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PROBLEM SET 7

DUE: 12/04/2015



c) $x(t) = \sum_{k=-\infty}^{\infty} C_k e^{j\frac{2\pi}{T}kt}$

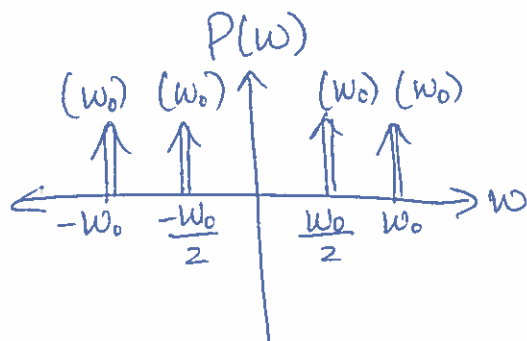
$$\begin{aligned} X(\omega) &= \int_{-\infty}^{\infty} x(t) e^{j\omega t} dt \\ &= \int_{-\infty}^{\infty} \sum_{k=-\infty}^{\infty} C_k e^{j\frac{2\pi}{T}kt} e^{j\omega t} dt \\ &= \sum_{k=-\infty}^{\infty} C_k \int_{-\infty}^{\infty} e^{j(\frac{2\pi}{T}k - \omega)t} dt \\ &= \sum_{k=-\infty}^{\infty} C_k 2\pi \delta(2\omega - \frac{2\pi}{T}k) \end{aligned}$$

Sidenote:

$$\begin{aligned} e^{j\omega t} &\rightarrow 2\pi \delta(\omega - \omega_0) \\ \omega - (\frac{2\pi}{T}k - \omega) &= 2\omega - \frac{2\pi}{T}k \end{aligned}$$

e) Changing T:

→ as $T \rightarrow \infty$, $P(\omega)$ and $p(t)$ approach 0
 → as $T \rightarrow 0$, $P(\omega)$ and $p(t)$ approach ∞



b) Fourier Series Representation

$$\sum_{k=-\infty}^{\infty} \delta(t - kT) = p(t)$$

$$\begin{aligned} C_k &= \frac{1}{T} \int_{-T/2}^{T/2} p(t) e^{-j\frac{2\pi}{T}kt} dt \\ &= \frac{1}{T} \int_{-T/2}^{T/2} \sum_{k=-\infty}^{\infty} \delta(t - kT) e^{-j\frac{2\pi}{T}kt} dt \\ &= \frac{1}{T} \int_{-T/2}^{T/2} \delta(t) e^{-j\frac{2\pi}{T}kt} dt \\ &= \frac{1}{T} \end{aligned}$$

d) $P(\omega) = \frac{2\pi}{T} \sum_{k=-\infty}^{\infty} \delta(2\omega - \frac{2\pi}{T}k)$

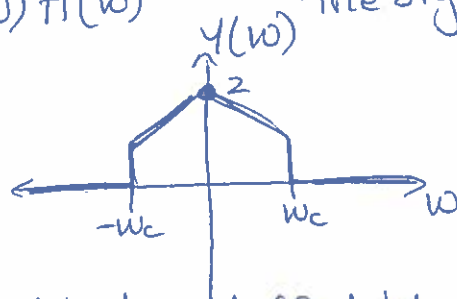
$$\omega_0 = \frac{2\pi}{T}$$

$$P(\omega) = \omega_0 \sum_{k=-\infty}^{\infty} \delta(2\omega - \omega_0 k)$$

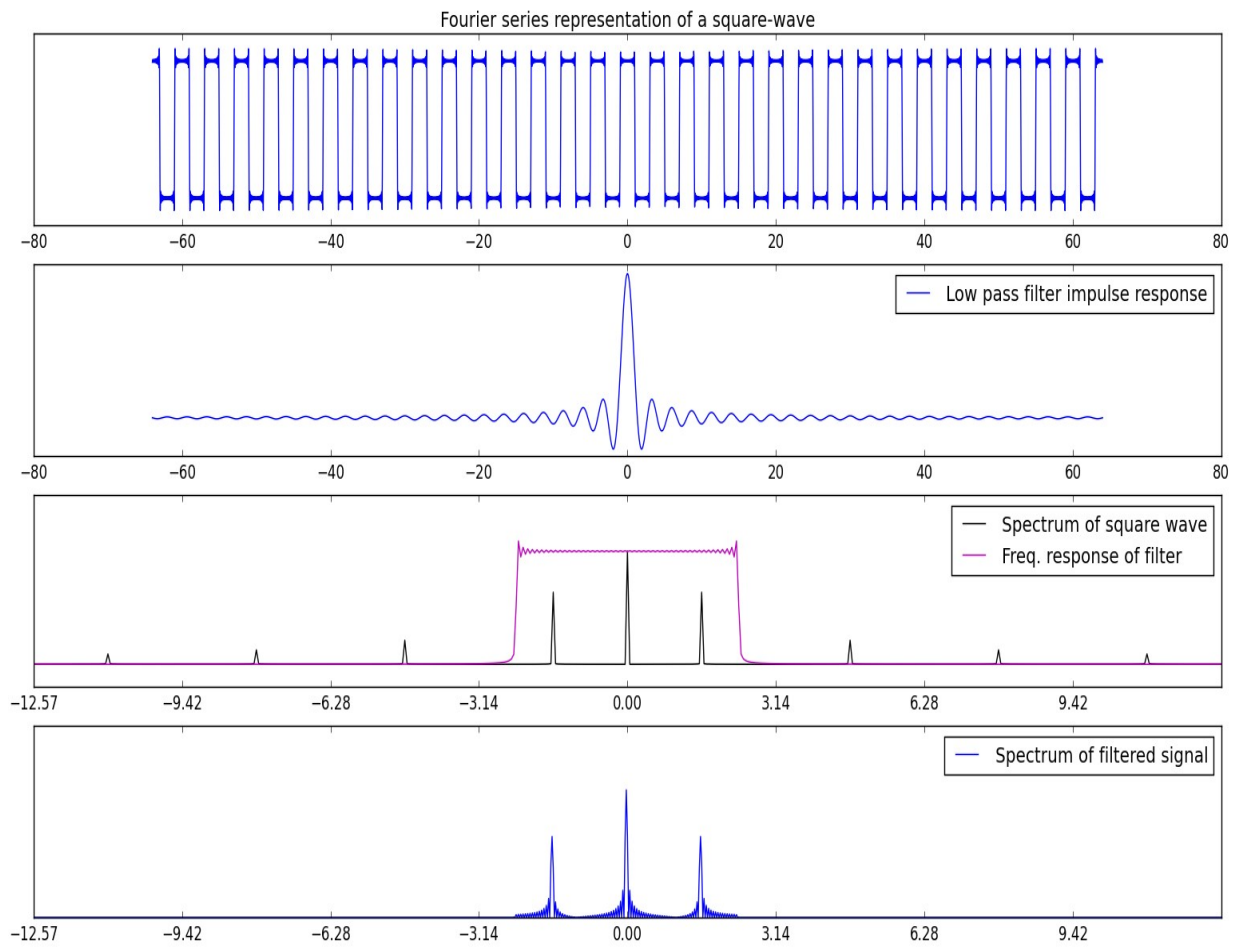
(2) a) $h(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} H(\omega) e^{j\omega t} d\omega$ $\left. \begin{array}{l} -\omega_c < \omega < \omega_c \rightarrow H(\omega) = 1 \\ \text{These are equal expressions} \\ \text{because the signal is 0 after } \omega_c \text{ and} \\ \text{before } -\omega_c \end{array} \right\}$
 $h(t) = \frac{1}{2\pi} \int_{-\omega_c}^{\omega_c} e^{j\omega t} d\omega$
 $h(t) = \frac{1}{2\pi} \left[\frac{1}{j t} e^{j\omega t} \right]_{-\omega_c}^{\omega_c}$
 $h(t) = \frac{1}{\pi t} \left(\frac{1}{2j} e^{j\omega_c t} - \frac{1}{2j} e^{-j\omega_c t} \right)$
 $h(t) = \frac{1}{\pi t} \sin(\omega_c t)$

b) $y(t) = x * h(t)$
 $Y(\omega) = X(\omega) H(\omega)$

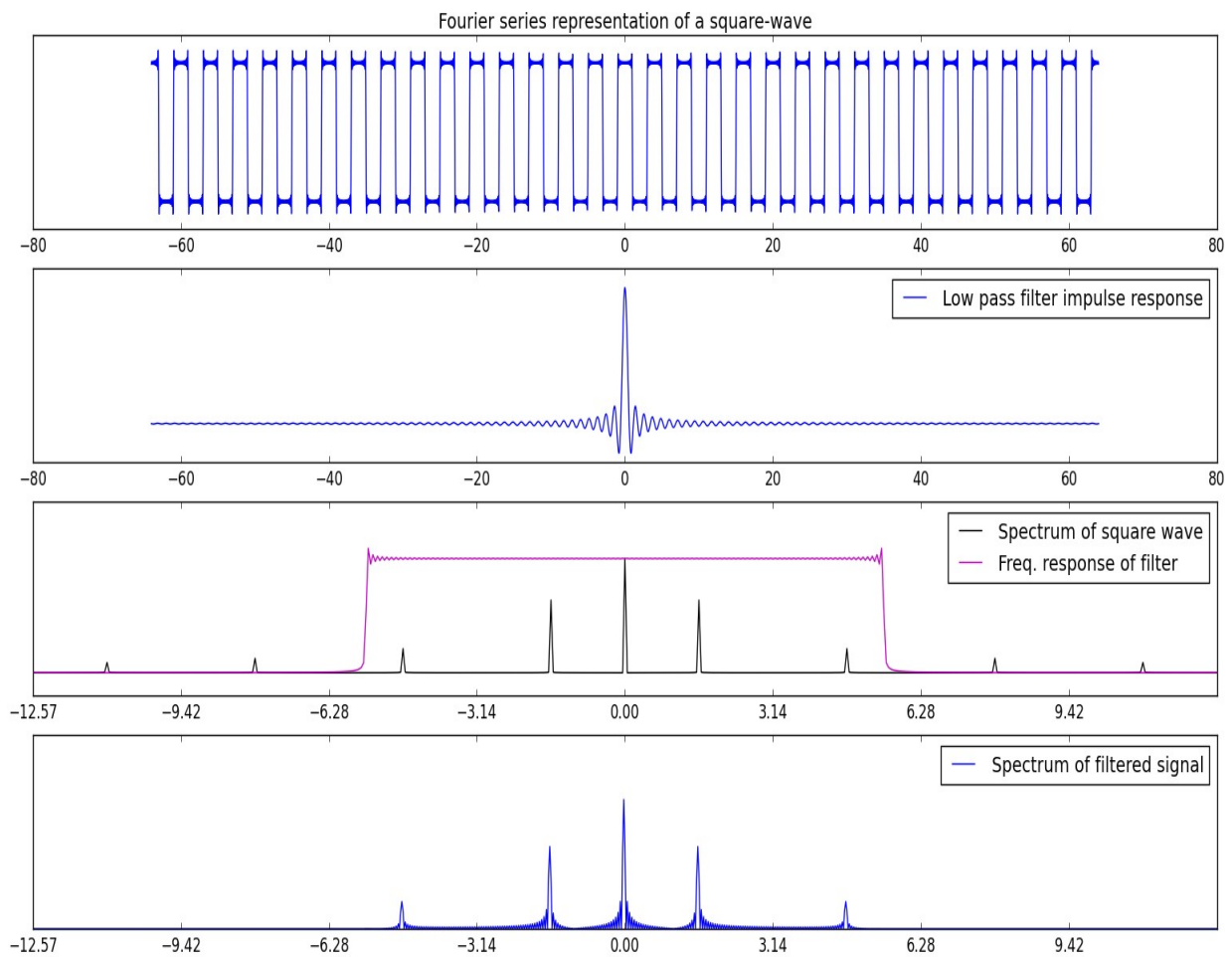
\therefore Between $-\omega_c$ and ω_c , $Y(\omega)$ is the same as $X(\omega)$. Otherwise, the signal is 0.



- c) Because it is able to cutoff higher frequencies ($> |\omega_c|$) while keeping the original signal the same.
 d) Plots are attached (next pages)



The graphs above show my results when I input my $h(t)$ function calculated earlier on in the question. The value for the cutoff frequency is set to $0.75 \cdot \pi$, which is validated by the third graph where you can see that the frequency cuts off slightly under the value of π .



I inputted the $h(t)$ function calculated earlier on in the question with the value of the cutoff frequency set to 1.75π , which can be seen in the third graph. You can see that the frequency cuts off slightly between the values of π and 2π (closer to the value of 2π).

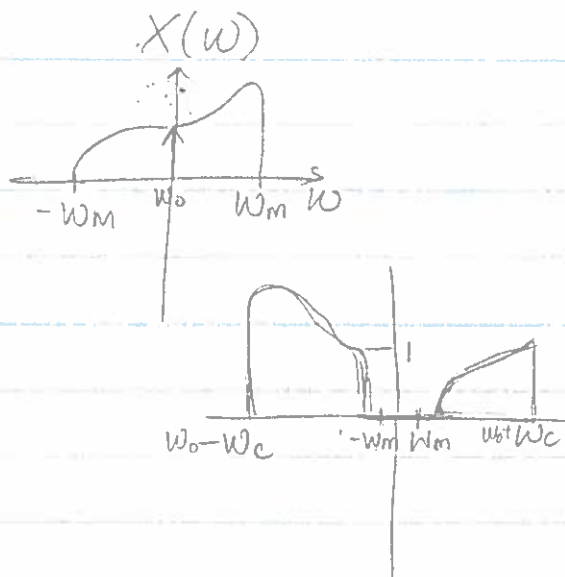
(3) $y(t) = x(t) \cos(\omega_c t)$

Multiplication in the time domain is convolution in the frequency domain.

$$Y(\omega) = \frac{1}{2\pi} X * H(\omega)$$

$$\cos(\omega_c t) \rightarrow \pi \delta(\omega - \omega_c) + \pi \delta(\omega + \omega_c)$$

$$H(\omega) = \pi \delta(\omega - \omega_c) + \pi \delta(\omega + \omega_c)$$



Looking at the graphs drawn, since $\omega_c \gg \omega_m$, and assuming that ^{the points between $-\omega_m$ and ω_m} ω_m don't overlap within the graph, $X(\omega) = Y(\omega)$ but if the graph overlaps then the graph will be scaled by 0.5.

Radio

$X(\omega)$ - radio signal being sent

$Y(\omega)$ - signal received

$H(\omega)$ - reproduces many signals' at different frequency levels