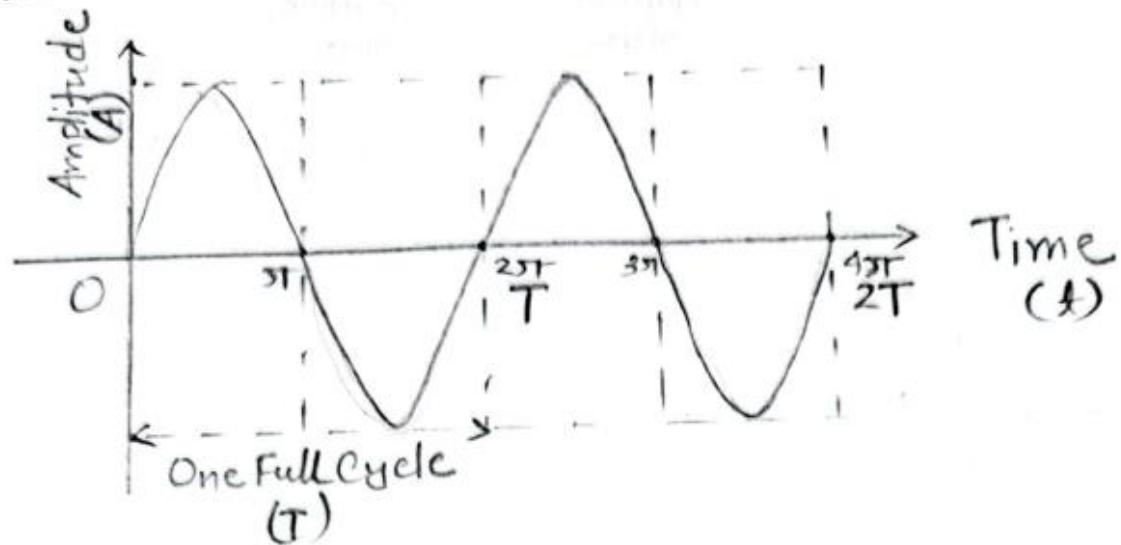


Different Types of waves:

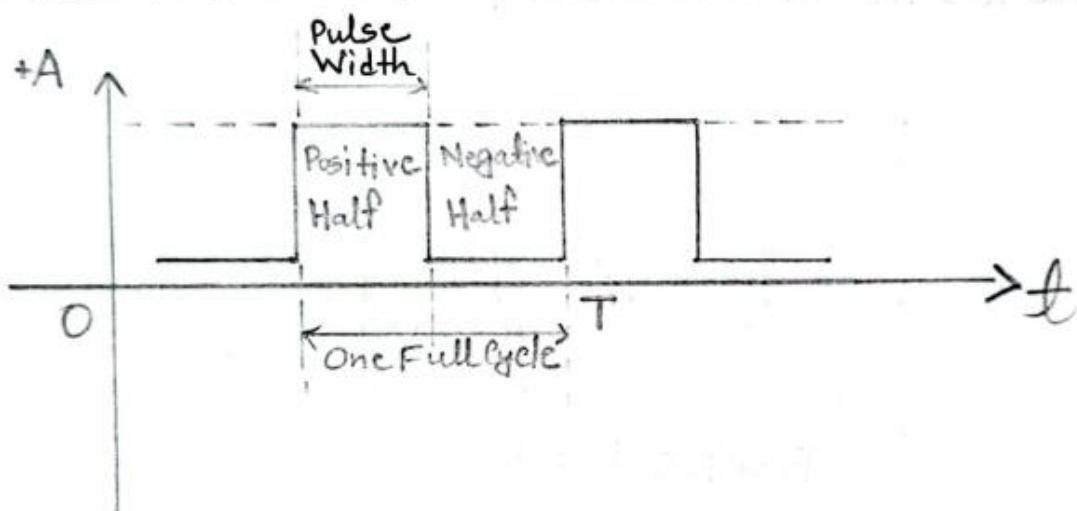
Sin waveform:

Sine Waveform:

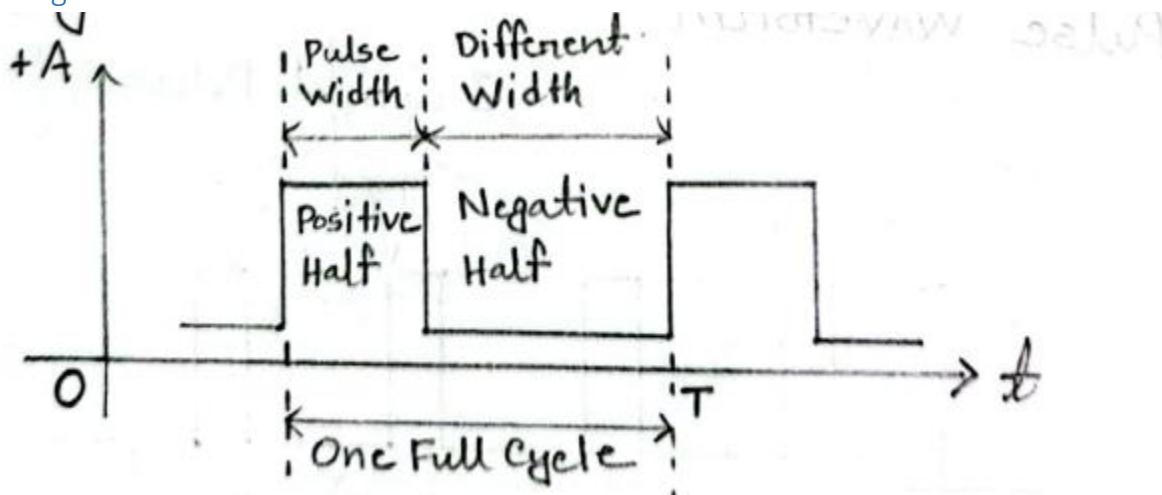


Square waveform:

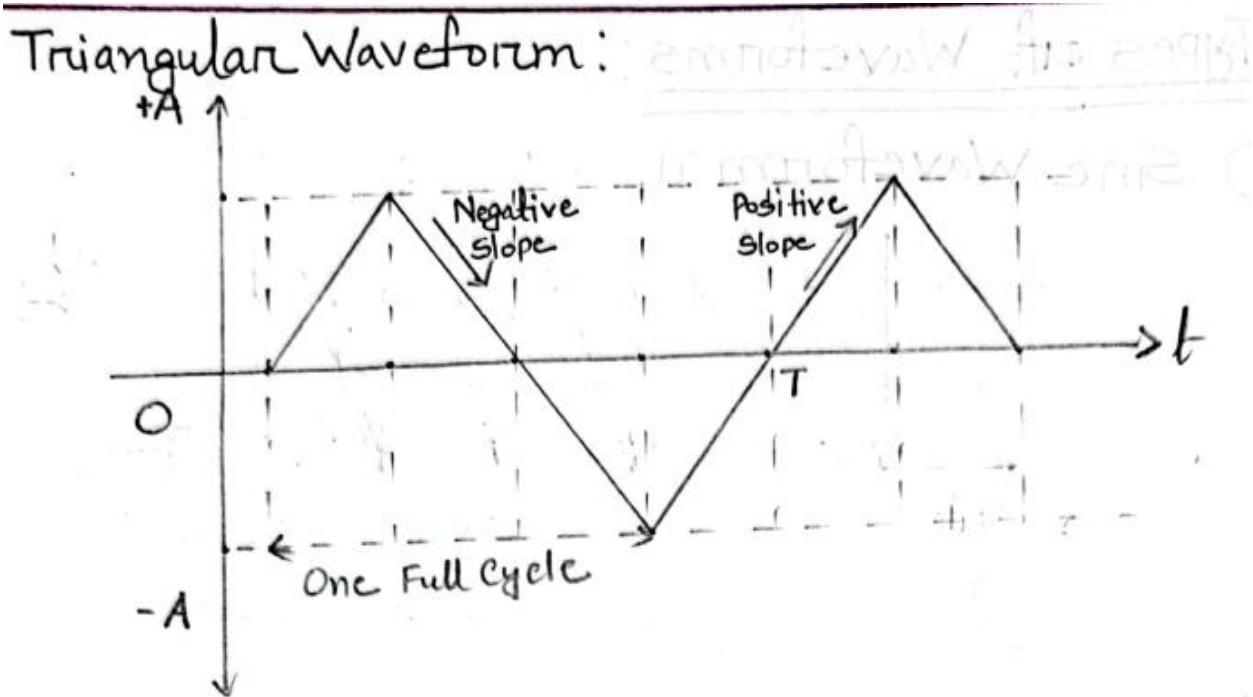
Square Waveform:



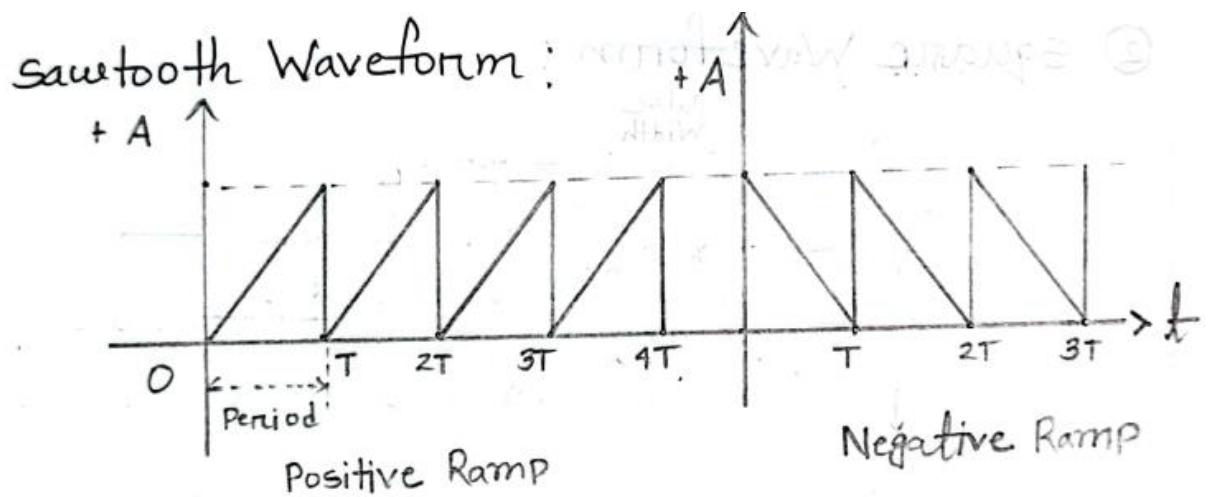
Rectangular waveform:



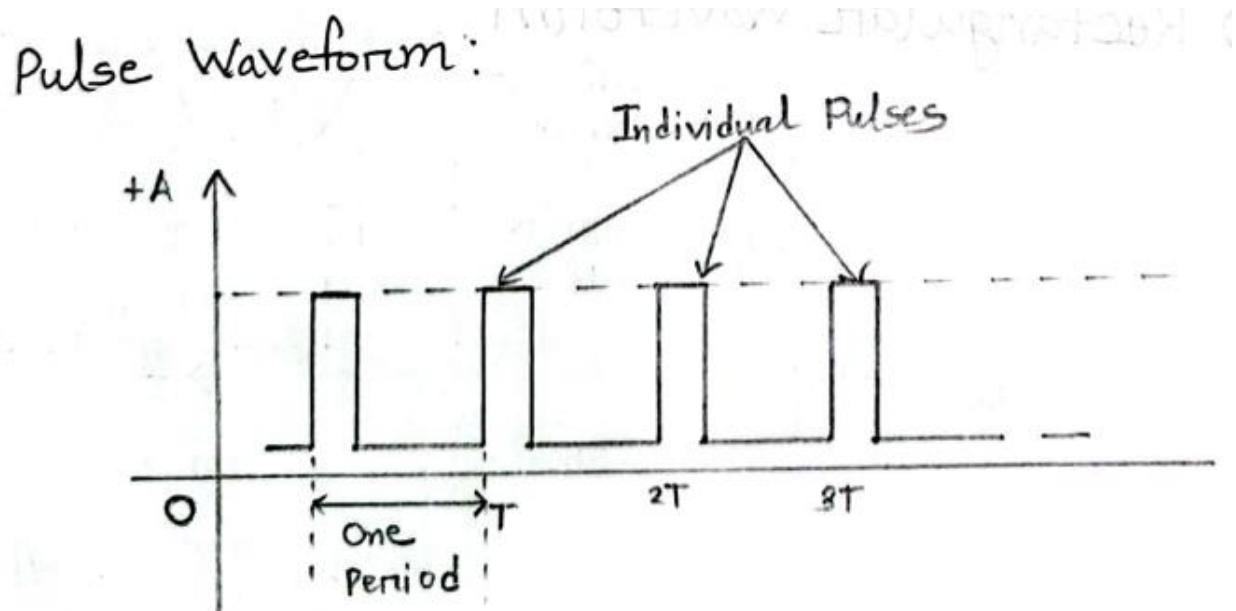
Triangular waveform:



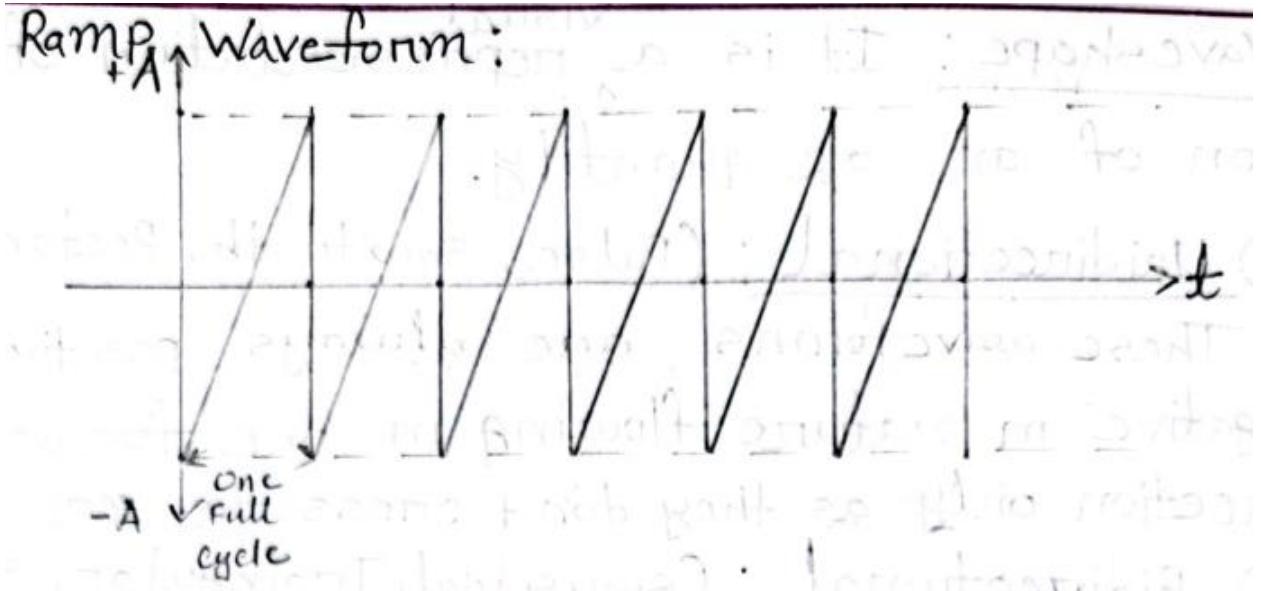
Sawtooth waveform:



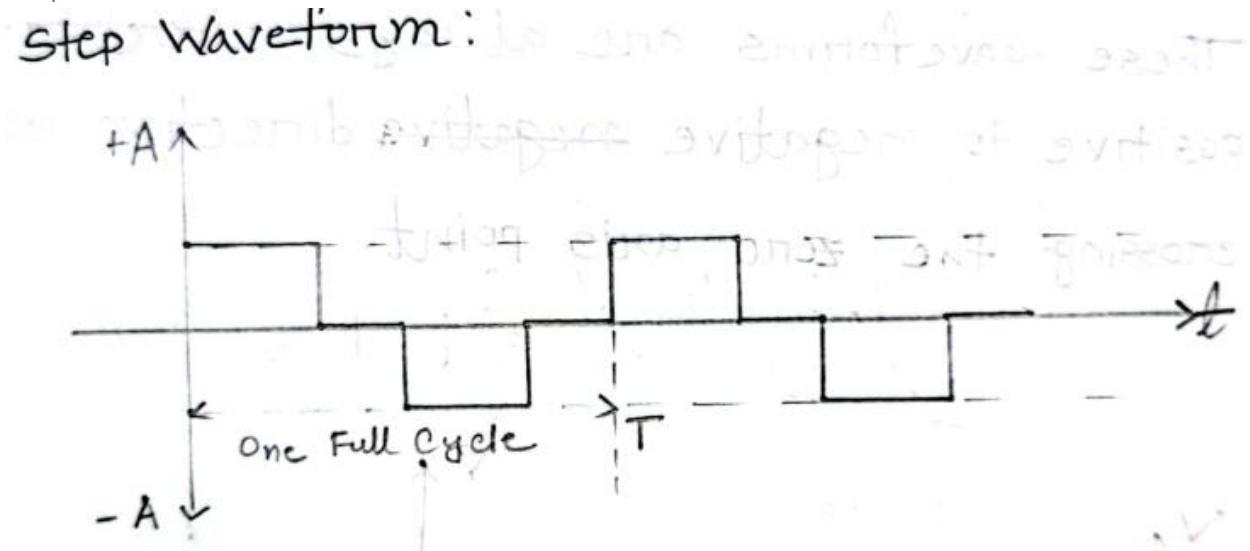
Pulse waveform:



Ramp Waveform:



Step waveform:



Waveshape:

Waveshape is the visual representation of variation of an AC quantity.

It's of two types:

Unidirectional:

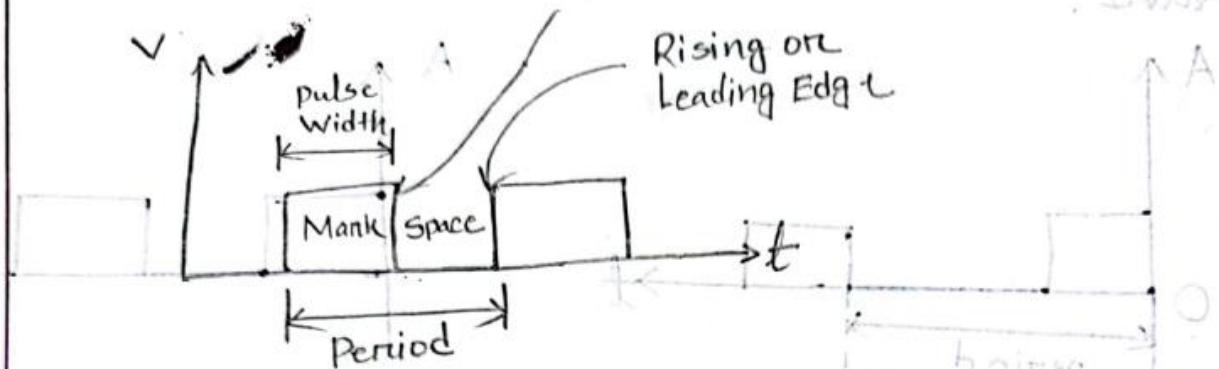
These waveforms are either always positive or always negative, and flows in one forward direction as they don't cross the zero axis point. E.g. **pulse, sawtooth, rectangular**.

Bidirectional:

Always alters between positive and negative as they always cross the zero axis point. E.g. **Sinusoidal, Triangular, Square.**

Parts of square waveforms:

Square Waveform: Falling on Trailing Edge



$$T = \text{Mark Time} + \text{Space Time}$$

Duty Cycle  $\rightarrow$  Percentage

Duty cycle:

In digital circuit, the percentage of the positive pulse or mark time is called the duty cycle.

Math:

In a waveshape, the pulse width (mark time) is 20ms and duty cycle is 33%. Calculate its frequency.

Solution: positive =  $\pi$   $\rightarrow$  20ms

negative =  $2\pi$   $\rightarrow$  40ms

$$T = 20 + 40 = 60\text{ms}$$

$$f = \frac{1}{60 \times 10^{-3}} \text{ Hz} \quad (\text{Ans.})$$

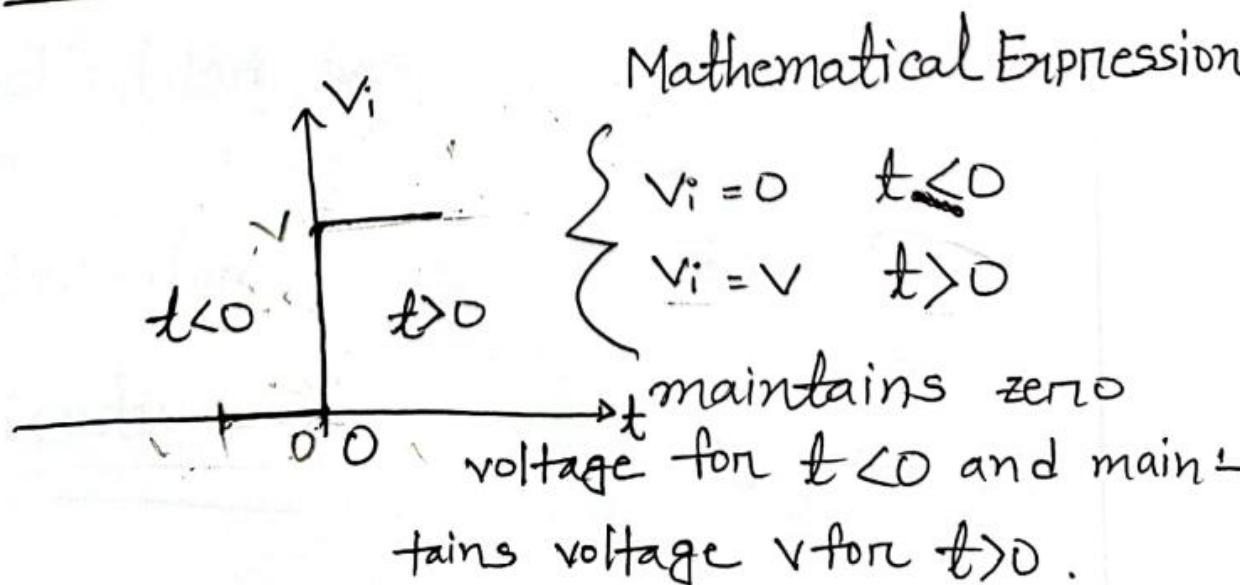
Difference between Rectangular and square waves:

- Square has same space and mark time, thus 50% duty cycle.
- Square is symmetrical.
- Square is used in electronics and micro electronic devices. Used to generate clock signals and timing circuits.

- Rectangular wave is used for varying any electrical quantity by varying the power supplied to a load.

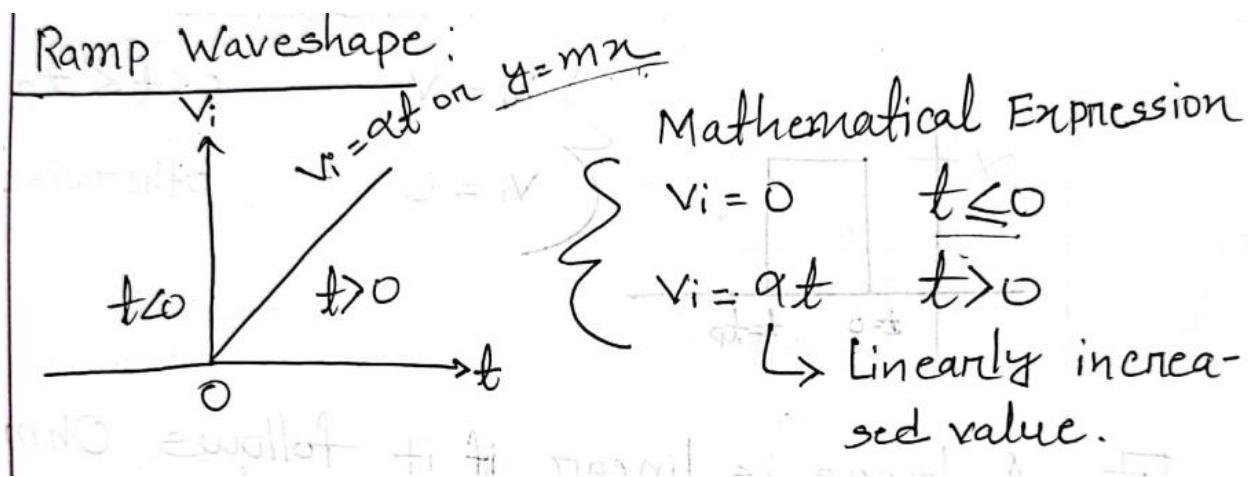
### Step waveshape:

A step voltage is one which maintains voltage 0 when  $t < 0$  and voltage  $v$  when  $t > 0$ .



### Ramp waveshape:

A ramp voltage is one where voltage 0 is maintained when  $t \leq 0$  and voltage linearly increased when  $t > 0$ .

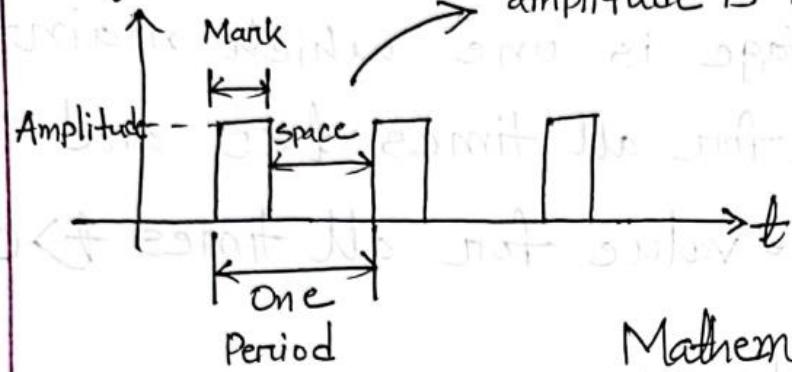


### Pulse waveshape:

A pulse voltage is when voltage is maintained at  $v$  when  $0 \leq t \leq t_p$ , otherwise maintained at  $v=0$ .

Here time duration is too small for mark, but amplitude is high.

## Pulse:



Time duration is too little but amplitude is high.

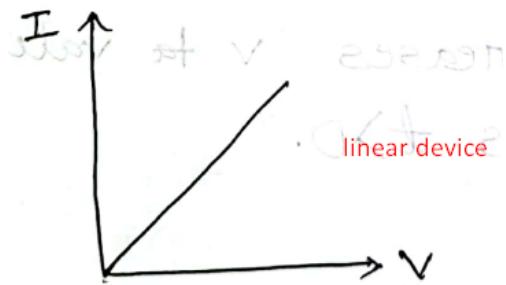
Differentiator provides these types of waveshapes

Mathematical Expression:

$$\left\{ \begin{array}{l} v_i = V \quad 0 \leq t \leq t_p \\ v_i = 0 \quad \text{otherwise} \end{array} \right.$$

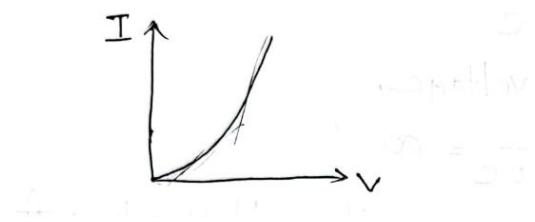
## V-I characteristics:

A device is linear when it follows ohm's law i.e. R, L, C.



Diode and transistors are nonlinear devices:

Diode V - I characteristic:



## Linear waveshaping:

Linear elements such as resistors, capacitors and inductors are employed to shape a signal in this linear wave shaping.

## Filtering:

Filtering is the process where unwanted portion of a signal is removed.

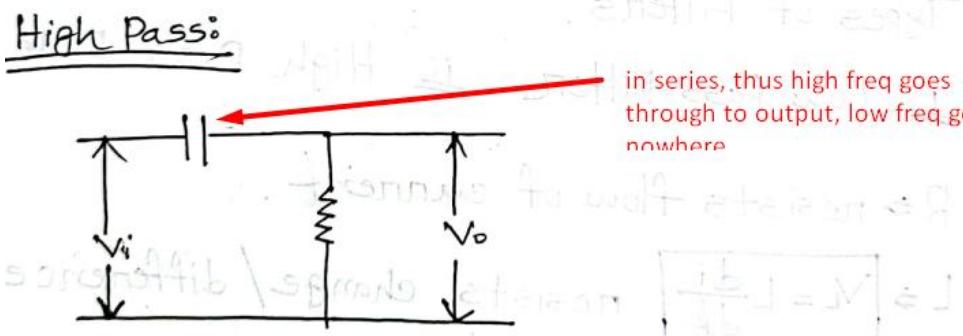
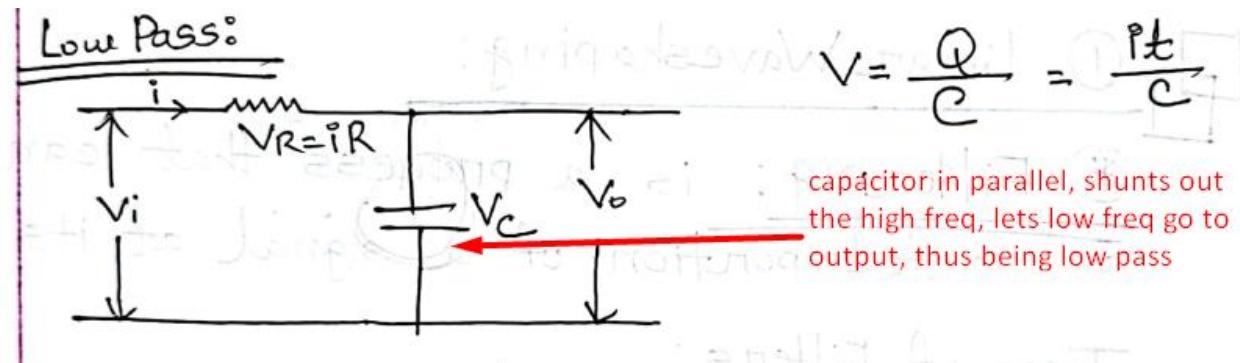
1. Low pass filter
2. High pass filter

Capacitive reactance:

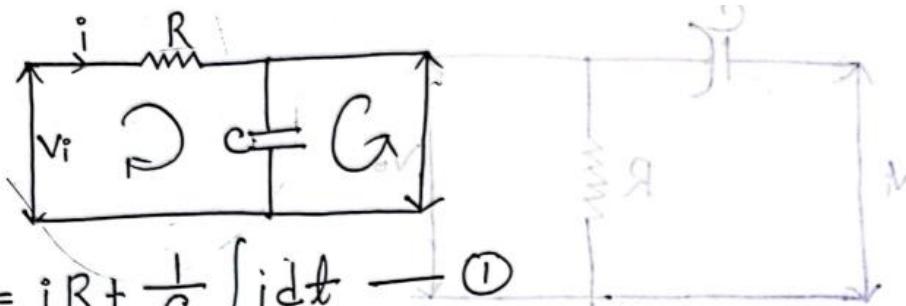
Capacitor blocks DC current. In DC, frequency is zero, thus reactance is infinite, thus blocking it. For AC, the higher the frequency or  $f$ , lower reactance, thus letting higher frequencies pass and blocking low frequencies.

$$X_C = \frac{1}{2\pi f C}$$

Here:



Low pass filter as integrator:



$$Vi = iR + \frac{1}{C} \int idt \quad \text{--- (1)}$$

$$Vo = \frac{1}{C} \int idt \quad \text{--- (2)}$$

$$\text{since } \frac{1}{C} \int idt \ll iR \quad \text{--- (3)} \quad \rightarrow R_i = \infty$$

from (1)

$$Vi = iR$$

$$\Rightarrow i = \frac{Vi}{R}$$

$$(2) \quad Vo = \frac{1}{C} \int idt$$

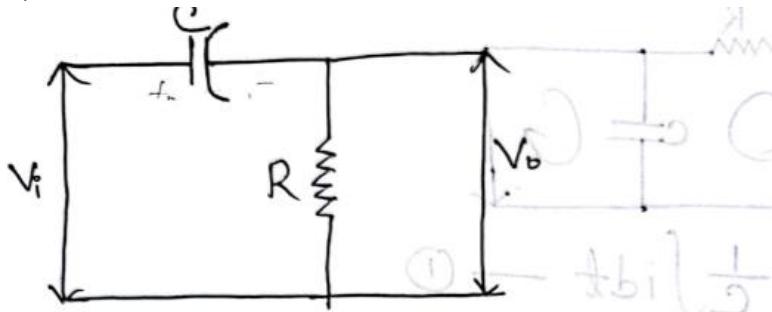
$$= \frac{1}{C} \int \frac{Vi}{R} dt$$

$$\therefore Vo = \frac{1}{RC} \int Vi dt$$

$$\therefore Vo \propto \int Vi dt \text{ as } \frac{1}{RC} \text{ is constant}$$

low pass, so DC is way higher  
iR represents DC

High pass filter as differentiator:



$$V_i = \frac{1}{C} \int i dt + iR \quad \textcircled{1}$$

$$V_o = iR \quad \textcircled{2}$$

since,  $iR \ll \frac{1}{C} \int i dt$

$$\text{From } \textcircled{2} \Rightarrow i = \frac{V_o}{R}$$

$$\text{From } \textcircled{1} \Rightarrow$$

$$V_i = \frac{1}{C} \int i dt$$

$$= \frac{1}{C} \int \frac{V_o}{R} dt$$

$$V_i = \frac{1}{RC} \int V_o dt$$

differentiating both sides,

$$\frac{dV_i}{dt} = \frac{V_o}{RC} \quad \text{as } \frac{1}{RC} \approx \frac{1}{\omega_c} \Rightarrow \frac{dV_i}{dt} \propto \omega_c V_o$$

$$\text{or, } V_o = RC \frac{dV_i}{dt}$$

$\therefore V_o \propto \frac{dV_i}{dt}$  as  $RC$  is constant.

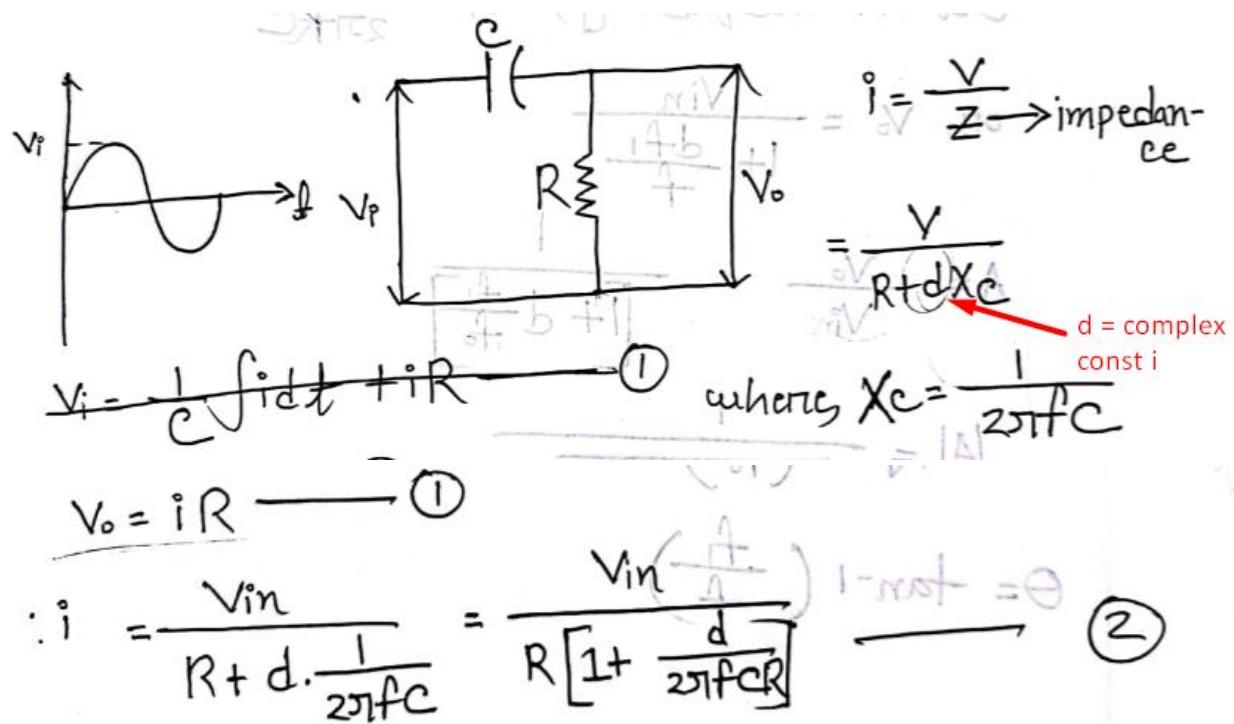
Sinusoidal input to a High pass filter:

Cut off frequency:

This "Cut-off", "Corner" or "Breakpoint" frequency is defined as being the frequency point where the capacitive reactance and resistance are equal. Capacitive resistance increases with frequency.

$$f_c = \frac{1}{2\pi RC}$$

**Calculation:**



From ①  $\Rightarrow$

$$V_o = \frac{V_{in}}{R \left[ 1 + \frac{d}{2\pi f C R} \right]} R$$

or,  $V_o = \frac{V_{in}}{1 + \frac{d}{2\pi f R C}}$

$$X_C = \frac{1}{2\pi f C}$$

$$f = \frac{1}{2\pi X_C C}$$



Let;

Cut off frequency,  $f_1 = \frac{1}{2\pi R C}$

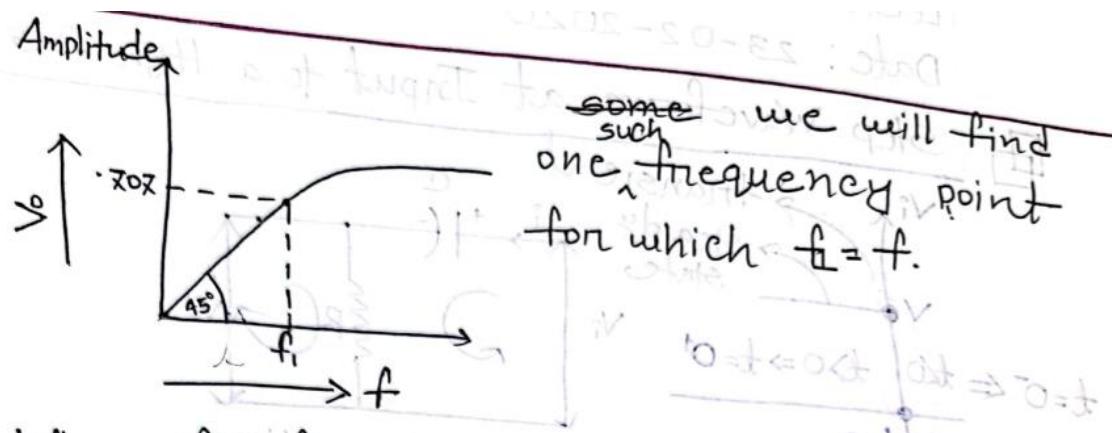
or,  $V_o = \frac{V_{in}}{1 + \frac{d f_1}{f}}$

$$A = \frac{V_o}{V_{in}} = \frac{1}{1 + d \frac{f_1}{f}}$$

$$|A| = \frac{1}{\sqrt{1 + \left( \frac{f_1}{f} \right)^2}}$$

it is being calculated in complex plane

$$\theta = \tan^{-1} \left( \frac{f_1}{f} \right)$$



When  $f_i = f$ ,

$$\text{Amplitude} = \frac{1}{\sqrt{1+(1)^2}} = \frac{1}{\sqrt{2}} = 0.707$$

$$\theta = \tan^{-1}(1) = 45^\circ$$

The point where  $f_i = f$  is called 3dB point

Step input to a high pass filter:

Step input is when for  $t > 0$   $V_{in} = v$  and  $t < 0$   $V_{in} = 0$ .

The transition between two voltage occurs  
at  $t=0$

$$V_i = 0 \text{ when } t = 0^- \quad \text{Eq} = (1) \text{ for } t = 0^-$$

$$V_i = v \text{ when } t = 0^+ \text{ it starts rising}$$

$$V_i = \frac{1}{C} \int i dt + iR \quad \text{--- ①}$$

$$V_o = iR \quad \text{--- ②} \Rightarrow i = \frac{V_o}{R} \quad \text{--- ③}$$

From ①  $\Rightarrow$

$$V_i = \frac{1}{C} \int \frac{V_o}{R} dt + \frac{V_o}{R} \cdot R$$
$$= \frac{1}{RC} \int V_o dt + V_o$$

Differentiating both sides  $\rightarrow$  constant value

$$\frac{dV_i}{dt} = \left( \frac{1}{RC} \times V_o \right) + \frac{dV_o}{dt}$$

steady state response  $\rightarrow$  transient state

If  $V_i = V$  [V is  $\neq$  a constant]

$$\frac{dV_i}{dt} = \frac{dV}{dt} = 0 \quad \underline{\underline{Q = [Initial] V}}$$

From ③  $\Rightarrow$

$$\frac{1}{RC} V_o + \frac{dV_o}{dt} = 0 \quad \rightarrow \text{Differential Eqn}$$

The general solution will be,

$$V_o = \text{complementary function} + \text{particular integral} \quad ⑤$$

Complementary Equation:  $\text{O} = \sqrt{\frac{b}{4b}}$

$$\frac{dV_i}{dt} = -\frac{V_o}{RC} + \frac{d}{dt}V_o \quad \text{--- (6)}$$

Let us consider,

$$D = \frac{d}{dt}$$

Then (6) is,

$$DV_i = \frac{V_o}{RC} + DV_o$$

$$\text{or, } DV_i = V_o \left( D + \frac{1}{RC} \right)$$

L.H.S.

R.H.S.

To find complementary function.

$$V_o \left[ D + \frac{1}{RC} \right] = 0$$

$$\Rightarrow D = -\frac{1}{RC}$$

$$\therefore \text{Complementary Function} = K e^{+yt}$$

$$= K \cdot e^{-\frac{t}{RC}}$$

Particular Integral:

$$\Rightarrow \frac{d}{dt} V = 0$$

$$\therefore \text{Particular Integral} = 0$$

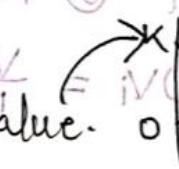
From ⑤  $\Rightarrow$

$$V_o = K e^{-\frac{t}{RC}} + 0 \quad (\text{exponential curve})$$

$\rightarrow$  we don't know the value

of "K" yet.

Curve will degrade from K value. of  $t$



$$V_o = V e^{-t/RC}$$

i) At,  $t = RC$

$$V_o = V e^{-RC/RC} = 0.3678 V$$

ii) At,  $t = 2RC$   $\Rightarrow$  to position where  $V_o$  is equal to  $V$ .

$$V_o = V e^{-2RC/RC} = 0.135 V$$

$$iii) At, t = 3RC, V_o = V e^{-3RC/RC} = 0.049 V$$

$$iv) At, t = 5RC, V_o = V e^{-5RC/RC} = 0.007 V.$$

$\therefore$  The output voltage  $V_o$  becomes practically zero after 5 time constants.

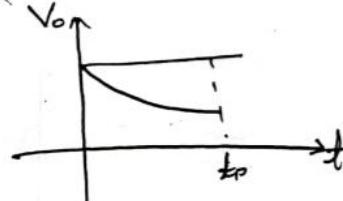
Pulse waveform as input to a HPF:

May be considered as the sum of two step wave,  $+v$  when  $t=0$  and  $-v$  when  $t=t_p$ .

There are three situations to this output:

For ① If the pulse is applied to the High Pass Filter, the response for time  $t < t_p$  is same as step voltage input.

$$V_p = V_o = V e^{-\frac{t_p}{RC}}$$



where  $t < t_p$

② At the end of the pulse, the input falls quickly by the amount  $V$ .

So, the output voltage also decreases by the same amount  $V$ .

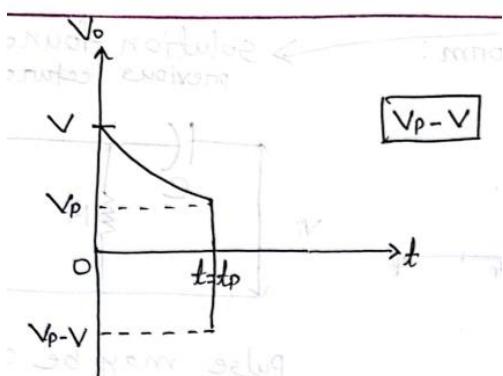
At  $t = t_p$ ,

$$V_o = V_p - V$$

where  $t = t_p$

$$\text{or, } V_o = V e^{-\frac{t_p}{RC}} - V$$

$$= \textcircled{O} V \left[ e^{-\frac{t_p}{RC}} - 1 \right]$$



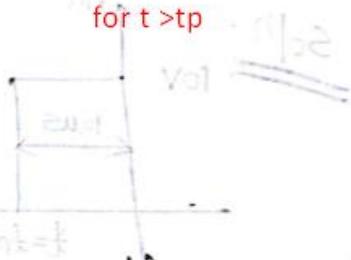
③  $t=t_p$  or  $t > t_p$

$$V_o = (V_p - V) e^{-\frac{(t-t_p)}{RC}}$$

$$= \left( V e^{-\frac{t_p}{RC}} - V \right) e^{-\frac{t-t_p}{RC}}$$

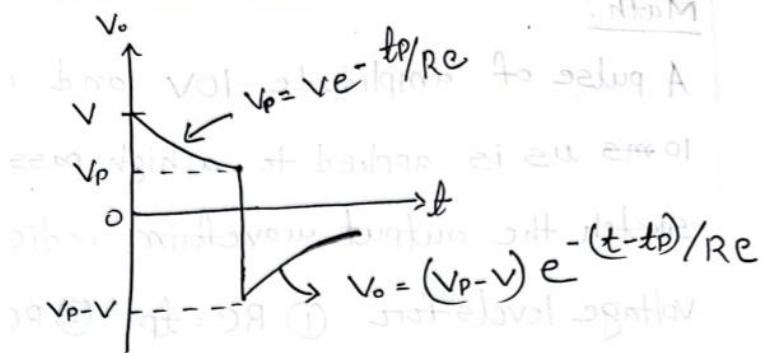
$$\frac{dt}{ds} = \frac{1}{RC} \quad ③$$

for  $t > t_p$



If we know  $t_p$ ,  $V_p$  and  $V_p - V$  then we can draw the curve

$$V_o = V e^{-\frac{t-t_p}{RC}} = V_p - (V_p - V) e^{-\frac{t-t_p}{RC}}$$



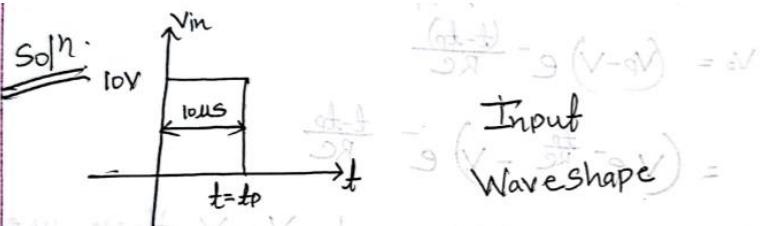
If we know  $t_p$ ,  $V_p$ , and  $V_p - V$  then we can draw the curve.

MATH:

A pulse of amplitude, 10V and duration 10ms us is applied to a high pass filter.

Sketch the output waveform indicating the voltage levels for ①  $RC = t_p$  ②  $RC = 0.5t_p$

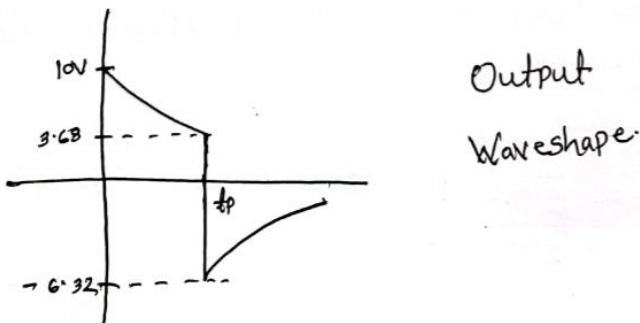
③  $RC = 2t_p$



① When  $RC = t_p = 10ms$

$$V_p = V e^{-\frac{t_p}{RC}} = 10 \cdot e^{-1} = 3.68V$$

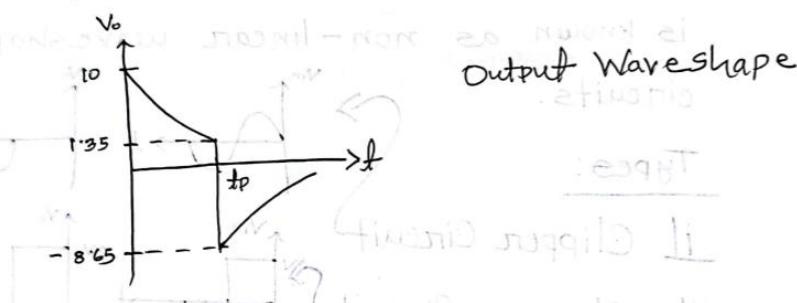
$$V_p - V = 3.68 - 10 = -6.32$$



② When  $RC = 0.5t_p$

$$V_p = V e^{-\frac{t_p}{RC}} = 10 \cdot e^2 = 1.35$$

$$V_p - V = 1.35 - 10 = -8.65$$



## **Nonlinear waveshaping:**

The circuits where the output waves are non-sinusoidal for sinusoidal input, are called non-linear waveshaping.

1. Clipper
2. Clamper

### **Application of clipper circuit:**

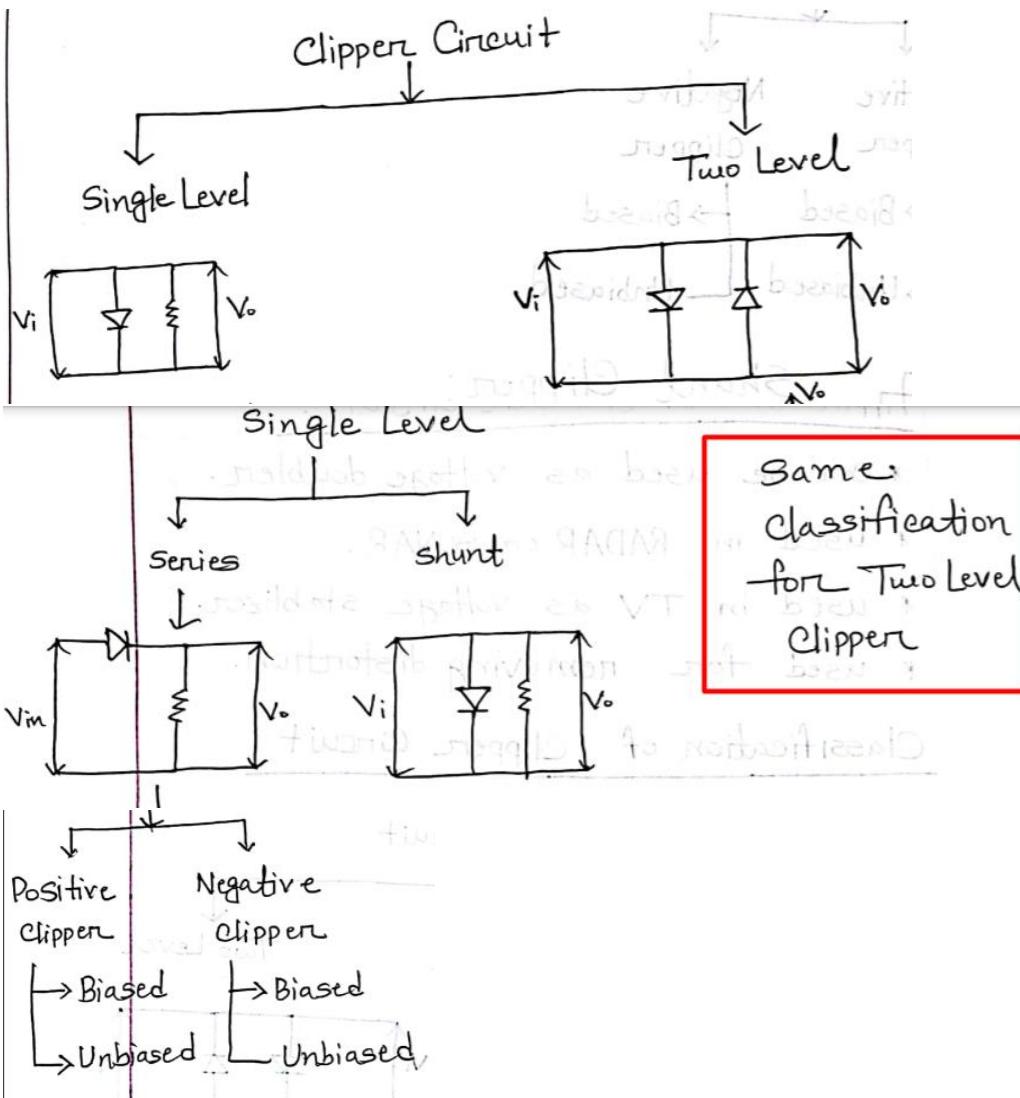
1. For shaping of the wave
2. For removing noise peak
3. For limiting the magnitude of voltage

### **Application of clamper:**

1. Can be used as voltage doubler
2. Used in SONAR and RADAR
3. Used in TV as voltage stabilizer
4. Used for removing distortion

## **Clipper:**

### **Classifications:**



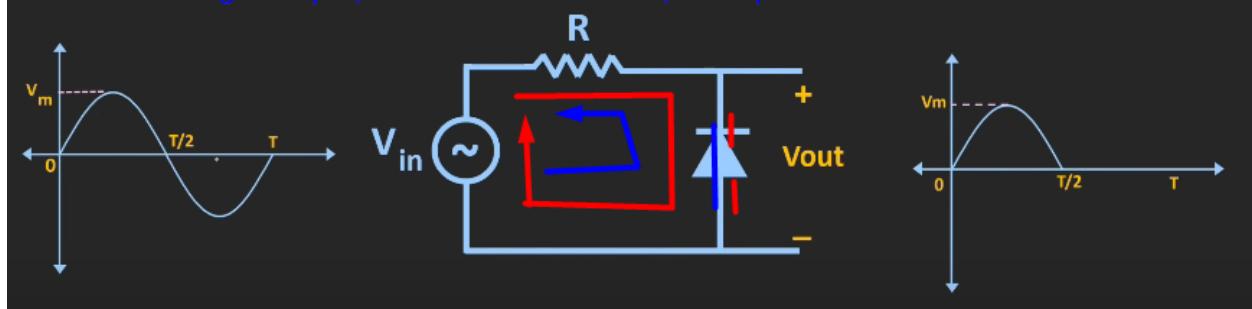
EXAMPLE:

## Parallel Clipper Circuit

when red of positive cycle, diode is like open circuit, so output points and input points are same, so  $V_o = V_i$

When negative cycle, diode is like short circuit, so output is zero

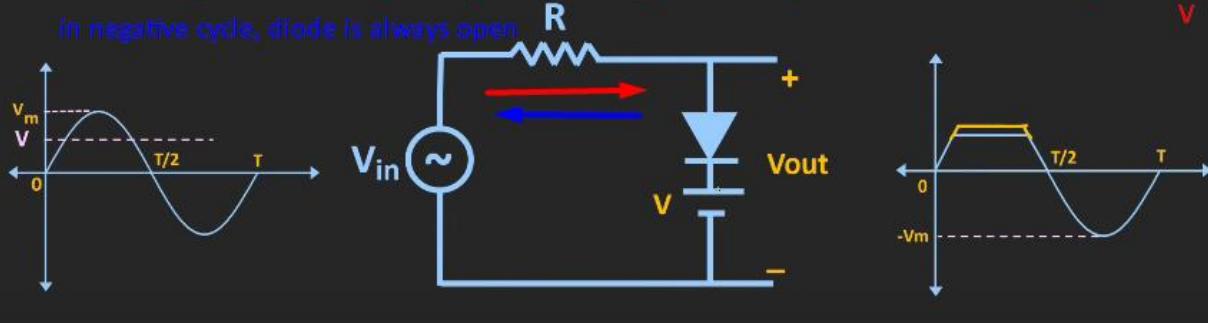
$$V_o = V_i$$



If diode voltage is given, when in forward bias, there will be a voltage drop, so output will not be exact zero, it will be  $-V_d$ .

## Clipper Circuit with Biasing Voltage

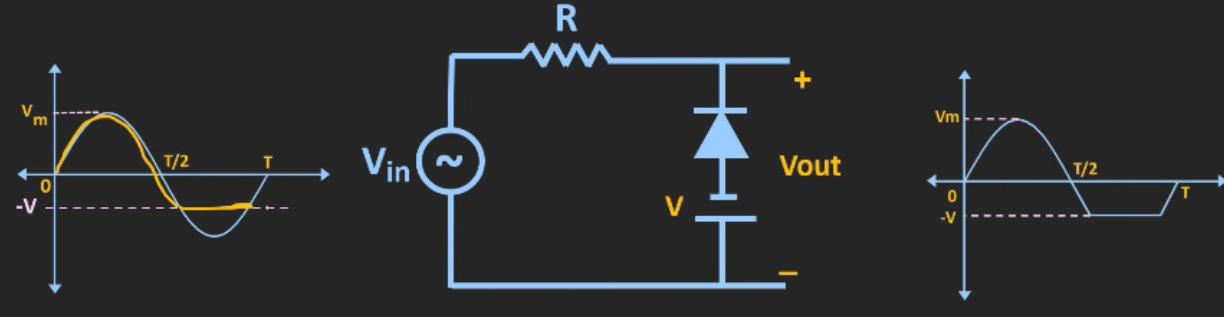
when  $V_{in}$  keeps forward bias,  $V$  keeps reverse bias, until  $V_{in} > V$ , after that diode is short, so  $V_{out}$  is  $V$ .  
in negative cycle, diode is always open.



While biasing in clipper, the  $V$  is always makes the diode reverse biased if wavy shape is on axis.

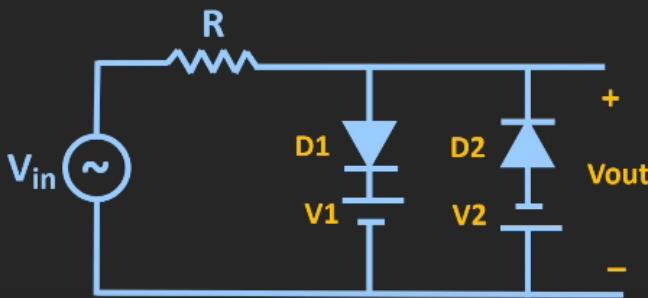
Here is Negative biased clipper:

## Clipper Circuit with Biasing Voltage

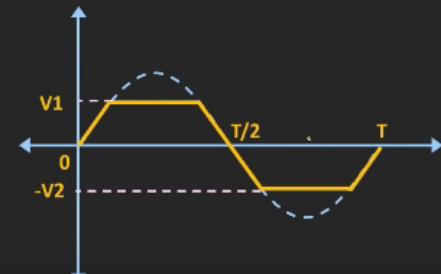


Two ways:

## Clipper Circuit with Biasing Voltage



In positive cycle, D1 is shorted after  $V_{in} > V_1$ , D2 is always open. Vice versa for negative cycle.

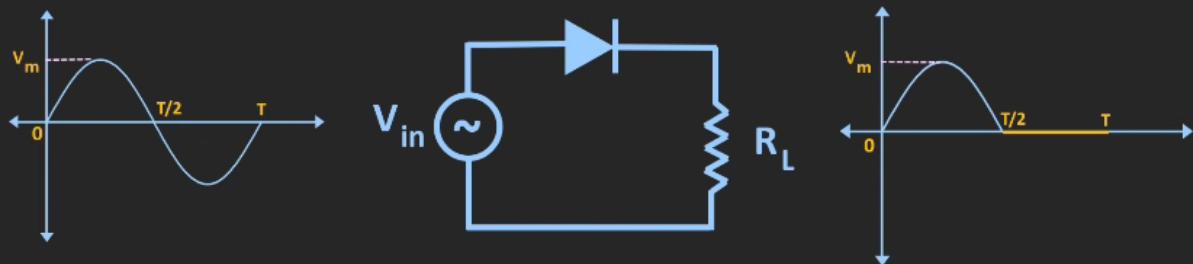


**Series Clippers:**

Are simple, one cycle goes other doesn't.

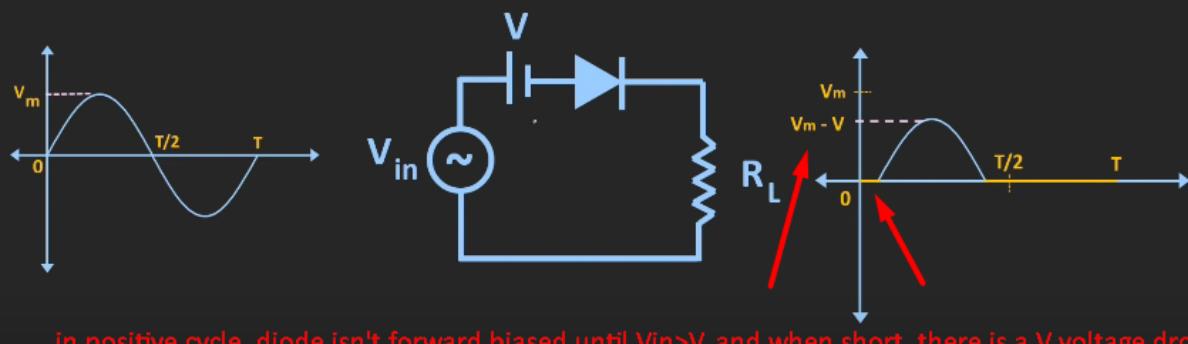
## Series Clipper Circuit

i



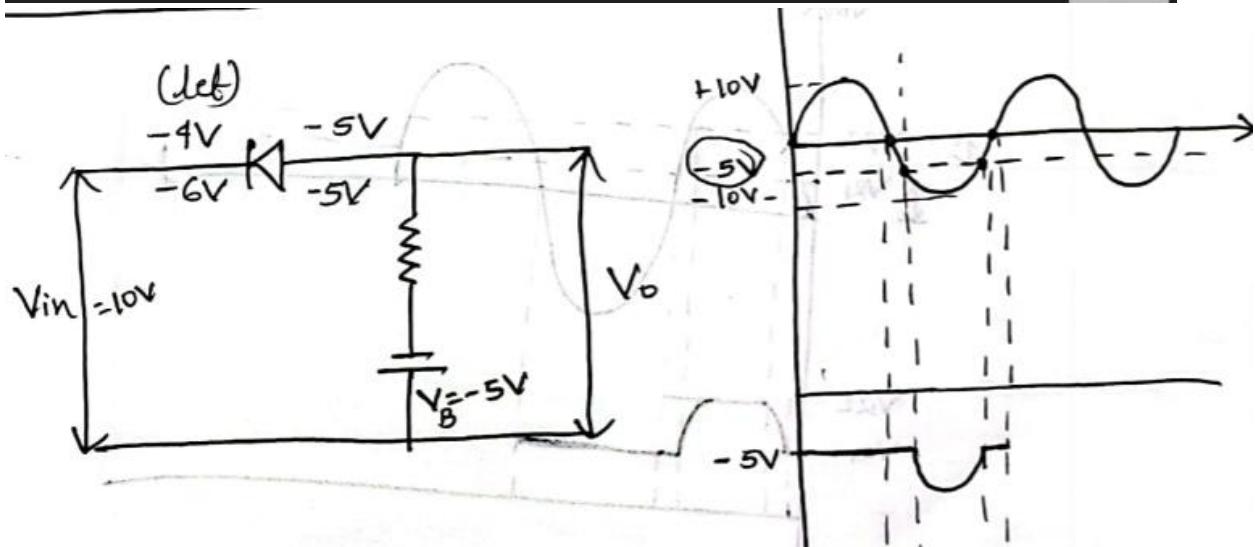
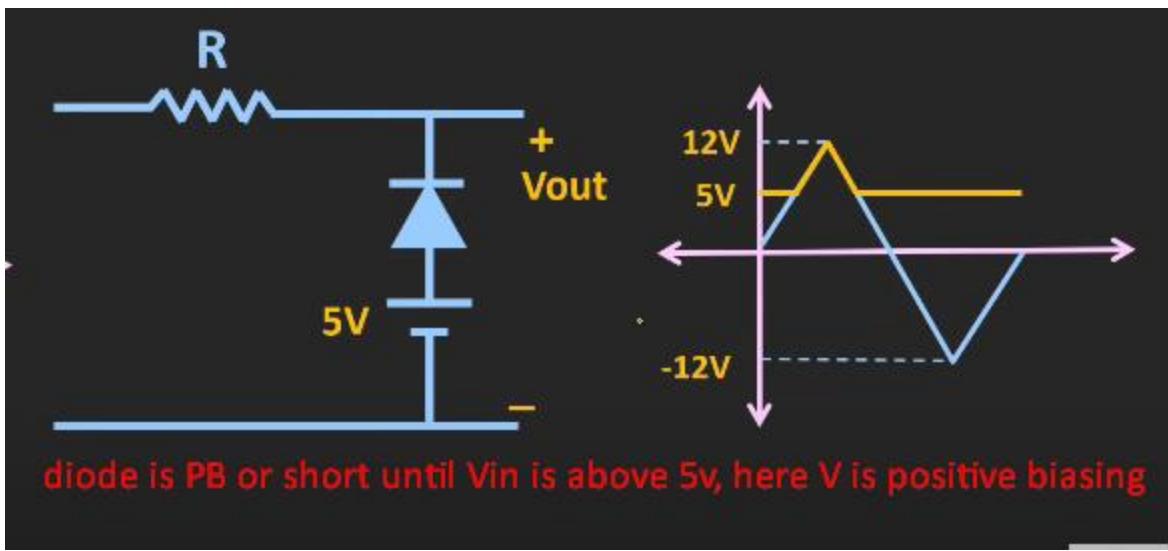
## Series Clipper Circuit with Biasing Voltage

i



in positive cycle, diode isn't forward biased until  $V_{in} > V$ , and when short, there is a  $V$  voltage drop



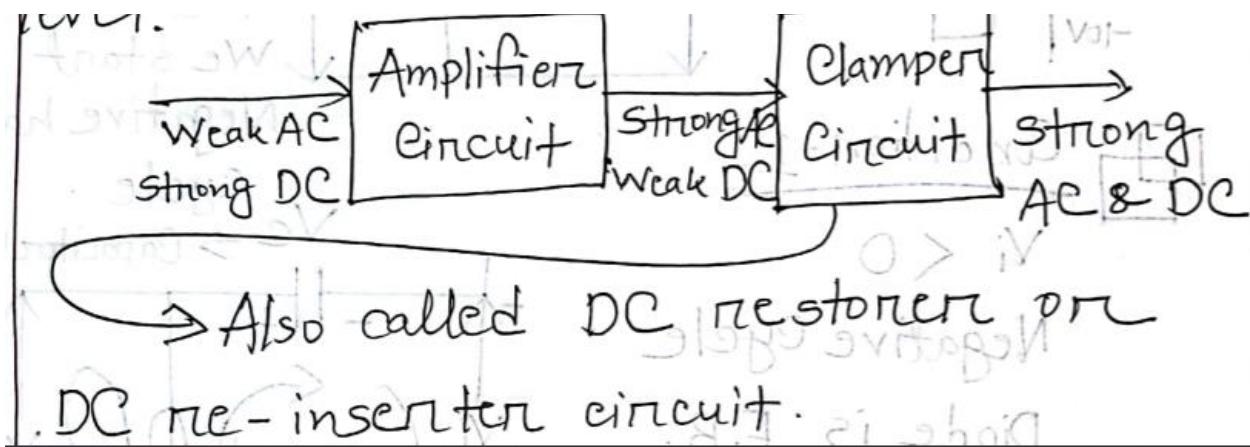


Here output is always -5v until there is a higher voltage source, meaning, when  $V_{in} < -5v$

### Clamper:

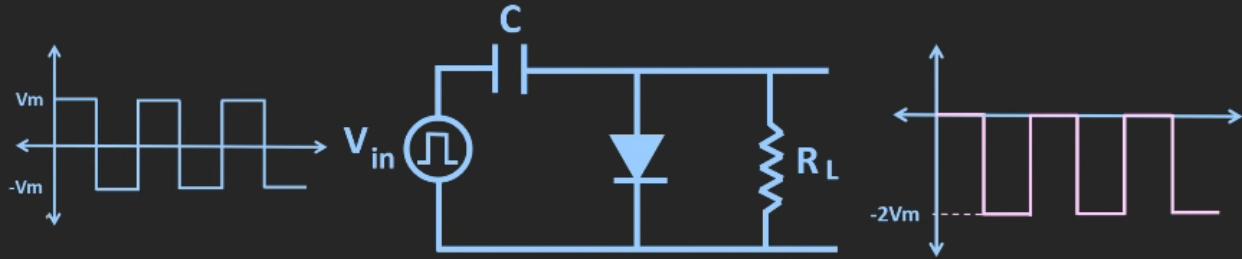
A clamping circuit is one which takes an input waveform and provides a faithful replica of its shape, but has one edge tightly clamped to zero voltage reference point.

**Why use:** Clamping is used to introduce a DC component to a signal after going through a capacitor coupling.



EXAMPLE:

## Negative Clamper Circuit

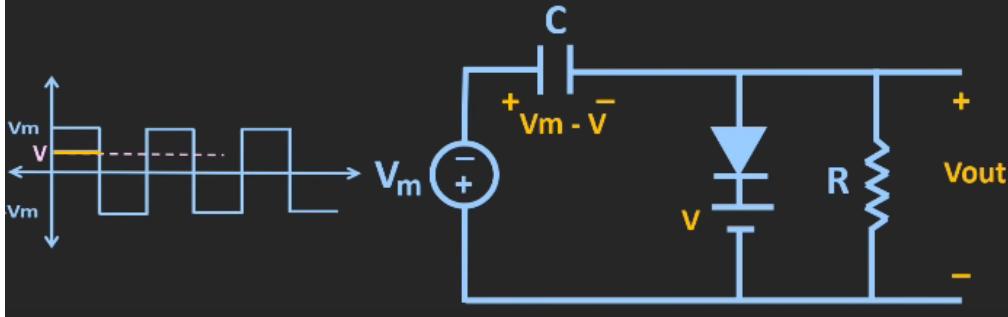


In positive cycle, diode is shorted, so output is zero.  $C$  is charged to  $V_m$ .

In negative cycle, diode is open, so output is input + capacitor discharge, meaning  $-2V_m$

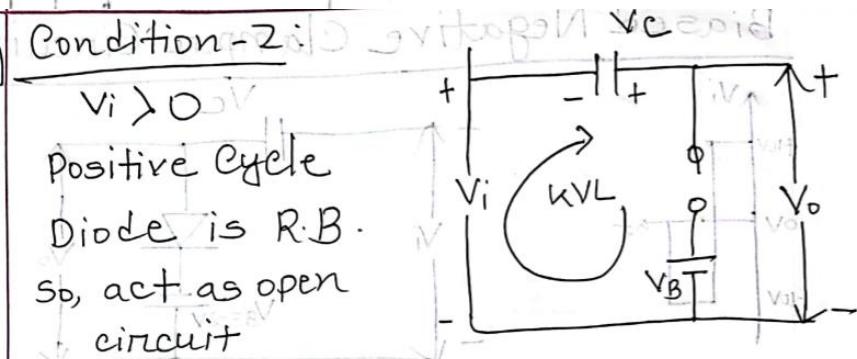
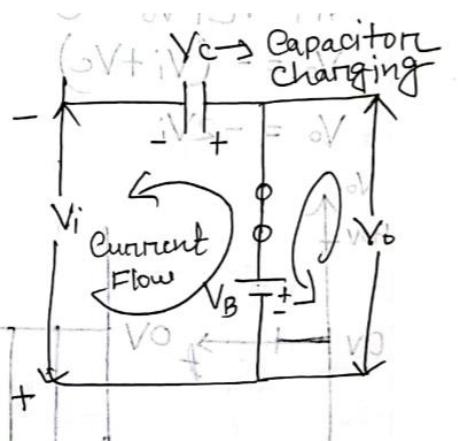
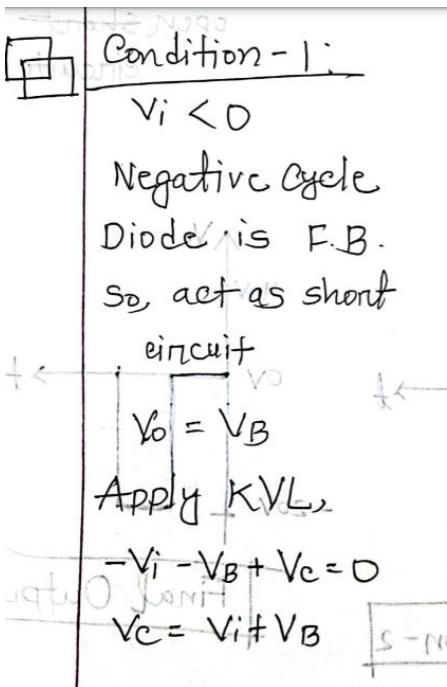


## Clamper Circuit with Biasing Voltage



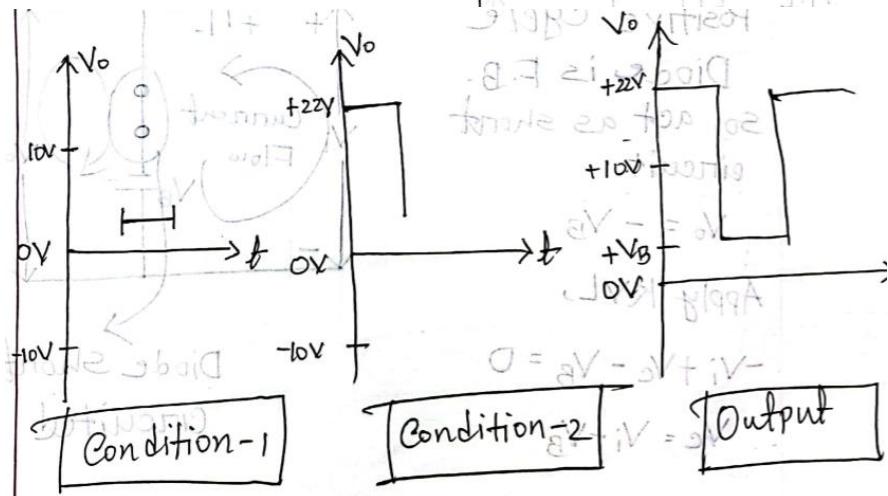
Here, the capacitor is charged to  $V_m - V$  in positive half cycle, and diode is shorted, so output is  $V$ . In negative half, diode is open. So output is input + discharge. So  $-2V_m + V$ .

Calculation is given below.



Apply KVL, ~~At the positive note~~  
 $-V_i - V_c + V_o = 0$   
 or,  $V_o = V_i + V_c$   
 $\therefore V_o = 2V_i + V_B$

Condition 1  
Condition 2



## Switching circuit:

A circuit which can turn on or off in an electrical circuit is known as a switching circuit.

Has two parts:

1. A switch:
  - a. Mechanical switch
  - b. Electro-mechanical switch

c. Electronic switch

2. Associated circuitry

**Characteristics of different switches:**

**Electro-mechanical switch:**

A mechanical switch that is operated electrically.

Pros:

- Requires very small amount of current
- Can turn on or off from a distance
- No danger of sparking

Cons:

- Has moving parts, thus has friction
- Less than 5 operations per second at maximum.

**Electrical switches:**

Switches that can turn on or off electrically with help of diode, transistors etc.

Pros:

- Cheaper and smaller
- No noise and friction
- $10^9$  operations per second
- Trouble free service because of solid state

**Diode as switching device:**

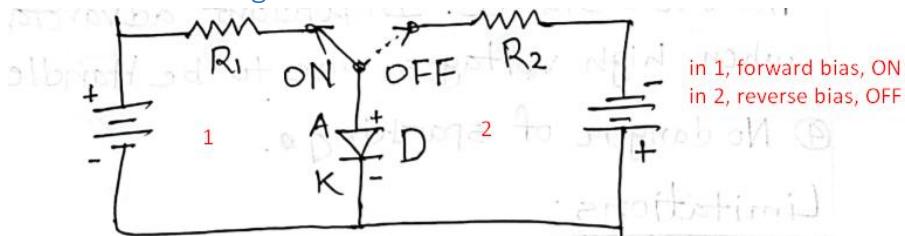
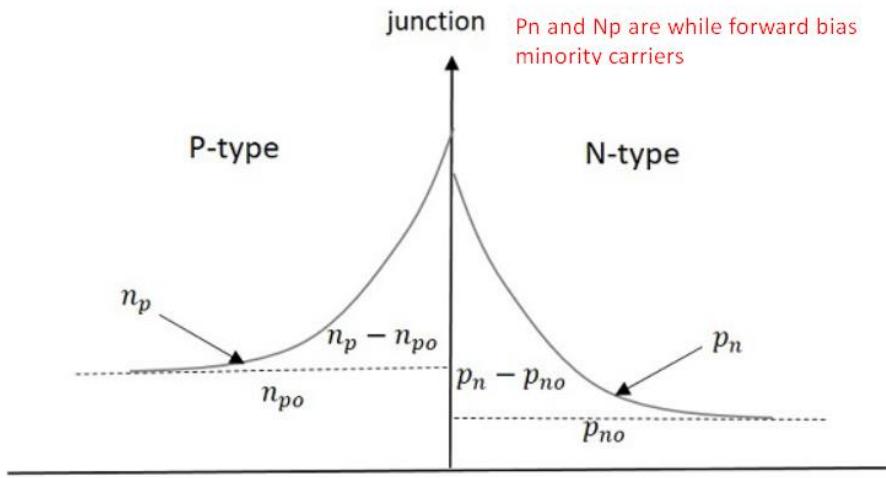


Fig: switching circuit using diode.

If,

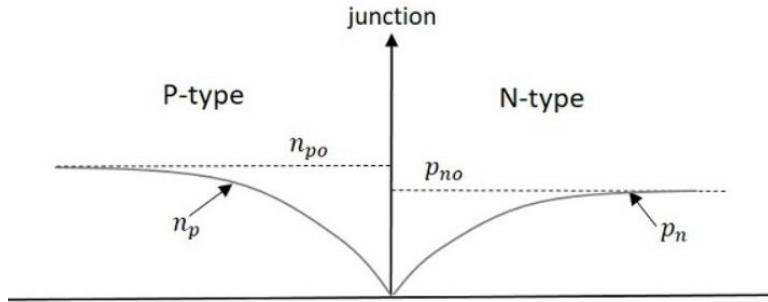
- The majority carriers in P-type holes =  $P_{po}$
- The majority carriers in N-type electrons =  $N_{no}$
- The minority carriers in P-type electrons =  $N_{po}$
- The majority carriers in N-type holes =  $P_{no}$

**During forward bias:**

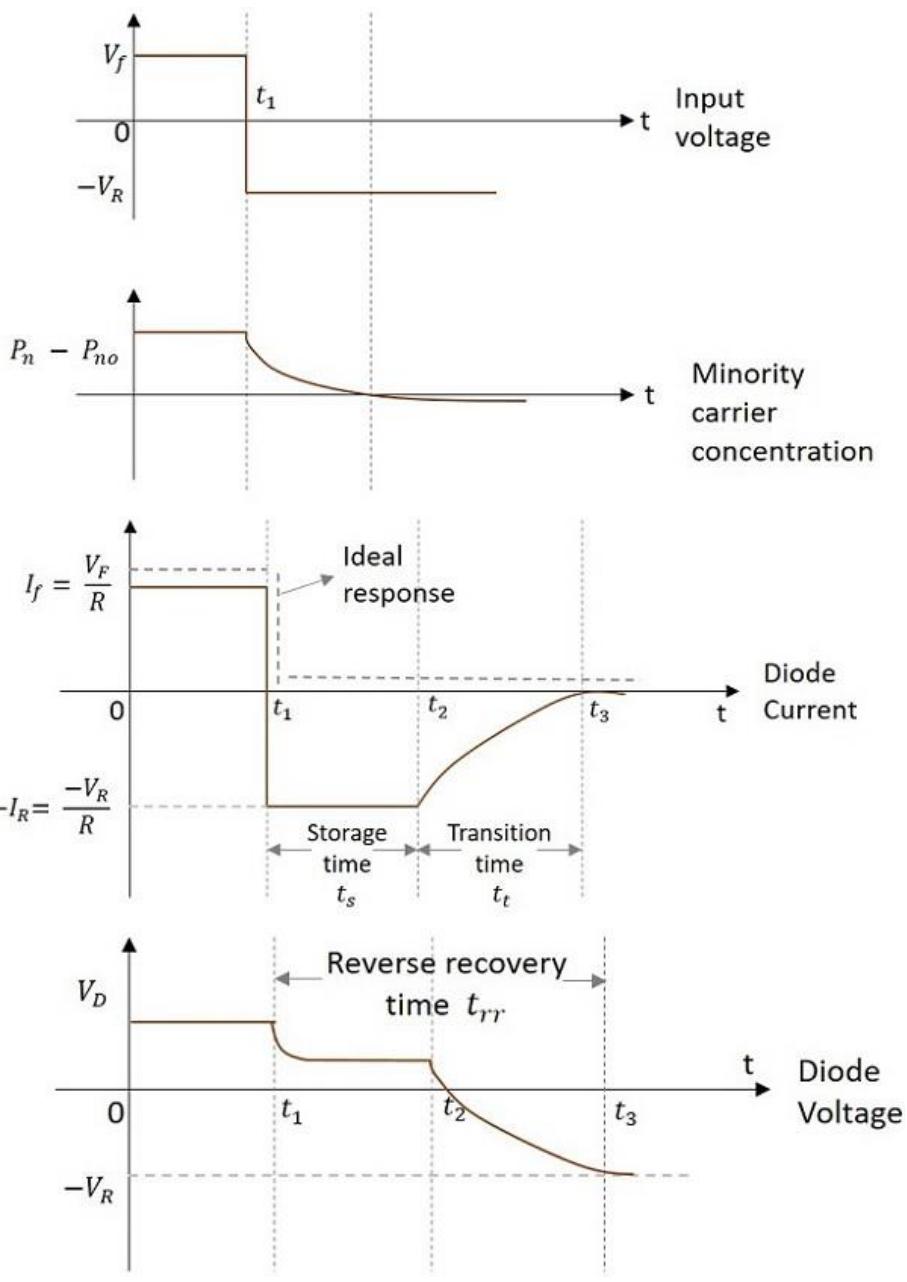


Number of minority carriers increase. So, Excess minority carrier charge in P-type =  $P_n - P_{no}$  with  $P_{no}$  steady state value, Excess minority carrier charge in N-type =  $N_p - N_{po}$  with  $N_{po}$  steady state value

Then after giving **Reverse bias**, opposite happens:



As the current due to minority charge carriers is large enough to conduct, the circuit will be ON until this excess charge is removed.



In minority carrier concentration graph, because of the excess minority charge, diode doesn't immediately turn off.

So for diode current, ideal would be zero current immediately, but instead goes negative. For storage time.

Diode voltage goes to negative infinity to become an open circuit, but needs Reverse recovery time which is storage time + transition time.

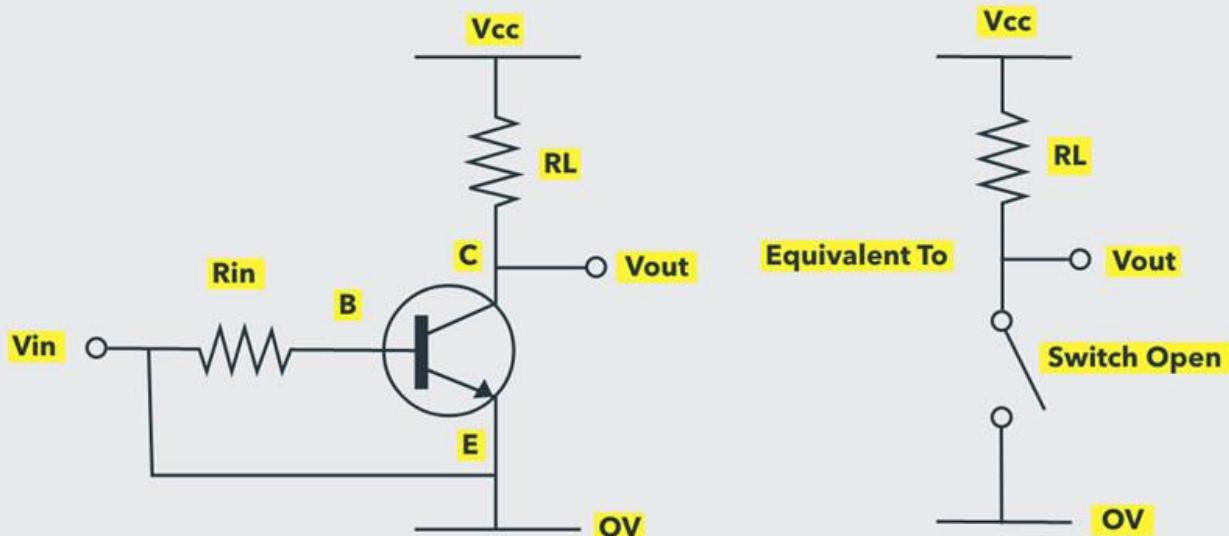
**Storage time** – The time period for which the diode remains in the conduction state even in the reverse biased state, is called as Storage time.

**Transition time** – The time elapsed in returning back to the state of non-conduction, i.e. steady state reverse bias, is called Transition time.

**Reverse recovery time** – The time required for the diode to change from forward bias to reverse bias is called as Reverse recovery time.

Transistor as switching device:

### NPN Transistor In Cut-Off Region

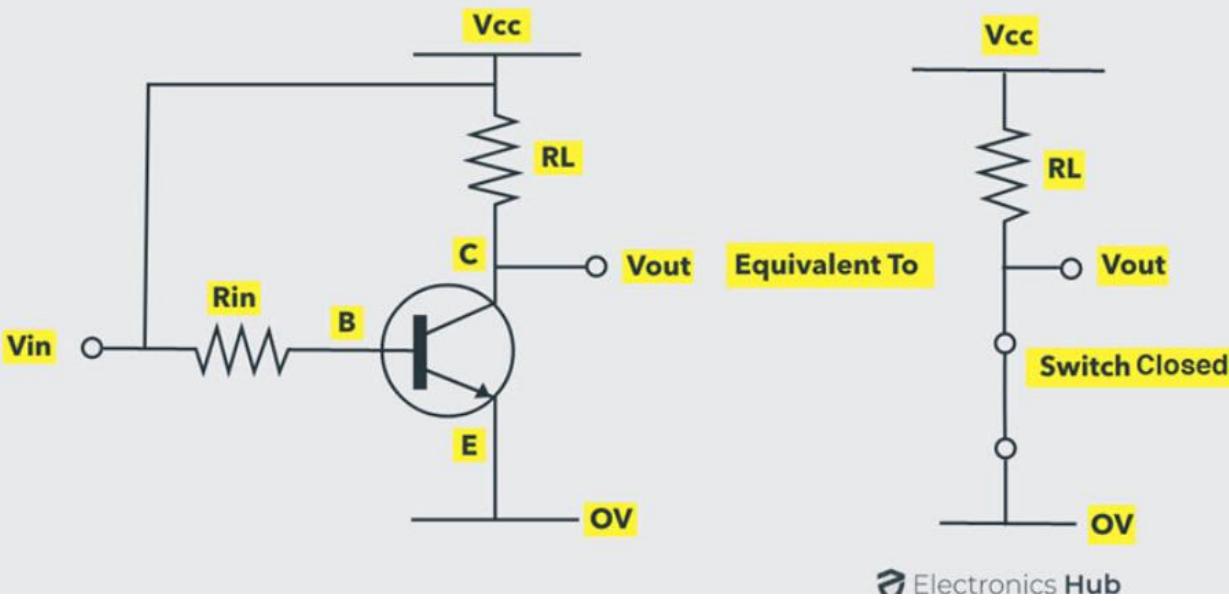


Electronics Hub

Here,  $I_c$  or collector current is zero.  $V_{cc} = V_{ce}$

Both P-N junctions are reverse biased.

### NPN Transistor In Saturation Region



Electronics Hub

Here,  $I_c = V_{cc}/R_L$

Both P-N junctions are forward biased.

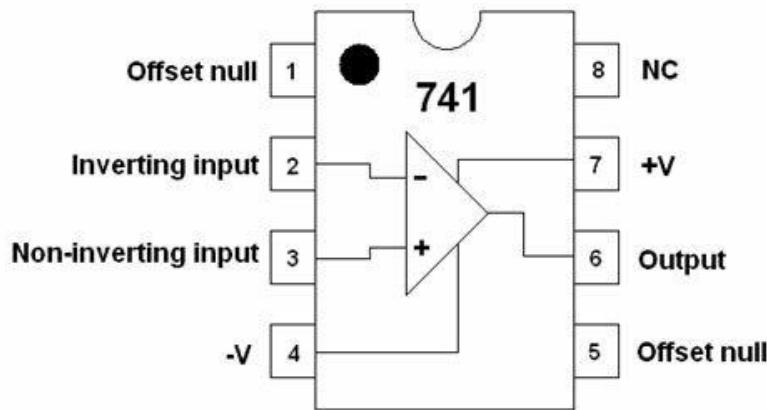
OP-AMP as comparator:

OP-AMP Usages:

1. Amplification
2. Math operation

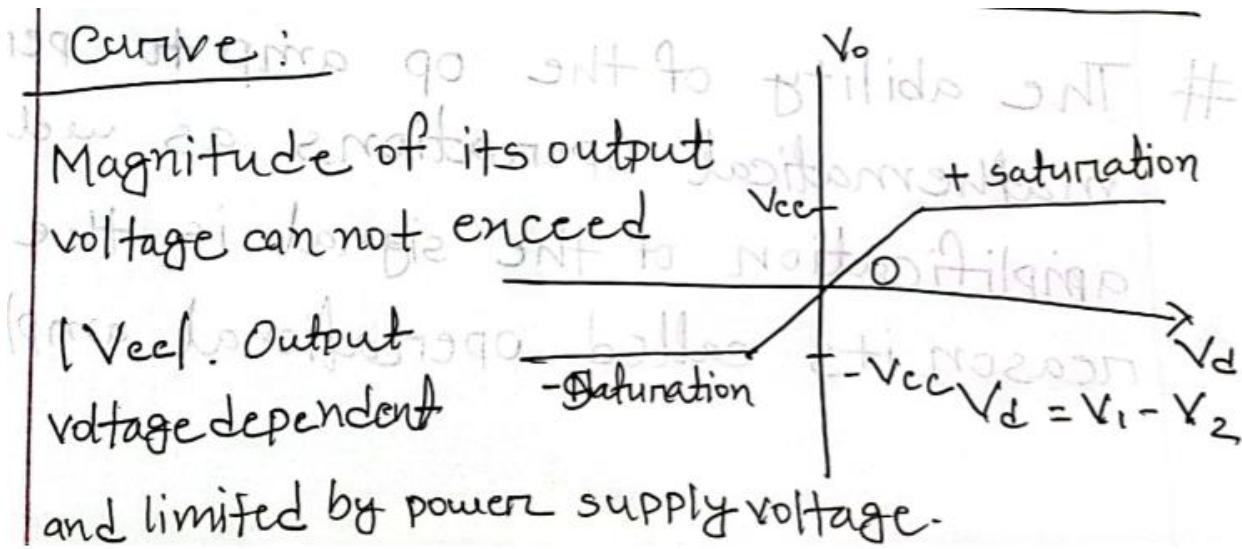
### 3. Signal shaping (Comparator)

Named OP-AMP from OP(mathematical operation) – AMP(amplification)



-V and +V are biasing voltage

**Characteristic curve:**



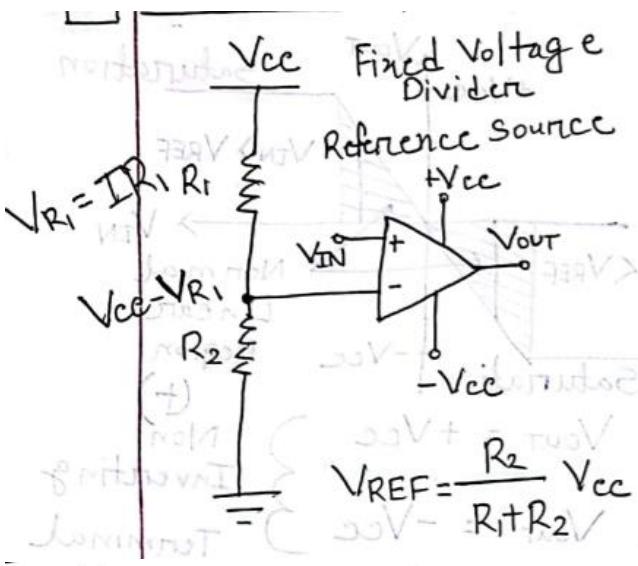
**Comparator:**

Basically, Output is High/Low if input is higher than a reference voltage, and Low/High if lower than reference voltage. (Non-inverting/ Inverting)

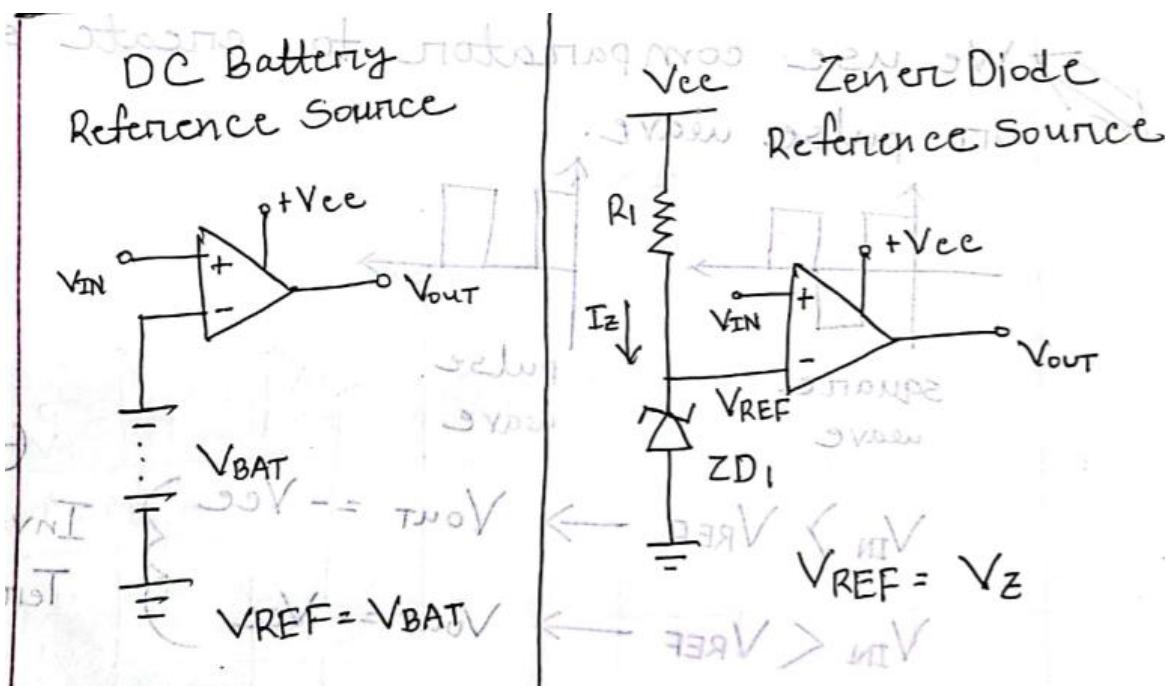
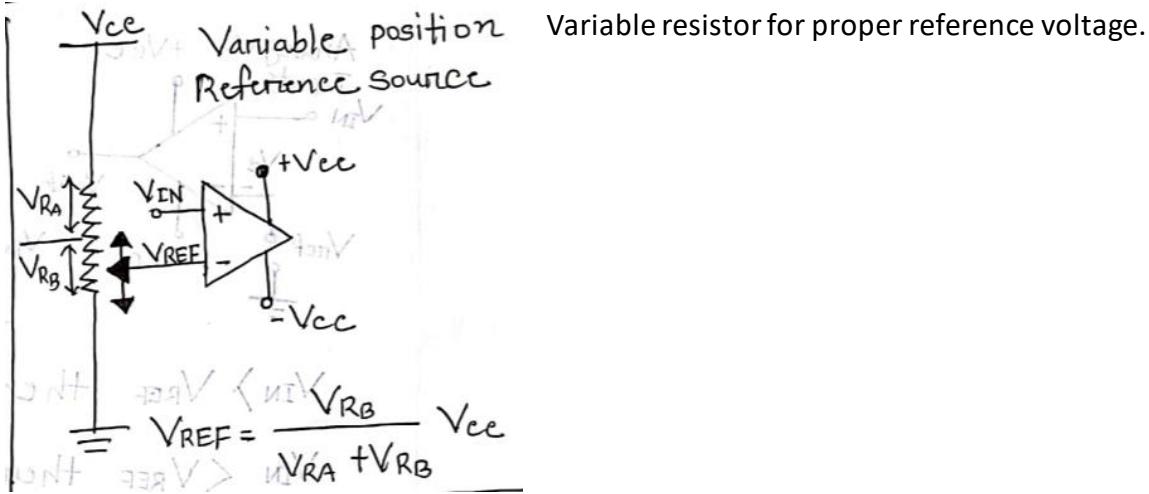
Comparator cause compares with reference voltage.

Comparator is used to generate square or pulse waves.

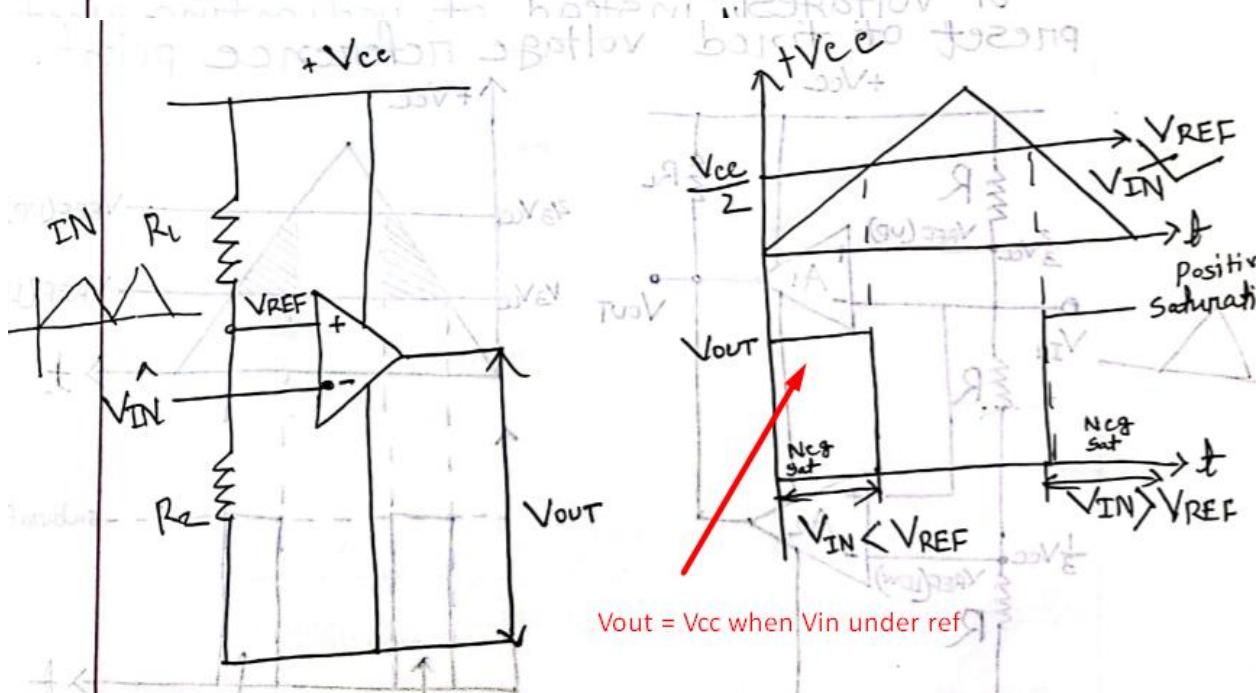
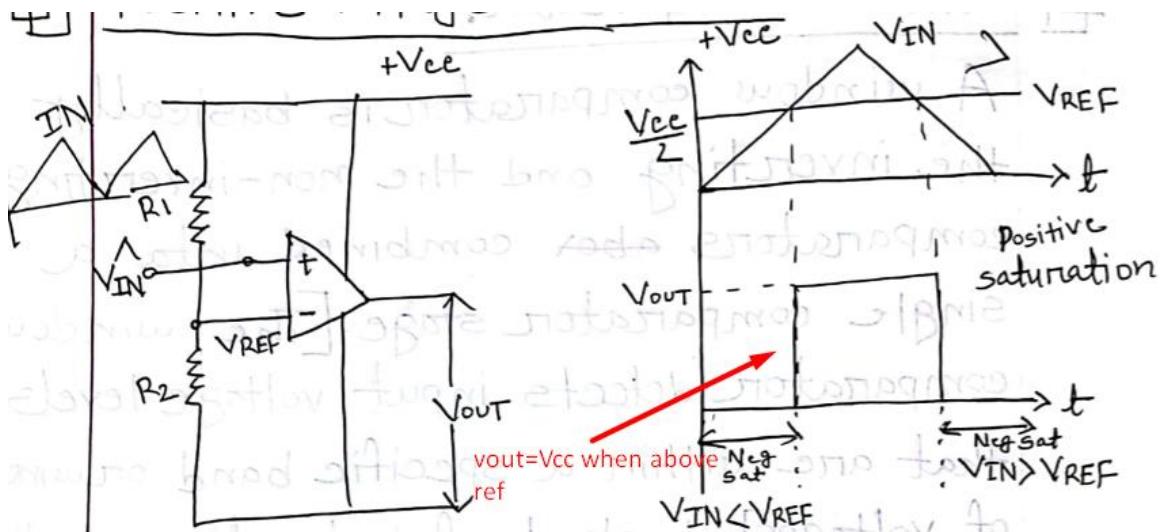
**Comparator reference voltage methods:**



$R_1$  and  $R_2$  are controlled to give proper reference voltage as input.

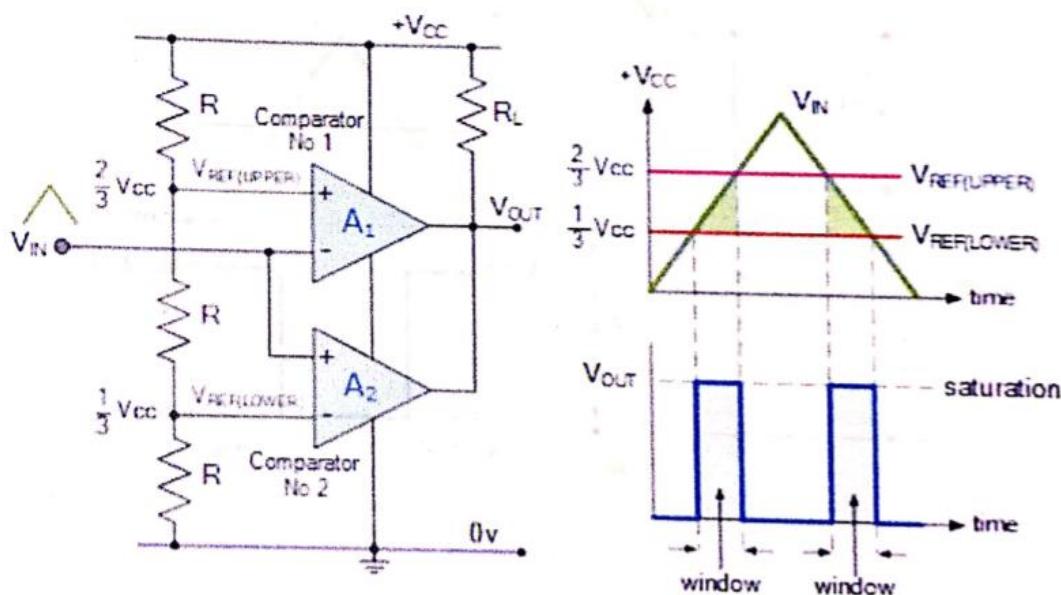


Positive and negative voltage comparator:



### Window comparator:

Window comparator is when two comparators, one negative and one positive comparator is used to detect voltage input within a window of voltages.



### Negative Comparator (Comparator 1)

$V_{IN} < V_{REF} \implies V_{OUT} = \text{Positive Saturation}$

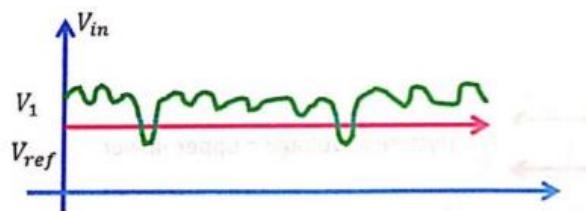
$V_{IN} > V_{REF} \implies V_{OUT} = \text{Negative Saturation}$

### Positive Comparator (Comparator 2)

$V_{IN} > V_{REF} \implies V_{OUT} = \text{Positive Saturation}$

$V_{IN} < V_{REF} \implies V_{OUT} = \text{Negative Saturation}$

### Problem with OP-AMP comparator:

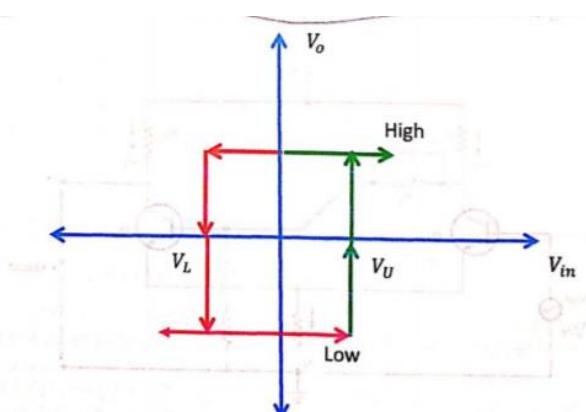


Because of noise, OP-AMP can sometimes get false triggered, creating problems. To solve this, something with a threshold for noise should be used, like Schmitt trigger.

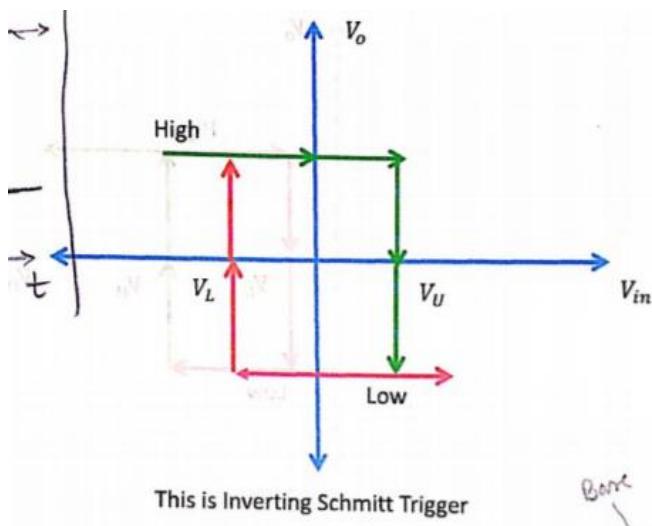


### Transfer characteristics graph:

When  $V_{in}$  is lower than  $V_{low}$ ,  $V_o$  is low, then when  $V_{in}$  is higher than  $V_{up}$ ,  $V_o$  is high. For inverting, its just reversed.

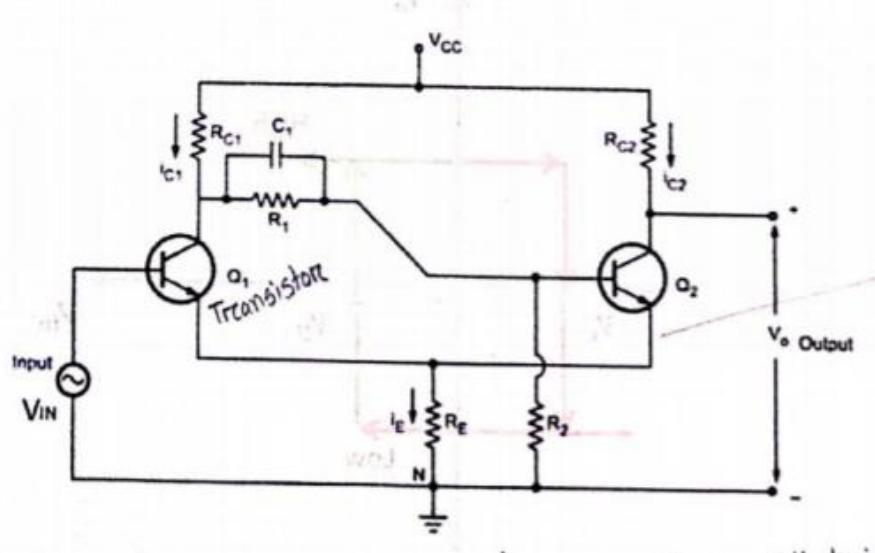


This is non-inverting Schmitt Trigger



### Schmitt trigger using transistors:

Schmitt triggers work like window comparator by providing upper and lower thresholds before counting input as a trigger.



- When  $V_{IN}$  is zero,  $V_{CC}$  turns on  $Q_2$ , so  $Q_2$  is saturated thus works as short circuit, thus  $V_o$  being low or zero.
- There is a voltage across  $R_E$  through the  $R_{C2} \rightarrow R_E \rightarrow$  ground path. It is denoted by,

$$V_{RE} = I \cdot R_E = \frac{V_{CC}}{(R_{C2} + R_E)} \times R_E$$

*nsitors*

- Which is also the emitter voltage of  $Q_1$ , so to turn on  $Q_1$ ,  $V_{IN}$  should be  $V_{RE} + 0.7$
- When  $V_{IN}$  turns on  $Q_1$ , as long the voltage is higher, output is high. The  $V_{RE}$  has a new value now,

$$V_{IN} = (V_{CC} \times R_E) / (R_E + R_{C1}) + 0.7$$

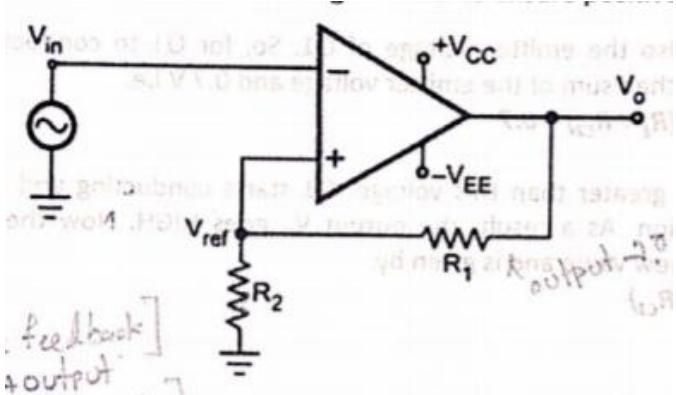
This is how it works, the lower threshold and upper threshold is denoted by,

$$V_{LT} = (V_{CC} \times R_E) / (R_E + R_{C1}) + 0.7$$

$$V_{UT} = (V_{CC} \times R_E) / (R_E + R_{C2}) + 0.7$$

## OP-AMP based Schmitt trigger:

### Inverting Schmitt trigger circuit:



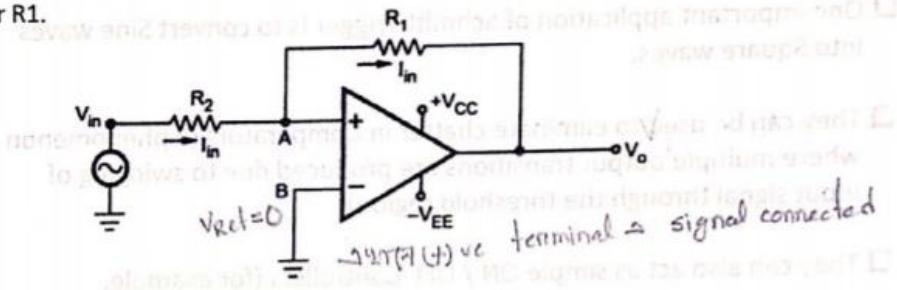
$$V_{REF} = (V_{SAT} \times R_2) / (R_1 + R_2)$$

$$\text{e.g., } -V_{REF} = (-V_{SAT} \times R_2) / (R_1 + R_2)$$

### Non inverting Schmitt trigger:

#### Non-Inverting Schmitt Trigger Circuit

Coming to Non-Inverting Schmitt Trigger, the input in this case is applied to the non-inverting terminal of the Op-Amp. The output voltage is fed back to the non-inverting terminal through the resistor  $R_1$ .

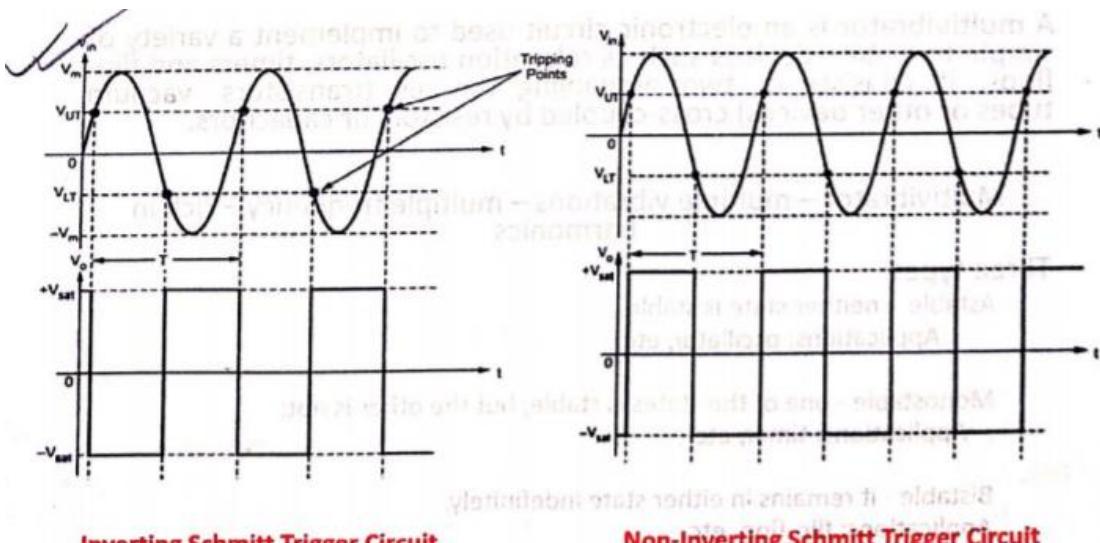


Let us assume that initially, the output voltage is at  $V_{SAT}$ . Until  $V_{IN}$  becomes less than  $V_{LP}$ , the output stays at this saturation level. Once the input voltage crosses the lower threshold voltage level, the output changes state to  $-V_{SAT}$ .

The output remains at this state until the input rises beyond the upper threshold voltage.

### OUTPUT WAVES:

1. A threshold is introduced by using the  $V_o$  as a positive feedback.
2.  $V_o$  is either  $+V_{CC}$  or  $-V_{EE}$ . That is taken and then controlled using  $R_1$  and  $R_2$  to use as  $V_{ref}$ .
3. When  $+V_{CC}$  is  $V_o$ , it is upper threshold and vice versa.



Inverting Schmitt Trigger Circuit

Non-Inverting Schmitt Trigger Circuit

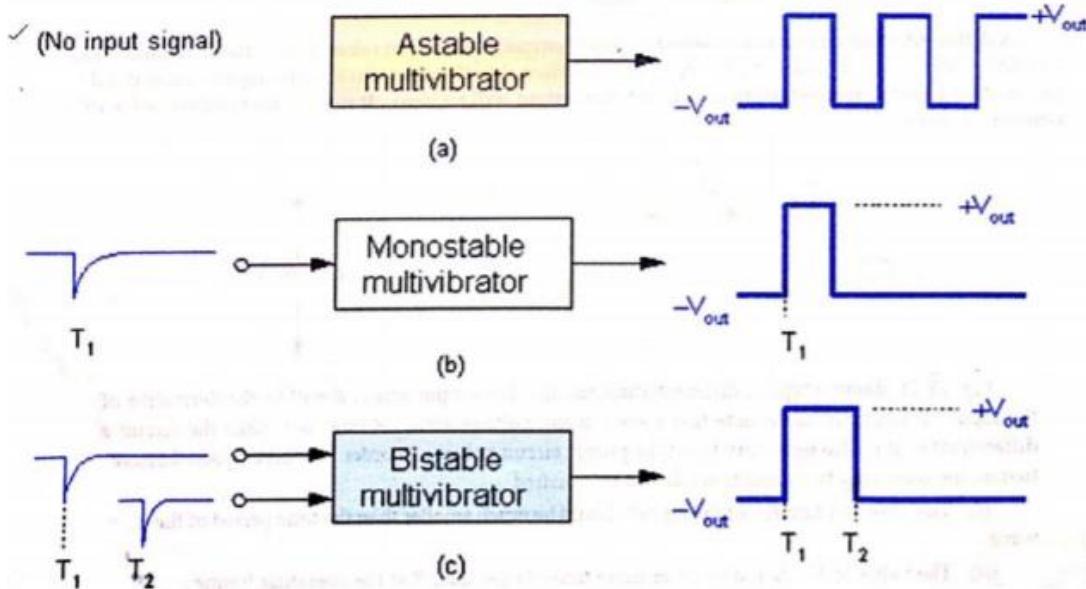
## Multi-Vibrator:

A multivibrator is an IC that is used to create different two state devices. It consists of two amplifying devices that are cross coupled.

Three types:

1. Astable – Oscillator
2. Monostable – Timer
3. Bistable – flipflop

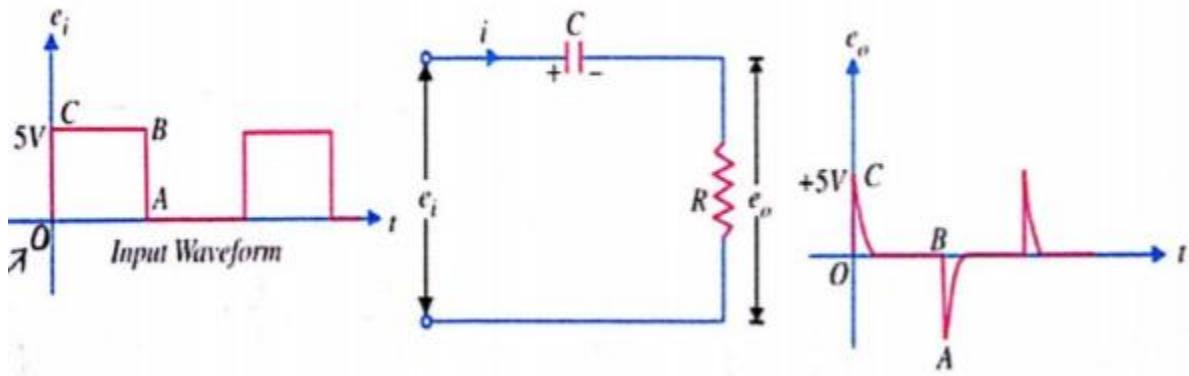
## Waveshapes:



## Trigger pulse generation:

A circuit where the output is derivative of the input

$$V_o = d/dt(V_{in})$$

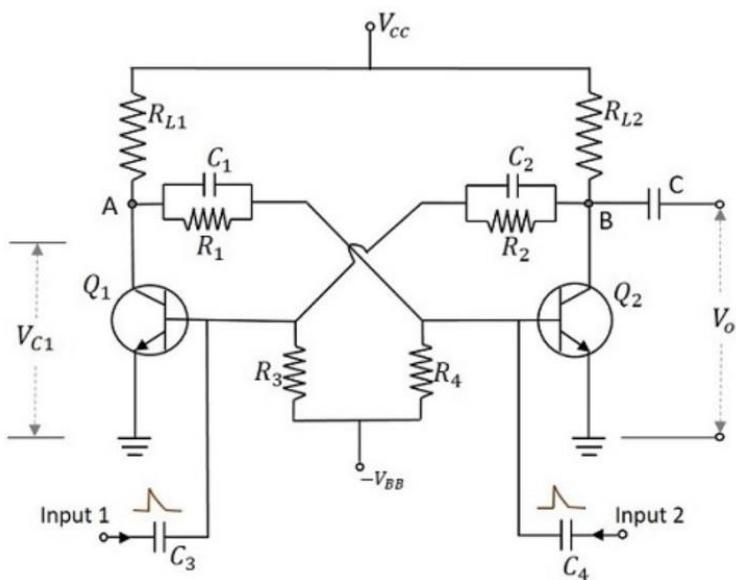


## Requirements:

1. RC circuit time constant  $.69RC$  should be much smaller than input wave period.
2. Capacitive resistance should be 10 or more times higher than  $R$ .

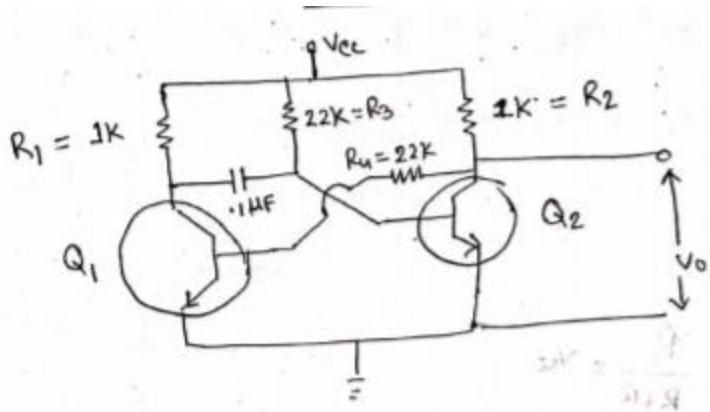
## Bistable:

### What is Bistable Multivibrator



1. When turned on, one of either of the transistors are on, lets say  $Q_2$ .
2. When given trigger to  $Q_1$ , collector current of  $Q_1$  is quickly increased, and collector voltage decreases.
3. Which is applied to the  $Q_2$  base, thus decreasing its collector current and increase base voltage, which is then applied to  $Q_1$  base, thus creating a loop. Eventually collector current of  $Q_2$  is zero and  $Q_1$  is on.

## Monostable:



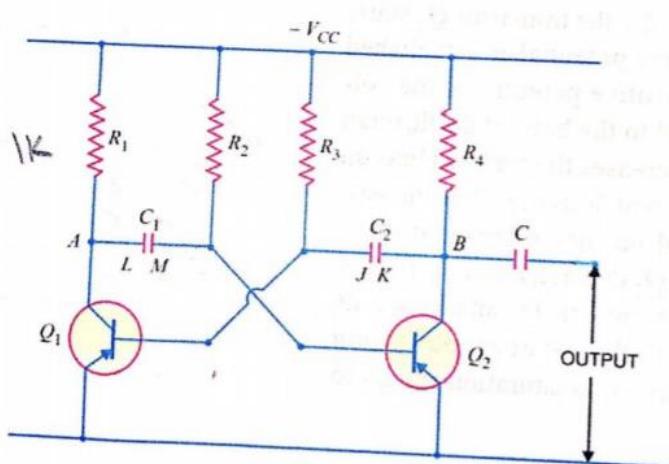
When applied initial  $V_{cc}$ ,  $Q_1$  is off so goes  $R_3 > Q_2$  Base, which is the lowest resistance path.

$Q_2$  is turned on, so  $V_{cc}$  now goes through  $R_2$  to ground. So  $V_o$  is low as short circuit. Capacitor is being charged.

When applied trigger,  $Q_1$  is turned on, so  $V_{cc}$  goes through  $R_1$  to ground.  $Q_2$  is off. Then after a time capacitor is discharged, turning on  $Q_2$ , which in result turns off  $Q_1$ .  $Q_2$  is stable state transistor.

### Astable:

It is the non-stable part of Monostable x2. Doesn't need any trigger to switch state.



$V_{cc}$  first turns on either, let say  $Q_1$ .

$C_1$  is being charged. After some time,  $C_1$  discharges, turning on  $Q_2$  and starting charging of  $C_2$ .

The on off time is determined by RC constant.

~~ON or OFF time.~~ The time for which either transistor remains ON or OFF is given by :

ON time for  $Q_1$  (or OFF time for  $Q_2$ ) is

$$T_1 = 0.694 R_2 C_1$$

OFF time for  $Q_1$  (or ON time for  $Q_2$ ) is

$$T_2 = 0.694 R_3 C_2$$

Total time period of the square wave is

$$T = T_1 + T_2 = 0.694 (R_2 C_1 + R_3 C_2)$$

As  $R_2 = R_3 = R$  and  $C_1 = C_2 = C$ ,

$$\therefore T = 0.694 (RC + RC) \approx 1.4 RC \text{ seconds}$$

Frequency of the square wave is

$$f = \frac{1}{T} \approx \frac{0.7}{RC} \text{ Hz}$$