# Noise propagation on the Golub-Kahan iterative bidiagonalization process

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The problem

#### Ax = b

■ We want to solve the problem:

$$Ax = b$$
,  $A \in \mathbb{R}^{n \times n}$ ,  $b = b^{exact} + b^{noise}$ .

- $b^{noise}$  is a white noise vector with unknown noise level  $\delta_{noise}$ .
- Information on how the noise propagates can be used to develop a stopping criteria for hybrid methods (iterate and regularize).
- The Golub-Kahan iterative bidiagonalization process provides a way to analyze how the noise spreads into the core problem.

Golub-Kahan Iterative Bidiagonalization process

# Golub-Kahan Iterative Bidiagonalization process

■ Given  $w_0 = 0$ ,  $s_1 = b/\beta_1$ , where  $\beta_1 = ||b|| \neq 0$ :

$$\alpha_{j} w_{j} = A^{T} s_{j} - \beta_{j} w_{j-1}, \qquad ||w_{j}|| = 1, 
\beta_{j+1} s_{j+1} = A w_{j} - \alpha_{j} s_{j}, \qquad ||s_{j+1}|| = 1$$

until  $\alpha_j = 0$  or  $\beta_{j+1} = 0$ , or we reach the dimensionality of the problem.

■ The vectors  $s_k$  and  $w_k$  are orthonormal.

#### Matrix form

■ Let  $S_k = [s_1, ..., s_k]$  and  $W_k = [w_1, ..., w_k]$ , then:

$$A^T S_k = W_k L_k^T$$
,  $AW_k = [S_k, s_{k+1}]L_{k+}$ 

where

$$L_{k} = \begin{bmatrix} \alpha_{1} & & & & \\ \beta_{2} & \alpha_{2} & & & \\ & \ddots & \ddots & \\ & & \beta_{k} & \alpha_{k} \end{bmatrix}, \quad L_{k+} = \begin{bmatrix} L_{k} \\ \beta_{k+1} e_{k}^{T} \end{bmatrix}$$

This bidiagonal decomposition can be used to calculate the SVD of A in a stable way.  $A = S_k L_k W_k^T = S_k U \Sigma V^T W_k^T = U_k \Sigma V_k^T$ , with  $U_k$  and  $V_k$  unitary.

Golub-Kahan Iterative Bidiagonalization process

# Lanczos tridiagonalization of $AA^T$

Using the starting vector s<sub>1</sub> = b/β<sub>1</sub>, β<sub>1</sub> = ||b||, yields in k steps:

$$AA^{T}S_{k} = S_{k}T_{k} + \alpha_{k}\beta_{k+1}s_{k+1}e_{k}^{T},$$

and

$$T_{k} = L_{k}L_{k}^{T} = \begin{bmatrix} \alpha_{1}^{2} & \alpha_{1}\beta_{2} & & & \\ \alpha_{1}\beta_{2} & \alpha_{2}^{2} + \beta_{2}^{2} & \ddots & & \\ & \ddots & \ddots & \alpha_{k-1}\beta_{k} \\ & & \alpha_{k-1}\beta_{k} & \alpha_{k}^{2} + \beta_{k}^{2} \end{bmatrix}$$

The matrix L<sub>k</sub> from the GKIB is a Cholesky factor of the matrix T<sub>k</sub>. Golub-Kahan Iterative Bidiagonalization process

# The core problem

- Consider  $x_k = W_k y_k$  as an approximation for the solution of  $Ax_k \approx b$ .
- Looking to the residual:

$$r_k = b - AW_k y_k = S_{k+1}(\beta_1 e_1 - L_{k+} y_k)$$
  
=  $S_k(\beta_1 e_1 - L_k y_k) - (\beta_{k+1} e_k^T y_k) s_{k+1}$ ,

we get two subproblems:

- 1  $L_k y_k = \beta_1 e_1$ , for  $r_k$  orthogonal to  $S_k$  (CGME);
- 2  $y_k = \min_y \|L_{k+}y \beta_1 e_1\|$ , if we minimize  $r_k$  (LSQR or CGLS).

Summary

# Summarizing

- The bidiagonalization concentrates the useful information of the main problem on its bidiagonal block.
- In presence of noise, those subproblems might be polluted by the noise.
- An investigation on how the noise propagates might aid on solving the ill-posed problem. (When to stop the iteration procedure).
- We will use two main approaches: one looks to the behavior of the high frequencies of the vectors  $s_k$  and the other its normalized cumulative periodogram.

Picard condition

- The test problem utilized for the numerical experiments was shaw (400), which satisfy the discrete Picard condition on average.
- We used the following noise levels  $\delta_{noise} = 10^{-14}, 10^{-8}, 10^{-4}$ . Where

$$\delta_{noise} = \frac{\|b^{noise}\|}{\|b^{exact}\|}.$$

Picard condition

# Picard plot

lacksquare  $u_j$  are the left singular vectors of A.

Finding k<sub>noise</sub>

# Spectral coefficients



Finding knoise

# Left singular vectors

Finding k<sub>noise</sub>

# Analizing the spectral coefficients

Using the SVD of A and the Lanczos tridiagonalization of AA<sup>T</sup>:

$$\Sigma^{2}(U^{T}S_{k}) = (U^{T}S_{k})(L_{k}L_{k}^{T}) + \alpha_{k}\beta_{k+1}(U^{T}S_{k+1})e_{k}^{T}.$$

Setting k = 1 we can see how the noise spreads to  $s_2$  (looking to the last column):

$$\alpha_1\beta_2(U^Ts_2) = (\Sigma^2 - \alpha_1^2I)U^Ts_1$$

The matrix acts like a filter on the lower frequencies

$$\frac{\sigma_i}{\alpha_1\beta_2} - \frac{\alpha_1}{\beta_2}$$

- The noise level on the high frequencies will get larger on 🕏

Finding knoise

# Generalizing

For the general case we can write:

$$U^T s_{k+1} = \phi_k(\Sigma^2) U^T s_1$$

- $\phi_k(\lambda)$  is the Lanczos polynomial with root  $(\theta_l^{(k)})^2$ , l = 1, ..., k (Ritz values).
- The large Ritz values  $(\theta_k^{(k)})^2$ ,  $(\theta_k^{(k)})^2$ , ... closely approximate the singular values  $\sigma_1^2$ ,  $\sigma_2^2$ , ....
- Note that the constant term

$$\phi_k(0) = \prod_{j=1}^k \frac{\alpha_j}{\beta_{j+1}} = \rho_k^{-1}$$

•  $\phi_k$  acts like a filter on low frequencies and the constant term  $\phi_k(0) = \rho_k^{-1}$  causes the amplification of the high frequency noise present in the noisy vector  $s_1$ .

# Signal and noise spaces

- $span\{u_1, ..., u_{k+1}\}$  is the signal subspace.
- $span\{u_{k+2},...,u_n\}$  is the noise subspace.

Behavior of  $s_k^{exact}$  and  $s_k^{noise}$ 

# Behavior of $s_k^{exact}$ and $s_k^{noise}$

■ To further illustrate the noise amplification we look to  $s_k^{exact}$  and  $s_k^{noise}$ . Let  $s_1 = s_1^{exact} + s_1^{noise}$  and

$$\begin{split} \beta_{k+1} s_{k+1}^{\textit{exact}} &= \textit{Aw}_k - \alpha_k s_k^{\textit{exact}}, \\ \beta_{k+1} s_{k+1}^{\textit{noise}} &= -\alpha_k s_k^{\textit{noise}}, \\ s_{k+1} &= s_{k+1}^{\textit{exact}} + s_{k+1}^{\textit{exact}}, \quad \beta_{k+1} s_{k+1} = \textit{Aw}_k - \alpha_k s_k. \end{split}$$

- They aren't the true exact and noise data, but they give good approximations to the euclidean norm. See [1].
- Note that,

$$s_{k+1}^{noise} = -\frac{\alpha_k}{\beta_{k+1}} s_k^{noise} = (-1)^k \rho_k^{-1} s_1^{noise}.$$

Normalized Cumulative Periodogram

### Normalized Cumulative Periodogram (NCP)

Definition:

$$\mathbf{c}_j = |dft(\mathbf{y})_j|^2, \quad z_j = \frac{\sum_{i=1}^J \mathbf{c}_i}{\sum_{i=1}^q \mathbf{c}_i}, \quad j = 1, \ldots, q.$$

- Kolmogorov-Smirnoff at 5% significance level: the NCP curve must lie between the limits  $\pm 1.35q^{-1/2}$  of the straight line (diagonal).
- There are other kind of measures for testing if a distribution is white-noise like, e.g., total deviation.

Normalized Cumulative Periodogram

# NCP plot

Normalized Cumulative Periodogram

#### White-noise measures

NCPs vs "GKIB"

# Comparing the methods to find $k_{noise}$

problem	shaw(400)			
$\delta_{ extit{noise}}$	$10^{-14}$	$10^{-8}$	$10^{-4}$	$10^{-2}$
k <sub>noise</sub> (GKIB)	17	11	7	3
k <sub>noise</sub> (NCP)	27	26	8	23
$\tilde{\delta}_{noise}$ (GKIB)	5.9 <i>E</i> – 15	1.2 <i>E</i> – 8	5.3 <i>E</i> – 5	1.6 <i>E</i> – 2
$\tilde{\delta}_{noise}$ (NCP)	1.5 <i>E</i> – 15	8.1 <i>E</i> – 9	3.7 <i>E</i> – 4	3.6 <i>E</i> – 3

lacksquare  $\tilde{\delta}_{\textit{noise}}$  indicates the estimated noise level.



Expresion for R<sup>#</sup>

# Resolution matrix of the GKIB process

$$\blacksquare R^{\sharp} = W_k(L_k^T L_k)^{-1} L_k^T S_k^T A.$$

Show the resolution matrix in Matlab.

#### Expresion for R<sup>#</sup>



Hnetynkova, I. and Plesinger, M.. *The regularizing effect of the Golub-Kahan iterative bidiagonalization and revealing the noise level in the data.* BIT Numer Math (2009) 49: 669-696.