## **PROBLEMS**

**B–6–1.** Plot the root loci for the closed-loop control system with

$$G(s) = \frac{K(s+1)}{s^2}, \quad H(s) = 1$$

**B–6–2.** Plot the root loci for the closed-loop control system with

$$G(s) = \frac{K}{s(s+1)(s^2+4s+5)}, \qquad H(s) = 1$$

**B-6-3.** Plot the root loci for the system with

$$G(s) = \frac{K}{s(s+0.5)(s^2+0.6s+10)}, \qquad H(s) = 1$$

B-6-4. Show that the root loci for a control system with

$$G(s) = \frac{K(s^2 + 6s + 10)}{s^2 + 2s + 10}, \qquad H(s) = 1$$

are arcs of the circle centered at the origin with radius equal to  $\sqrt{10}$ .

**B-6-5.** Plot the root loci for a closed-loop control system with

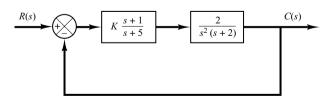
$$G(s) = \frac{K(s+0.2)}{s^2(s+3.6)}, \qquad H(s) = 1$$

B-6-6. Plot the root loci for a closed-loop control system with

$$G(s) = \frac{K(s+9)}{s(s^2+4s+11)}, \qquad H(s) = 1$$

Locate the closed-loop poles on the root loci such that the dominant closed-loop poles have a damping ratio equal to 0.5. Determine the corresponding value of gain K.

**B–6–7.** Plot the root loci for the system shown in Figure 6–100. Determine the range of gain K for stability.



**Figure 6–100** Control system.

**B–6–8.** Consider a unity-feedback control system with the following feedforward transfer function:

$$G(s) = \frac{K}{s(s^2 + 4s + 8)}$$

Plot the root loci for the system. If the value of gain *K* is set equal to 2, where are the closed-loop poles located?

**B–6–9.** Consider the system whose open-loop transfer function is given by

$$G(s)H(s) = \frac{K(s - 0.6667)}{s^4 + 3.3401s^3 + 7.0325s^2}$$

Show that the equation for the asymptotes is given by

$$G_a(s)H_a(s) = \frac{K}{s^3 + 4.0068s^2 + 5.3515s + 2.3825}$$

Using MATLAB, plot the root loci and asymptotes for the system.

**B–6–10.** Consider the unity-feedback system whose feed-forward transfer function is

$$G(s) = \frac{K}{s(s+1)}$$

The constant-gain locus for the system for a given value of *K* is defined by the following equation:

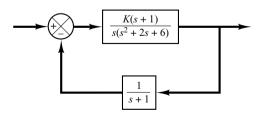
$$\left| \frac{K}{s(s+1)} \right| = 1$$

Show that the constant-gain loci for  $0 \le K \le \infty$  may be given by

$$\left[\sigma(\sigma+1)+\omega^2\right]^2+\omega^2=K^2$$

Sketch the constant-gain loci for K = 1, 2, 5, 10, and 20 on the s plane.

**B–6–11.** Consider the system shown in Figure 6–101. Plot the root loci with MATLAB. Locate the closed-loop poles when the gain K is set equal to 2.



**Figure 6–101** Control system.

**B–6–12.** Plot root-locus diagrams for the nonminimum-phase systems shown in Figures 6–102(a) and (b), respectively.

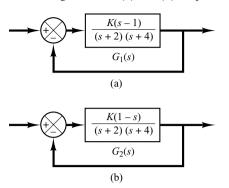


Figure 6–102 (a) and (b) Nonminimum-phase systems.

**B–6–13.** Consider the mechanical system shown in Figure 6–103. It consists of a spring and two dashpots. Obtain the transfer function of the system. The displacement  $x_i$  is the input and displacement  $x_o$  is the output. Is this system a mechanical lead network or lag network?

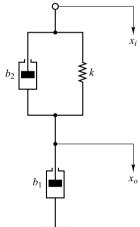


Figure 6–103 Mechanical system.

**B–6–14.** Consider the system shown in Figure 6–104. Plot the root loci for the system. Determine the value of K such that the damping ratio  $\zeta$  of the dominant closed-loop poles is 0.5. Then determine all closed-loop poles. Plot the unit-step response curve with MATLAB.

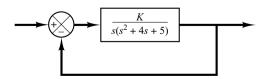


Figure 6-104 Control system.

**B–6–15.** Determine the values of K,  $T_1$ , and  $T_2$  of the system shown in Figure 6–105 so that the dominant closed-loop poles have the damping ratio  $\zeta = 0.5$  and the undamped natural frequency  $\omega_n = 3$  rad/sec.

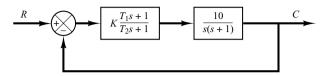


Figure 6–105 Control system.

**B–6–16.** Consider the control system shown in Figure 6–106. Determine the gain K and time constant T of the controller  $G_c(s)$  such that the closed-loop poles are located at  $s = -2 \pm j2$ .

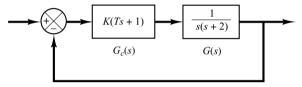


Figure 6–106 Control system.

**B–6–17.** Consider the system shown in Figure 6–107. Design a lead compensator such that the dominant closed-loop poles are located at  $s = -2 \pm j2\sqrt{3}$ . Plot the unit-step response curve of the designed system with MATLAB.

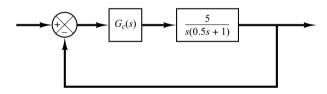


Figure 6–107 Control system.

**B–6–18.** Consider the system shown in Figure 6–108. Design a compensator such that the dominant closed-loop poles are located at  $s = -1 \pm j1$ .

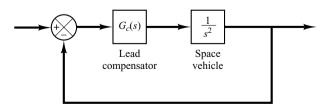
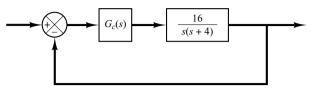


Figure 6–108 Control system.

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**B–6–19.** Referring to the system shown in Figure 6–109, design a compensator such that the static velocity error constant  $K_v$  is  $20 \sec^{-1}$  without appreciably changing the original location  $(s = -2 \pm j2\sqrt{3})$  of a pair of the complex-conjugate closed-loop poles.

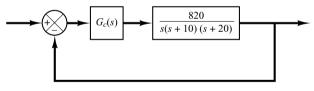


**Figure 6–109** Control system.

**B–6–20.** Consider the angular-positional system shown in Figure 6–110. The dominant closed-loop poles are located at  $s = -3.60 \pm j4.80$ . The damping ratio  $\zeta$  of the dominant closed-loop poles is 0.6. The static velocity error constant  $K_v$  is 4.1 sec<sup>-1</sup>, which means that for a ramp input of 360°/sec the steady-state error in following the ramp input is

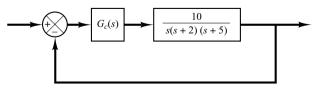
$$e_v = \frac{\theta_i}{K_v} = \frac{360^{\circ}/\text{sec}}{4.1 \text{ sec}^{-1}} = 87.8^{\circ}$$

It is desired to decrease  $e_v$  to one-tenth of the present value, or to increase the value of the static velocity error constant  $K_v$  to  $41~{\rm sec}^{-1}$ . It is also desired to keep the damping ratio  $\zeta$  of the dominant closed-loop poles at 0.6. A small change in the undamped natural frequency  $\omega_n$  of the dominant closed-loop poles is permissible. Design a suitable lag compensator to increase the static velocity error constant as desired.



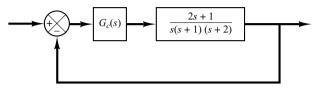
**Figure 6–110** Angular-positional system.

**B–6–21.** Consider the control system shown in Figure 6–111. Design a compensator such that the dominant closed-loop poles are located at  $s = -2 \pm j2\sqrt{3}$  and the static velocity error constant  $K_n$  is 50 sec<sup>-1</sup>.



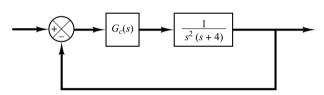
**Figure 6–111** Control system.

**B–6–22.** Consider the control system shown in Figure 6–112. Design a compensator such that the unit-step response curve will exhibit maximum overshoot of 30% or less and settling time of 3 sec or less.



**Figure 6–112** Control system.

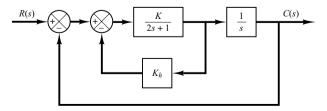
**B–6–23.** Consider the control system shown in Figure 6–113. Design a compensator such that the unit-step response curve will exhibit maximum overshoot of 25% or less and settling time of 5 sec or less.



**Figure 6–113** Control system.

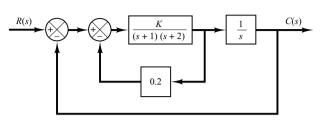
**B–6–24.** Consider the system shown in Figure 6–114, which involves velocity feedback. Determine the values of the amplifier gain K and the velocity feedback gain  $K_h$  so that the following specifications are satisfied:

- 1. Damping ratio of the closed-loop poles is 0.5
- 2. Settling time  $\leq 2 \sec$
- **3.** Static velocity error constant  $K_v \ge 50 \text{ sec}^{-1}$
- **4.**  $0 < K_h < 1$



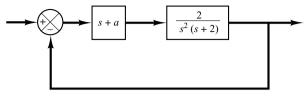
**Figure 6–114** Control system.

**B–6–25.** Consider the system shown in Figure 6–115. The system involves velocity feedback. Determine the value of gain K such that the dominant closed-loop poles have a damping ratio of 0.5. Using the gain K thus determined, obtain the unit-step response of the system.



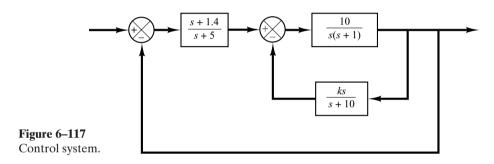
**Figure 6–115** Control system.

**B–6–26.** Consider the system shown in Figure 6–116. Plot the root loci as a varies from 0 to  $\infty$ . Determine the value of a such that the damping ratio of the dominant closed-loop poles is 0.5.



**Figure 6–116** Control system.

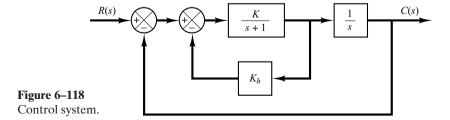
**B–6–27.** Consider the system shown in Figure 6–117. Plot the root loci as the value of k varies from 0 to  $\infty$ . What value of k will give a damping ratio of the dominant closed-loop poles equal to 0.5? Find the static velocity error constant of the system with this value of k.



**B–6–28.** Consider the system shown in Figure 6–118. Assuming that the value of gain K varies from 0 to  $\infty$ , plot the root loci when  $K_h = 0.1, 0.3$ , and 0.5.

Compare unit-step responses of the system for the following three cases:

- $(1) K = 10, K_h = 0.1$
- (2)  $K = 10, K_h = 0.3$
- (3)  $K = 10, K_h = 0.5$



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