## 1. IIR Digital Lattice Filter

(a) Using tf2sos subroutine, obtaining coefficients to second-order-system cascaded filter for H(z)

>> gain =

0.0609

where the transfer function is in the form

$$0.0609 \frac{1 - 1.881z^{-1} + z^{-2}}{1 + 0.7493z^{-1} + 0.3790z^{-2}} \frac{1 - 1.3129z^{-1} + z^{-2}}{1 - 0.2219z^{-1} + 0.7562z^{-2}} \frac{1 - 1.0724z^{-1} + z^{-2}}{1 - 0.6471z^{-1} + 0.9241z^{-2}} \frac{1 - 0.9177z^{-1} + z^{-2}}{1 - 0.7763z^{-1} + 0.9824z^{-2}}$$
(1)

(b) Calculating the first SOS section by hand:

$$K_1 = \frac{a(1)}{1 + a(2)} = \frac{0.7493}{1 + 0.379} = 0.5434 \tag{2}$$

$$K_2 = 0.379 (3)$$

$$V_3 = 1 \tag{4}$$

$$V_2 = b(1) - (K_1 K_2 + K_1) V_3 = -2.6303$$
(5)

$$V_1 = b(0) - K_1 V_2 - K_2 V_3 = 2.0503 (6)$$

(c) Calculate all the SOS lattice coefficients using MATLAB

```
17
   % loop through all and calculate all coefficients
  for i = 1:rows
19
20
       b_{new(i, :)} = sos(i, 1:3);
       a_new(i, :) = sos(i, 4:end);
21
22
        [k_new(:, i), v_new(:, i)] = tf2latc(b_new(i, :), a_new(i, :));
23
   end
24
25
   % display
26
27 disp(k_new');
   disp(v_new');
```

```
>> k_new' =
    0.5433
              0.3790
   -0.1264
              0.7562
   -0.3363
              0.9241
   -0.3916
              0.9824
>> v_new' =
    2.0501
             -2.6302
                         1.0000
             -1.0909
    0.1059
                         1.0000
   -0.0671
             -0.4253
                         1.0000
   -0.0378
             -0.1413
                         1.0000
```

where each section is represented by each row of the displayed output, and each coefficient index is represented by the column. We can see that our hand calculation in part (b) was the same as the MATLAB result:

```
>> k_new(:, 1)' =
    0.5433    0.3790

>> v_new(:, 1)' =
    2.0501    -2.6302    1.0000
```

## Hand drawn block diagram

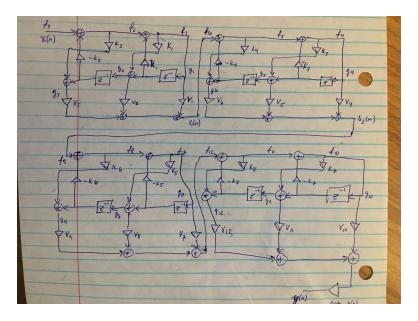


Figure 1: Handdrawn Block Diagram of the IIR Lattice implementation

At the bottom, it got cut off, but the gain term is taken into account right before y(n)

(d) Difference equations from block diagram as well as the coefficients from (c) are given by

$$f_3(n) = x(n) \tag{7}$$

$$g_3(n) = 0.3790 f_2(n) + g_2(n-1)$$
(8)

$$f_2(n) = -0.3790g_2(n-1) + f_3(n) \tag{9}$$

$$g_2(n) = 0.5433f_1(n) + g_1(n-1)$$
(10)

$$f_1(n) = -0.5433g_1(n-1) + f_2(n)$$
(11)

$$g_1(n) = f_1(n) \tag{12}$$

$$s_1(n) = 2.0501g_1(n) - 2.6302g_2(n) + g_3(n)$$
(13)

$$f_6(n) = s_1(n) \tag{14}$$

$$g_6(n) = 0.7562 f_5(n) + g_5(n-1)$$
(15)

$$f_5(n) = -0.7562g_5(n-1) + f_6(n)$$
(16)

$$g_5(n) = -0.1264f_4(n) + g_4(n-1)$$
(17)

$$f_4(n) = 0.1264g_4(n-1) + f_5(n)$$
(18)

$$g_4(n) = f_4(n) \tag{19}$$

$$s_2(n) = 0.1059g_4(n) - 1.0909g_5(n) + g_6(n)$$
(20)

$$f_9(n) = s_2(n) \tag{21}$$

$$g_9(n) = 0.9241 f_8(n) + g_8(n-1)$$
(22)

$$f_8(n) = -0.9241g_8(n-1) + f_9(n)$$
(23)

$$g_8(n) = -0.3363f_7(n) + g_7(n-1)$$
(24)

$$f_7(n) = 0.3363g_7(n-1) + f_8(n)$$
(25)

$$g_7(n) = f_7(n) \tag{26}$$

$$s_3(n) = -0.0671g_7(n) - 0.4253g_8(n) + g_9(n)$$
(27)

$$f_{12}(n) = s_3(n) (28)$$

$$g_{12}(n) = 0.9824f_{11}(n) + g_{11}(n-1)$$
(29)

$$f_{11}(n) = -0.9824g_{11}(n-1) + f_{12}(n) \tag{30}$$

$$g_{11}(n) = -0.3916f_{10}(n) + g_{10}(n-1)$$
(31)

$$f_{10}(n) = 0.3916g_{10}(n-1) + f_{11}(n)$$
(32)

$$g_{10}(n) = f_{10}(n) (33)$$

$$s_4(n) = -0.0378g_{10}(n) - 0.1413g_{11}(n) + g_{12}(n)$$
(34)

$$y(n) = 0.0609s_4(n) \tag{35}$$

- 2. Lattice coefficients for various SOS are given
  - (a) MATLAB function to implement the second order IIR lattice filter

```
function y = my_sos(k, v, x)

%my_sos is a function for an individual SOS to implement a digital ...

IIR filter

NPUTS: k - k coefficients

N - v coefficients

N - v coefficients

N - input data

a = ones(1, 3);

b = ones(1, 3);

c = ones(1, 3);

c = ones(1, 3);

a = ones(1, 3);

c = ones
```

```
17 b(2) = v(2) + (k(1)*k(2) + k(1))*b(3);

18 b(1) = v(1) + v(2)*k(1) + v(3)*k(2);

19

20 % appropriately filtering

21 y = filter(b, a, x);

22 end
```

(b) Downloaded sampdata from Moodle

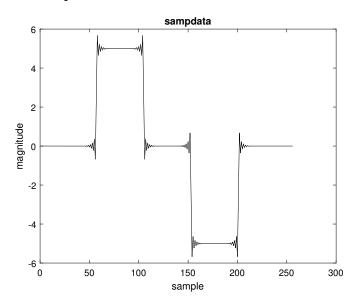


Figure 2: sampdata

(c) Filtering using my\_sos function from part (a)

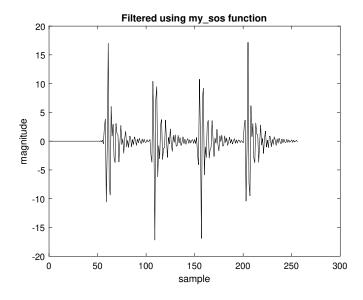


Figure 3: Using my function to filter sampdata

```
% get data
   data = sampdata;
   % coefficients
4
5
   k1 = [0.4545, 0.5703];
   v1 = [1.2459, 0.7137, -1];
   k2 = [-0.1963, 0.8052];
   v2 = [-0.0475, -1.2347, 1];
9
10
   k3 = [0.8273, 0.8835];
11
   v3 = [-0.1965, 0.3783, 1];
12
13
   % cascade through my function
14
15
   y_cascade = my_sos(k1, v1, data);
   y_cascade = my_sos(k2, v2, y_cascade);
16
   y_cascade = my_sos(k3, v3, y_cascade);
17
18
   % plot and save
19
20
   figure(1);
21 plot(y_cascade);
  xlabel("sample");
  ylabel("magnitude");
24
   title("Filtered using my\_sos function");
   print -deps 2c_mysos
25
```

## (d) Filtering using latc2tf function from MATLAB

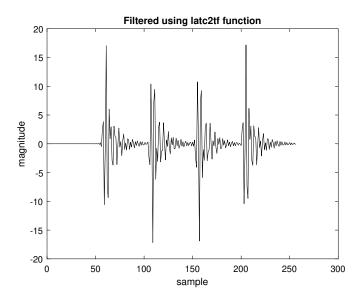


Figure 4: Using latc2tf to filter sampdata

```
1 % get data
2 data = sampdata;
3
4 % coefficients
5 k1 = [0.4545, 0.5703];
```

```
v1 = [1.2459, 0.7137, -1];
   k2 = [-0.1963, 0.8052];
        [-0.0475, -1.2347, 1];
10
   k3 = [0.8273, 0.8835];
11
   v3 = [-0.1965, 0.3783, 1];
13
14
   % find numerator and denominators
   [b1, a1] = latc2tf(k1, v1);
15
   [b2, a2] = latc2tf(k2, v2);
16
17
   [b3, a3] = latc2tf(k3, v3);
18
19
   % cascade through the filter each time
   y_cascade_latc = filter(b1, a1, data);
20
   y_cascade_latc = filter(b2, a2, y_cascade_latc);
   y_cascade_latc = filter(b3, a3, y_cascade_latc);
22
23
24
   % plot and save
  figure(2);
25
  plot(y_cascade_latc);
   xlabel("sample");
27
   ylabel("magnitude");
28
   title("Filtered using latc2tf function");
   print -deps 2d_latc
```

(e) The two plots from (c) and (d) are identical (with a minute absolute error). This means that the MATLAB function latc2tf is a simple routine that can be implemented in a couple lines, like shown with my routine of  $my\_sos$ . I have calculated the absolute difference, which has a maximum error on the order of  $10^{-14}$ 

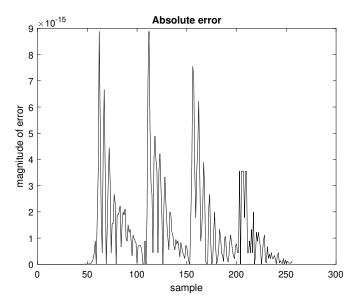


Figure 5: Absolute error between my\_sos and latc2tf

(f) The plots are seen above, and can also be reproduced using the MATLAB script below

```
1 %% PROBLEM 2
3 data = sampdata;
4
5 figure(1)
6 plot(sampdata);
7 xlabel("sample");
8 ylabel("magnitude");
9 title("sampdata");
10 print -deps sampdata
12 	 k1 = [0.4545, 0.5703];
v1 = [1.2459, 0.7137, -1];
14
15 \text{ k2} = [-0.1963, 0.8052];
v2 = [-0.0475, -1.2347, 1];
18 	 k3 = [0.8273, 0.8835];
v3 = [-0.1965, 0.3783, 1];
y_cascade = my_sos(k1, v1, data);
y_cascade = my_sos(k2, v2, y_cascade);
y_cascade = my_sos(k3, v3, y_cascade);
24
25 figure (2);
26 plot(y_cascade);
27 xlabel("sample");
28 ylabel("magnitude");
29 title("Filtered using my\_sos function");
30 print -deps 2c_mysos
31
32 [b1, a1] = latc2tf(k1, v1);
33 [b2, a2] = latc2tf(k2, v2);
   [b3, a3] = latc2tf(k3, v3);
y_cascade_latc = filter(b1, a1, data);
y-cascade-latc = filter(b2, a2, y-cascade-latc);
y-cascade_latc = filter(b3, a3, y-cascade_latc);
40 figure(3);
41 plot(y_cascade_latc);
42 xlabel("sample");
43 ylabel("magnitude");
44 title("Filtered using latc2tf function");
45 print -deps 2d_latc
47 figure (4);
48 plot(abs(y_cascade - y_cascade_latc));
49 xlabel("sample");
50 ylabel("magnitude of error");
51 title("Absolute error");
52 print -deps 2e_diff
   function y = my_sos(k, v, x)
   my-sos is a function for an individual SOS to implement a digital ...
55
       IIR filter
                   k - k coefficients
   응
       INPUTS:
56
                   v - v coefficients
57
                   x - input data
58
60 % allocating memory for coefficients
```

```
61 a = ones(1, 3);
62 b = ones(1, 3);
63
64 % calculating denominator coefficients
65 a(3) = k(2);
66 a(2) = k(1) * (1 + a(3));
67
68 % calculating numerator coefficients
69 b(3) = v(3);
70 b(2) = v(2) + (k(1)*k(2) + k(1))*b(3);
71 b(1) = v(1) + v(2)*k(1) + v(3)*k(2);
72
73 % appropriately filtering
74 y = filter(b, a, x);
75 end
```

3. The ideal frequency response is given by

$$H_d(\omega) = 1, \quad \frac{\pi}{4} < |\omega| \le \frac{3\pi}{4} \tag{36}$$

and 0 elsewhere between  $-\pi \le \omega \le \pi$ 

The Blackman window is given by

$$w(n) = 0.42 + 0.5\cos\left(\frac{2\pi n}{2M}\right) + 0.08\cos\left(\frac{4\pi n}{2M}\right), \quad -M \le n \le M$$
 (37)

- (a) Determine the coefficients for the linear phase filter with 61 filter coefficients
  - i. The required samples, M, of the impulse response is given by

$$M = \frac{61 - 1}{2} = 30\tag{38}$$

In addition, the filter coefficients are determined by taking the DTFT of the window  $H_d(\omega)$ 

$$DTFT\{H_d(\omega)\} = \frac{1}{2\pi} \int_{\omega = -\pi}^{\pi} H_d(\omega) e^{j\omega n} d\omega$$
 (39)

$$\frac{1}{2\pi} \int_{\omega = -\frac{3\pi}{4}}^{-\frac{\pi}{4}} e^{j\omega n} d\omega + \frac{1}{2\pi} \int_{\omega = \frac{\pi}{4}}^{\frac{3\pi}{4}} e^{j\omega n} d\omega \tag{40}$$

Which boils down to the following after applying Euler's identities

$$\frac{1}{\pi n} \left[ \sin \left( \frac{3\pi}{4} n \right) - \sin \left( \frac{\pi}{4} n \right) \right], \quad n \neq 0$$
 (41)

$$\frac{1}{2}, \quad n = 0 \tag{42}$$

- ii. Write MATLAB scripts for
  - A. Computing the required 61 samples for the desired impulse response

```
1
   응응 3
2
3 % length
4 len = 61;
5 \text{ range} = (len - 1) / 2;
7 % two angular frequencies
8 \text{ w1} = \text{pi} / 4;
  w2 = 3 * pi / 4;
9
10
^{11} % get negative values of impulse response
13 bneg = (1 ./ (pi * n1)).*(sin(w1 * n1) - sin(w2 * n1));
14
15 % get positive values of impulse response
16  n2 = 1:range;
17 bpos = (1 ./ (pi * n2)).*(sin(w1 * n2) - sin(w2 * n2));
19 % value of impulse response at n = 0
20 \text{ bmid} = 0.5;
21 b = [bneg bmid bpos];
```

B. Computing the required 61 samples for the Blackman window

```
1 function wind = myblackman(n)
2 % wind = myblackman(n);
3
  % This routine returns a Blackman window
5 % of length n
7 nmod = mod(n, 2);
8
  c = 2 * pi/(n - 1);
9
  if (nmod == 1)
10
       m = fix(0.5 * (n - 1));
       k = -m:1:m;
12
       wind = 0.42 + 0.5*\cos(k * c) + 0.08*\cos(2 * k * c);
13
14 elseif (nmod == 0)
      m = fix(0.5 * n);
15
16
       k = -m:1:(m - 1);
       k = k + 0.5;
17
       wind = 0.42 + 0.5*\cos(k * c) + 0.08*\cos(2 * k * c);
18
19 end
20
21 return
```

and then in the main script, the following is called

```
1 % get 61 samples of blackman window
2 myblack = myblackman(len);
```

C. Multiplying the two sequences together

```
1 % product of the impulse response and blackman window
2 % to get filter coefficients
3 b_prod = b .* myblack;
4
```

```
5 % plotting and saving
6 figure(1); plot(b_prod);
7 title("Filter designed using my MATLAB scripts");
8 print -deps 3aiic
```

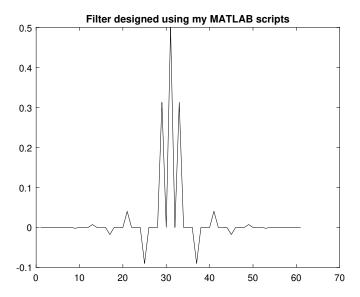


Figure 6: Plot of the filter coefficients

D. Magnitude and phase response of the designed filter

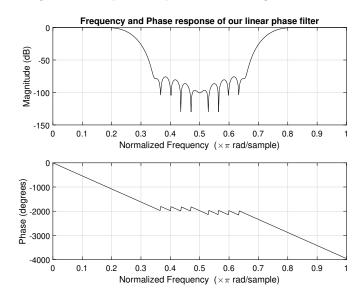


Figure 7: Magnitude and Phase response of filter designed above

iii. Design the same digital filter as above, but using built-in MATLAB routine fir1 within the signal processing toolbox. Use Blackman window function as well

```
1 % order of 60
2 order = len - 1;
3 % pi/4 and 3pi/4
4 wn = [0.25 0.75];
5 % get the blackman window for 61 values
6 w = window(@blackman, order + 1);
7 % get the filter coefficients using fir1
8 b_black = fir1(order, wn, 'stop', w);
```

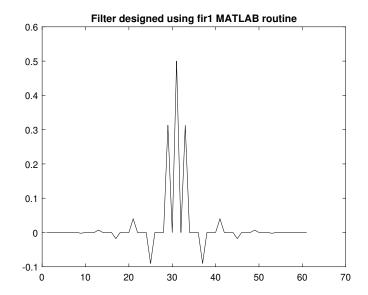


Figure 8: Filter coefficients from MATLAB fir1 routine plotted

iv. Magnitude and phase response of the newly designed filter

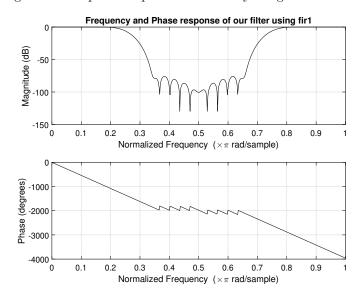
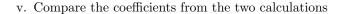


Figure 9: Magnitude and Phase response of filter designed above



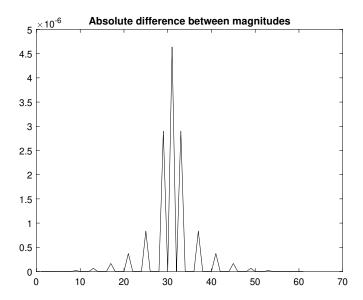


Figure 10: Absolute difference between coefficients

As one can see, the coefficients of the two methods are near identical, with a maximum error on the magnitude of  $10^{-6}$ 

- 4. Designing a linear phase FIR filter with length N=61, which satisfies the conditions given in the HW7 document
  - (a) The equation to determine the causal frequency response H(k) is given by

$$H(k) = \begin{cases} e^{\frac{-j2\pi k(N-1)}{2N}}, & 0 \le k \le 12\\ 0.9e^{\frac{-j2\pi k(N-1)}{2N}}, & k = 13\\ 0, & 14 \le k \le 47\\ 0.9e^{\frac{j2\pi k(N-1)}{2N}}, & k = 48\\ e^{\frac{j2\pi k(N-1)}{2N}}, & 49 \le k \le 60 \end{cases}$$

$$(43)$$

Due to the circular property in the frequency domain

(b) MATLAB script to determine the 61 filter coefficients for this filter

```
1 %% 4
2 % number of coefficients
3 N = 61;
4 M = 0.5 * (N - 1);
5 % indicies 0->12 are constant magnitude of 1
6 k1 = 0:12;
7
8 th = -1i * 2 * pi * k1 * M / N;
9 h = exp(th);
10
11 % at k = 13, 0.9 in magnitude
12 h(14) = 0.9 * exp(-1i * 2 * pi * 13 * M / N);
```

```
13
14 % at 14 \le k \le 30, 0 in magnitude
15 h(15:31) = 0;
16
17 % flip and complex conjugate the remaining ones
18 h(32:61) = fliplr(conj(h(2:31)));
19
20 % determine the coefficients
21 b = real(ifft(h));
22 a = 1;
```

(c) Plotting magnitude and phase response of the filter

```
1 % plotting mag and phase response with freqz
2 freqz(b, a);
3 title("Magnitude and Phase response of the FIR filter");
4 print -deps 4c
```

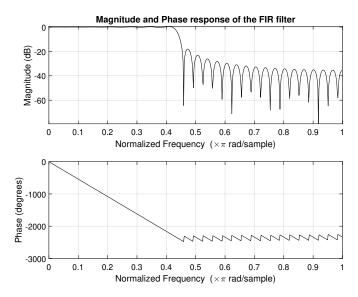


Figure 11: Magnitude and phase response of the filter we designed

(d) At k=13, the frequency  $\omega=\frac{2\pi 13}{61}\approx 1.339$ . If instead of plotting the magnitude in terms of dB, but we plot the magnitude with respect to the phase, at  $\omega\approx 1.339$  we find that

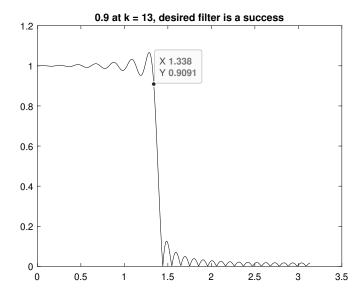


Figure 12: Magnitude of filter plotted against angular frequency

And you can observe that for lower values of k, the magnitude equals roughly 1, and for higher values of k, the magnitude equals roughly zero. The filter is a success!