

Reduction of Multiple Subsystems

(Nise's Textbook Ch. 4)

Components used in Block Diagram Representations

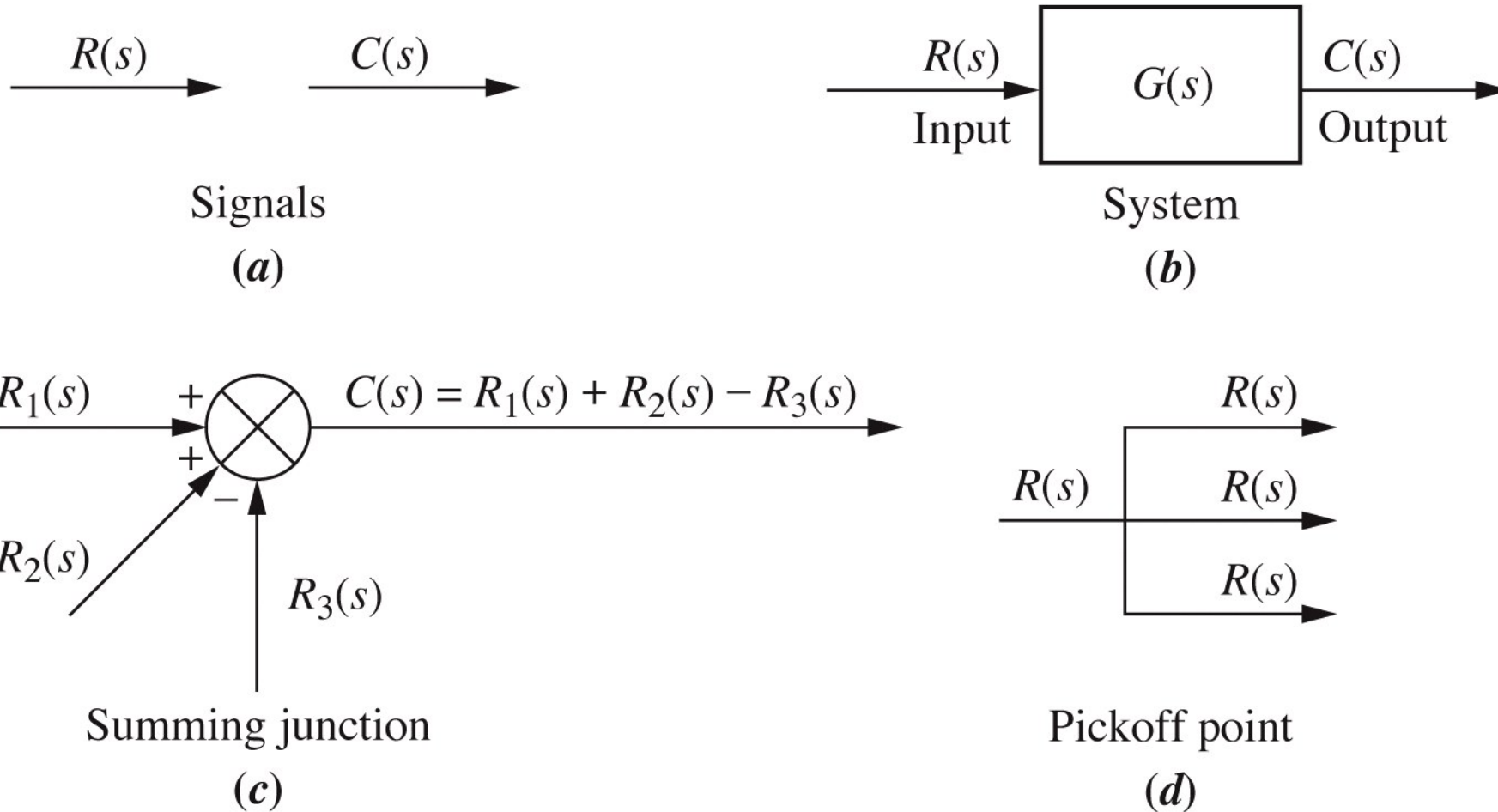


Figure 5.2
© John Wiley & Sons, Inc. All rights reserved.

Cascade Connection

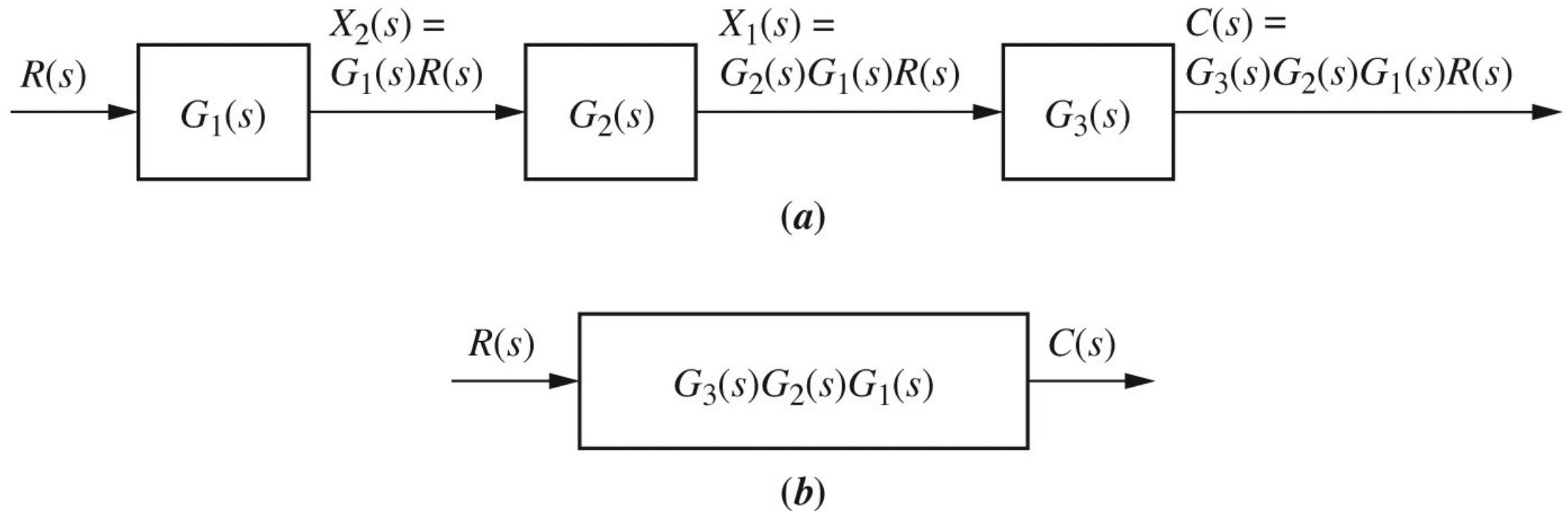
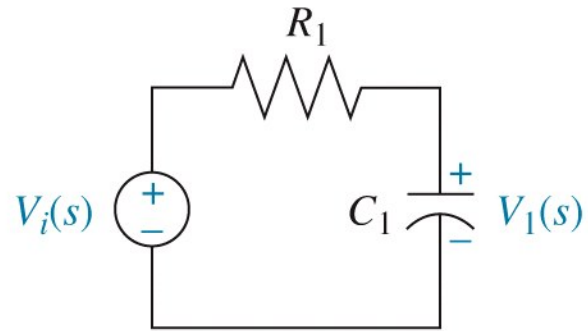


Figure 5.3

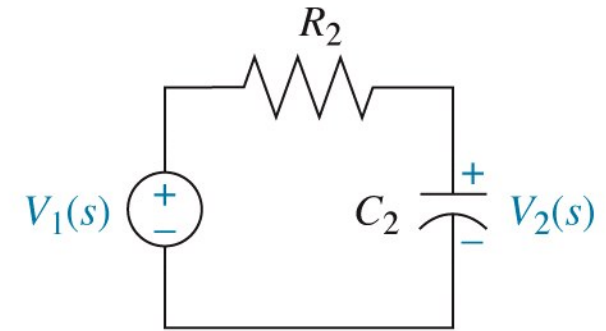
© John Wiley & Sons, Inc. All rights reserved.

Unloading Issue Explained



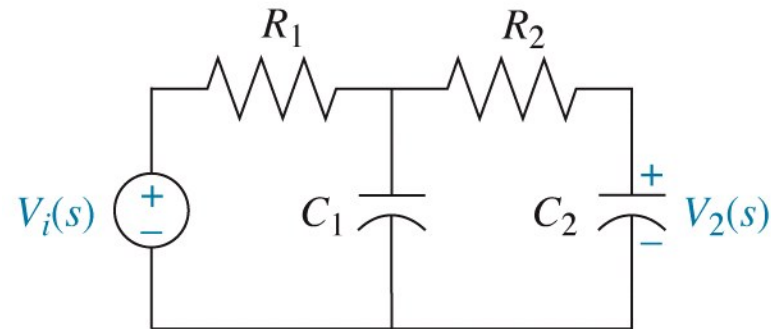
$$G_1(s) = \frac{V_1(s)}{V_i(s)}$$

(a)



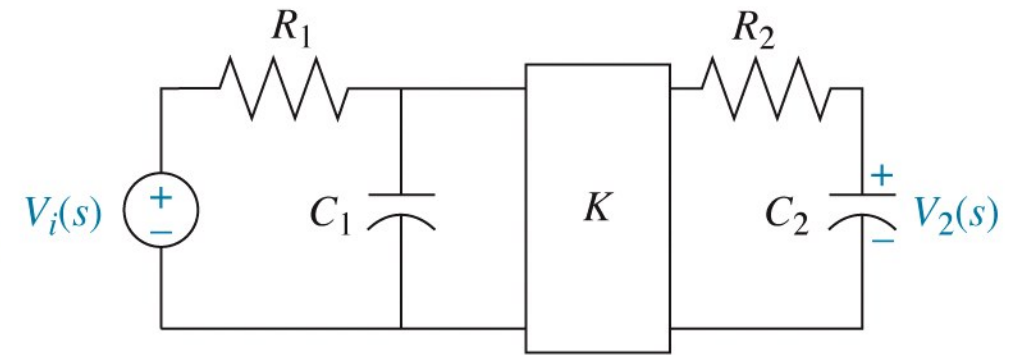
$$G_2(s) = \frac{V_2(s)}{V_1(s)}$$

(b)



$$G_T(s) = \frac{V_2(s)}{V_i(s)} \neq G_2(s)G_1(s)$$

(c)



$$G_T(s) = \frac{V_2(s)}{V_i(s)} = KG_2(s)G_1(s)$$

(d)

Figure 5.4
© John Wiley & Sons, Inc. All rights reserved.

Parallel Connection

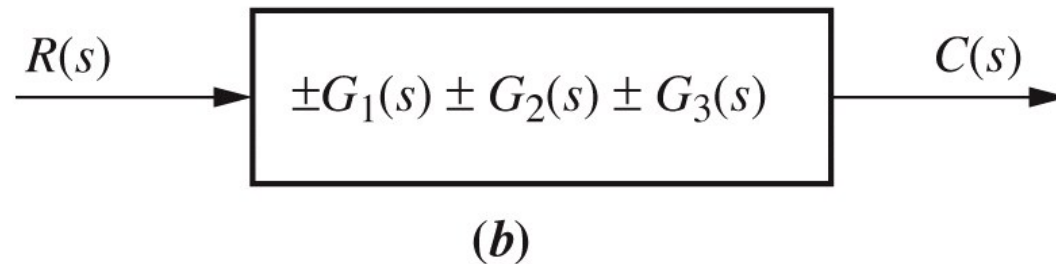
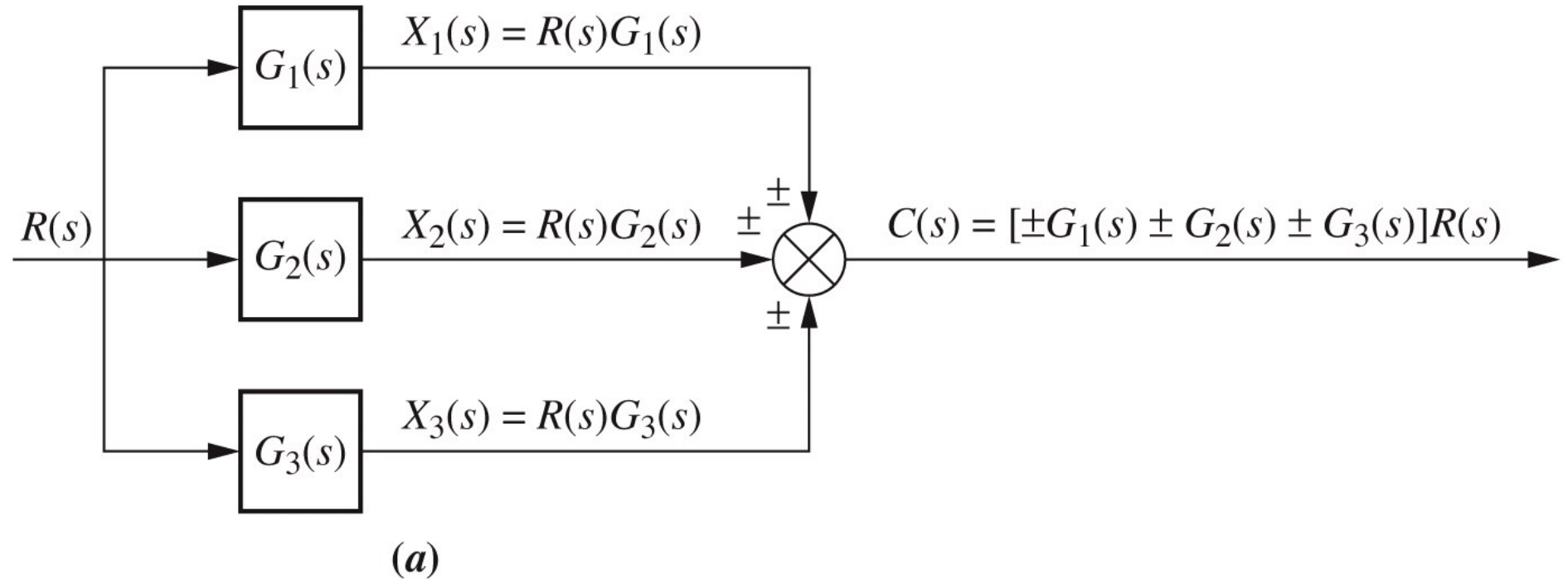
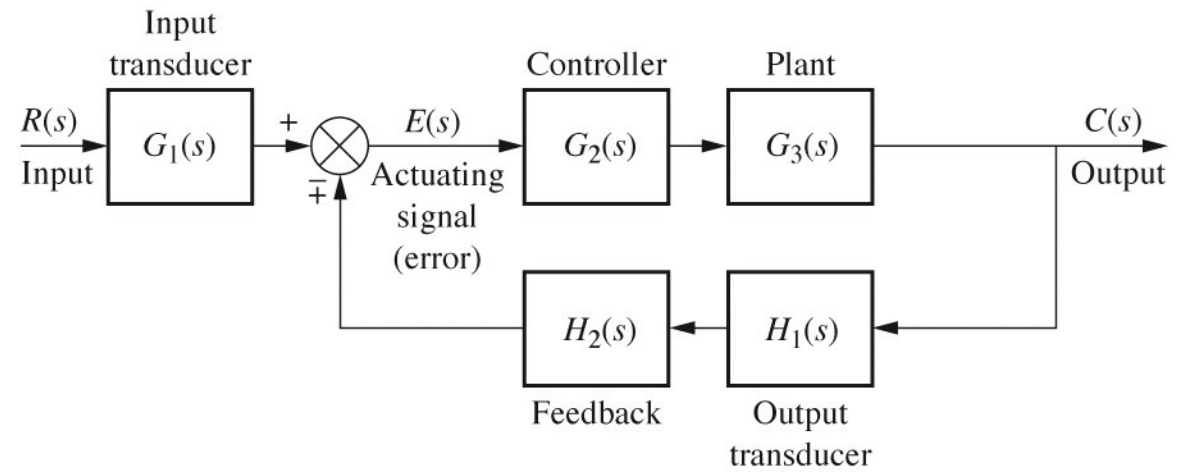
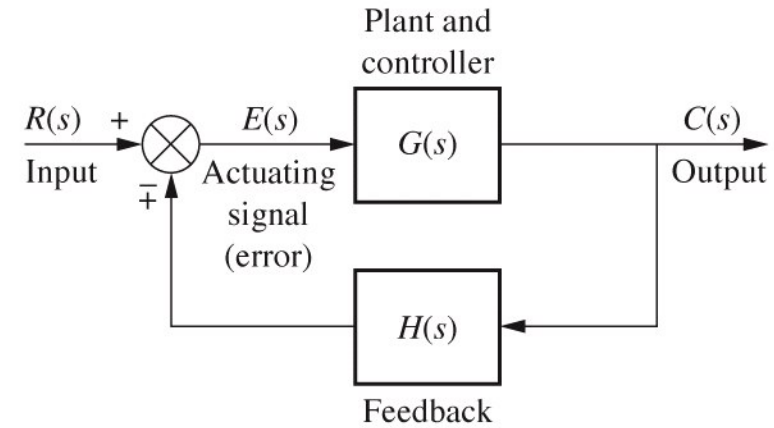


Figure 5.5
© John Wiley & Sons, Inc. All rights reserved.

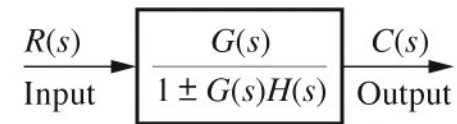
Feedback Connection



(a)



(b)

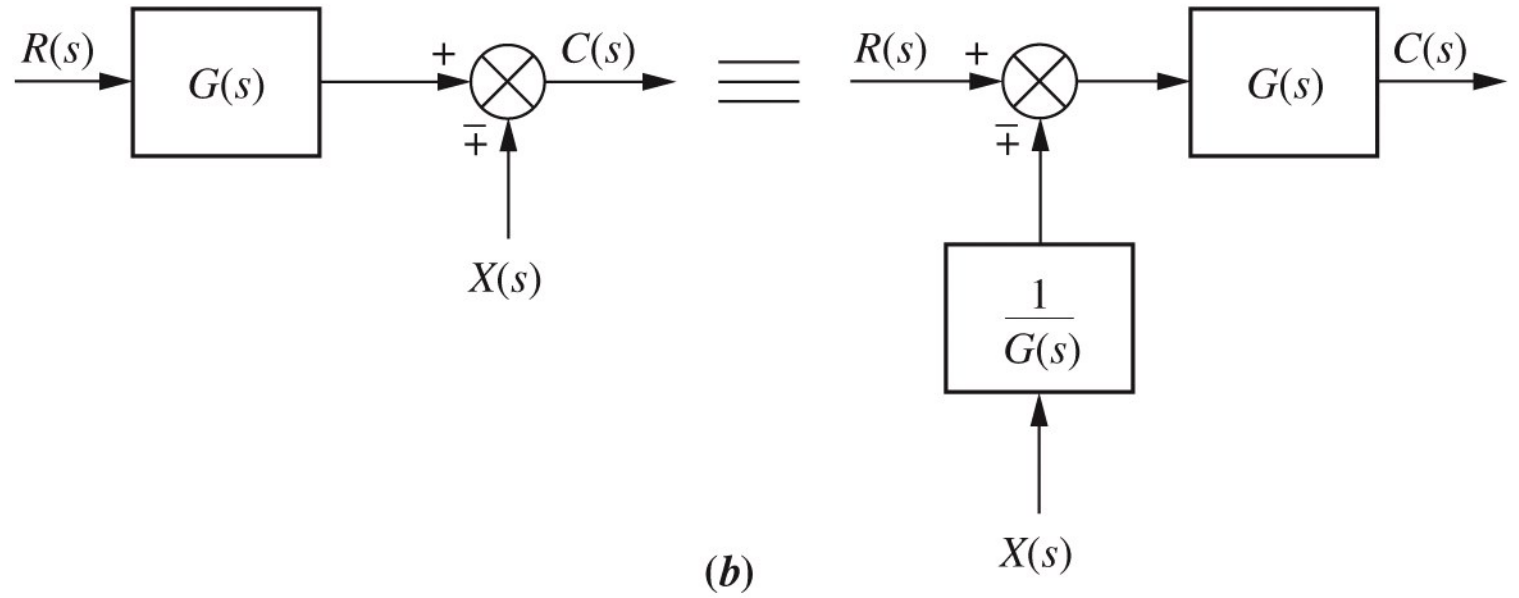
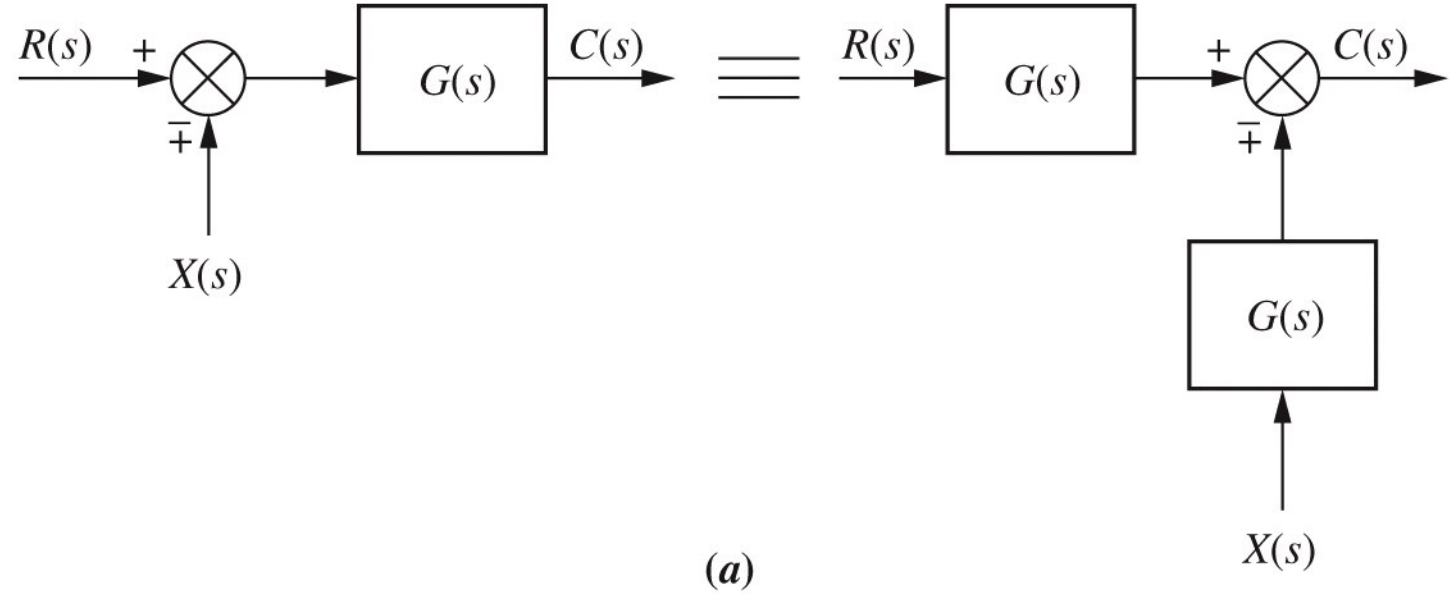


(c)

Figure 5.6

© John Wiley & Sons, Inc. All rights reserved.

Moving Blocks to get Familiar Forms-1



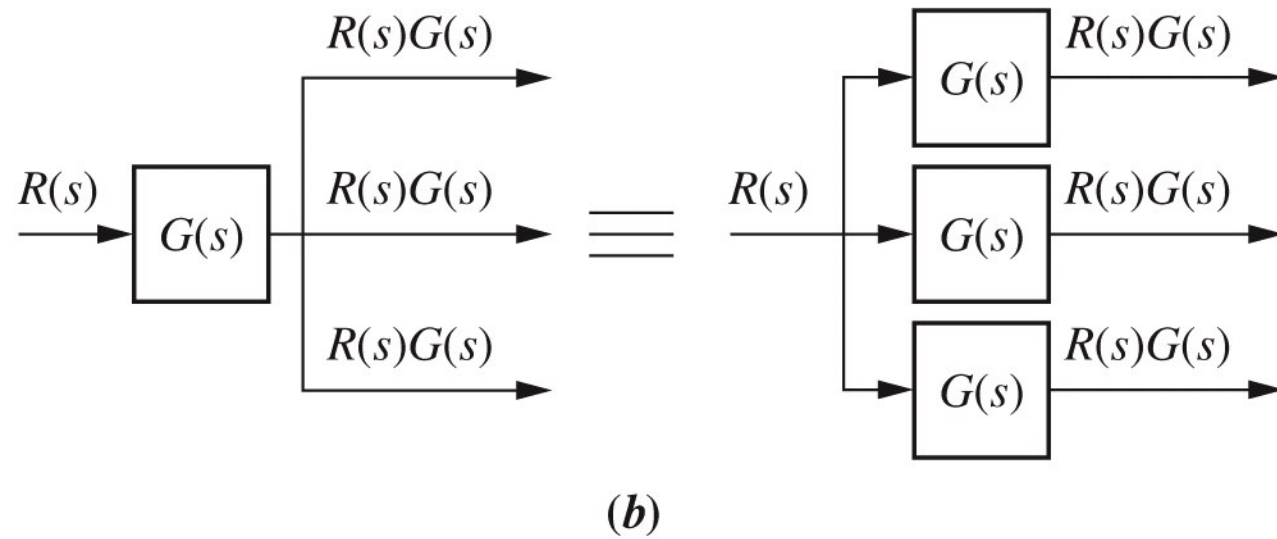
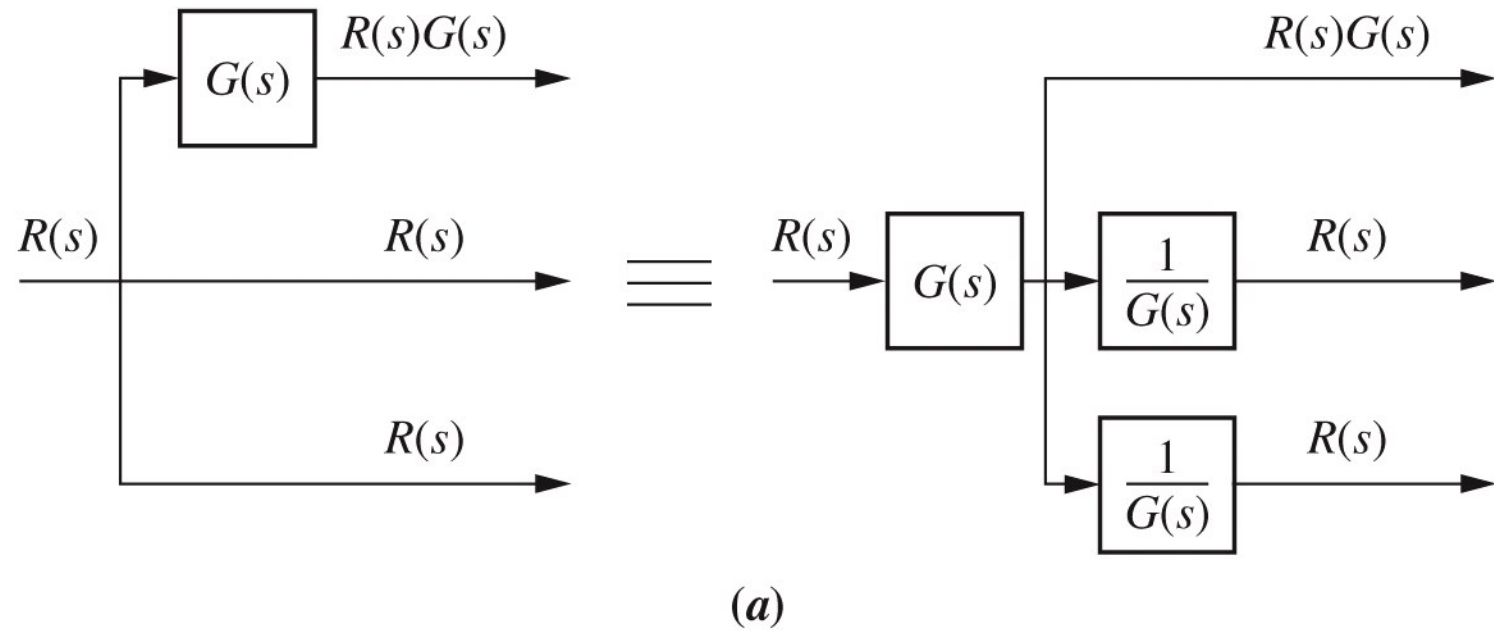
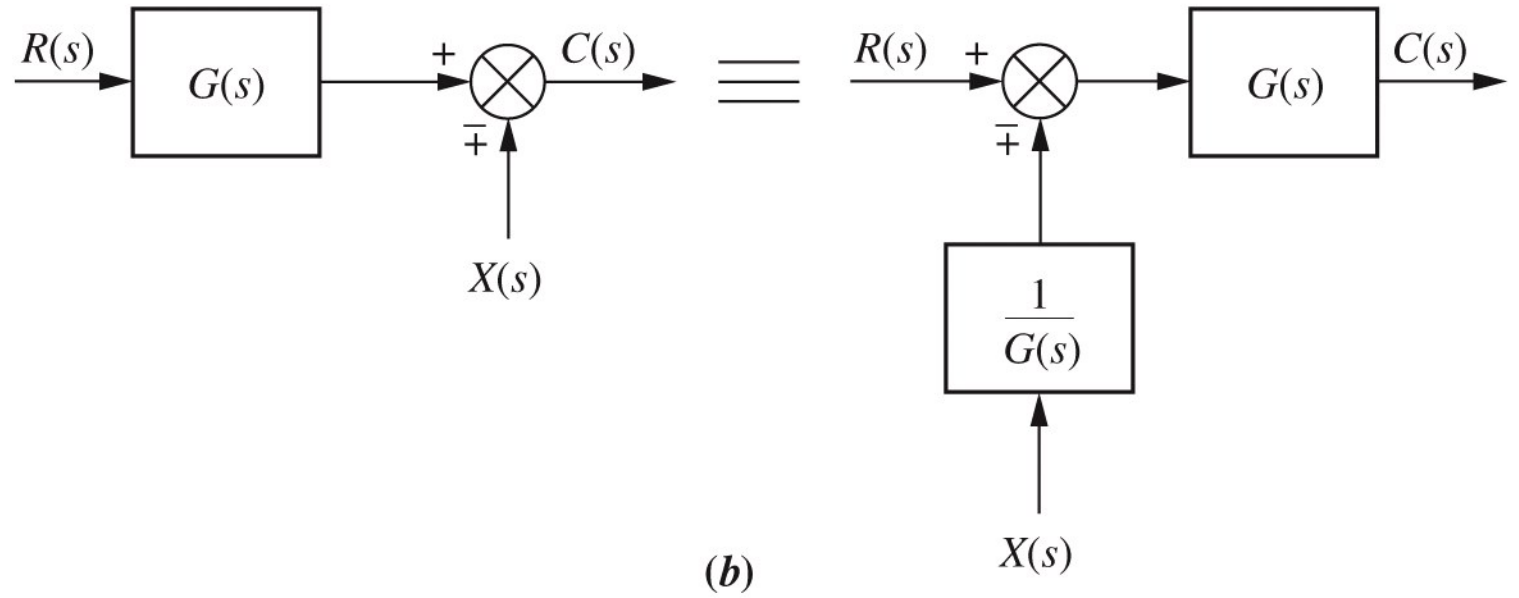
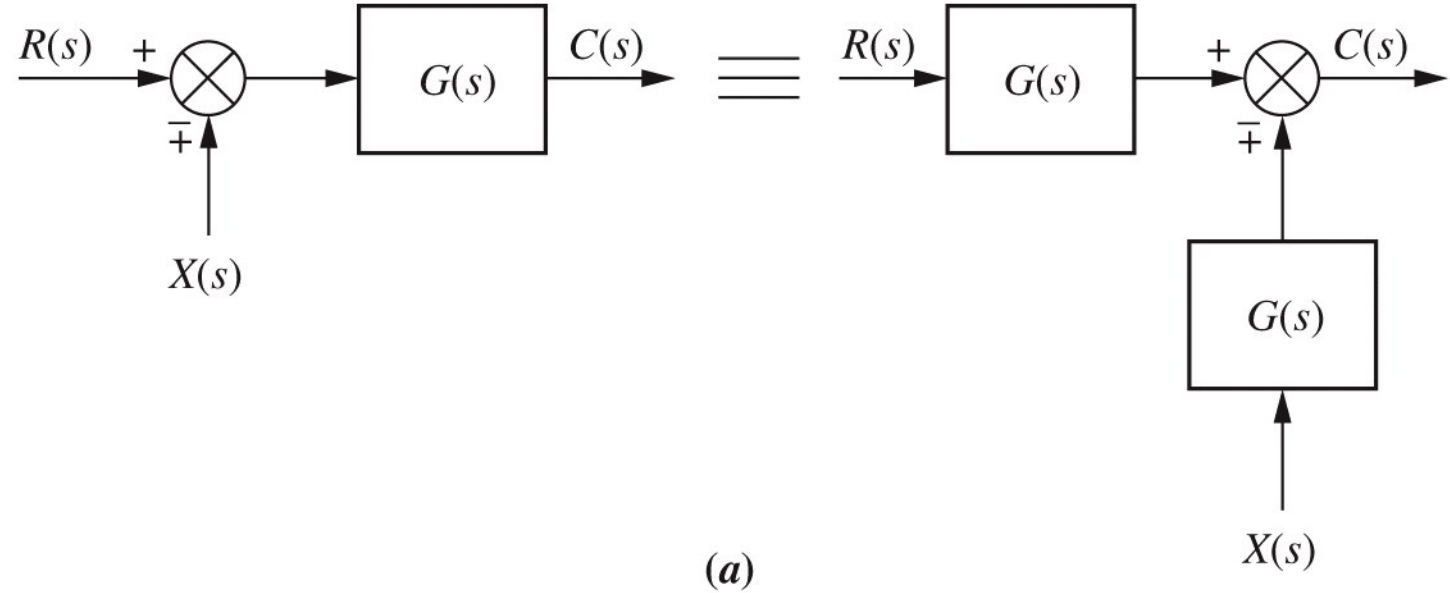


Figure 5.8
 © John Wiley & Sons, Inc. All rights reserved.

Moving Blocks to get Familiar Forms-2



Example-1

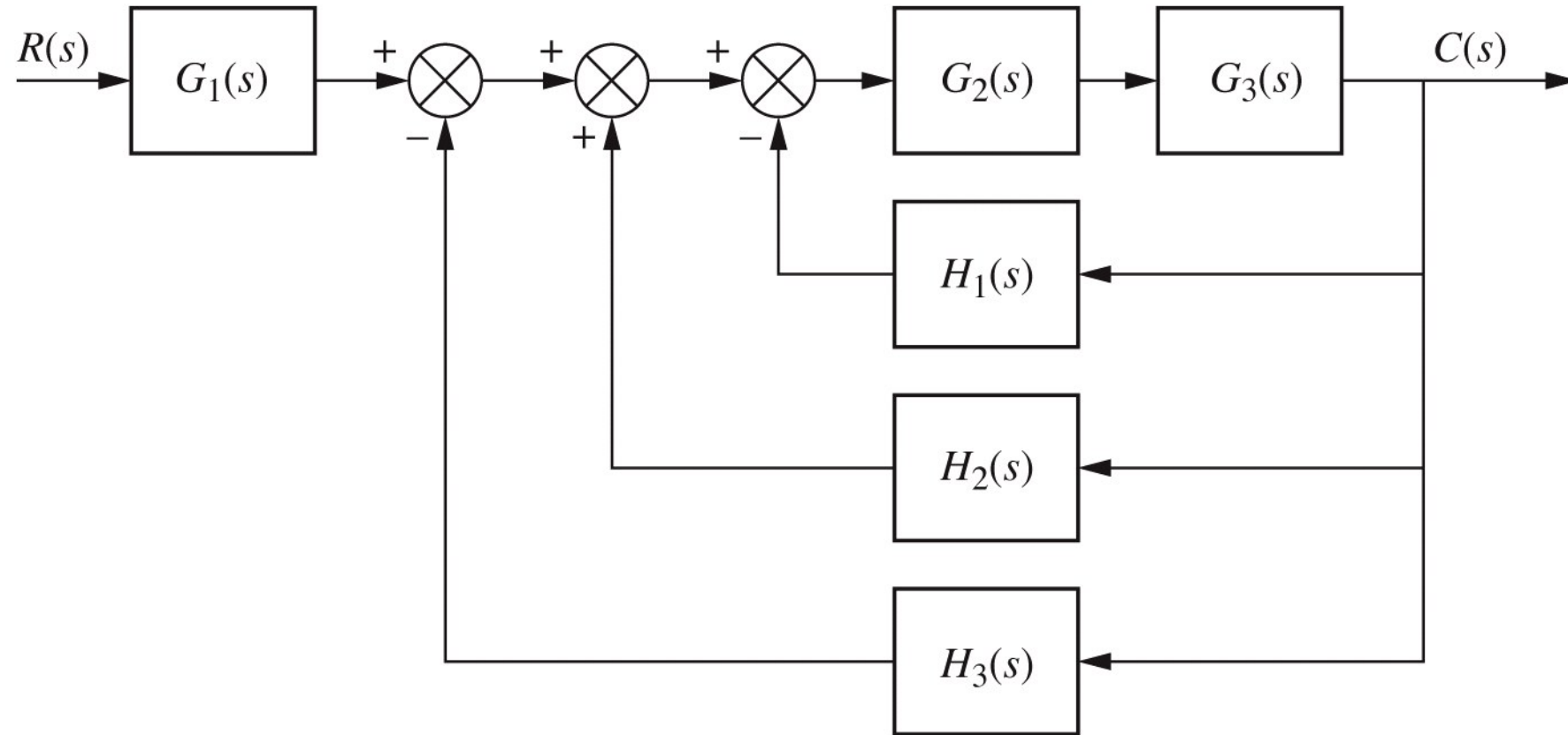
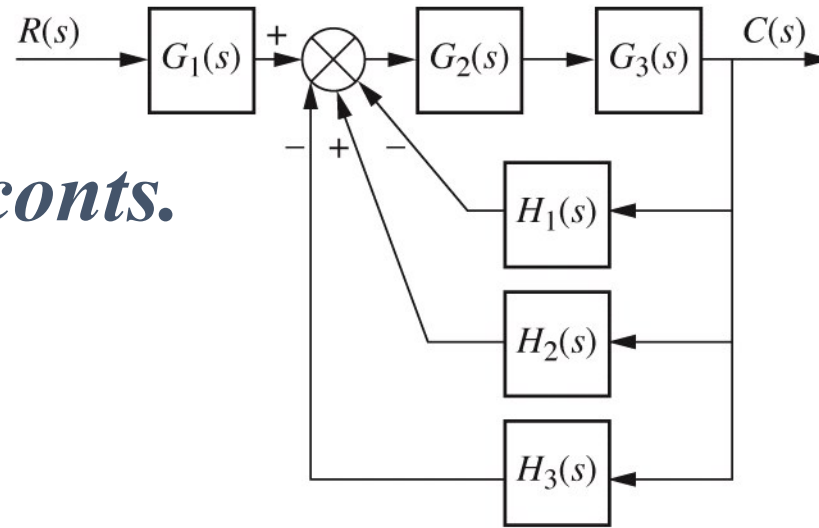
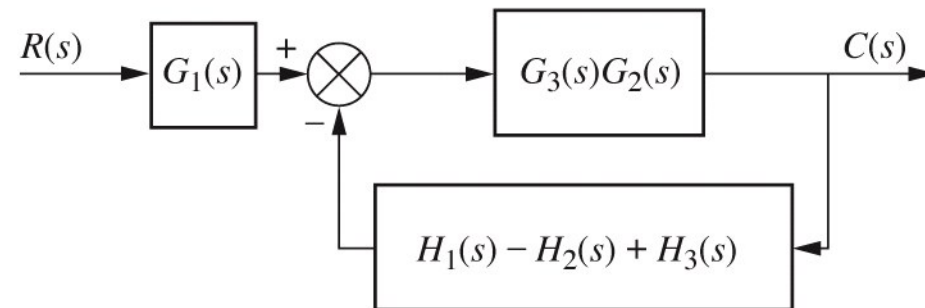


Figure 5.9
© John Wiley & Sons, Inc. All rights reserved.

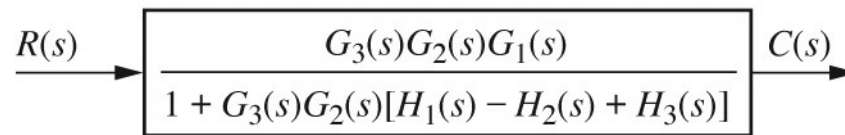
Example-1, *conts.*



(a)



(b)



(c)

Figure 5.10
© John Wiley & Sons, Inc. All rights reserved.

Example-2

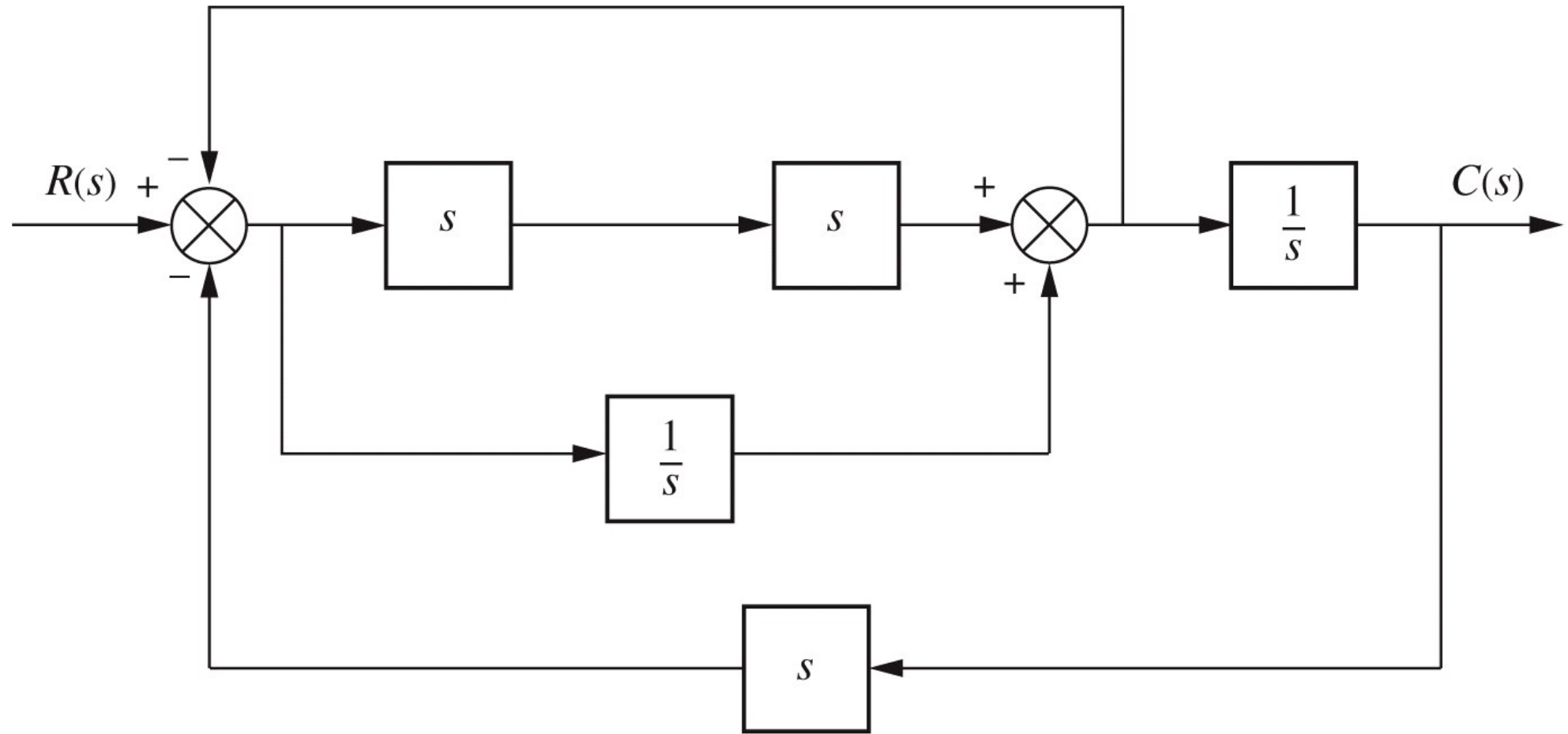
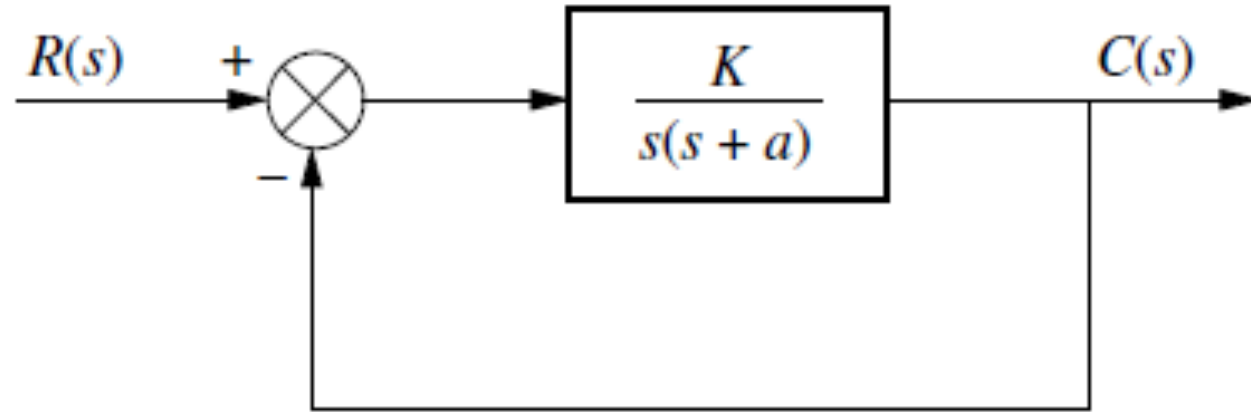


Figure 5.13
© John Wiley & Sons, Inc. All rights reserved.

Analysis and Design of Feedback Systems: An Introduction

- Consider the system shown in the figure below, which can model a control system such as the antenna azimuth position control system.



- The closed-loop transfer function is

$$T(s) = \frac{K}{s^2 + as + K}$$

where K models the amplifier gain, that is, the ratio of the output voltage to the input voltage.

Analysis and Design of Feedback Systems: An Introduction *cntd.*

$$T(s) = \frac{K}{s^2 + as + K}$$

For K between 0 and $a^2/4$, the system is overdamped with real poles located at

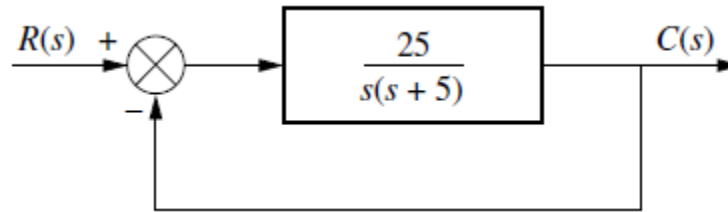
$$s_{1,2} = -\frac{a}{2} \pm \frac{\sqrt{a^2 - 4K}}{2}$$

For gains above $a^2/4$, the system is underdamped, with complex poles located at

$$s_{1,2} = -\frac{a}{2} \pm j \frac{\sqrt{4K - a^2}}{2}$$

For gains above $a^2/4$, as K increases, the real part remains constant and the imaginary part increases. Thus, the peak time decreases and the percent overshoot increases, while the settling time remains constant.

Example-6



For the system find the peak time, percent overshoot, and settling time.

The closed-loop transfer function of the system is

$$T(s) = \frac{25}{s^2 + 5s + 25}$$

$$\omega_n = \sqrt{25} = 5$$

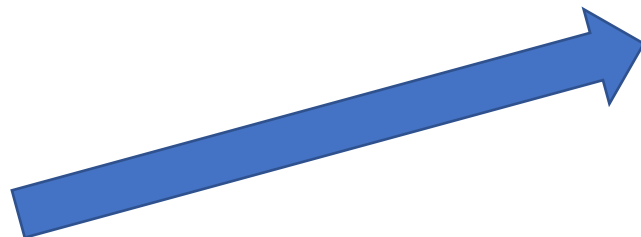
$$2\zeta\omega_n = 5$$

$$\zeta = 0.5$$

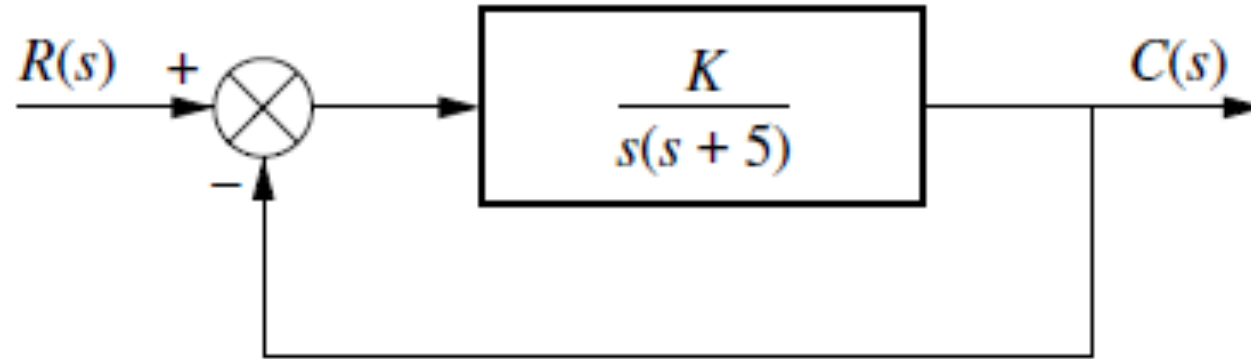
$$T_p = \frac{\pi}{\omega_n \sqrt{1 - \zeta^2}} = 0.726 \text{ second}$$

$$\%OS = e^{-\zeta\pi / \sqrt{1 - \zeta^2}} \times 100 = 16.303$$

$$T_s = \frac{4}{\zeta\omega_n} = 1.6 \text{ seconds}$$



Example-7



Design the value of gain, K , for the feedback control system so that the system will respond with a 10% overshoot.

The closed-loop transfer function of the system is

$$T(s) = \frac{K}{s^2 + 5s + K} \quad 2\zeta\omega_n = 5 \quad \omega_n = \sqrt{K} \quad \zeta = \frac{5}{2\sqrt{K}}$$

A 10% overshoot implies that $\zeta = 0.591$. Thus, $K = 17.9$.

Although we are able to design for percent overshoot in this problem, we could not have selected settling time as a design criterion because, regardless of the value of K , the real parts, -2.5, of the poles of the system remain the same.