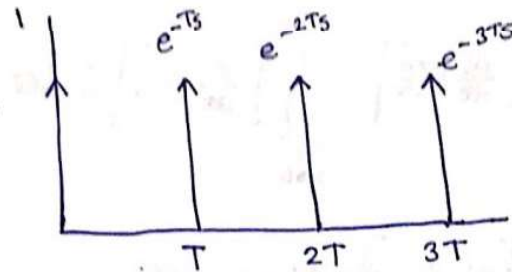


● Find LT of the following—



Laplace Transform of 1st Cycle = 1.

∴ using periodic signal formula

$$= 1 \cdot \frac{1}{1 - e^{-Ts}}$$

$$= \frac{1}{1 - e^{-Ts}}$$

### Initial Value and Final Value Theorem

$f(0) \rightarrow$  Value at 0 (Initial Value)

$f(\infty) \rightarrow$  Final Value

If given in s domain convert to time domain using Inverse Laplace.

Then apply above method.

This is the classical method.

But it can <sup>be</sup> directly found as well.

$$L[f(t)] = F(s)$$

$$\lim_{s \rightarrow \infty} L\left[\frac{d}{dt}f(t)\right] = \int_0^{\infty} \left[\frac{d}{dt}f(t)\right] e^{-st} dt$$

$$\Rightarrow \lim_{s \rightarrow \infty} [sF(s) - f(0)] = 0$$

$$\Rightarrow \boxed{\lim_{s \rightarrow \infty} sF(s) = \lim_{s \rightarrow \infty} f(0) = f(0)}$$

→ Initial Value Theorem

$$\lim_{s \rightarrow 0} \mathcal{L} \left[ \frac{d}{dt} f(t) \right] = \int_0^{\infty} \left[ \frac{d}{dt} f(t) \right] e^{-st} dt$$

$$\Rightarrow \lim_{s \rightarrow 0} [sF(s) - f(0)] = f(\infty) - f(0)$$

$$\Rightarrow \boxed{\lim_{s \rightarrow 0} sF(s) = f(\infty)}$$

→ Final Value Theorem

1.  $F(s) = \frac{1}{s(s+2)}$

Find out the initial value and final value of the equation.

∴ final value

$$= \lim_{s \rightarrow 0} sF(s)$$

$$= \lim_{s \rightarrow 0} \frac{1}{(s+2)}$$

$$= \frac{1}{2} \text{ (Ans)}$$

∴ initial value

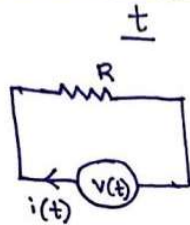
$$= \lim_{s \rightarrow \infty} sF(s)$$

$$= \lim_{s \rightarrow \infty} \frac{1}{(s+2)}$$

$$= 0 \text{ (Ans)}$$

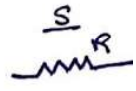
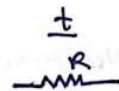
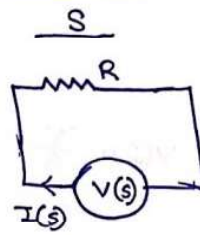
# Laplace Domain Representation of R, L, C

1) R

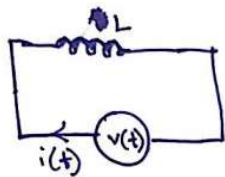


$$v(t) = R i(t)$$

$$V(s) = R I(s)$$



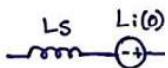
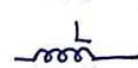
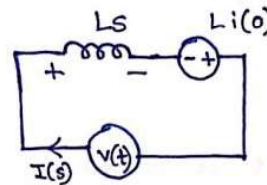
2) L



$$v(t) = L \frac{di(t)}{dt}$$

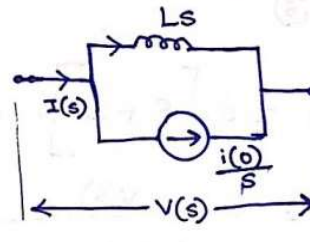
$$V(s) = L [s I(s) - i(0)]$$

$$= L s I(s) - L i(0)$$

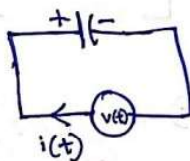


Current Source

$$I(s) = \frac{V(s)}{Ls} + \frac{L i(0)}{Ls}$$



3) C

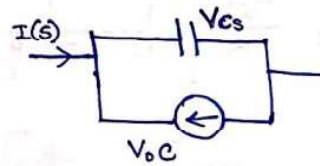
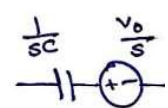
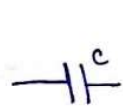
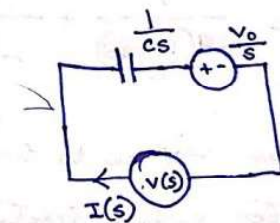


$$v(t) = \frac{1}{C} \int i(t) dt + V_0$$

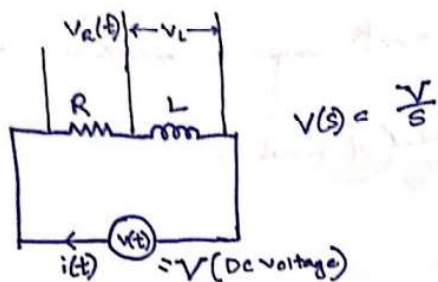
$$V(s) = \frac{1}{C} \frac{I(s)}{s} + \frac{V_0}{s}$$

Current Source

$$I(s) = C s V(s) - V_0 C$$



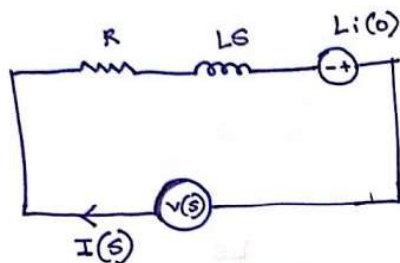
## Transient Response



O/P

- 1)  $i(t)$
- 2)  $V_R(t) = i(t)R$
- 3)  $V_L(t)$

Convert from  $t$  to  $s$  domain:



Using KVL,

$$V(s) = I(s) [R + LS] - Li(0)$$

$$\Rightarrow I(s) = \frac{V(s)}{R + LS} + \frac{Li(0)}{R + LS}$$

$$= \frac{V}{LS \left( \frac{R}{L} + s \right)} + \frac{Li(0)}{L \left( s + \frac{R}{L} \right)} \quad \left[ \because V(s) = \frac{V}{s} \right]$$

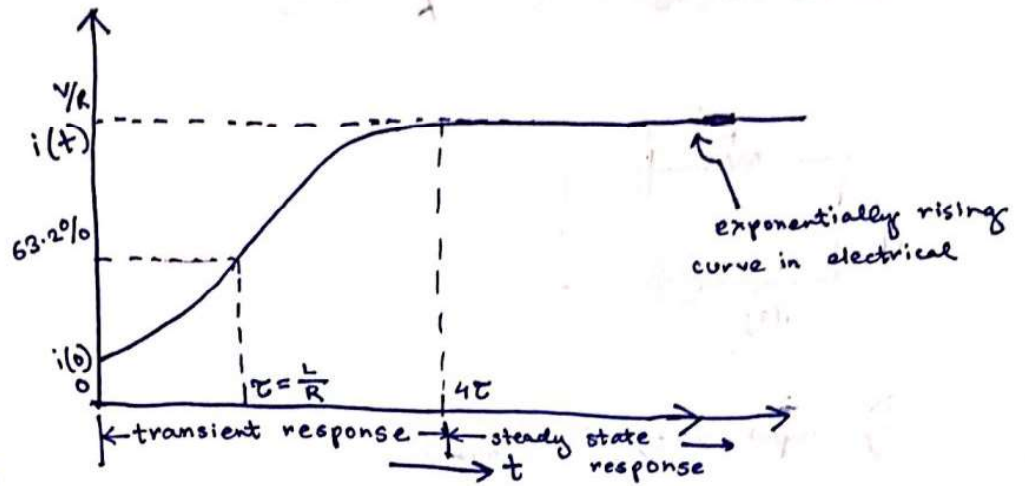
Use inverse Laplace to find  $i(t)$ .

Convert to partial fraction form.  $\rightarrow \frac{A}{s} + \frac{B}{s + \frac{R}{L}}$

$$A = \frac{V}{R} \quad B = -\frac{V}{R}$$

$$I(s) = \frac{V}{R} \left[ \frac{1}{s} - \frac{1}{s + \frac{R}{L}} \right] + \frac{i(0)}{s + \frac{R}{L}}$$

$$i(t) = \frac{V}{R} \left[ 1 - e^{-\frac{R}{L}t} \right] + i(0) e^{-\frac{R}{L}t}$$



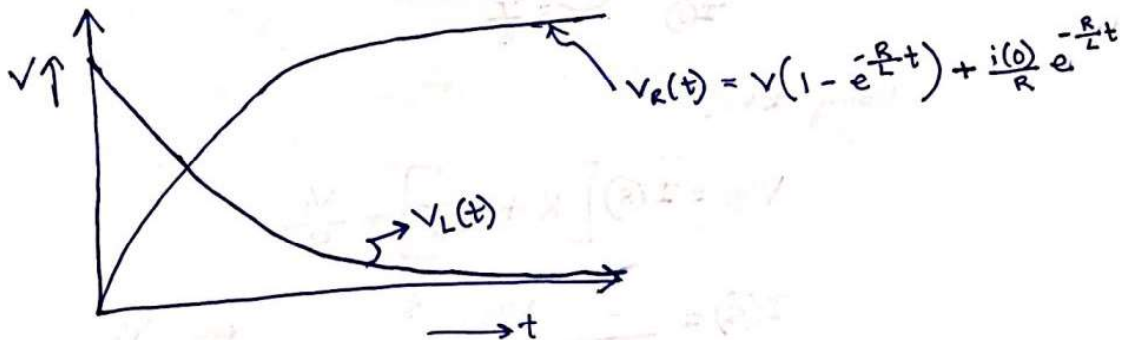
Transient response  $\rightarrow 0$  to  $4\tau$

Beyond  $4\tau \rightarrow$  Steady state response

Time constant of RL Circuit: At time  $\tau = \frac{L}{R}$ , the circuit response will be 63.2% of its steady state / <sup>final</sup> response. This time is ~~not~~ called Time constant. Unit of  $\tau = \text{sec}$

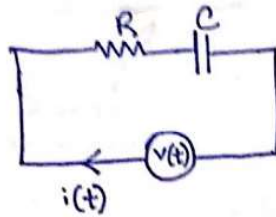
Imp

- At time  $t=0$  it acts as an open circuit, at  $t=\infty$ , it acts as a short circuit.





# Series Transient Response of RC Circuit



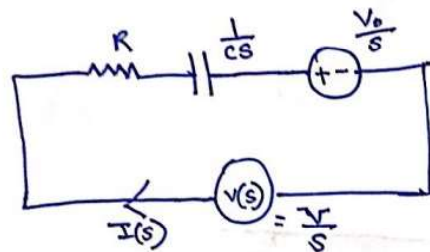
1) input  $v(t) = V(\text{const})$

2) output i)  $i(t)$

ii)  $V_R(t) = i(t)R$

iii)  $V_C(t) = V - V_R(t)$

Converting from  $t$  to  $s$  domain,



Using KVL,

$$V(s) = I(s) \left[ R + \frac{1}{Cs} \right] + \frac{V_0}{s}$$

$$I(s) = \frac{V(s) - \frac{V_0}{s}}{R + \frac{1}{Cs}} = \frac{\frac{V}{s} - \frac{V_0}{s}}{R + \frac{1}{Cs}}$$

$$= \frac{\frac{V - V_0}{s} \cdot Cs}{RCS + 1} = \frac{(V - V_0)Cs}{s(RCS + 1)}$$

$$= \frac{V - V_0}{R} \cdot \frac{1}{\left(s + \frac{1}{RC}\right)} = \frac{V - V_0}{R} \cdot \frac{1}{\left(s + \frac{1}{RC}\right)}$$

##

Using Inverse Laplace,

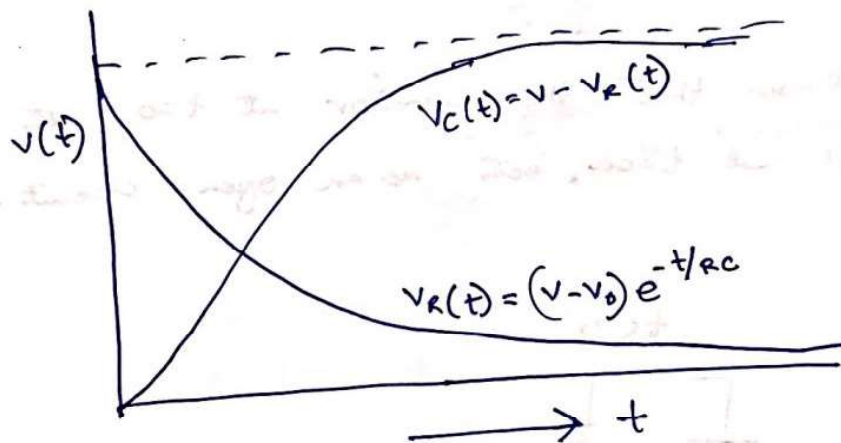
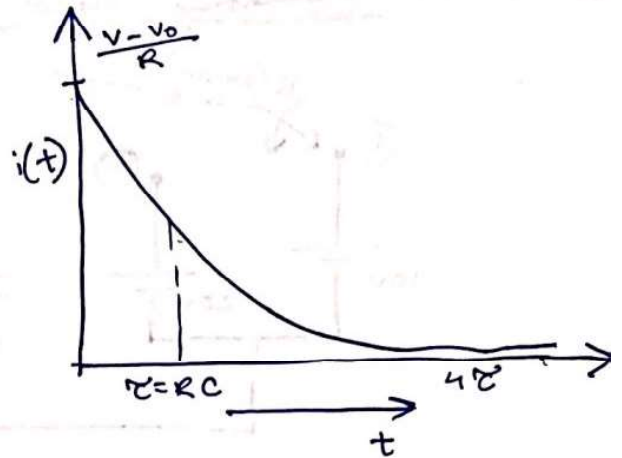
$$\therefore i(t) = \left( \frac{V - V_0}{R} \right) e^{-\frac{t}{RC}} \quad \therefore \text{exponentially decreasing}$$

- At  $t = RC$ , the current response will be 36.7% of its initial response.

- Steady State current = 0.

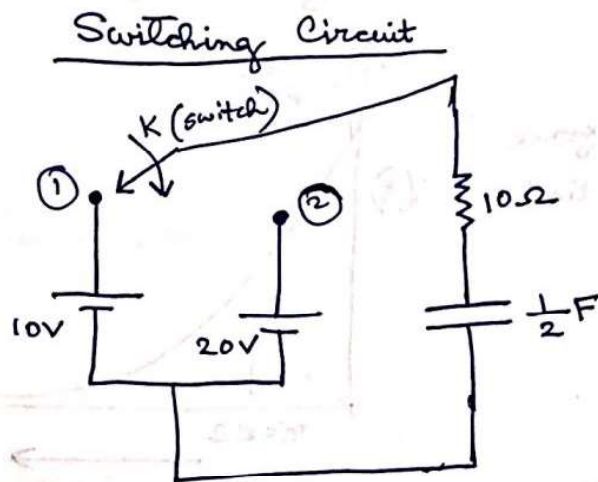
- At Time  $t=0$ , Capacitor acts as a short circuit.

At  $t = \infty$ , Capacitor acts as an open circuit.



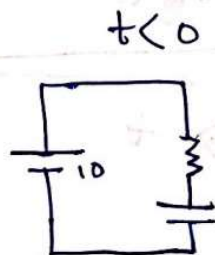
Now, if a gate function,  is applied.

i) input  $v(t) = V[u(t) - u(t-T)]$



1. Suppose switch  $K$  is at position 1 for a long time. Now at  $t=0$ , switch is moved from position 1 to position 2. Now find out the transient current through the capacitor.

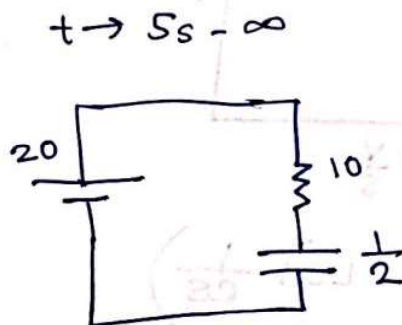
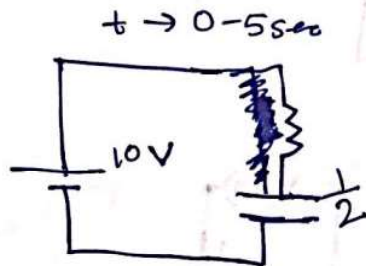
We know that the capacitor at  $t=0$  acts as a short circuit and at  $t=\infty$ , acts as an open circuit.



In this condition, the circuit remains ~~a~~ for a long time.  $\therefore$  Steady state is ~~is~~ reached.

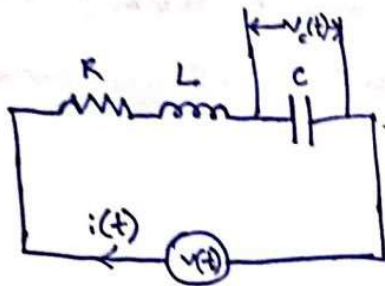


2. At time  $t=0$ , the switch is at position 1. After 5 sec, the switch is moved from position 1 to position 2. Now find out the Transient current through the circuit and the find out the Transient voltage across the capacitor.

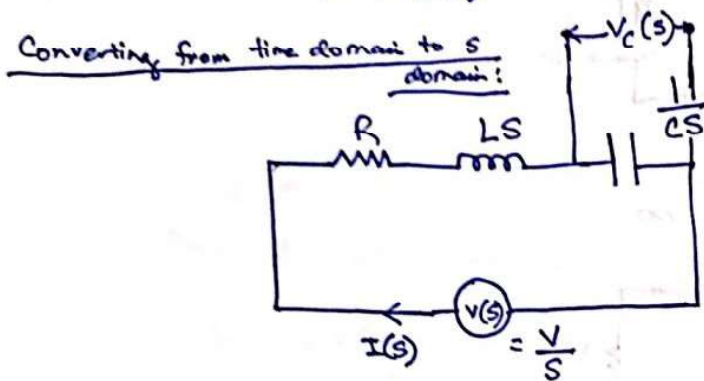


~~3.67 V~~  $\rightarrow$  Initial Voltage  
3.67 V

## Transient Response of RLC Series Circuit



O/P  $\rightarrow v_c(t)$



$$V(s) = I(s) \left( R + LS + \frac{1}{CS} \right)$$

$$\Rightarrow I(s) = \frac{V(s)}{R + LS + \frac{1}{CS}}$$

$$\Rightarrow V_c(s) = I(s) \frac{1}{CS} = \frac{V(s) \cancel{CS}}{(RCS + LCS^2 + 1)} \times \frac{1}{\cancel{CS}}$$

$$\Rightarrow \frac{V_c(s)}{V(s)} = \frac{1}{LC \left( s^2 + \frac{R}{L}s + \frac{1}{LC} \right)}$$

$$\Rightarrow \frac{V_c(s)}{V(s)} = \frac{\left( \frac{1}{\sqrt{LC}} \right)^2}{s^2 + \frac{R}{L}s + \left( \frac{1}{\sqrt{LC}} \right)^2}$$

System Transfer Function

This equation is called a ~~2~~ second order system.

# Transfer Function of any 2nd order system

$$\frac{\omega_n^2}{s^2 + 2 \xi \omega_n s + \omega_n^2}$$

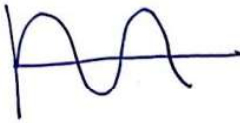
$\omega_n \rightarrow$  natural frequency

$\xi \rightarrow$  damping constant

1)  $\xi = 0$  Inverse Laplace Transform of  $\frac{\omega_n^2}{s^2 + \omega_n^2}$

$\omega_n \sin \omega_n t$

(Undamped)



2)  $0 < \xi < 1$

$$\frac{\omega_n^2}{s^2 + 2 \xi \omega_n s + \omega_n^2} \Rightarrow \frac{a}{(s+b)^2 + a^2}$$

$$\Rightarrow \frac{\omega_n^2}{(s + \xi \omega_n)^2 - \xi^2 \omega_n^2 + \omega_n^2}$$

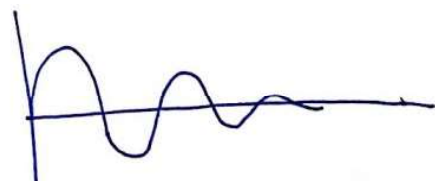
$$\Rightarrow \frac{\omega_n^2}{(s + \xi \omega_n)^2 + (\omega_n \sqrt{1 - \xi^2})^2}$$

$$\Rightarrow \frac{1}{\sqrt{1 - \xi^2}} \frac{\omega_n \omega_n \sqrt{1 - \xi^2}}{(s + \xi \omega_n)^2 + (\omega_n \sqrt{1 - \xi^2})^2}$$

Inverse  
Taking Laplace Transform,

$$\frac{\omega_n}{\sqrt{1 - \xi^2}} e^{-\xi \omega_n t} \sin \left[ (\omega_n \sqrt{1 - \xi^2}) t \right]$$

(Underdamped System)



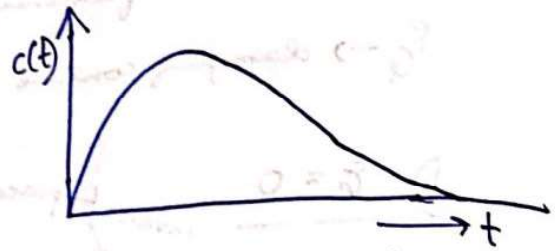
3)  $\xi = 1$

Taking  $\xrightarrow{\text{Inverse Laplace Transform}}$

$$t e^{-\omega_n t}$$

At  $t=0$ , 0

At  $t=\infty$ , 0



(Critically Damped)

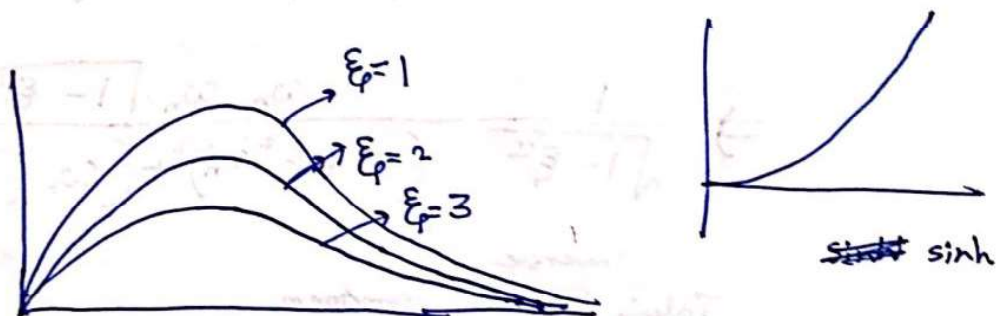
4)  $\xi > 1$

Like ②,

$$\frac{1}{\sqrt{\xi^2 - 1}} \frac{\omega_n \omega_n \sqrt{\xi^2 - 1}}{(s + \xi \omega_n)^2 - (\omega_n \sqrt{\xi^2 - 1})^2}$$

Taking Inverse Laplace Transform,

$$\frac{\omega_n}{\sqrt{\xi^2 - 1}} e^{-\xi \omega_n t} \sinh \left[ \omega_n \sqrt{(\xi^2 - 1)} t \right]$$



(Overdamped)

$$\text{Now, } \omega_n = \sqrt{\frac{1}{LC}}$$

$$\xi = \frac{R}{2} \sqrt{\frac{C}{L}}$$