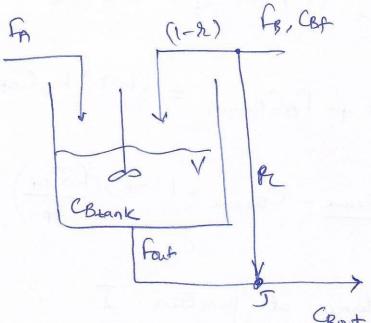
Week 8

Analysis of (p.g) order system.



Assure

(1) fa >>> fr

(3) of const: Acronghost the system

Sa & fr & frank & fout

Court, Fortlet

overall mans balance over the tank:

fa >> fs ; V constt

overall man balance @ junction]:

Component balance over the tank for B VdCBtank = (1-8) FBCBg - FACBtank VolCBeant + FACBEANK = (1-8) FBCBF (V) of CB+ank = (1-8) (FaCB) - O FA) of CB+ank = (1-8) (FaCB) - O component balance at junction: J FA CBLANK + Y FRCBF = FACBORDA CBont = CBtankt & (FBCBF) - (D) Let CBtank - CBtank, 95 = 2 CBout - CBout Ss = y FBCBF - (FBCBF) = U eq O can be rewritten as V dz + x = (1-8) u
FA dt equ (1) can be rewritten as y-2+29/4 -

$$\left(\frac{V}{FA}S+4\right)\overline{x}(S)=(1-8)\overline{u}(S)$$
 taking laplace of eq. 3

$$\frac{\overline{\chi}(s)}{\overline{u}(s)} = \frac{1-r}{\left(\frac{V}{f_{A}}\right)^{S}+1}$$

$$\frac{\overline{y}(s)}{\overline{F_A}} = \frac{(1-r)\overline{u}(s)}{\overline{v}(s)} + r \overline{u}(s)$$

$$\frac{y(s)}{u(s)} = \frac{(\frac{Vr}{f_A})s + 1}{(\frac{V}{f_A})s + 1}$$

$$\frac{\tilde{y}(s)}{\tilde{u}(s)} = \underbrace{zS+1}_{TS+1}$$

Comment on dynamics of eystem !!

$$\mathcal{J}_{(S)} = \frac{K}{\text{CS+L}} \Rightarrow y(t) = AK(1-e^{-t/e})$$

$$(I^{*} \text{ order})$$

Response of pure 2 voler system will never cross response of pure 18th order system for any value of I.

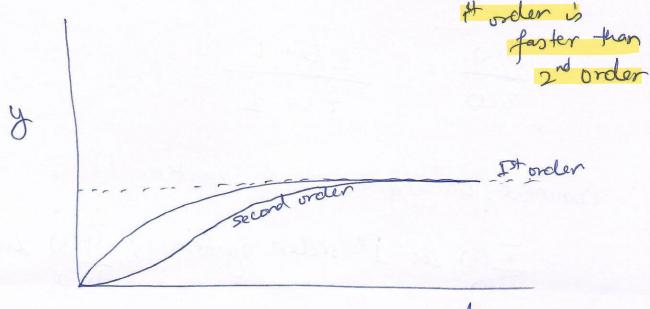
when subject to step furction of mag. A.

when open

to a color

uce) is step

function of



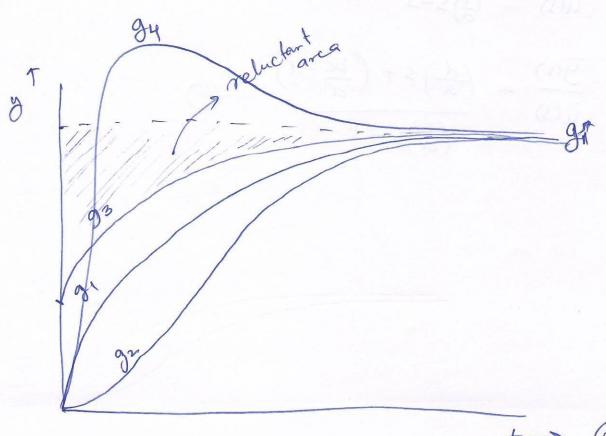
zeros of mmerator = zeros zeros of denominator = poles.

Addetion of pole makes response sluggish

$$93 = \frac{K(3S+1)}{TS+1} = \frac{\text{lead Lag system}}{T = \text{lag comit}}$$
 $T = \frac{\text{lag comit}}{T}$
 $T = \frac{\text{lag comit}}{T}$

if we add zero, reluctant once her in comparson to reluctant area of 1st order system

Addition of 2000 makes a response faster



$$g_4 = \frac{K(\Xi S + 1)}{(\Xi S + 1)(\Xi S + 1)}$$
 order (2,1)

for distribution column

for SISO

$$\frac{dx}{dt} = ax + by - 6$$

$$y = cx + dy - 6$$

$$\frac{\overline{\lambda}(s)}{\overline{u}(s)} = \frac{b/a}{(\frac{1}{a})s-1} - \overline{3}$$

$$\frac{\overline{y}(s)}{\overline{u}(s)} = \frac{\left(\frac{d}{a}\right)s + \left(\frac{bc}{a} - d\right)}{\left(\frac{d}{a}\right)s - 1}$$

Two Ip Two ofp de = a1121 + a1222 + b11U1 + b12U2 drz = a21 21 + and2 + b214, + b2242 - @ y1 = C11 X1 + C12 X2 + d11 U1 + d12 U2 - 3 J2= Cy X1 + G2 X2 +d2 4 4, +d24_ - 0 laplace of eg' (1) 2 (2) (S-an) 2(s) - an22(s) = bull(s) + bi242(s) - 9x 74(5) + (S-022) 72(57 = b2, 4, (5) + b2242(5) $\begin{bmatrix} S-911 & -a_{12} \\ -a_{21} & S-a_{22} \end{bmatrix} \begin{bmatrix} \widehat{x}_1(s) \\ \widehat{x}_2(s) \end{bmatrix} = \begin{bmatrix} b_{11} & b_{12} \\ b_{21} & b_{22} \end{bmatrix} \begin{bmatrix} u_1(s) \\ u_2(s) \end{bmatrix}$ $\begin{vmatrix} s \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} - \begin{bmatrix} a_{11} & a_{12} \\ a_{21} & a_{22} \end{vmatrix} \begin{bmatrix} \overline{a}_{1}(s) \\ \overline{a}_{2}(s) \end{bmatrix} = \begin{bmatrix} b_{11} & b_{12} \\ b_{21} & b_{22} \end{bmatrix} \begin{bmatrix} \overline{u}_{1}(s) \\ \overline{u}_{2}(s) \end{bmatrix}$ (SI-A) X(s) = BU(s) X(s) = [SI-A] B U(s) - B

4

laplace for eg^ (3) & (4).

$$\overline{Y}_{(s)} = \underline{C} \, \underline{X}_{(s)} + \underline{D} \, \underline{u}_{(s)}$$

for M input, P output, you will get P*M matrix

dx = ax+ bu gate space model y = extdy Inverse laplace Transform domain BLOCK DIAGRAM FOR DOMAIN TRANSPORM be +d. STATE SPACE DOMAIN da = aat bu y = cx+dy pink box -Const Value u - Ap signal

mrense laplace

$$\frac{dy}{dt} - \frac{dy}{dt} = \frac{d\left(\frac{dy}{dt}\right) + \left(\frac{bc}{ad}\right) u}{dt}$$

Transform domain -> state space domain

block diagram for following in both steady state Transform domain day = anx1+ a12x2+ b14

dx2 = a2x1 + a2x2+ b24

y = Cp2, + C222 + du

da

Response system (TS+1) - (TS+1) Given g(s) = I/p - o/p Block diagram of g(s) let û be step FIP & magnitude A y(s)= AK - AK S(Trs+1) Inverse laplace AK, (1-e") - AK2 (1-e") behaviour of system:-(K1-K2)

(6)

 $\frac{dy}{dt} = A \left(\frac{k_1}{4} - \frac{k_2}{\tau_L} \right) = 0 \quad \text{if} \quad \frac{k_1}{\tau_L} \times \frac{k_2}{\tau_L}$ $= \frac{1}{4} \quad \text{system of order (2,1)}$ $= \frac{1}{4} \quad \text{such that}$ $= \frac{k_1 \tau_L - k_1 \tau_J}{k_1 - k_2} \quad \text{condition of inverse}$ $= \frac{k_1 \tau_L - k_1 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_1 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$ $= \frac{k_1 \tau_L - k_2 \tau_J}{k_1 - k_2} \quad \text{formula}$

