

[4]

- ii. For the network shown in figure 8, determine the current  $i(t)$  when the switch is closed at  $t = 0$  with zero initial conditions.

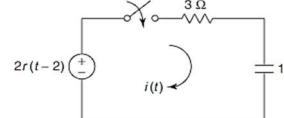


Figure 8

7 03 02 02 01

Total No. of Questions: 6

Total No. of Printed Pages: 4

- OR iii. In the network of figure 9, the switch is moved from the position 1 to 2 at  $t = 0$ , steady-state condition having been established in the position 1. Determine  $i(t)$  for  $t > 0$ .

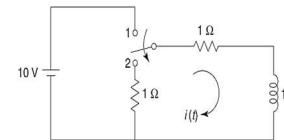


Figure 9

7 03 02 02 1

Programme: B.Tech.

Branch/Specialisation: EC

**Maximum Marks: 60**

- Q.5 i. What does  $y_{22}$  represent in Y-parameters?  
ii. Find Z-parameters for the network shown in figure 10.

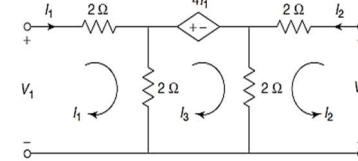


Figure 10

2 02 01 05 01  
8 03 02 05 1

- OR iii. Find h-parameters for the network shown in figure 11.

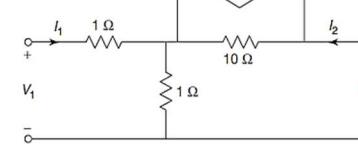


Figure 11

8 03 02 05 01

- Q.6 i. What is the significance of pole-zero analysis in determining the stability of a system?  
ii. Determine the Foster-I form realisation of the RC impedance function-

$$Z(s) = \frac{(s+1)(s+3)}{s(s+2)(s+4)}$$

- OR iii. Determine the Cauer-I form of the RL impedance function-

$$Z(s) = \frac{s(s+4)(s+8)}{(s+1)(s+6)}$$

\*\*\*\*\*



**Duration: 3 Hrs.**

Note: All questions are compulsory. Internal choices, if any, are indicated. Answers of Q.1 (MCQs) should be written in full instead of only a, b, c or d. Assume suitable data if necessary. Notations and symbols have their usual meaning.

- |   | Marks | BL | PO | CO | PSO |
|---|-------|----|----|----|-----|
| Q.1 i. A dependent (controlled) source in a circuit depends on:   | 1     | 01 | 01 | 01 | 01  |
| (a) The value of another voltage or current in the circuit  |       |    |    |    |     |
| (b) The internal resistance of the source   |       |    |    |    |     |
| (c) Only the input power supply   |       |    |    |    |     |
| (d) The type of circuit configuration   |       |    |    |    |     |
| ii. In mesh analysis, which of the following is primarily calculated?   | 1     | 02 | 01 | 01 | 01  |
| (a) Node voltages      (b) Branch voltages  |       |    |    |    |     |
| (c) Loop currents      (d) Equivalent resistance  |       |    |    |    |     |
| iii. In the network shown in figure 1, the switch is closed at $t=0$ . With the capacitor uncharged, value of current $i$ at $t=0+$ . | 1     | 03 | 02 | 02 | 01  |
|   |       |    |    |    |     |
| iv. (a) 0 A    (b) 0.1 A    (c) 100 A    (d) 1000 A   |       |    |    |    |     |
| The maximum power transfer theorem states that maximum power is transferred from a source to a load when:                             | 1     | 02 | 01 | 03 | 01  |
| (a) The load resistance is maximum  |       |    |    |    |     |
| (b) The load resistance equals the source resistance  |       |    |    |    |     |
| (c) The load resistance is zero   |       |    |    |    |     |
| (d) The load resistance is double the source resistance   |       |    |    |    |     |

Figure 1

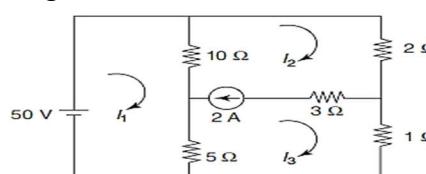


Figure 3

[4]

- 1 02 01 04 01

[3]

- iii. For the circuit shown in figure 4, draw its graph and write the fundamental cutset matrix.

5 04 02 01 02

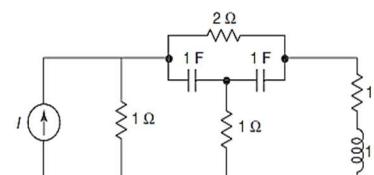


Figure 4

- OR iv. For the circuit shown in figure 5, draw its graph and write the tieset matrix.

5 04 02 01 02

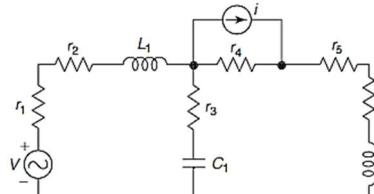


Figure 5

Q.3 i. Define Tellegen's Theorem.

2 02 01 03 01

ii. Derive the expression for the power delivered to the load in terms of load and source resistance and use it to find the condition for maximum power.

3 03 01, 03

iii. Use Superposition theorem to find the current through the  $6\ \Omega$  resistor shown in figure 6.

5 03 02 02 01

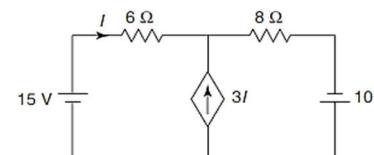


Figure 6

- OR iv. Use Thevenin's theorem to find the current through the  $1\ \Omega$  resistor shown in figure 7.

5 03 02 02 01

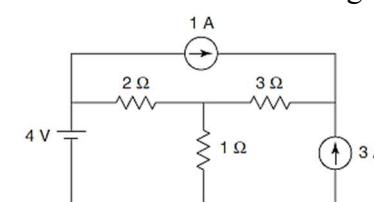


Figure 7

- Q.4** i. State the initial value theorem and the final value theorem in Laplace transforms and their significance.

3 02 01 04 01

## Marking Scheme

### EC3CO05 (T) Circuit Analysis & Synthesis (T)

- Q.1**
- i) A dependent (controlled) source in a circuit depends on: **1**
  - a) The value of another voltage or current in the circuit.
  - ii) In mesh analysis, which of the following is primarily calculated? **1**
  - c) Loop currents
  - iii) In the network shown in Figure 1, the switch is closed at  $t=0$ . With the capacitor uncharged, value of current  $i$  at  $t=0+$ . **1**

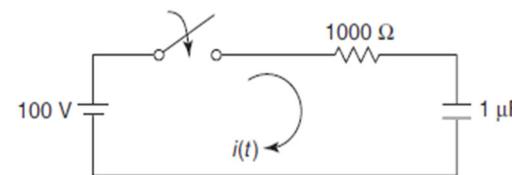


Figure 1

- b) 0.1 A
- iv) **The Maximum Power Transfer Theorem states that maximum power is transferred from a source to a load when:** **1**
- b) The load resistance equals the source resistance.
- v) Which property of the Laplace transform states that  $L[f(t-T)u(t-T)] = e^{-Ts} F(s)$ ? **1**
- d) Time Shifting
- vi) The Laplace transform of a unit impulse function  $\delta(t-T)$  is **1**
- a)  $e^{-Ts}$
- vii) The value of  $Z_{11}$  for the network shown in Figure 2. **1**

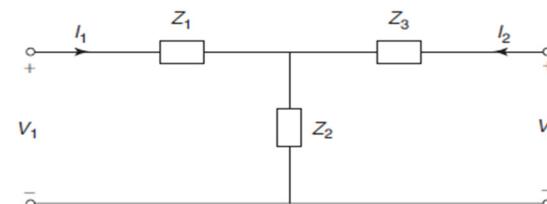


Figure 2

- c)  $Z_1 + Z_2$
- viii) **In a two-port network with ABCD parameters, the condition of symmetry is given by:** **1**
- a)  $A=D$
- ix) **A Hurwitz polynomial is defined as a polynomial where:** **1**
- c) All roots lie in the left-half of the s-plane.
- x) **What is the primary application of the Foster-I form in network** **1**

**synthesis?**

- d) To achieve the desired impedance characteristics.

- Q.2**
- i. Source transformation is the process of converting a voltage source in series with a resistor into an equivalent current source in parallel with the same resistor, and vice versa. It simplifies circuit analysis. **2**

- ii. **Solution** Applying KVL to Mesh 1,
- $$50 - 10(I_1 - I_2) - 5(I_1 - I_3) = 0 \quad \dots(i)$$
- $$15I_1 - 10I_2 - 5I_3 = 50$$

Meshes 2 and 3 will form a supermesh as these two meshes share a common current source of 2 A.  
Writing current equation for the supermesh,

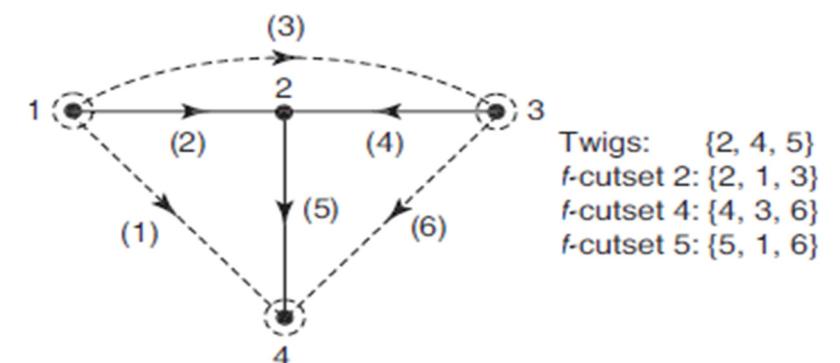
$$\begin{aligned} I_2 - I_3 &= 2 & \dots(ii) \\ -10(I_2 - I_1) - 2I_2 - 1I_3 - 5(I_3 - I_1) &= 0 \\ -15I_1 + 12I_2 + 6I_3 &= 0 & \dots(iii) \end{aligned}$$

Solving Eqs (i), (ii) and (iii),

$$\begin{aligned} I_1 &= 20 \text{ A} \\ I_2 &= 17.33 \text{ A} \\ I_3 &= 15.33 \text{ A} \end{aligned}$$

Current through the 5  $\Omega$  resistor =  $I_1 - I_3 = 20 - 15.33 = 4.67 \text{ A}$

iii.

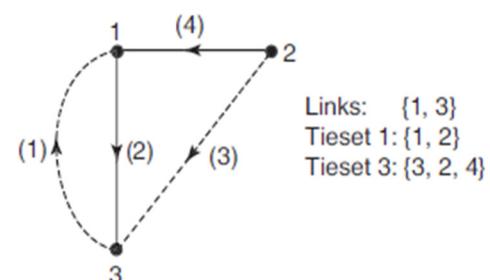
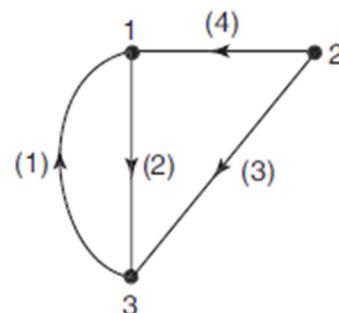


Fundamental Cutset Matrix ( $Q$ )

$$Q = \begin{bmatrix} 1 & 2 & 3 & 4 & 5 & 6 \\ 2 & 1 & 1 & 1 & 0 & 0 \\ 4 & 0 & 0 & -1 & 1 & 0 \\ 5 & 1 & 0 & 0 & 0 & 1 \end{bmatrix}$$

[2]

OR iv.



$$B = \begin{bmatrix} 1 & 2 & 3 & 4 \\ 1 & 1 & 0 & 0 \\ 0 & -1 & 1 & -1 \end{bmatrix}$$

5

[3]

$$I' = I_1$$

... (i)

Meshes 1 and 2 will form a supermesh.  
Writing current equation for the supermesh,

$$I_2 - I_1 = 3I' = 3I_1$$

$$4I_1 - I_2 = 0$$

... (ii)

Applying KVL to the outer path of the supermesh,

$$15 - 6I_1 - 8I_2 = 0$$

$$6I_1 + 8I_2 = 15$$

Solving Eqs (ii) and (iii),

$$I_1 = 0.39\text{ A}$$

$$I_2 = 1.57\text{ A}$$

$$I' = I_1 = 0.39\text{ A} (\rightarrow)$$

$$I'' = I_1$$

... (i)

Meshes 1 and 2 will form a supermesh.  
Writing current equation for the supermesh,

$$I_2 - I_1 = 3I'' = 3I_1$$

$$4I_1 - I_2 = 0$$

... (ii)

Applying KVL to the outer path of the supermesh,

$$-6I_1 - 8I_2 + 10 = 0$$

$$6I_1 + 8I_2 = 10$$

Solving Eqs (ii) and (iii),

$$I_1 = 0.26\text{ A}$$

$$I_2 = 1.05\text{ A}$$

$$I'' = I_1 = 0.26\text{ A} (\rightarrow)$$

**Step III** By superposition theorem,

$$I = I' + I'' = 0.39 + 0.26 = 0.65\text{ A}$$

- Q.3 i. Tellegen's Theorem states that for any electrical network, the sum of power across all branches (the sum of products of voltage and current for each branch) is zero.

ii. Answer: The power delivered to the load  $P_L$  is given by:

2

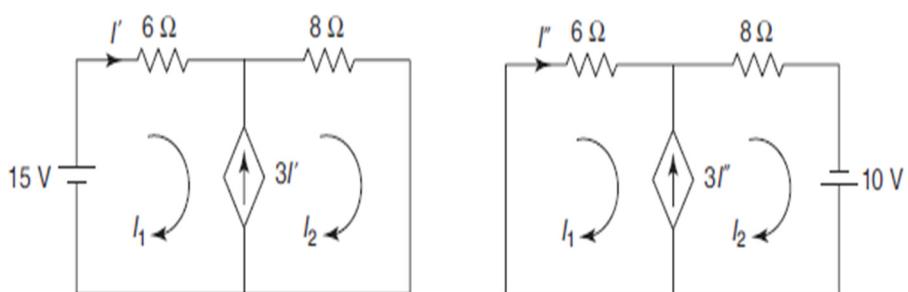
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where  $V_s$  is the source voltage. Simplifying, we get:

$$P_L = \frac{V_s^2 R_L}{(R_s + R_L)^2}$$

To find the maximum power, take the derivative of  $P_L$  with respect to  $R_L$  and set it to zero. This yields  $R_L = R_s$  as the condition for maximum power transfer.

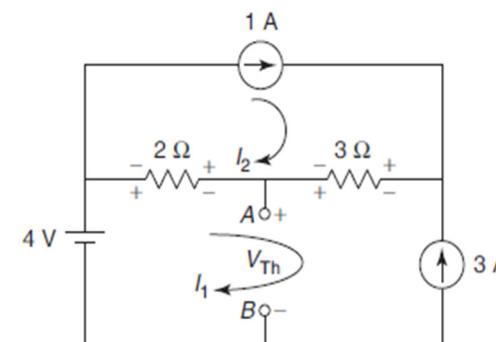
iii.



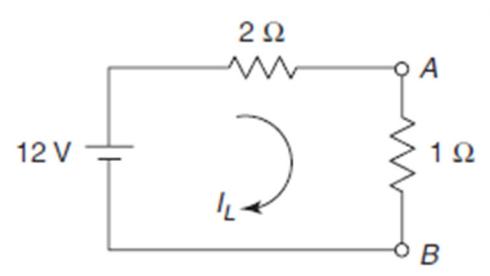
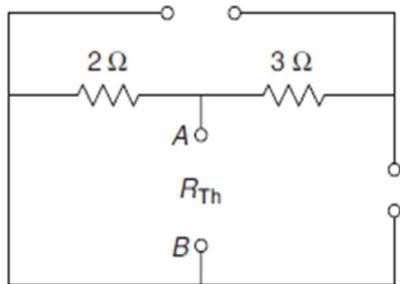
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OR iv.

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[2]

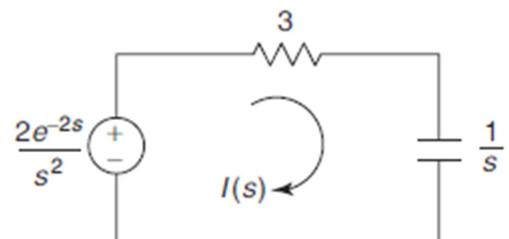


$$I_L = \frac{12}{2+1} = 4 \text{ A}$$

Q.4 i. Answer:

- The **Initial Value Theorem** states:  $\lim_{t \rightarrow 0^+} f(t) = \lim_{s \rightarrow \infty} sF(s)$ . It provides the initial value of the time-domain function without needing to compute the inverse Laplace Transform.
- The **Final Value Theorem** states:  $\lim_{t \rightarrow \infty} f(t) = \lim_{s \rightarrow 0} sF(s)$ . It helps determine the steady-state or long-term behavior of the function. These theorems are useful for quickly finding initial and final values in system analysis.

ii.



[3]

$$\begin{aligned} \frac{2e^{-2s}}{s^2} - 3I(s) - \frac{1}{s}I(s) &= 0 \\ \left(3 + \frac{1}{s}\right)I(s) &= \frac{2e^{-2s}}{s^2} \\ I(s) &= \frac{2e^{-2s}}{s^2 \left(3 + \frac{1}{s}\right)} = \frac{0.67e^{-2s}}{s(s+0.33)} \end{aligned}$$

By partial-fraction expansion,

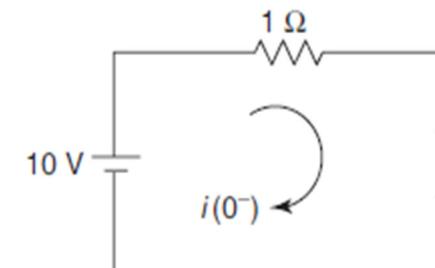
$$\begin{aligned} \frac{0.67}{s(s+0.33)} &= \frac{A}{s} + \frac{B}{s+0.33} \\ A &= \left. \frac{0.67}{s+0.33} \right|_{s=0} = 2 \\ B &= \left. \frac{0.67}{s} \right|_{s=-0.33} = -2 \\ I(s) &= e^{-2s} \left( \frac{2}{s} - \frac{2}{s+0.33} \right) = 2 \frac{e^{-2s}}{s} - 2 \frac{e^{-2s}}{s+0.33} \end{aligned}$$

Taking inverse Laplace transform,

$$i(t) = 2u(t-2) - 2e^{-0.33(t-2)}u(t-2) \quad \text{for } t > 0$$

OR iii.

7



$$i(0^-) = \frac{10}{1} = 10 \text{ A}$$

Since the current through the inductor cannot change instantaneously,

$$i(0^+) = 10 \text{ A}$$

7

P.T.O.

[2]

Applying KVL to the mesh for  $t > 0$ ,

$$\begin{aligned} -I(s) - I(s) - sI(s) + 10 &= 0 \\ I(s)(s+2) &= 10 \\ I(s) &= \frac{10}{s+2} \end{aligned}$$

Taking inverse Laplace transform,

$$i(t) = 10e^{-2t} \quad \text{for } t > 0$$

- Q.5 i.  **$y_{22}$ , the output admittance**, is the ratio of output current  $I_2$  to output voltage  $V_2$  when the input voltage  $V_1$  is zero. It is given by

$$y_{22} = I_2/V_2 \text{ when } V_1 = 0$$

ii.

**Solution** Applying KVL to Mesh 1,

$$\begin{aligned} V_1 &= 2I_1 + 2(I_1 - I_3) \\ &= 4I_1 - 2I_3 \end{aligned}$$

Applying KVL to Mesh 2,

$$\begin{aligned} V_2 &= 2I_2 + 2(I_2 + I_3) \\ &= 4I_2 + 2I_3 \end{aligned}$$

Applying KVL to Mesh 3,

$$-2(I_3 - I_1) - 4I_1 - 2(I_3 + I_2) = 0$$

$$I_1 + I_2 = -2I_3$$

Substituting Eq. (iii) in Eq. (i),

$$\begin{aligned} V_1 &= 4I_1 + I_1 + I_2 \\ &= 5I_1 + I_2 \end{aligned}$$

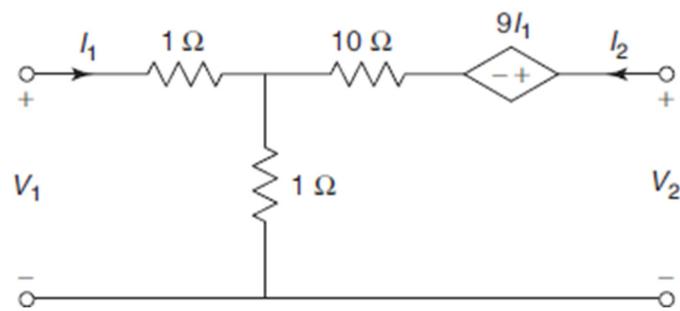
Substituting Eq. (iii) in Eq. (ii),

$$\begin{aligned} V_2 &= 4I_2 - I_1 - I_2 \\ &= -I_1 + 3I_2 \end{aligned}$$

Comparing Eqs (iv) and (v) with Z-parameter equations,

$$\begin{bmatrix} Z_{11} & Z_{12} \\ Z_{21} & Z_{22} \end{bmatrix} = \begin{bmatrix} 5 & 1 \\ -1 & 3 \end{bmatrix}$$

OR iii.



8

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[3]

$$V_1 = 2I_1 + I_2 \quad \dots(i)$$

Applying KVL to Mesh 2,

$$\begin{aligned} V_2 &= 9I_1 + 10I_2 + I(I_1 + I_2) \\ &= 10I_1 + 11I_2 \end{aligned} \quad \dots(ii)$$

Comparing Eqs (i) and (ii) with Z-parameter equations,

$$\begin{bmatrix} Z_{11} & Z_{12} \\ Z_{21} & Z_{22} \end{bmatrix} = \begin{bmatrix} 2 & 1 \\ 10 & 11 \end{bmatrix}$$

**$h$ -parameters**

$$\begin{aligned} h_{11} &= \frac{\Delta Z}{Z_{22}} = \frac{12}{11} \Omega, & h_{12} &= \frac{Z_{12}}{Z_{22}} = \frac{1}{11} \\ h_{21} &= -\frac{Z_{21}}{Z_{22}} = -\frac{10}{11}, & h_{22} &= \frac{1}{Z_{22}} = \frac{1}{11} \Omega \end{aligned}$$

Hence,  $h$ -parameters are

$$\begin{bmatrix} h_{11} & h_{12} \\ h_{21} & h_{22} \end{bmatrix} = \begin{bmatrix} \frac{12}{11} & \frac{1}{11} \\ -\frac{10}{11} & \frac{1}{11} \end{bmatrix}$$

- Q.6 i. Pole-zero analysis involves examining the poles and zeros of a system's transfer function. For a system to be stable, all poles must be in the left half of the complex plane (for continuous systems), indicating that system responses decay over time without oscillations.

2

ii.

**Solution**

**Foster I Form** The Foster I form is obtained by the partial-fraction expansion of the impedance function  $Z(s)$ .

By partial-fraction expansion,

$$Z(s) = \frac{K_0}{s} + \frac{K_1}{s+2} + \frac{K_2}{s+4}$$

where

$$K_0 = sZ(s)|_{s=0} = \frac{(1)(3)}{(2)(4)} = \frac{3}{8}$$

$$K_1 = (s+2)Z(s)|_{s=-2} = \frac{(-2+1)(-2+3)}{(-2)(-2+4)} = \frac{(-1)(1)}{(-2)(2)} = \frac{1}{4}$$

$$K_2 = (s+4)Z(s)|_{s=-4} = \frac{(-4+1)(-4+3)}{(-4)(-4+2)} = \frac{(-3)(-1)}{(-4)(-2)} = \frac{3}{8}$$

$$Z(s) = \frac{\frac{3}{8}}{s} + \frac{\frac{1}{4}}{s+2} + \frac{\frac{3}{8}}{s+4}$$

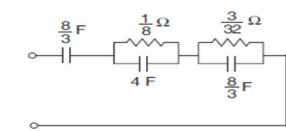
The first term represents the impedance of a capacitor of  $\frac{8}{3}$  F. The remaining terms represent the impedance of a parallel RC circuit for which

$$Z_{RC}(s) = \frac{\frac{1}{C_i}}{s + \frac{1}{R_i C_i}}$$

By direct comparison,

$$R_1 = \frac{1}{8} \Omega, \quad C_1 = 4 \text{ F}$$

$$R_2 = \frac{3}{32} \Omega, \quad C_2 = \frac{8}{3} \text{ F}$$



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[2]

OR iii.

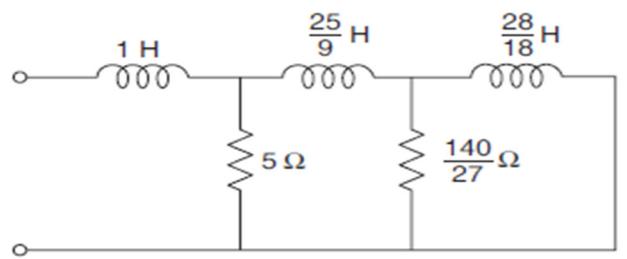
**Cauer I Form** The Cauer I form is obtained by continued fraction expansion of  $Z(s)$  about the pole at infinity.

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$$Z(s) = \frac{s^3 + 12s^2 + 32s}{s^2 + 7s + 6}$$

By continued fraction expansion,

$$\begin{aligned} & s^2 + 7s + 6 \Big) s^3 + 12s^2 + 32s \left( s \leftarrow Z \right. \\ & \quad \frac{s^3 + 7s^2 + 6s}{5s^2 + 26s} \Big) s^2 + 7s + 6 \left( \frac{1}{5} \leftarrow Y \right. \\ & \quad \frac{s^2 + \frac{26}{5}s}{\frac{9}{5}s + 6} \Big) 5s^2 + 26s \left( \frac{25}{9}s \leftarrow Z \right. \\ & \quad \frac{5s^2 + \frac{50}{3}s}{\frac{28}{3}s} \Big) \frac{9}{5}s + 6 \left( \frac{27}{140} \leftarrow Y \right. \\ & \quad \frac{\frac{9}{5}s}{0} \Big) \frac{28}{3}s \left( \frac{28}{18}s \leftarrow Z \right. \\ & \quad \frac{\frac{28}{3}s}{0} \end{aligned}$$



[3]

P.T.O.