1. Given that

$$x(n) = \{4, -3, 2, -1, -5, 3, -1, 0\}, \quad n = 0 \to 7$$

(a) Using only 2-point FFTs, calculate the FFT of x(n)

Heres the MATLAB script. Implemented for any size sequence x(n) where N is a power of 2

```
%% 1 part a
1
   % sequence x(n)
4 \times = [4, -3, 2, -1, -5, 3, -1, 0];
   % get N, constant WN
7 N = length(x);
8 \text{ WN} = \exp(-1i * 2 * pi / N);
   % result of FFT is result_x
10
   result_x = x;
12 % temporary array holds values, go from one 2-point FFT to another
13 hold_x = bitrevorder(result_x);
14
   % array to hold exponents for WN
15
   exp\_array = zeros(1, length(x) / 2);
   % number of sections for corresponding FFTs
17
  blocks = N / 2;
18
19
   % loop through each layer of fast DFT algorithm
20
   for n = 0:log2(N) - 1
^{21}
22
       % populate array for exponents of WN
       for k = 0:2^n:(N / 2) - 1
24
            exp_array(k + 1: k + 2^n) = 0:(N / 2^n + 1):(N / 2) - 1;
25
26
       end
27
       % a = index of exponents for WN
       a = 1;
29
       % other algorithm variables, will explain in a figure later
30
       idx = 1:
31
       for m = 1:blocks
32
           for idx2 = 0:(2^n) - 1
33
               index = idx + idx2;
34
35
                % do the appropriate 2-point FFT
36
                radix_res = fft([hold_x(index), (WN^exp_array(a)) * ...
37
                    hold_x(index + 2^n)], 2);
38
39
                % store results into result_x
               result_x(index) = radix_res(1);
40
                result_x(index + 2^n) = radix_res(2);
41
42
43
                a = a + 1;
44
           end
           idx = idx + 2^{n}(n + 1);
45
46
47
       blocks = blocks / 2;
48
49
       hold_x = result_x;
50 end
```

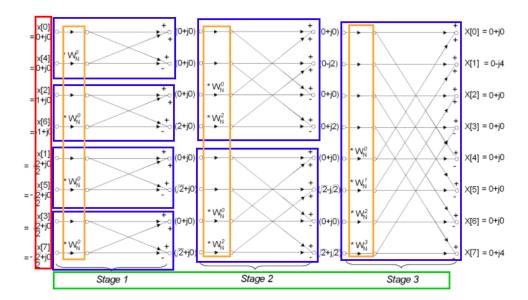


Figure 1: Annotated Radix-2 FFT diagram

Figure courtesy of

https://riptutorial.com/algorithm/example/27088/radix-2-fft.

In my algorithm shown above, the whole for loop is represented by the green sections, blocks is represented by the blue sections, the initial bitrevorder(x) is done on x to represent the red section, and the exp_array is represented by the orange sections. As for the indicies, the inner for loop loops through each block to appropriately compute the 2-point DFT.

Here is the resulting plot of result_x in MATLAB

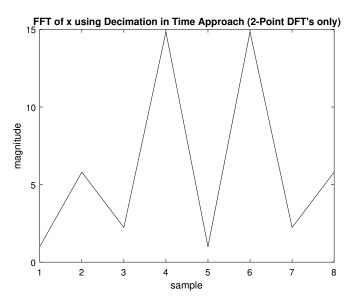


Figure 2: Radix-2 FFT magnitude result

(b) The standard 8-point DFT of x(n) using the MATLAB routine fft looks as such

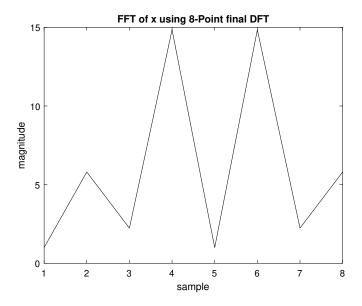


Figure 3: Standard 8-Point FFT magnitude result

(c) The results are identical, with an absolute difference on the magnitude of 10^{-15}

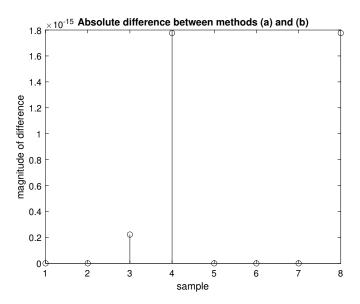


Figure 4: Absolute difference of results

2. (a) Get H(z) in the form

$$H(z) = E_1(z^4) + z^{-1}E_2(z^4) + z^{-2}E_3(z^4) + z^{-3}E_4(z^4)$$

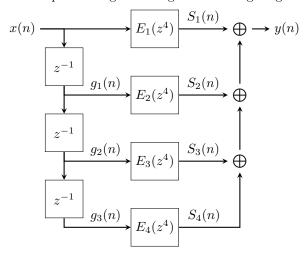
$$E_1(z) = 0.0168 + 0.0287z^{-1} + 0.2453z^{-2} - 0.0409z^{-3}$$
 (1)

$$E_2(z) = 0.0264 - 0.1205z^{-1} - 0.1205z^{-2} + 0.0264z^{-3}$$
 (2)

$$E_3(z) = -0.0409 + 0.2453z^{-1} + 0.0287z^{-2} + 0.0168z^{-3}$$
(3)

$$E_4(z) = 0.0334 + 0.6694z^{-1} + 0.0334z^{-2}$$
(4)

(b) Appropriate difference equation is given using the following diagram



$$g_1(n) = x(n-1) \tag{5}$$

$$g_2(n) = g_1(n-1) (6)$$

$$g_3(n) = g_2(n-1) (7)$$

$$S_1(n) = 0.0168x(n) + 0.0287x(n-4) + 0.2453x(n-8) - 0.0409x(n-12)$$
 (8)

$$S_2(n) = 0.0264g_1(n) - 0.1205g_1(n-4) - 0.1205g_1(n-8) + 0.0264g_1(n-12)$$
 (9)

$$S_3(n) = -0.0409g_2(n) + 0.2453g_2(n-4) + 0.0287g_2(n-8) + 0.0168g_2(n-12)$$
 (10)

$$S_4(n) = 0.0334g_3(n) + 0.6694g_3(n-4) + 0.0334g_3(n-8)$$
(11)

$$y(n) = S_1(n) + S_2(n) + S_3(n) + S_4(n)$$
(12)

3. (a) The 2N-point sequence y(n) is defined as

$$y(n) = \begin{cases} x(\frac{n+1}{2}), & n \text{ odd} \\ 0, & n \text{ even} \end{cases}$$

where X(k) is the N-point DFT of the sequence x(n) for $n \in \{0, N-1\}$ The 2N-point DFT of y(n) is defined as

$$Y(k) = \sum_{n=0}^{2N-1} y(n)e^{\frac{-j2\pi nk}{2N}}$$
 (13)

And since $\forall n_{\text{even}}$, y(n) = 0, the sequence y(n) under the condition that n_{odd} summated equals the summated sequence x(n). The only difference in the DFT would be the n in the exponential term, where it ranges from $n = 0 \to N - 1 \ \forall n_{\text{odd}}$, meaning

$$\sum_{n=0}^{2N-1} y(n)e^{\frac{-j2\pi nk}{2N}} = \sum_{n=0}^{N-1} x(n)e^{\frac{-j2\pi(2n-1)k}{2N}}$$
(14)

where 2n-1 represents the circular odd integers from $0 \to N-1$. Keep in mind, that $k \in \{0, 2N-1\}$, so the second equation will be circularly evaluated twice So that

$$Y(k) = e^{\frac{j2\pi k}{2N}} \sum_{n=0}^{N-1} x(n)e^{\frac{-j2\pi nk}{N}} = e^{\frac{j2\pi k}{2N}} X((k))_N, \quad k \in \{0, 2N-1\}$$
 (15)

(b) Given

$$X_3[k] = \frac{1}{N} \sum_{l=0}^{N-1} X_1[l] X_2[((k-l))_N]$$

show that $x_3[n] = x_1[n]x_2[n]$

First, we know from sampling as well as the convolution theorem, that

$$N \sum_{r=-\infty}^{\infty} \delta[n-rN] \longleftrightarrow \sum_{k=-\infty}^{\infty} \delta(f-k/N)$$
 (16)

so that

$$\frac{1}{N} \sum_{l=0}^{N-1} X_1[l] X_2[((k-l))_N] = \sum_{l=0}^{N-1} \sum_{n=0}^{N-1} \left(x_1[n] e^{\frac{-j2\pi nl}{N}} x_2[n] e^{\frac{-j2\pi n(k-l)}{N}} \right)$$
(17)

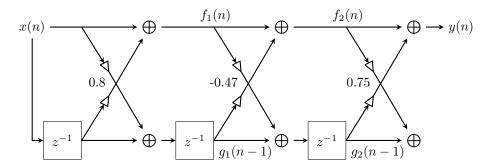
$$\sum_{n=0}^{N-1} x_3[n] e^{\frac{-j2\pi nk}{N}} = \sum_{l=0}^{N-1} \sum_{n=0}^{N-1} \left(x_1[n] x_2[n] e^{\frac{-j2\pi nl}{N}} e^{\frac{j2\pi nl}{N}} e^{\frac{-j2\pi nk}{N}} \right)$$
(18)

$$\sum_{n=0}^{N-1} x_3[n] e^{\frac{-j2\pi nk}{N}} = \sum_{n=0}^{N-1} x_1[n] x_2[n] e^{\frac{-j2\pi nk}{N}}$$
(19)

and then assuming element-wise equivalence, it is shown that

$$x_3[n] = x_1[n]x_2[n], \quad n = 0, 1, \dots, N-1$$
 (20)

- 4. Consider an FIR lattice filter with coefficients $K_1=0.8,\,K_2=-0.47,\,K_3=0.75$
 - (a) Draw the FIR lattice filter



(b) The difference equations are

$$y(n) = f_2(n) + 0.75g_2(n-1)$$
(21)

$$g_2(n) = g_1(n-1) - 0.47f_1(n)$$
(22)

$$f_2(n) = f_1(n) - 0.47g_1(n-1)$$
(23)

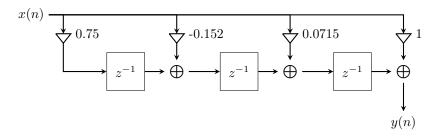
$$g_1(n) = x(n-1) + 0.8x(n)$$
(24)

$$f_1(n) = x(n) + 0.8x(n-1)$$
(25)

which gets reduced down to

$$y(n) = x(n) + 0.0715x(n-1) - 0.152x(n-2) + 0.75x(n-3)$$
(26)

(c) Draw the equivalent direct-form structure



5. Given

$$H(z) = 0.1325 - 0.0867z^{-1} + 0.4205z^{-2} + 1.3592z^{-3} + 0.4205z^{-4} - 0.0867z^{-5} + 0.1325z^{-6}$$

(a) We can split up the transfer function into multiple sections by knowing its roots. Since all roots are complex, the transfer function H(z) can be split up into multiple 2nd-order sections

$$H(z) = \frac{1}{0.1325} S_1(z) S_2(z) S_3(z)$$
 (27)

where

$$S_1(z) = z^2 - 1.0184z + 31.0874 (28)$$

$$S_2(z) = z^2 + 0.3968z + 1 (29)$$

$$S_3(z) = z^2 - 0.0328z + 0.0322 (30)$$

Where we can take any sections $S_k(z)$ and $S_l(z)$ where $k \neq l$ to form either the first or second part of the cascaded filter. For this example, we will create the first filter F_1 out of $S_1(z)$ and $S_2(z)$ while the second filter F_2 will be $S_3(z)$. Using the algorithm in class, we find (by hand) that

$$F_1: K_1 = 0.1984, K_2 = 1, K_3 = -0.0317, K_4 = 31.0874$$
 (31)

$$F_2: K_1 = -0.0317, K_2 = 0.0322$$
 (32)

We can confirm this in MATLAB, where I wrote a function to do just this

```
function [Ks, gain] = fir_lattice(Hz)
   % Author: Arpad Voros
   % fir_lattice() routine takes in coefficients of a transfer function
   % and determines the FIR Lattice coefficients
       INPUT:
                  Hz - array of H(z) coefficients
       OUTPUT:
                  Ks - coefficents of lattice, ordered from input to ...
       output cell array, if needs to be split up (special cases)
7
                    gain - the amount of gain for a given lattice
  % initialize some variables
9
  order = length(Hz) -1;
11
   % initialize outputs
12
  Ks = \{\};
13
  gain = \{\};
14
   % check special case
16
   if Hz(1) == Hz(order + 1)
17
       % include first gain
18
       if Hz(1) \neq 1
19
20
           gain\{end + 1\} = Hz(1);
21
22
       % split the roots
23
       rootsHz = roots(Hz);
25
       split = ceil(length(rootsHz) / 2);
       if rootsHz(split) == conj(rootsHz(split + 1))
26
27
           split = split + 1;
       end
28
       % recursive call
30
31
       [Ks\{end + 1\}, gain\{end + 1\}] = fir_lattice(poly(rootsHz(1:split)));
       [Ks{end + 1}, gain{end + 1}] = fir_lattice(poly(rootsHz(split + ...
32
           1:end)));
33
   else
       % normalize, place into Az
34
35
       if Hz(1) \neq 1
           gain\{end + 1\} = Hz(1);
36
           Az = Hz / Hz(1);
37
       else
38
39
           Az = Hz;
40
41
       % temporary coefficients to be appended to Ks
42
43
       K = zeros(order, 1);
44
45
       for n = order:-1:3
46
           % get last coefficient
```

```
K(n) = Az(n + 1);
48
            % reverse coefficients
50
            Bz = flip(Az);
51
            % get new value of Az
           Az = (Az - K(n) *Bz) / (1 - K(n)^2);
           Az = Az(1:end - 1);
55
56
       % final two coefficients
57
       K(2) = Az(3);
       K(1) = Az(2) / (1 + K(2));
59
60
       % append to rest of coefficients
61
       Ks\{end + 1\} = K;
62
63
64
65
   % delete empty cells
   gain = gain(¬cellfun('isempty', gain));
```

So by calling fir_lattice in the script below, we get

```
1 % coefficients of transfer function
2 h = [0.1325, -0.0867, 4.205, 1.3592, 4.205, -0.0867, 0.1325];
3
4 % display gain and K coefficients
5 [result, gain] = fir_lattice(h);
6 celldisp(gain);
7 celldisp(result);
```

(b) Superscripts in this section do not indicate exponentiation, but rather help distinguish between F_1 and F_2 , where f_k^1 and g_k^1 correspond to F_1 while f_k^2 and g_k^2 correspond to F_2 . The appropriate difference equations for the cascaded filter above are

$$y(n) = f_1^2(n) + 0.0322g_1^2(n-1)$$
(33)

$$g_1^2(n) = f_4^1(n-1) - 0.0317f_4^1(n) \tag{34}$$

$$f_1^2(n) = f_4^1(n) - 0.0317 f_4^1(n-1)$$
(35)

$$f_4^1(n) = f_3^1(n) + 31.0874g_3^1(n-1)$$
(36)

$$g_3^1(n) = g_2^1(n-1) - 0.0317f_2^1(n)$$
(37)

$$f_3^1(n) = f_2^1(n) - 0.0317g_2^1(n-1)$$
(38)

$$g_2^1(n) = g_1^1(n-1) + f_1^1(n)$$
(39)

$$f_2^1(n) = f_1^1(n) + g_1^1(n-1)$$
(40)

$$g_1^1(n) = x(n-1) + 0.1984x(n)$$
(41)

$$f_1^1(n) = x(n) + 0.1984x(n-1)$$
(42)