## 4.2 Integral Control (Answers)

## 4.2.1 Closed-Loop Transfer Function with Integral Control

The control signal for integral control is:  $u_m(t) = k_i \int_0^t \left( r(\tau) - \omega_m(\tau) \right) d\tau$ 

In the Laplace domain:  $U_m(s) = rac{k_i}{s} \left( R(s) - \Omega_m(s) 
ight)$ 

Substitute  $U_m(s)$  into the open-loop transfer function:  $\Omega_m(s) = rac{K}{ au s + 1} U_m(s)$ 

This leads to:  $\Omega_m(s) = rac{K}{ au s + 1} rac{k_i}{s} \left( R(s) - \Omega_m(s) 
ight)$ 

Rearranging for  $\Omega_m(s)$  :  $\Omega_m(s)( au s^2 + s + K k_i) = K k_i R(s)$ 

Thus, the closed-loop transfer function  $G_I(s)$  is:  $G_I(s)=rac{\Omega_m(s)}{R(s)}=rac{Kk_i}{ au s^2+s+Kk_i}$ 

## 4.2.2 Location of Poles as a Function of k\_i

The characteristic equation is:  $au s^2 + s + K k_i = 0$ 

The poles are the roots of this quadratic equation:

$$s = rac{-1 \pm \sqrt{1 - 4 au K k_i}}{2 au}$$

As  $k_i$  increases, the poles shift. For small  $k_i$ , the system has a slower response. As  $k_i$  increases, the system responds faster, but increasing  $k_i$  too much may lead to oscillations or instability if the discriminant becomes negative (complex poles).

## 4.2.3 Steady-State Value Using Final Value Theorem

For a step input  $r(t)=r_0$ , the Laplace transform is:  $R(s)=rac{r_0}{s}$ 

Using the closed-loop transfer function:  $\Omega_m(s) = rac{K k_i}{ au s^2 + s + K k_i} \cdot rac{r_0}{s}$ 

Applying the Final Value Theorem:

$$\omega_m(\infty) = \lim_{s o 0} s \cdot \Omega_m(s)$$

At s=0:  $\omega_m(\infty)=r_0$ . Thus, with integral control, the steady-state output matches the input exactly, eliminating steady-state error.