

[4]

- ii. For the network shown in figure 8, determine the current $i(t)$ when the switch is closed at $t = 0$ with zero initial conditions.

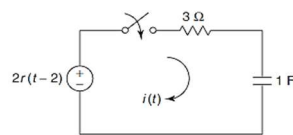


Figure 8

- OR iii. In the network of figure 9, the switch is moved from the position 1 to 2 at $t = 0$, steady-state condition having been established in the position 1. Determine $i(t)$ for $t > 0$.

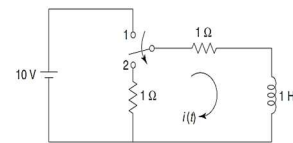


Figure 9

- Q.5 i. What does y_{22} represent in Y-parameters? **2** 02 01 05 01
 ii. Find Z-parameters for the network shown in figure 10. **8** 03 02 05 1

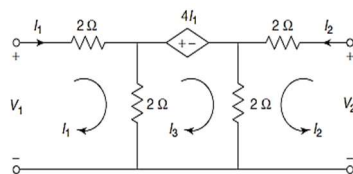


Figure 10

- OR iii. Find h-parameters for the network shown in figure 11. **8** 03 02 05 01

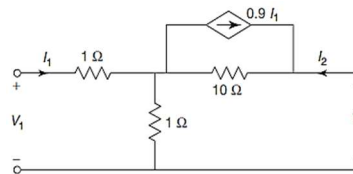


Figure 11

- Q.6 i. What is the significance of pole-zero analysis in determining the stability of a system? **2** 02 01 05 01
 ii. Determine the Foster-I form realisation of the RC impedance function-

$$Z(s) = \frac{(s+1)(s+3)}{s(s+2)(s+4)}$$

- OR iii. Determine the Cauer-I form of the RL impedance function-

$$Z(s) = \frac{s(s+4)(s+8)}{(s+1)(s+6)}$$

Total No. of Questions: 6

Total No. of Printed Pages: 4

Enrollment No.....



Faculty of Engineering
 End Sem Examination Dec 2024
 EC3CO05 Circuit Analysis & Synthesis

Programme: B.Tech.

Branch/Specialisation: EC

Duration: 3 Hrs.

Maximum Marks: 60

Note: All questions are compulsory. Internal choices, if any, are indicated. Answers of Q.1 (MCQs) should be written in full instead of only a, b, c or d. Assume suitable data if necessary. Notations and symbols have their usual meaning.

- | | Marks | BL | PO | CO | PSO |
|--|-------|----|----|----|-----|
| Q.1 i. A dependent (controlled) source in a circuit depends on:
(a) The value of another voltage or current in the circuit
(b) The internal resistance of the source
(c) Only the input power supply
(d) The type of circuit configuration | 1 | 01 | 01 | 01 | 01 |
| ii. In mesh analysis, which of the following is primarily calculated?
(a) Node voltages (b) Branch voltages
(c) Loop currents (d) Equivalent resistance | 1 | 02 | 01 | 01 | 01 |
| iii. In the network shown in figure 1, the switch is closed at $t=0$. With the capacitor uncharged, value of current i at $t=0+$. | 1 | 03 | 02 | 02 | 01 |

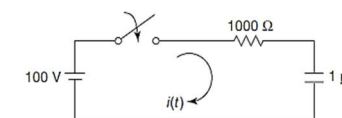


Figure 1

- | | | | | | |
|---|---|----|----|----|----|
| (a) 0 A (b) 0.1 A (c) 100 A (d) 1000 A | | | | | |
| iv. The maximum power transfer theorem states that maximum power is transferred from a source to a load when:
(a) The load resistance is maximum
(b) The load resistance equals the source resistance
(c) The load resistance is zero
(d) The load resistance is double the source resistance | 1 | 02 | 01 | 03 | 01 |

- [2]
- v. Which property of the Laplace transform states that $L[f(t-T)u(t-T)] = e^{-Ts}F(s)$?
 (a) Linearity (b) Convolution
 (c) Frequency Shifting (d) Time Shifting
- vi. The Laplace transform of a unit impulse function $\delta(t-T)$ is-
 (a) e^{-Ts} (b) $s e^{-Ts}$ (c) $T e^{-Ts}$ (d) $s^T e^{-Ts}$
- vii. The value of Z_{11} for the network shown in figure 2-

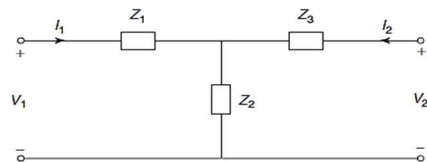


Figure 2

- (a) Z_2+Z_3 (b) Z_1 (c) Z_1+Z_2 (d) Z_2
- viii. In a two-port network with ABCD parameters, the condition of symmetry is given by:
 (a) $A=D$ (b) $B=C$ (c) $AD-BC=0$ (d) $A=B$
- ix. A Hurwitz polynomial is defined as a polynomial where:
 (a) All coefficients are positive
 (b) All roots lie in the right-half of the s-plane
 (c) All roots lie in the left-half of the s-plane
 (d) All roots are real and positive
- x. What is the primary application of the Foster-I form in network synthesis?
 (a) To design active filters
 (b) To ensure stability of the network
 (c) To create low-pass and high-pass filters
 (d) To achieve the desired impedance characteristics

- Q.2 i. What is source transformation in circuit analysis? 2 02 01 01 01
 ii. Find the current in the 5Ω resistor of the network shown in figure 3. 3 03 02 02 01

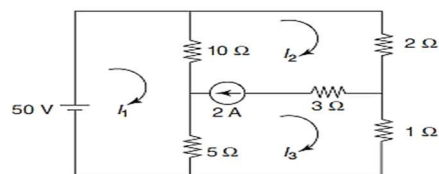


Figure 3

- [3]
- iii. For the circuit shown in figure 4, draw its graph and write the fundamental cutset matrix. 5 04 02 01 02

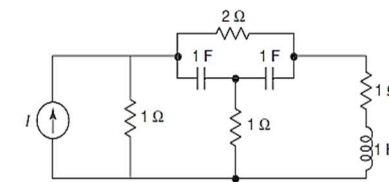


Figure 4

- OR iv. For the circuit shown in figure 5, draw its graph and write the tieset matrix. 5 04 02 01 02

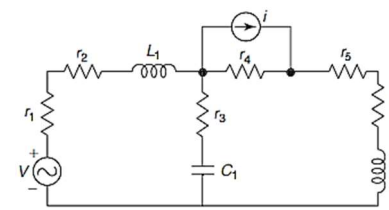


Figure 5

- Q.3 i. Define Tellegen's Theorem. 2 02 01 03 01
 ii. Derive the expression for the power delivered to the load in terms of load and source resistance and use it to find the condition for maximum power transfer. 3 03 01, 02 03 01
 iii. Use Superposition theorem to find the current through the 6Ω resistor shown in figure 6. 5 03 02 02 01

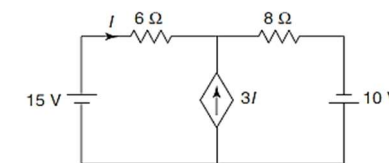


Figure 6

- OR iv. Use Thevenin's theorem to find the current through the 1Ω resistor shown in figure 7. 5 03 02 02 01

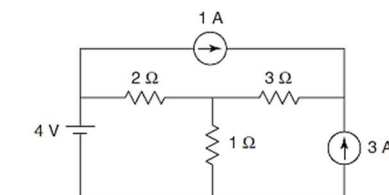


Figure 7

- Q.4 i. State the initial value theorem and the final value theorem in Laplace transforms and their significance. 3 02 01 04 01

Marking Scheme

EC3CO05 (T) Circuit Analysis & Synthesis (T)

- Q.1 i) A dependent (controlled) source in a circuit depends on: 1
 a) The value of another voltage or current in the circuit.
 ii) In mesh analysis, which of the following is primarily calculated? 1
 c) Loop currents
 iii) In the network shown in Figure 1, the switch is closed at $t=0$. With the capacitor uncharged, value of current i at $t=0+$. 1

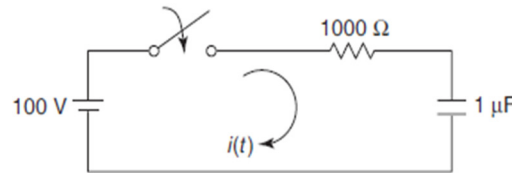


Figure 1

- b) 0.1 A
 iv) **The Maximum Power Transfer Theorem states that maximum power is transferred from a source to a load when:** 1
 b) The load resistance equals the source resistance.
 v) Which property of the Laplace transform states that $L[f(t-T)u(t-T)] = e^{-Ts}F(s)$? 1
 d) Time Shifting
 vi) The Laplace transform of a unit impulse function $\delta(t-T)$ is 1
 a) e^{-Ts}
 vii) The value of Z_{11} for the network shown in Figure 2. 1

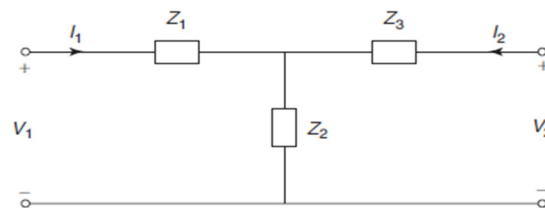


Figure 2

- c) $Z_1 + Z_2$
 viii) **In a two-port network with ABCD parameters, the condition of symmetry is given by:** 1
 a) $A=D$
 ix) **A Hurwitz polynomial is defined as a polynomial where:** 1
 c) All roots lie in the left-half of the s-plane.
 x) **What is the primary application of the Foster-I form in network** 1

synthesis?

d) To achieve the desired impedance characteristics.

- Q.2 i. Source transformation is the process of converting a voltage source in series with a resistor into an equivalent current source in parallel with the same resistor, and vice versa. It simplifies circuit analysis. 2

- ii. **Solution** Applying KVL to Mesh 1,
 $50 - 10(I_1 - I_2) - 5(I_1 - I_3) = 0$
 $15I_1 - 10I_2 - 5I_3 = 50$...(i)

Mesheres 2 and 3 will form a supermesh as these two meshes share a common current source of 2 A.
 Writing current equation for the supermesh,

$$I_2 - I_3 = 2 \quad \text{...(ii)}$$

Applying KVL to the outer path of the supermesh,

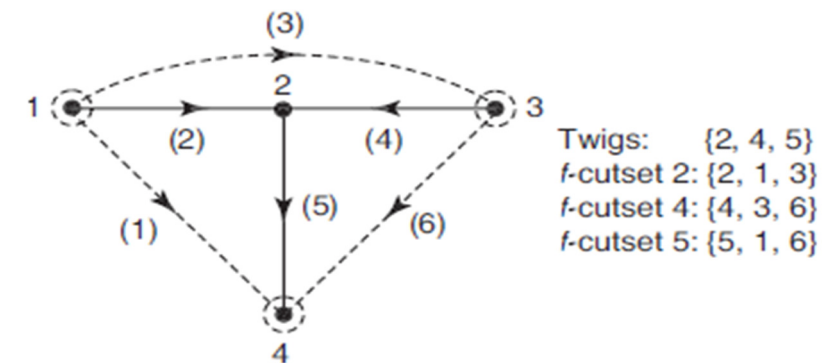
$$\begin{aligned} -10(I_2 - I_1) - 2I_2 - 1I_3 - 5(I_3 - I_1) &= 0 \\ -15I_1 + 12I_2 + 6I_3 &= 0 \quad \text{...(iii)} \end{aligned}$$

Solving Eqs (i), (ii) and (iii),

$$\begin{aligned} I_1 &= 20 \text{ A} \\ I_2 &= 17.33 \text{ A} \\ I_3 &= 15.33 \text{ A} \end{aligned}$$

Current through the 5Ω resistor $= I_1 - I_3 = 20 - 15.33 = 4.67 \text{ A}$

- iii. 5

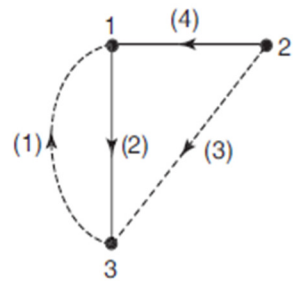
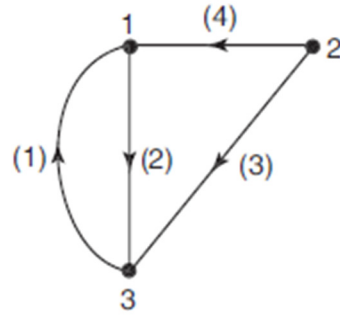


Fundamental Cutset Matrix (Q)

$$Q = \begin{matrix} & \begin{matrix} 1 & 2 & 3 & 4 & 5 & 6 \end{matrix} \\ \begin{matrix} 2 \\ 4 \\ 5 \end{matrix} & \begin{bmatrix} 1 & 1 & 1 & 0 & 0 & 0 \\ 0 & 0 & -1 & 1 & 0 & 1 \\ 1 & 0 & 0 & 0 & 1 & 1 \end{bmatrix} \end{matrix}$$

OR iv.

[2]



Links: {1, 3}
Tieset 1: {1, 2}
Tieset 3: {3, 2, 4}

$$B = \frac{1}{3} \begin{bmatrix} 1 & 1 & 0 & 0 \\ 0 & -1 & 1 & -1 \end{bmatrix}$$

5

Q.3 i. Tellegen's Theorem states that for any electrical network, the sum of power across all branches (the sum of products of voltage and current for each branch) is zero.

2

ii. Answer: The power delivered to the load P_L is given by:

3

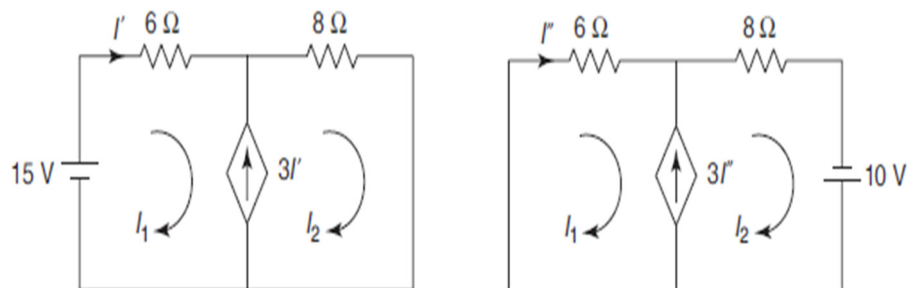
$$P_L = I^2 R_L = \left(\frac{V_s}{R_s + R_L} \right)^2 R_L$$

where V_s is the source voltage. Simplifying, we get:

$$P_L = \frac{V_s^2 R_L}{(R_s + R_L)^2}$$

To find the maximum power, take the derivative of P_L with respect to R_L and set it to zero. This yields $R_L = R_s$ as the condition for maximum power transfer.

iii.



5

[3]

$$I' = I_1 \quad \dots(i)$$

Mesheres 1 and 2 will form a supermesh.
Writing current equation for the supermesh,

$$I_2 - I_1 = 3I' = 3I_1$$

$$4I_1 - I_2 = 0 \quad \dots(ii)$$

Applying KVL to the outer path of the supermesh,

$$15 - 6I_1 - 8I_2 = 0$$

$$6I_1 + 8I_2 = 15$$

Solving Eqs (ii) and (iii),

$$I_1 = 0.39A$$

$$I_2 = 1.57A$$

$$I' = I_1 = 0.39A \quad (\rightarrow)$$

$$I'' = I_1 \quad \dots(i)$$

Mesheres 1 and 2 will form a supermesh.
Writing current equation for the supermesh,

$$I_2 - I_1 = 3I'' = 3I_1$$

$$4I_1 - I_2 = 0 \quad \dots(ii)$$

Applying KVL to the outer path of the supermesh,

$$-6I_1 - 8I_2 + 10 = 0$$

$$6I_1 + 8I_2 = 10$$

Solving Eqs (ii) and (iii),

$$I_1 = 0.26A$$

$$I_2 = 1.05A$$

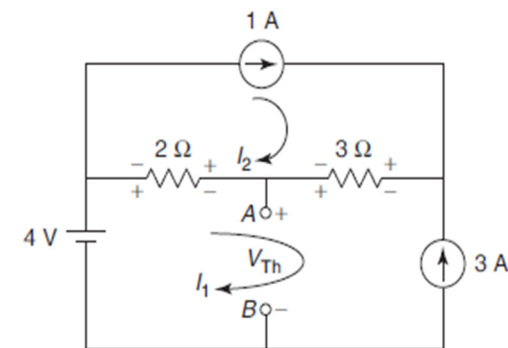
$$I'' = I_1 = 0.26A \quad (\rightarrow)$$

Step III By superposition theorem,

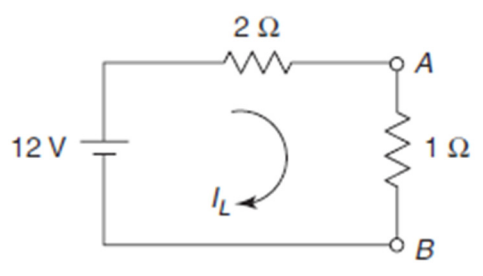
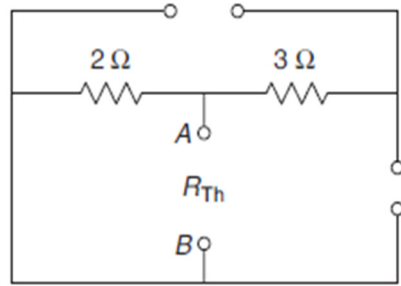
$$I = I' + I'' = 0.39 + 0.26 = 0.65A$$

OR iv.

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[2]



$$I_L = \frac{12}{2+1} = 4 \text{ A}$$

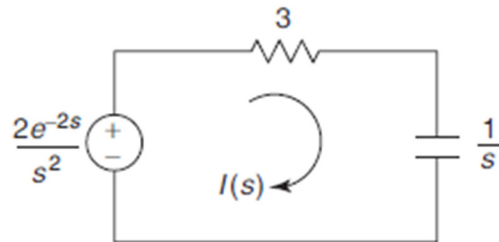
Q.4 i.

Answer:

- The **Initial Value Theorem** states: $\lim_{t \rightarrow 0^+} f(t) = \lim_{s \rightarrow \infty} sF(s)$. It provides the initial value of the time-domain function without needing to compute the inverse Laplace Transform.
- The **Final Value Theorem** states: $\lim_{t \rightarrow \infty} f(t) = \lim_{s \rightarrow 0} sF(s)$. It helps determine the steady-state or long-term behavior of the function. These theorems are useful for quickly finding initial and final values in system analysis.

3

ii.



7

[3]

$$\begin{aligned} \frac{2e^{-2s}}{s^2} - 3I(s) - \frac{1}{s}I(s) &= 0 \\ \left(3 + \frac{1}{s}\right)I(s) &= \frac{2e^{-2s}}{s^2} \\ I(s) &= \frac{2e^{-2s}}{s^2\left(3 + \frac{1}{s}\right)} = \frac{0.67e^{-2s}}{s(s+0.33)} \end{aligned}$$

By partial-fraction expansion,

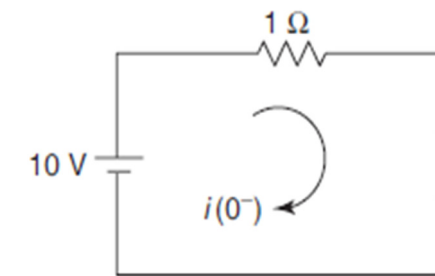
$$\begin{aligned} \frac{0.67}{s(s+0.33)} &= \frac{A}{s} + \frac{B}{s+0.33} \\ A &= \left. \frac{0.67}{s+0.33} \right|_{s=0} = 2 \\ B &= \left. \frac{0.67}{s} \right|_{s=-0.33} = -2 \\ I(s) &= e^{-2s} \left(\frac{2}{s} - \frac{2}{s+0.33} \right) = 2 \frac{e^{-2s}}{s} - 2 \frac{e^{-2s}}{s+0.33} \end{aligned}$$

Taking inverse Laplace transform,

$$i(t) = 2u(t-2) - 2e^{-0.33(t-2)}u(t-2) \quad \text{for } t > 0$$

OR iii.

7



$$i(0^-) = \frac{10}{1} = 10 \text{ A}$$

Since the current through the inductor cannot change instantaneously,

$$i(0^+) = 10 \text{ A}$$

[2]

Applying KVL to the mesh for $t > 0$,

$$\begin{aligned} -I(s) - I(s) - sI(s) + 10 &= 0 \\ I(s)(s+2) &= 10 \\ I(s) &= \frac{10}{s+2} \end{aligned}$$

Taking inverse Laplace transform,

$$i(t) = 10e^{-2t} \quad \text{for } t > 0$$

- Q.5 i. **y_{22} , the output admittance**, is the ratio of output current I_2 to output voltage V_2 when the input voltage V_1 is zero. It is given by $y_{22} = I_2/V_2$ when $V_1=0$ 2
- ii. 8

Solution Applying KVL to Mesh 1,

$$\begin{aligned} V_1 &= 2I_1 + 2(I_1 - I_3) \\ &= 4I_1 - 2I_3 \end{aligned}$$

Applying KVL to Mesh 2,

$$\begin{aligned} V_2 &= 2I_2 + 2(I_2 + I_3) \\ &= 4I_2 + 2I_3 \end{aligned}$$

Applying KVL to Mesh 3,

$$\begin{aligned} -2(I_3 - I_1) - 4I_1 - 2(I_3 + I_2) &= 0 \\ I_1 + I_2 &= -2I_3 \end{aligned}$$

Substituting Eq. (iii) in Eq. (i),

$$\begin{aligned} V_1 &= 4I_1 + I_1 + I_2 \\ &= 5I_1 + I_2 \end{aligned}$$

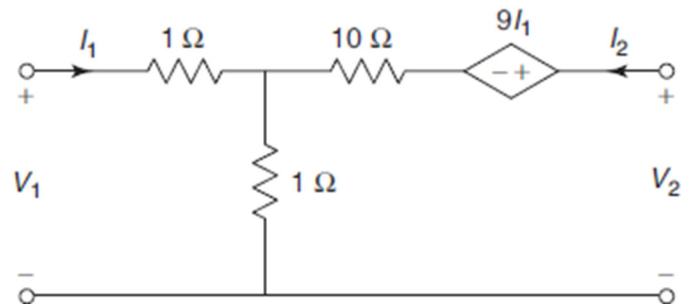
Substituting Eq. (iii) in Eq. (ii),

$$\begin{aligned} V_2 &= 4I_2 - I_1 - I_2 \\ &= -I_1 + 3I_2 \end{aligned}$$

Comparing Eqs (iv) and (v) with Z-parameter equations,

$$\begin{bmatrix} Z_{11} & Z_{12} \\ Z_{21} & Z_{22} \end{bmatrix} = \begin{bmatrix} 5 & 1 \\ -1 & 3 \end{bmatrix}$$

OR iii.



8

[3]

$$V_1 = 2I_1 + I_2 \quad \dots(i)$$

Applying KVL to Mesh 2,

$$\begin{aligned} V_2 &= 9I_1 + 10I_2 + 1(I_1 + I_2) \\ &= 10I_1 + 11I_2 \end{aligned} \quad \dots(ii)$$

Comparing Eqs (i) and (ii) with Z-parameter equations,

$$\begin{bmatrix} Z_{11} & Z_{12} \\ Z_{21} & Z_{22} \end{bmatrix} = \begin{bmatrix} 2 & 1 \\ 10 & 11 \end{bmatrix}$$

h-parameters

$$\begin{aligned} h_{11} &= \frac{\Delta Z}{Z_{22}} = \frac{12}{11} \Omega, & h_{12} &= \frac{Z_{12}}{Z_{22}} = \frac{1}{11} \\ h_{21} &= -\frac{Z_{21}}{Z_{22}} = -\frac{10}{11}, & h_{22} &= \frac{1}{Z_{22}} = \frac{1}{11} \text{ S} \end{aligned}$$

Hence, h-parameters are

$$\begin{bmatrix} h_{11} & h_{12} \\ h_{21} & h_{22} \end{bmatrix} = \begin{bmatrix} \frac{12}{11} & \frac{1}{11} \\ -\frac{10}{11} & \frac{1}{11} \end{bmatrix}$$

- Q.6 i. Pole-zero analysis involves examining the poles and zeros of a system's transfer function. For a system to be stable, all poles must be in the left half of the complex plane (for continuous systems), indicating that system responses decay over time without oscillations. 2
- ii. 8

Solution

Foster I Form The Foster I form is obtained by the partial-fraction expansion of the impedance function $Z(s)$.

By partial-fraction expansion,

$$Z(s) = \frac{K_0}{s} + \frac{K_1}{s+2} + \frac{K_2}{s+4}$$

where

$$K_0 = sZ(s)|_{s=0} = \frac{(1)(3)}{(2)(4)} = \frac{3}{8}$$

$$K_1 = (s+2)Z(s)|_{s=-2} = \frac{(-2+1)(-2+3)}{(-2)(-2+4)} = \frac{(-1)(1)}{(-2)(2)} = \frac{1}{4}$$

$$K_2 = (s+4)Z(s)|_{s=-4} = \frac{(-4+1)(-4+3)}{(-4)(-4+2)} = \frac{(-3)(-1)}{(-4)(-2)} = \frac{3}{8}$$

$$Z(s) = \frac{3}{8s} + \frac{1}{4(s+2)} + \frac{3}{8(s+4)}$$

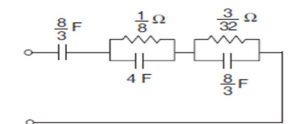
The first term represents the impedance of a capacitor of $\frac{8}{3}$ F. The remaining terms represent the impedance of a parallel RC circuit for which

$$Z_{RC}(s) = \frac{\frac{1}{C_i}}{s + \frac{1}{R_i C_i}}$$

By direct comparison,

$$R_1 = \frac{1}{8} \Omega, \quad C_1 = 4 \text{ F}$$

$$R_2 = \frac{3}{32} \Omega, \quad C_2 = \frac{8}{3} \text{ F}$$



P.T.O.

[2]

[3]

OR iii. **Cauer I Form** The Cauer I form is obtained by continued fraction expansion of $Z(s)$ about the pole at infinity. **8**

$$Z(s) = \frac{s^3 + 12s^2 + 32s}{s^2 + 7s + 6}$$

By continued fraction expansion,

$$\begin{aligned} & s^2 + 7s + 6 \Big) s^3 + 12s^2 + 32s \Big(s \leftarrow Z \\ & \quad \underline{s^3 + 7s^2 + 6s} \\ & \quad \quad 5s^2 + 26s \Big) s^2 + 7s + 6 \Big(\frac{1}{5} \leftarrow Y \\ & \quad \quad \quad \underline{s^2 + \frac{26}{5}s} \\ & \quad \quad \quad \quad \frac{9}{5}s + 6 \Big) 5s^2 + 26s \Big(\frac{25}{9}s \leftarrow Z \\ & \quad \quad \quad \quad \quad \underline{5s^2 + \frac{50}{3}s} \\ & \quad \quad \quad \quad \quad \quad \frac{28}{3}s \Big) \frac{9}{5}s + 6 \Big(\frac{27}{140} \leftarrow Y \\ & \quad \quad \quad \quad \quad \quad \quad \underline{\frac{9}{5}s} \\ & \quad \quad \quad \quad \quad \quad \quad \quad 6 \Big) \frac{28}{3}s \Big(\frac{28}{18}s \leftarrow Z \\ & \quad \quad \quad \quad \quad \quad \quad \quad \quad \underline{\frac{28}{3}s} \\ & \quad \quad \quad \quad \quad \quad \quad \quad \quad \quad 0 \end{aligned}$$

