where } kn = kn' \cdot (W/L) = \text{gain factor}$$

$$\uparrow Id = kn \cdot (V_{gs} - V_t) V_{ds} \rightarrow \text{linear } f(V_{ds})$$



Channel voltage = $V_{gs} - V_{ds}$

$$Id = kn \cdot [(V_{gs} - V_t)(V_{gs} - V_t) - (V_{gs} - V_t)^2/2] \\ \text{Where } kn = kn' \cdot (W/L) = \text{gain factor}$$

$$Id = kn/2 \cdot (V_{gs} - V_t)^2 \rightarrow (\text{no more linear } f(V_{ds}))$$

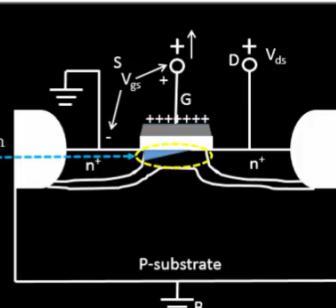
$$kn = kn' \cdot (W/L) = \mu_n \cdot Cox \cdot (W/L)$$

$$Id = \frac{Kn}{2} \cdot \frac{W}{L} \cdot (V_{gs} - V_t)^2$$

All constant. However L depends on V_{ds}

Replace V_{ds} by $V_{gs} - V_t$

$V(x)$ is voltage at a point 'x' along channel
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Effective conductive channel length is modulated by applied V_{ds}
 $V_{ds} \uparrow$ Depletion region at drain \uparrow
Effective channel length \downarrow

Looks like perfect current source i.e. current is constant.
Not 'correct'

$$Id = \frac{Kn}{2} \cdot \frac{W}{L} \cdot (V_{gs} - V_t)^2 [1 + \lambda V_{ds}]$$

More accurate equation, and λ = channel length modulation

