

c)
$$X_{p}(w) = \frac{1}{2\pi} X(w) * P(w)$$

$$= \frac{1}{2\pi} X(w) * \frac{2\pi}{T} \sum_{k=-\infty}^{\infty} \delta(w - \frac{2\pi}{T} k)$$

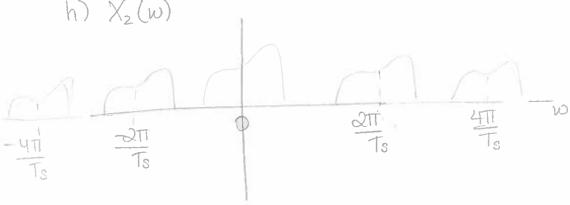
$$= \frac{1}{2\pi} X(w) * \frac{2\pi}{T} \sum_{k=-\infty}^{\infty} \delta(w - \frac{2\pi}{T} k)$$



e) Apply a filter-allowing = \frac{1}{\intersection \text{X(w-2TTK)}} \ values within the range \\
-wm to wm to pass through

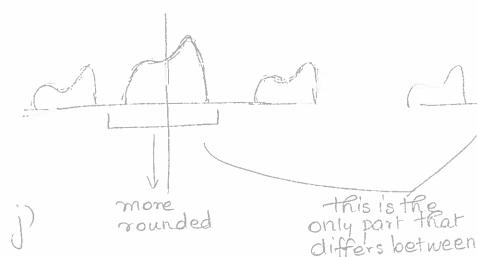
$$g(x_2(t) = x_p * 2(t)$$





i)
$$\overline{X}(w) = X_2(w) H(w)$$

$$\hat{X}(w) = X_{p}(w)H(w)$$



$$X_2(t) = X_p * 2(t) \rightarrow X_2(w) = X_p(w) Z(w)$$

 $X_p(t) = X(t) p(t) \rightarrow X_p(w) = 1 = X_p(w) Z(w)$
 $Z(w) = 1 = Cinc(w/2)$

$$\frac{\overline{X(w)}}{\widehat{X(w)}} = \frac{X_2(w) + (tw)}{X_p(w) + (tw)} = \frac{X_p(w) + Z(w)}{X_p(w)} = \frac{Z(w)}{X_p(w)}$$
Radio is $Z(w)$

the two.

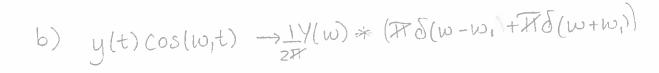
(2) a) Y(w)

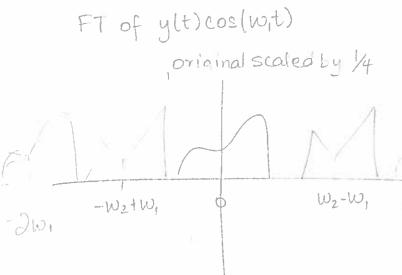
scaled by 1/2

cos(wot) -> TTS(w-wo)+TTS(w+wo)

 $y(t) = x_1(t) \cos(w_1t) + x_2(t) \cos(w_2t)$

 $\frac{1}{w_{2}^{1}} \frac{1}{Y(w)} = \frac{1}{2} X_{1}(w) \delta(w-w_{1}) + \delta(w+w_{1}) + \dots$





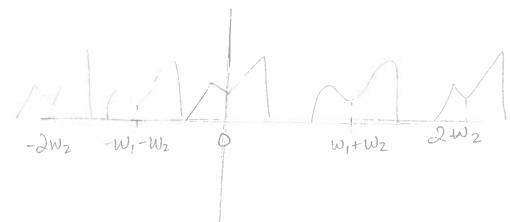
Since Wz > 2 wm + 10, there showdn't be any interference

 $\frac{1}{2\omega_1}\omega_1 + \omega_2 = \omega_1 + (2\omega_m + \omega_1)$ $= 2\omega_1 + 2\omega_m$

FT of y(t) cos (w2t)
original scaled by 1/4

 $w_1 \leq w_2 - 2w_m$ $w_1 + w_2 = w_2 - 2w_m + w_2$

22102-2Wm



c) Using filters with a bandwidth of 2 wm centered around 0 for each

$$=\frac{1}{\text{jwRC+(jw)}^2\text{LC+1}} = \frac{\text{imag}}{\text{real}} = \frac{\text{wRC}}{\text{w}^2\text{LC+1}}$$

c)
$$|H(w)| = \frac{1}{\sqrt{(w)^2(RC)^2 + (1-w^2(LC)^2)^2}}$$

 $\int [w)^2(RC)^2 + (1-w^2(LC)^2)^2$
 $\int [w]^2(RC)^2 + (1-w^2(LC)^2)^2$

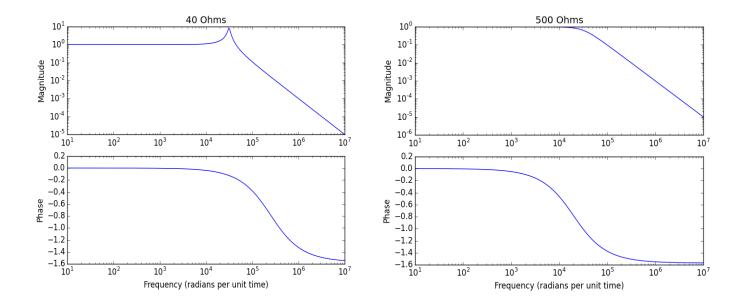
d) |H(w) | is maximized, when d|H(w) =0

$$\frac{d|H(\omega)|_{-}}{d\omega} = -\frac{1}{2} \left((RC\omega)^{2} + (1-\omega^{2}(LC)^{2})^{2} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(1-\omega^{2}(LC)^{2})(-2\omega)(1-\omega^{2}(LC)^{2}) \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(1-\omega^{2}(LC)^{2})(-2\omega)(1-\omega^{2}(LC)^{2}) \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(1-\omega^{2}(LC)^{2})(1-\omega^{2}(LC)^{2}) \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2}\omega + 2(LC)^{2} \right)^{-\frac{3}{2}} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2}\omega + 2(LC)^{2}\omega \right)^{-\frac{2}{2}} \right)^{-\frac{3}{2}} \left(2(RC)^{2}\omega + 2(LC)^{2}\omega \right)^{-\frac{3}{2}} \right)^$$

2(RX)2 w= 4 w/LX)2 +4w8(LX)4 2R=4L2+4w2L4c2

 $W = \sqrt{\frac{R - 2L}{2L^4C^2}}$

Both graphs display low pass filters, where the higher frequencies are cut off. However, changing the resistance changes where the frequency is cutoff.



By comparing the two graphs, we can see that the graph with the higher resistance (500 Ohms) is cutoff at a lower frequency than the lower resistance (40 Ohms). As a result, we can see that the cutoff frequency is inversely proportional to the resistance, when the capacitance and inductance are kept constant. Another trend we can see is that the phase shift corresponds to the cutoff frequency.

For the lower resistance, we can also see that the magnitude suddenly spikes before it drops. This relates to the maximum value for omega, the equation of which was calculated in the previous question.