Audio Filtering

EE23BTECH11009 - AROSHISH PRADHAN*

Contents

***Abstract*—This manual attempts digital signal processing of an audio file.**

1 Software Installation

Run the following commands

```
sudo apt-get update
sudo apt-get install libffi-dev libsndfile1 python3
    -scipy python3-numpy python3-matplotlib
sudo pip install cffi pysoundfile
```

2 Digital Filter

2.1 Download the sound file from

```
wget https://raw.githubusercontent.com/
    gadepall/
EE1310/master/filter/codes/Sound_Noise.wav
```

2.2 You will find a spectrogram at https://academo.org/demos/spectrum-analyzer.
Upload the sound file that you downloaded in Problem 2.1 in the spectrogram and play. Observe the spectrogram. What do you find?

Solution: The purple areas of the spectrogram represent frequencies with low intensities (noise) while the red-yellow regions represent frequnecies with high intensities (voice).

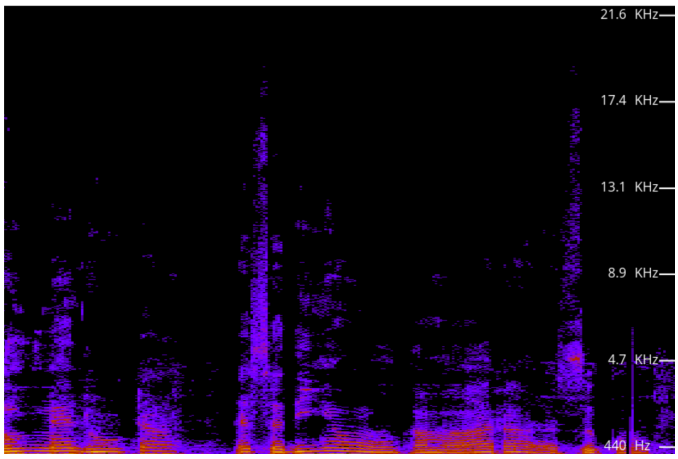

Fig. 2.2: Spectrogram before filtering

2.3 Write the python code for removal of out of band noise and execute the code.

Solution: Noise in the audio is filtered out using the following python code:

```
import soundfile as sf
from scipy import signal

# Read .wav file
input_signal, fs = sf.read('2.wav')

# Order of the filter
order = 4

# Cutoff frequency 6kHz
cutoff_freq = 6000.0

# Digital frequency
Wn = 2 * cutoff_freq / fs

# b and a are numerator and denominator
    polynomials, respectively
b, a = signal.butter(order, Wn, 'low')

print(a)
print(b)
```

```
output_signal = signal.lfilter(b,a,
    input_signal)
# Write the output signal into a .wav file
sf.write('2_fil.wav', output_signal, fs)
```

2.4 The output of the python script in Problem 2.3 is the audio file 2_fil.wav. Play the file in the spectrogram in Problem 2.2. What do you observe?

Solution:

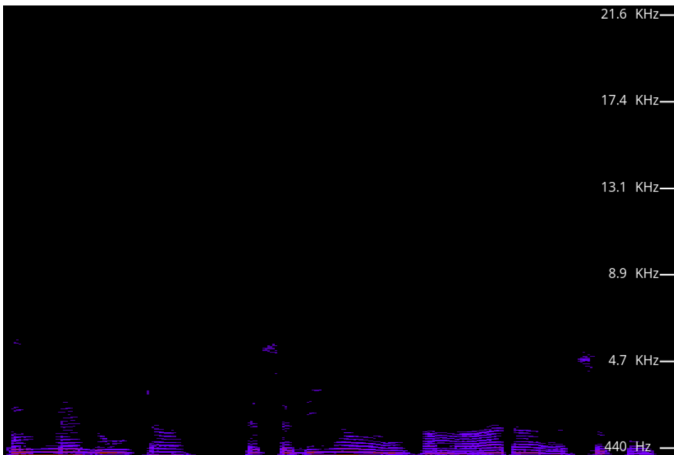

Fig. 2.4: Spectrogram after filtering

The background noise (low intensities) is subdued in the audio. Also, the signal is blank for frequencies above 6 kHz.

3 Difference Equation

3.1 Let

$$x(n) = \left\{ \underset{\uparrow}{1}, 2, 3, 4, 2, 1 \right\} \tag{3.1}$$

Sketch $x(n)$.

Solution: The following code yields Fig. 3.1.

```
wget https://github.com/gadepall/EE1310/raw/
    master/filter/codes/3.1.py
```

3.2 Let

$$y(n) + \frac{1}{2}y(n-1) = x(n) + x(n-2), \quad y(n) = 0, n < 0 \tag{3.2}$$

Sketch $y(n)$.

Solution: The following code yields Fig. 3.2.

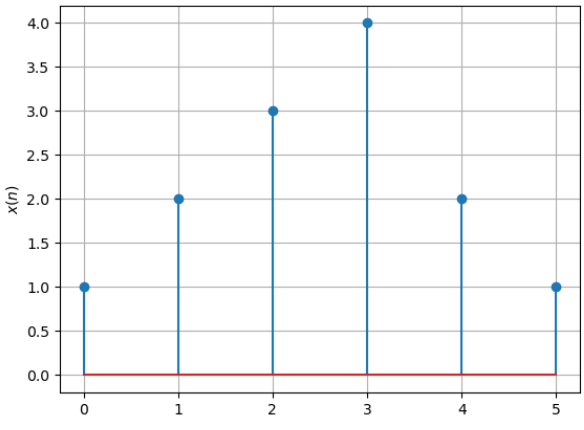

Fig. 3.1: Digital Filter Input $x(n)$

```
wget https://github.com/gadepall/EE1310/raw/
    master/filter/codes/3.2.py
```

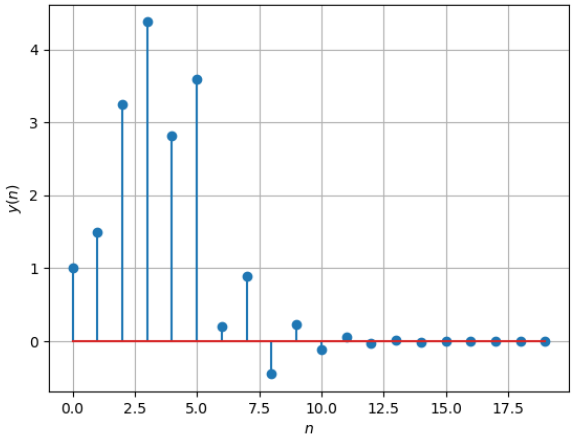

Fig. 3.2: Digital Filter Output $y(n)$

4 Z-transform

4.1 The Z-transform of $x(n)$ is defined as

$$X(z) = \mathcal{Z}\{x(n)\} = \sum_{n=-\infty}^{\infty} x(n) z^{-n} \tag{4.1}$$

Show that

$$\mathcal{Z}\{x(n-1)\} = z^{-1}X(z) \tag{4.2}$$

and find

$$\mathcal{Z}\{x(n-k)\} \tag{4.3}$$

Solution: From (4.1),

$$\mathcal{Z}\{x(n-1)\} = \sum_{n=-\infty}^{\infty} x(n-1)z^{-n} \tag{4.4}$$

$$n \longrightarrow n+1 \tag{4.5}$$

$$= \sum_{n=-\infty}^{\infty} x(n)z^{-n-1}$$

$$= z^{-1}\sum_{n=-\infty}^{\infty} x(n)z^{-n} \tag{4.6}$$

$$= z^{-1}X(z) \tag{4.7}$$

resulting in (4.2). Similarly, it can be shown that

$$\mathcal{Z}\{x(n-k)\} = \sum_{n=-\infty}^{\infty} x(n-k)z^{-n} \tag{4.8}$$

$$n \longrightarrow n+k$$

$$= \sum_{n=-\infty}^{\infty} x(n)z^{-n-k} \tag{4.9}$$

$$= z^{-k}\sum_{n=-\infty}^{\infty} x(n)z^{-n} \tag{4.10}$$

$$= z^{-k}X(z) \tag{4.11}$$

4.2 Find

$$H(z) = \frac{Y(z)}{X(z)} \tag{4.12}$$

from (3.2) assuming that the Z-transform is a linear operation.

Solution: Using (4.11) in (3.2),

$$Y(z) + \frac{1}{2}z^{-1}Y(z) = X(z) + z^{-2}X(z) \tag{4.13}$$

$$\implies \frac{Y(z)}{X(z)} = \frac{1+z^{-2}}{1+\frac{1}{2}z^{-1}} \tag{4.14}$$

4.3 Find the Z transform of

$$\delta(n) = \begin{cases} 1 & n = 0 \\ 0 & \text{otherwise} \end{cases} \tag{4.15}$$

and show that the Z-transform of

$$u(n) = \begin{cases} 1 & n \geq 0 \\ 0 & \text{otherwise} \end{cases} \tag{4.16}$$

is

$$U(z) = \frac{1}{1-z^{-1}}, \quad |z| > 1 \tag{4.17}$$

Solution:

$$\mathcal{Z}\{\delta(n)\} = \sum_{n=-\infty}^{\infty} \delta(n)z^{-n} \tag{4.18}$$

$$= \delta(0)z^{-0} \tag{4.19}$$

$$= 1 \tag{4.20}$$

and from (4.16),

$$U(z) = \sum_{n=0}^{\infty} z^{-n} \tag{4.21}$$

$$= \frac{1}{1-z^{-1}}, \quad |z| > 1 \tag{4.22}$$

using the formula for the sum of an infinite geometric progression.

4.4 Show that

$$a^n u(n) \longleftrightarrow \frac{1}{1-az^{-1}} \quad |z| > |a| \tag{4.23}$$

Solution:

$$\mathcal{Z}\{a^n u(n)\} = \sum_{n=-\infty}^{\infty} a^n u(n)z^{-n} \tag{4.24}$$

$$= \sum_{n=0}^{\infty} (az^{-1})^n \tag{4.25}$$

$$= \frac{1}{1-az^{-1}} \quad |z| > |a| \tag{4.26}$$

using the formula for the sum of an infinite geometric progression.

4.5 Let

$$H\left(e^{j\omega}\right) = H\left(z = e^{j\omega}\right). \tag{4.27}$$

Plot $\left|H\left(e^{j\omega}\right)\right|$. Comment. $H(e^{j\omega})$ is known as the *Discrete Time Fourier Transform* (DTFT) of $h(n)$.

Solution: The following code plots Fig. 4.5.

wget https://raw.githubusercontent.com/gadepall/EE1310/master/**filter**/codes/dtft.py

Substituting $z = e^{j\omega}$ in (4.14),

$$\left|H(e^{j\omega})\right| = \left|\frac{1+e^{-2j\omega}}{1+\frac{1}{2}e^{-j\omega}}\right| \tag{4.28}$$

$$= \frac{\sqrt{(1+\cos 2\omega)^2 + (\sin 2\omega)^2}}{\sqrt{\left(1+\frac{\cos\omega}{2}\right)^2 + \left(\frac{\sin\omega}{2}\right)^2}} \tag{4.29}$$

$$= \frac{4|\cos\omega|}{\sqrt{5+4\cos\omega}} \tag{4.30}$$

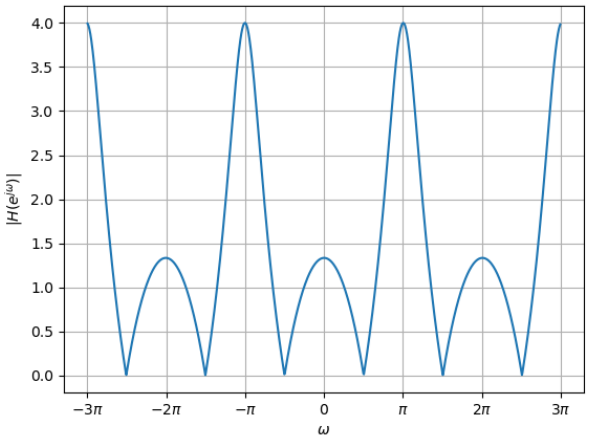

Fig. 4.5: Plot of $\left|H\left(e^{j\omega}\right)\right|$

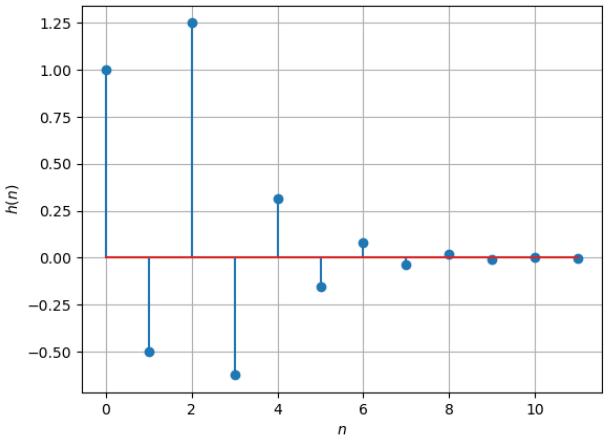

Fig. 5.2: Plot of $h(n)$

which has a fundamental period of 2π:

$$\left|H(e^{(j\omega+2\pi)})\right| = \frac{4|\cos(\omega+2\pi)|}{\sqrt{5+4\cos(\omega+2\pi)}} \tag{4.31}$$

$$= \frac{4|\cos\omega|}{\sqrt{5+4\cos\omega}} \tag{4.32}$$

$$= \left|H(e^{j\omega})\right| \tag{4.33}$$

This can be verified from the graph too.
The plot verifies the property of DTFT of a siganl that it is continuous and periodic.

5 Impulse Response

5.1 Find an expression for $h(n)$ using $H(z)$, given that

$$h(n) \longleftrightarrow H(z) \tag{5.1}$$

and there is a one to one relationship between $h(n)$ and $H(z)$. $h(n)$ is known as the *impulse response* of the system defined by (3.2).
Solution: From (4.14),

$$H(z) = \frac{1}{1+\frac{1}{2}z^{-1}} + \frac{z^{-2}}{1+\frac{1}{2}z^{-1}} \tag{5.2}$$

$$\implies h(n) = \left(-\frac{1}{2}\right)^n u(n) + \left(-\frac{1}{2}\right)^{n-2} u(n-2) \tag{5.3}$$

using (4.23) and (4.11).

5.2 Sketch $h(n)$. Is it bounded? Convergent?
Solution: The following code plots Fig. 5.2.

```
wget https://raw.githubusercontent.com/
    gadepall/EE1310/master/filter/codes/hn.py
```

From graph, it is visible that $h(n)$ is bounded. To check for convergence we can use the ratio test:

$$\lim_{n\to\infty}\left|\frac{h(n+1)}{h(n)}\right| = \left|\frac{\left(-\frac{1}{2}\right)^{n+1}+\left(-\frac{1}{2}\right)^{n-1}}{\left(-\frac{1}{2}\right)^{n}+\left(-\frac{1}{2}\right)^{n-2}}\right| \tag{5.4}$$

$$= \frac{1}{2} < 1 \tag{5.5}$$

Hence, $h(n)$ is convergent.

5.3 The system with $h(n)$ is defined to be stable if

$$\sum_{n=-\infty}^{\infty} h(n) < \infty \tag{5.6}$$

Is the system defined by (3.2) stable for the impulse response in (5.1)?
Solution: Sum of infinite terms of a convergent series is finite. From (5.5), we proved that $h(n)$ was convergent therefore

$$\sum_{n=-\infty}^{\infty} h(n) < \infty \tag{5.7}$$

Hence, the system with the impulse response $h(n)$ is a stable system.

5.4 Compute and sketch $h(n)$ using

$$h(n) + \frac{1}{2}h(n-1) = \delta(n) + \delta(n-2), \tag{5.8}$$

This is the definition of $h(n)$.
Solution: The following code plots Fig. 5.4. Note that this is the same as Fig. 5.2.

wget https://raw.githubusercontent.com/gadepall/EE1310/master/**filter**/codes/hndef.py

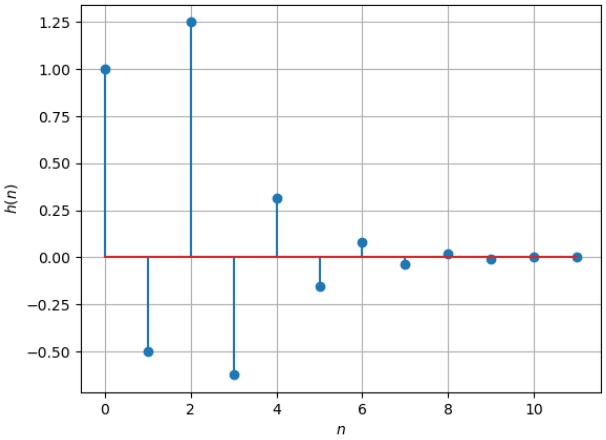

Fig. 5.4: Plot of $h(n)$ using definition

5.5 Compute

$$y(n) = x(n) * h(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k) \quad (5.9)$$

Comment. The operation in (5.9) is known as *convolution*.

Solution: The following code plots Fig. 5.5. Note that this is the same as $y(n)$ in Fig. 3.2.

wget https://raw.githubusercontent.com/gadepall/EE1310/master/**filter**/codes/ynconv.py

5.6 Show that

$$y(n) = \sum_{k=-\infty}^{\infty} x(n-k)h(k) \quad (5.10)$$

Solution: From (5.9),

$$y(n) = \sum_{k=-\infty}^{\infty} x(k)\,h(n-k) \quad (5.11)$$

Substitute $k \to n-k$

$$= \sum_{n-k=-\infty}^{\infty} x(n-k)\,h(k) \quad (5.12)$$

$$= \sum_{k=-\infty}^{\infty} x(n-k)\,h(k) \quad (5.13)$$

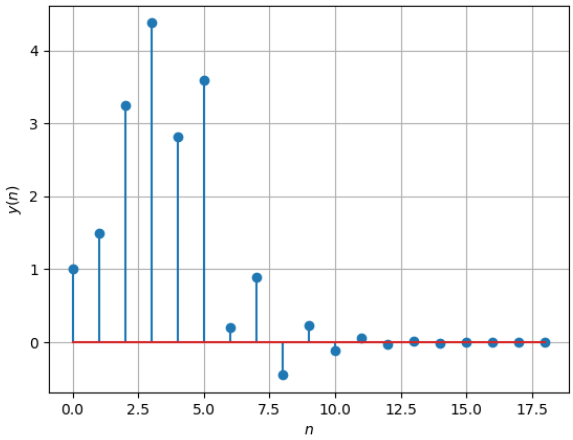

Fig. 5.5: $y(n)$ using convolution

as flipping limits does not change sum.

The following code plots Fig. 5.6. Note that this is the same as $y(n)$ in Fig. 5.5.

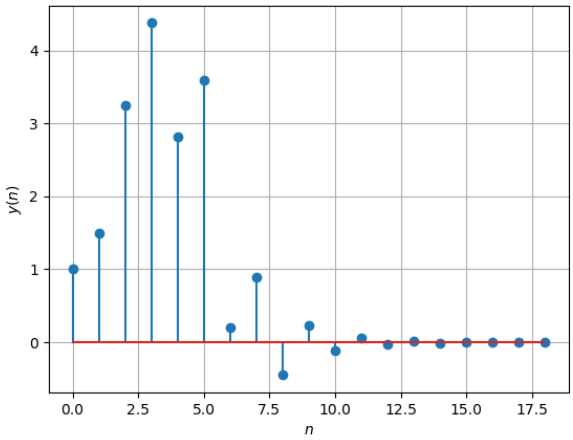

Fig. 5.6: Plot of $y(n)$ using (5.13)

6 DFT and FFT

6.1 Compute

$$X(k) \stackrel{\triangle}{=} \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \ldots, N-1 \quad (6.1)$$

and $H(k)$ using $h(n)$.

6.2 Compute

$$Y(k) = X(k)H(k) \quad (6.2)$$

6.3 Compute

$$y(n) = \frac{1}{N}\sum_{k=0}^{N-1} Y(k)\cdot e^{j2\pi kn/N}, \quad n = 0, 1, \ldots, N-1 \tag{6.3}$$

Solution: The following code plots Fig. 6.3 by taking *Inverse Discrete Fourier Transform* (IDFT) of $Y(k)$. Note that this is also the same as $y(n)$ in Fig. 3.2. It also prints out the values of $X(k)$, $H(k)$, $Y(k)$, and $y(n)$.

```
wget https://raw.githubusercontent.com/
    gadepall/EE1310/master/filter/codes/yndft.
    py
```

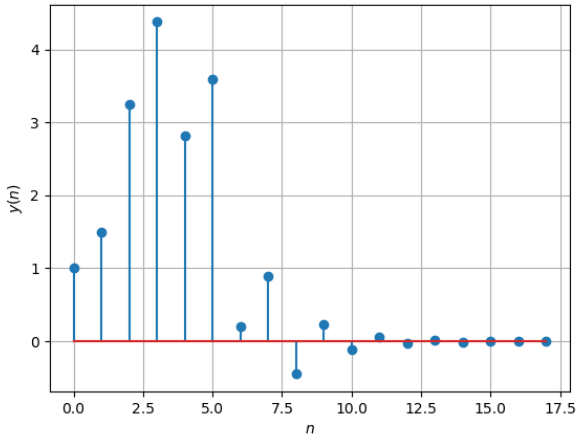

Fig. 6.3: $y(n)$ from IDFT

6.4 Repeat the previous exercise by computing $X(k), H(k)$ and $y(n)$ through FFT and IFFT.

Solution: The code below calculates $X(k)$, $H(k)$ using FFT and plots the graph of $y(n)$ using IFFT and IDFT both (to compare).

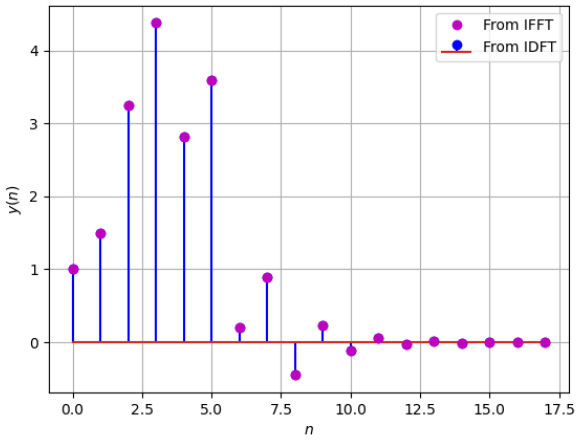

Fig. 6.4: Plot of $y(n)$ from IDFT and IFFT

6.5 Wherever possible, express all the above equations as matrix equations.

Solution: The DFT matrix is given by:

$$\mathbf{W} = \begin{pmatrix} \omega^0 & \omega^0 & \ldots & \omega^0 \\ \omega^0 & \omega^1 & \ldots & \omega^{N-1} \\ \vdots & \vdots & \ddots & \vdots \\ \omega^0 & \omega^{N-1} & \ldots & \omega^{(N-1)(N-1)} \end{pmatrix} \tag{6.4}$$

where $\omega = e^{-\frac{j2\pi}{N}}$. General DFT equation is given by:

$$\mathbf{X} = \mathbf{W}\mathbf{x} \tag{6.5}$$

where

$$\mathbf{x} = \begin{pmatrix} x(0) \\ x(1) \\ \vdots \\ x(n-1) \end{pmatrix} \tag{6.6}$$

$$\mathbf{X} = \begin{pmatrix} X(0) \\ X(1) \\ \vdots \\ X(n-1) \end{pmatrix} \tag{6.7}$$

Then from (6.2):

$$\mathbf{Y} = \mathbf{X} \odot \mathbf{H} = (\mathbf{W}\mathbf{x}) \odot (\mathbf{W}\mathbf{h}) \tag{6.8}$$

where $\odot$ represents the Hadamard product which multiplies corresponding elements of matrices of same size.
The below code computes $y(n)$ by DFT Matrix and then plots it.

https://github.com/dhanushnayakh03/EE1205/tree/main/Audio_%20Filter/codes/5.5.py

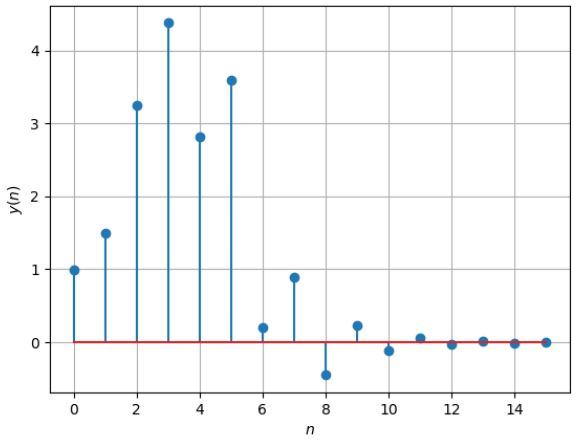

Fig. 6.5: Plot of $y(n)$ using matrix method

7 Exercises

Answer the following questions by looking at the python code in Problem 2.3.

7.1 The command

```
output_signal = signal.lfilter(b, a, input_signal)
```

in Problem 2.3 is executed through the following difference equation

$$\sum_{m=0}^{M} a(m)\, y(n-m) = \sum_{k=0}^{N} b(k)\, x(n-k) \quad (7.1)$$

where the input signal is $x(n)$ and the output signal is $y(n)$ with initial values all 0. Replace **signal.filtfilt** with your own routine and verify.

Solution: The code below plots the output of scipy.signal.lfilter and the output of custom function on the same graph.

https://github.com/dhanushnayakh03/EE1205/tree/main/Audio_%20Filter/codes/5.5.py

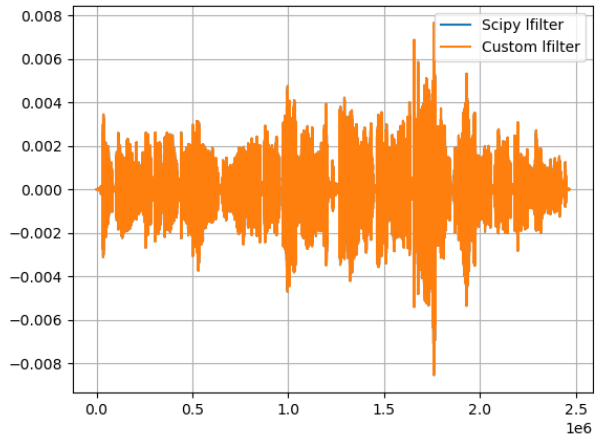

Fig. 7.1: Output of signal.lfilter and custom routine

Both the plots overlap, indicating that the custom filter code does the same function as scipy.signal.lfilter.

7.2 Repeat all the exercises in the previous sections for the above a and b.

Solution: The code in 2.3 calculates values of a and b and prints them:

$$a = [1.000 \quad -1.792 \quad 1.518 \quad -0.608 \quad 0.098]$$
$$b = [0.013 \quad 0.054 \quad 0.081 \quad 0.054 \quad 0.013]$$

Now, using (7.1), difference equation:

$$\begin{aligned} &a(0)\,y(n) + a(1)\,y(n-1) + a(2)\,y(n-2) \\ &+ a(3)\,y(n-3) + a(4)\,y(n-4) = b(0)\,x(n) \\ &+ b(1)\,x(n-1) + b(2)\,x(n-2) + b(3)\,x(n-3) \\ &+ b(4)\,x(n-4) \quad (7.2) \end{aligned}$$

Substituting,

$$\begin{aligned} &y(n) - 1.792y(n-1) + 1.518y(n-2) \\ &- 0.608y(n-3) + 0.098y(n-4) = 0.013x(n) \\ &+0.054x(n-1)+0.081x(n-2)+0.054x(n-3) \\ &+ 0.013x(n-4) \quad (7.3) \end{aligned}$$

The rational transfer function describing this filter in the z-transform domain is:

$$H(z) = \frac{b(0) + b(1)z^{-1} + b(2)z^{-2} + \ldots + b(N)z^{-N}}{a(0) + a(1)z^{-1} + a(2)z^{-2} + \ldots + a(M)z^{-M}} \quad (7.4)$$

$$= \frac{\sum_{k=0}^{N} b(k)z^{-k}}{\sum_{k=0}^{M} a(k)z^{-k}} \quad (7.5)$$

In our case, $M = N = 4$.

Now, the partial fraction of (7.5) is given by:

$$H(z) = \sum_i \frac{r(i)}{1 - p(i)z^{-1}} + \sum_j k(j)z^{-j} \quad (7.6)$$

Values of $r(i)$, $p(i)$ and $k(j)$ are calculated using scipy.signal.residuez, which returns the above mentioned series:

$r(i)$	$p(i)$	$k(i)$
$0.28018185 - 1.23886252j$	$0.38674749 + 0.17013423j$	0.13702919
$0.28018185 + 1.23886252j$	$0.38674749 - 0.17013423j$	–
$-0.3419458 + 0.19576406j$	$0.50928445 + 0.54087922j$	–
$-0.3419458 - 0.19576406j$	$0.50928445 - 0.54087922j$	–

TABLE 1: Values of $r(i)$, $p(i)$, $k(i)$

Inverse of (7.6) is given using:

$$a^n u(n) \longleftrightarrow \frac{1}{1 - az^{-1}} \quad (7.7)$$

$$\delta(n-k) \longleftrightarrow z^{-k} \quad (7.8)$$

$$\implies h(n) = \sum_i r(i)[p(i)]^n u(n) + \sum_j k(j)\delta(n-j) \quad (7.9)$$

Plot of h(n):

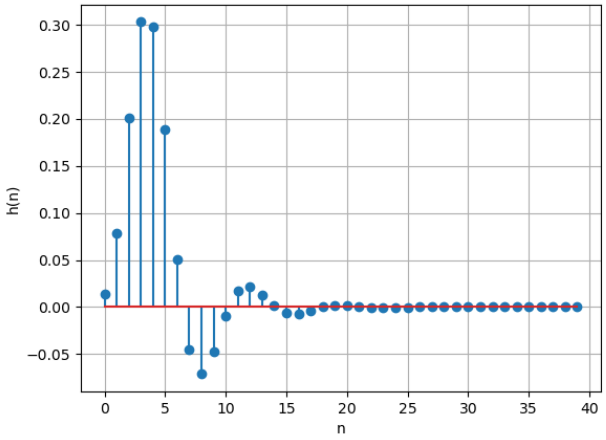

Fig. 7.2: Plot of $h(n)$

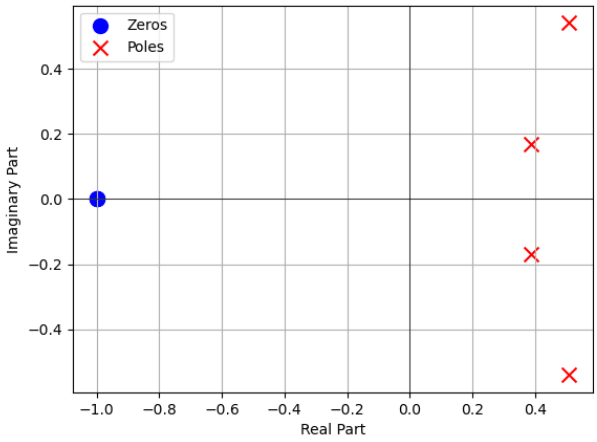

Fig. 7.2: Pole-Zero Plot

There are complex poles, so $h(n)$ has a damped sinusoidal form.

Stability of System

$$H(z) = \sum_{n=0}^{\infty} h(n) z^{-n} \quad (7.10)$$

$$\implies H(1) = \sum_{n=0}^{\infty} h(n) \quad (7.11)$$

$$= \frac{\sum_{k=0}^{N} b(k)}{\sum_{k=0}^{M} a(k)} < \infty \quad (7.12)$$

as both $a(k)$ and $b(k)$ are finite length sequences.
Then, (5.6) implies $h(n)$ is impulse response of a stable system.

Frequency Response of Butterworth Filter

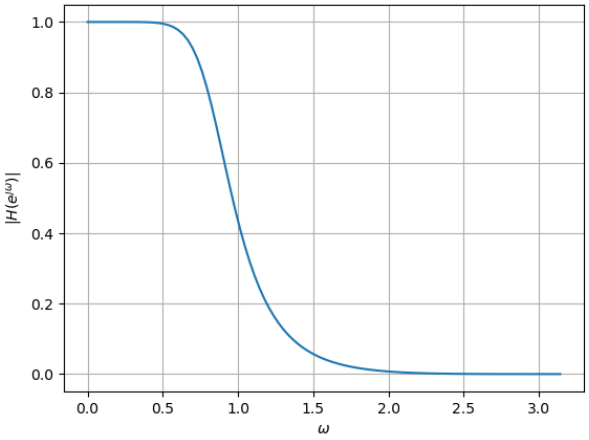

Fig. 7.2: Plot of Frequency Response

Frequency Response of Butterworth Filter in Analog Domain

To convert to analog domain, we can use the Bilinear Transform where we substitute:

$$z = \frac{1 + \frac{sT}{2}}{1 - \frac{sT}{2}} \tag{7.13}$$

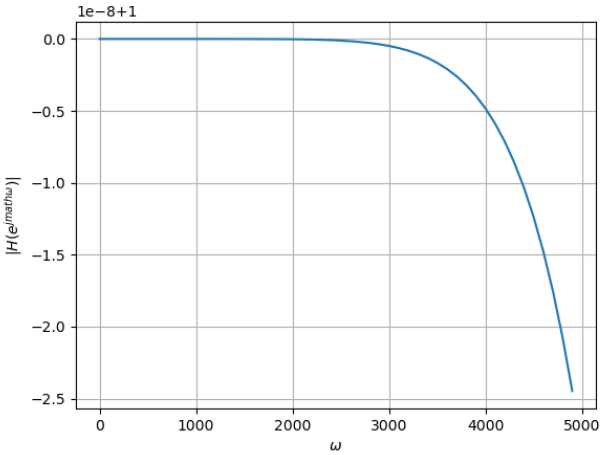

Fig. 7.2: Plot of Frequency Response in Analog Domain

7.3 Implement your own FFT routine in C and call this FFT in python
Solution: The below C code implements the FFT algorithm:

```
k
```

Run the following command to generate a shared library '7.3.so':

```
gcc −shared −o 7.3.so −fPIC 7.3.c
```

The C code is called in the following Python code and output is printed. It can be seen that the same output is printed through both codes.

7.4 Find the Time Complexities of computing $y(n)$ using FFT/IFFT and convolution and compare.
Solution: The below codes generate and plot the time complexities of computing $y(n)$ using FFT/IFFT and Convolution:

```
iva
```

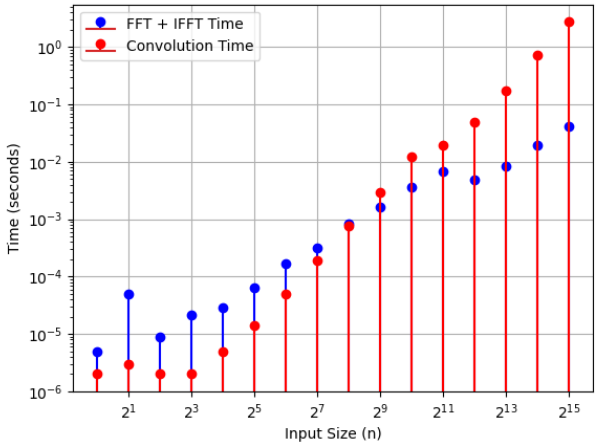

Fig. 7.4: Comparison of Time Complexities

Time complexity of FFT/IFFT method is $O(n\log(n))$ and that of Convolution method is $O(n^2)$.

7.5 What is the sampling frequency of the input signal?
Solution: Sampling frequency (fs) = 44100 Hz. It can be printed from 2.3.

7.6 What is type, order and cutoff-frequency of the above butterworth filter
Solution: The given Butterworth Filter is low pass with order=4 and cutoff-frequency = 6kHz.

7.7 Modifying the code with different input parameters and to get the best possible output.
Solution: The best filtered audio output was

obtained by setting order to 4 and keeping cutoff frequency at 6000 Hz. These parameters were used for the spectrogram output as well as for this section.