

Exercise 6.19

I Answer (b).

From circuit analysis, we have

$$v_0(t) = Ri(t) + \frac{1}{C} \int_{-\infty}^t i(\tau) d\tau + v_1(t) \quad \text{and} \quad (1)$$

$$i(t) = \frac{1}{L} \int_{-\infty}^t v_1(\tau) d\tau. \quad (2)$$

Taking the derivative of the preceding two equations (in order to eliminate the integral operators which would cause difficulties later), we have

$$(3) \quad \mathcal{D}v_0(t) = R\mathcal{D}i(t) + \frac{1}{C}i(t) + \mathcal{D}v_1(t) \quad \text{and} \quad \text{derivative of (1)}$$

$$(4) \quad \mathcal{D}i(t) = \frac{1}{L}v_1(t). \quad \text{derivative of (2)}$$

Taking the Fourier transform of these two equations, we obtain

$$(5) \quad j\omega V_0(\omega) = jR\omega I(\omega) + \frac{1}{C}I(\omega) + j\omega V_1(\omega) \quad \text{and} \quad \text{take FT of (3)}$$

$$j\omega I(\omega) = \frac{1}{L}V_1(\omega) \Rightarrow I(\omega) = \frac{1}{jL\omega}V_1(\omega). \quad \text{take FT of (4) and solve for I}$$

Combining the preceding two equations, we obtain

$$j\omega V_0(\omega) = jR\omega \left[\frac{1}{jL\omega}V_1(\omega) \right] + \frac{1}{C} \left[\frac{1}{jL\omega}V_1(\omega) \right] + j\omega V_1(\omega) \Rightarrow \text{Substitute (6) into (5)}$$

$$j\omega V_0(\omega) = \frac{R}{L}V_1(\omega) + \frac{1}{jLC\omega}V_1(\omega) + j\omega V_1(\omega). \quad \text{Simplify}$$

Rearranging the preceding equation, we have

$$[j\omega]V_0(\omega) = \left[\frac{R}{L} + \frac{1}{jLC\omega} + j\omega \right] V_1(\omega) \Rightarrow \text{factor out } V_0 \text{ and } V_1$$

$$[j\omega]V_0(\omega) = \left[\frac{jRC\omega + 1 - LC\omega^2}{jLC\omega} \right] V_1(\omega). \quad \text{common denominator}$$

Solving for $\frac{V_1(\omega)}{V_0(\omega)}$, we obtain

$$\frac{V_1(\omega)}{V_0(\omega)} = j\omega \left[\frac{jLC\omega}{jRC\omega + 1 - LC\omega^2} \right] \quad \text{multiply both sides by } \left(\frac{jLC\omega}{jRC\omega + 1 - LC\omega^2} \right) \left(\frac{1}{V_0(\omega)} \right)$$

$$= \frac{-LC\omega^2}{-LC\omega^2 + jRC\omega + 1} \quad \text{multiply}$$

$$= \frac{LC\omega^2}{LC\omega^2 - jRC\omega - 1}. \quad \text{flip sign of numerator and denominator}$$

Since the system is LTI, $H(\omega) = \frac{V_1(\omega)}{V_0(\omega)}$. Thus, we have

$$H(\omega) = \frac{LC\omega^2}{LC\omega^2 - jRC\omega - 1}. \quad (7) \quad H(\omega) = \frac{v_1(\omega)}{v_0(\omega)}$$

Computing $|H(0)|$ and $\lim_{|\omega| \rightarrow \infty} |H(\omega)|$, we obtain

$$\text{take magnitude of (7) and let } \omega=0 \Rightarrow |H(0)| = \left| \frac{0}{-1} \right| = 0 \quad \text{and} \quad \text{take magnitude of (7) and let } |\omega| \rightarrow \infty \Rightarrow \lim_{|\omega| \rightarrow \infty} |H(\omega)| = \lim_{|\omega| \rightarrow \infty} \left| \frac{LC\omega^2}{LC\omega^2} \right| = 1.$$

Since $|H(0)| = 0$ and $\lim_{|\omega| \rightarrow \infty} |H(\omega)| = 1$, the system best approximates a **highpass filter**.