

Lab 12: VHDL

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CSE 2301: Principles and Practice of Digital Logic Design

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Coded in L^AT_EX

Discussion

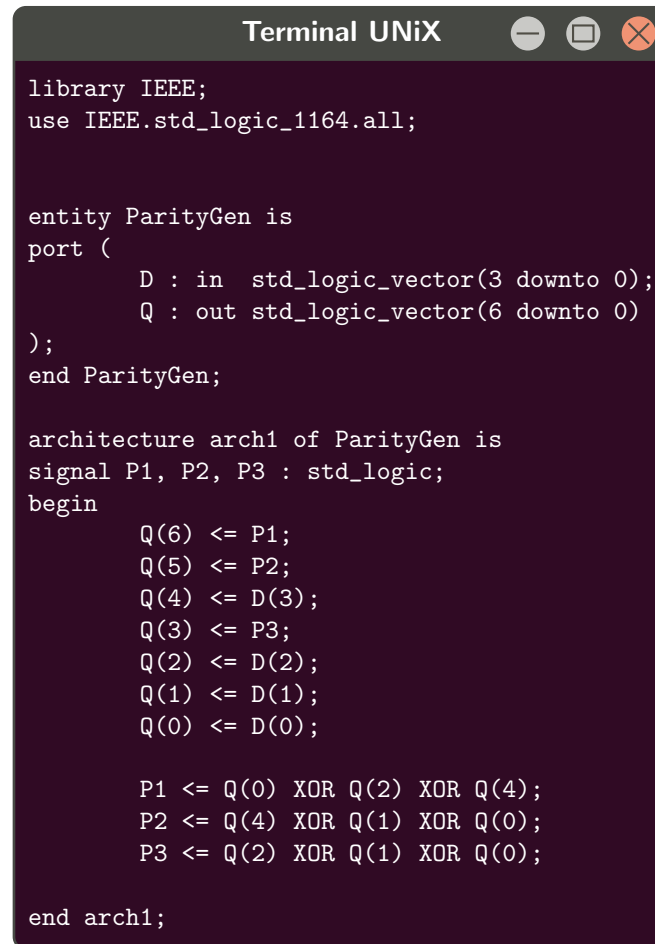
This lab introduced the concept of a Hardware Description Language (HDL) as a means to synthesize digital logic. VHDL, which stands for *Very High Speed Hardware Description Language*, was standardized by the IEEE and is widely used for digital circuit design and production. Unlike designing logic at the gate level, VHDL allows designers to focus on the high-level behavior or structure of a circuit, making complex systems more manageable and modular.

The task in this lab was to implement two circuits previously constructed in earlier exercises, this time using VHDL to emphasize the simplicity and power of hardware description languages. By abstracting the underlying gate-level implementations, VHDL streamlines the design process and highlights how real-world production environments operate.

The figures included are generated from tools commonly used in professional production environments. These tools provide a visualization of the synthesized designs but are not targeted toward specific physical specifications. This generality aligns with the educational objectives of the course, allowing students to focus on HDL design without delving into technology-specific constraints.

As usual, further “discussion” is embedded within each example to provide context and detail regarding the circuits and their implementations.

Example 1 The Hamming Parity Error Check in VHDL

A terminal window titled "Terminal UNiX" with standard window controls (minimize, maximize, close). The terminal displays VHDL code for an entity named ParityGen. The code includes library declarations, port definitions, and an architecture named arch1. The architecture defines signals P1, P2, and P3, and performs bit-level assignments to output Q and XOR operations to calculate parity signals.

```
library IEEE;
use IEEE.std_logic_1164.all;

entity ParityGen is
port (
    D : in  std_logic_vector(3 downto 0);
    Q : out std_logic_vector(6 downto 0)
);
end ParityGen;

architecture arch1 of ParityGen is
signal P1, P2, P3 : std_logic;
begin
    Q(6) <= P1;
    Q(5) <= P2;
    Q(4) <= D(3);
    Q(3) <= P3;
    Q(2) <= D(2);
    Q(1) <= D(1);
    Q(0) <= D(0);

    P1 <= Q(0) XOR Q(2) XOR Q(4);
    P2 <= Q(4) XOR Q(1) XOR Q(0);
    P3 <= Q(2) XOR Q(1) XOR Q(0);

end arch1;
```

This assignment required the completion of the assignments in the architecture definition for the entity.

In VHDL, several different styles of defining the operation of a circuit exist. This is a *behavioral* description, meaning that the signals are not implicitly connected to the gates (structural) nor does it use signals and operators to describe the dataflow of the entity.

In VHDL, the `<=` operator is used for signal assignments, ensuring that updates to Q and the parity signals (P1, P2, P3) occur concurrently within the architecture. Addressing specific bits in Q using parentheses allows for precise placement of the computed values.

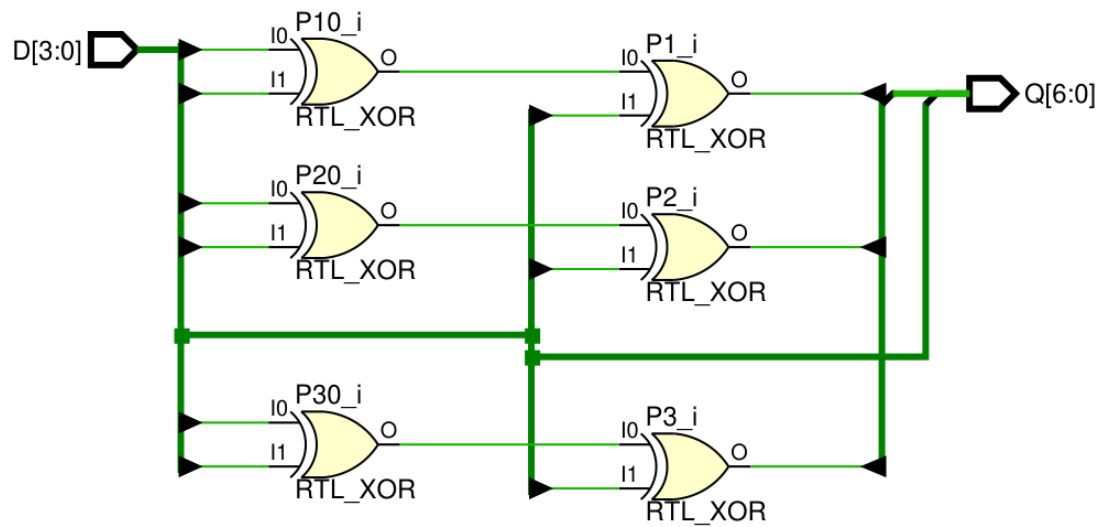


Figure 1: Synthesized Parity Checker

Example 2 The ALU Reprise

```

Terminal UNiX

library IEEE;
use IEEE.std_logic_1164.all;
use IEEE.numeric_std.all;

entity ALU2ElectricBoog is

    port(
        B      : in    std_logic_vector(1 downto 0);
        A      : in    std_logic_vector(1 downto 0);
        Q      : out   std_logic_vector(3 downto 0);
        Op     : in    std_logic_vector(1 downto 0)
    );

end ALU2ElectricBoog;

architecture arch1 of ALU2ElectricBoog is
begin
    process (A, B, Op)
    begin
        if Op = "00" then
            Q <= "00" & A and "00" & B;
        elsif Op = "01" then
            Q <= "00" & A or "00" & B;
        elsif Op = "10" then
            Q <= unsigned("00" & A) + unsigned("00" & B);
        elsif Op = "11" then
            Q <= unsigned(A) * unsigned(B);
        end if;
    end process;
end arch1;

```

This assignment differed from the last by having the required portion be filled under the **process** block. Here, we define an operation that is similar to a MUX utilizing the **if-elsif** construct from VHDL. This is still using the *behavioral* strategy of implementation with a **process** block.

The explicit use of casting the unsigned type and concatenating the signals with leading 00 is explained further. The synthesized schematic is as follows:

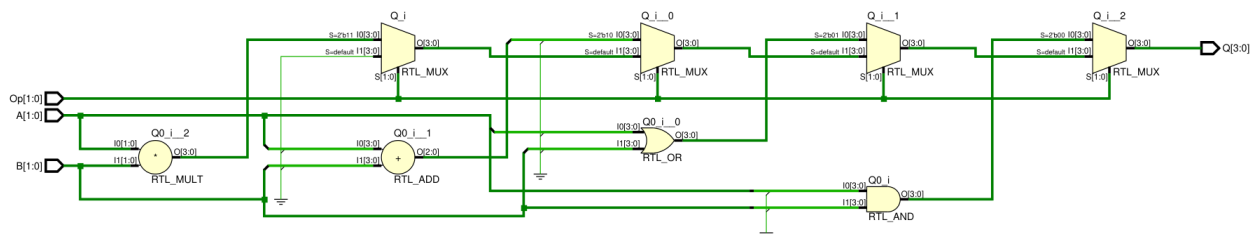


Figure 2: Synthesized Toy ALU

Example 3 The Use of `unsigned()` Cast

The `unsigned()` function in VHDL is used to cast a binary vector to an **UNSIGNED** type, treating the vector as a positive integer. This explicit casting allows VHDL to interpret binary data without ambiguity, particularly during arithmetic operations, ensuring that the result is processed as an unsigned value. In VHDL, operations between **SIGNED** and **UNSIGNED** types require conversions to maintain compatibility, as VHDL does not implicitly mix these types.

`unsigned()` ensures that **A** and **B** are interpreted as positive integers, allowing for correct behavior in operations like addition (+) without treating them as potentially negative numbers.

Example 4 Effect of Using `signed()` Instead of `unsigned()`

If you were to replace `unsigned()` with `signed()` in the ALU, the output could change significantly, particularly in cases where **A** or **B** have high bits set to 1 (interpreted as negative in 2's complement). For example:

- With `unsigned`, a binary vector such as "1100" would be interpreted as 12.
- With `signed`, the same vector "1100" would be interpreted as -4 in 2's complement form.

This change would result in different outputs for operations that involve addition or multiplication, where the result might switch from a positive to a negative value, depending on the high bits of the input.

Example 5 Circuit Element Emulated by `if` and `elsif` Statements

In the ALU, the `if` and `elsif` statements emulate the behavior of a **multiplexer (MUX)**. The `Op` input acts as a selector that determines which operation (**AND**, **OR**, **ADD**, or **MULTIPLY**) the ALU will perform based on its value. Each condition (e.g., `Op = "00"`, `Op = "01"`, etc.) corresponds to a specific operation, similar to how a multiplexer selects an output based on control signals.

In VHDL there are several other methods of defining a MUX like operation such as the use of the `when` and `else` operators.

Example 6 A Basic Combinational Logic Example

A basic diagram is given that describes

$$Q = ABC\bar{C} + A \oplus C$$

To create this in a VHDL simulation, the following behavioral entity is employed:

```

Terminal Mac

library IEEE;
use IEEE.std_logic_1164.all;
use IEEE.numeric_std.all;

entity examp6 is
  port (
    A, B, C : in std_logic;
    Q : out std_logic
  );
end examp6;

architecture rtl of examp6 is
begin
  Q <= (A and B and (not C)) or (A xor C);
end rtl;

```

This highlights how simple it is to assign complex combinational statements in a synthesizable form.

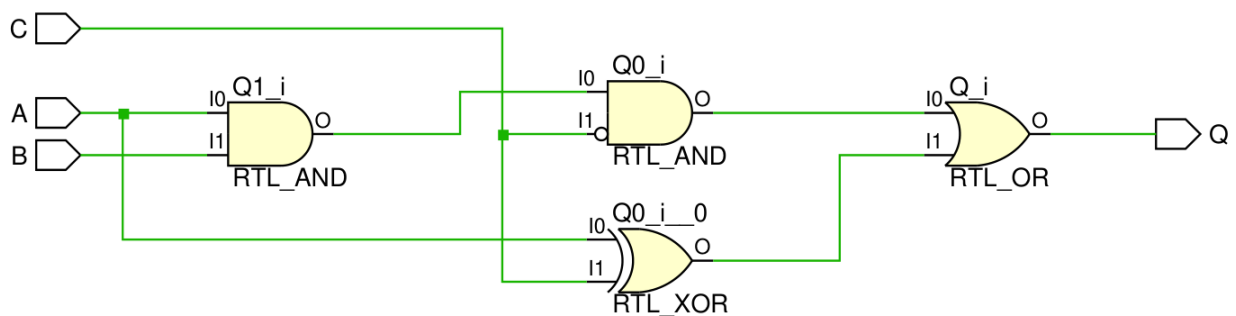


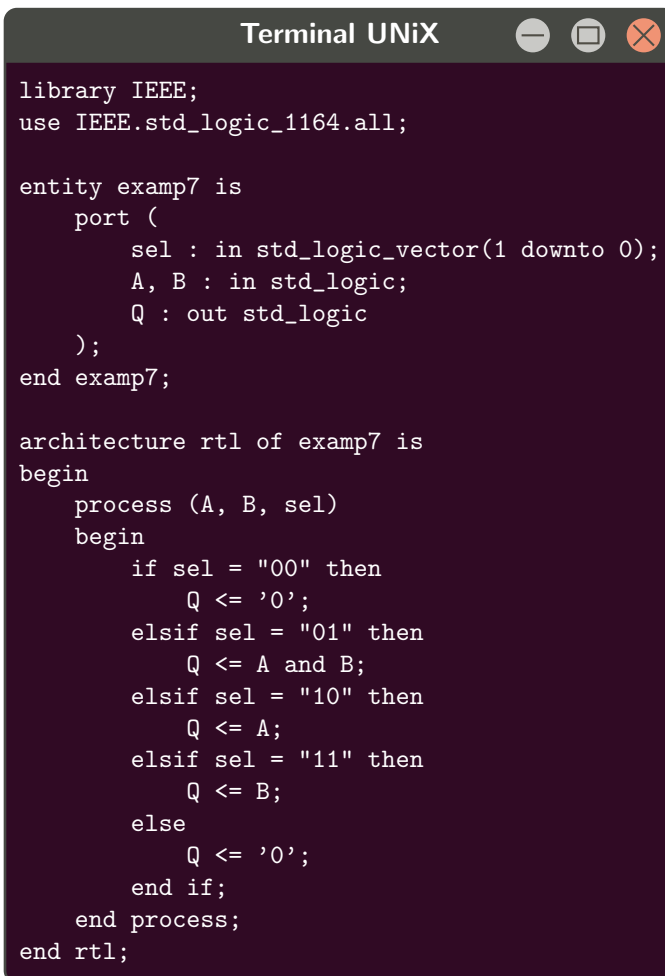
Figure 3: Synthesized Schematic from VHDL

Example 7 Another Multiplexor Example

As stated before, a MUX can be employed in VHDL in several different ways. This goes back to utilizing the `if-elsif` operators, which are straight forward to implement.

A 4 input MUX with a 2 bit selector is shown that uses two inputs, *A* and *B* to implement some sort of logical operation which as seen in previous labs, could be used to implement a larger machine such as a 12-person voting machine.

Since we only write synthesizable VHDL here and not other silly business, the code is completed as follows:

A screenshot of a terminal window titled "Terminal UNiX". The window has standard macOS window controls (minimize, maximize, close) in the top right corner. The terminal displays VHDL code for an entity named "examp7". The code includes a library declaration for IEEE, a port declaration for a 2-bit selector "sel" and two inputs "A" and "B", and a process block that implements a 4-input multiplexer using an if-elsif-else structure. The process assigns the output "Q" based on the value of "sel": "00" results in '0', "01" results in A and B, "10" results in A, "11" results in B, and any other value results in '0'.

```
library IEEE;
use IEEE.std_logic_1164.all;

entity examp7 is
  port (
    sel : in std_logic_vector(1 downto 0);
    A, B : in std_logic;
    Q : out std_logic
  );
end examp7;

architecture rtl of examp7 is
begin
  process (A, B, sel)
  begin
    if sel = "00" then
      Q <= '0';
    elsif sel = "01" then
      Q <= A and B;
    elsif sel = "10" then
      Q <= A;
    elsif sel = "11" then
      Q <= B;
    else
      Q <= '0';
    end if;
  end process;
end rtl;
```

This example is slightly more challenging due to the careful use of the `"` operators and the required formatting for each line according to the VHDL standard.

To ensure clarity and adherence to best practices, it is customary to write processes like this for readability. For example, consider the statement:

```
Q <= '0';
```

In this case, *Q* is of type `std_logic`, while *sel* is of type `std_logic_vector`. The assignment utilizes the `'` operators which are subtly different than the ones employed the line prior.

This distinction is crucial because attempting to assign a `std_logic_vector` directly to a `std_logic` would result in errors. This subtle type mismatch is a common source of bugs and can lead to hours of troubleshooting if not addressed early. (Ask me how I know!)

Additionally, the assignment code excluded something very important for synthesis, a default condition for the `sel` utilizing a terminating `else`. Without this, undesired outcomes may occur during synthesis as the software would place a latch on the output of the circuit, which is not what is intended. I have included this in my code above and assigned `Q` a 0 for all other edge cases. Again, this might not be pertinent to such a simple design, but can result in nasty, nasty bugs in the testbench.

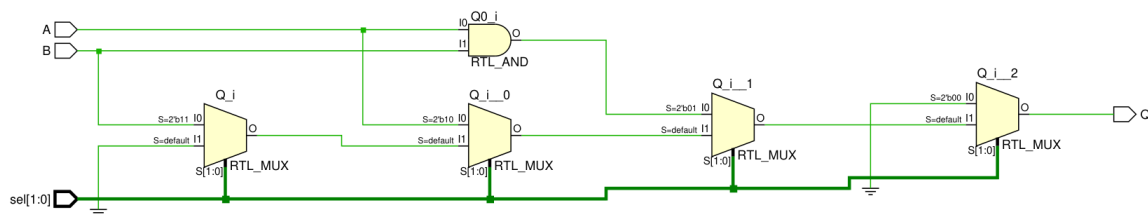


Figure 4: Synthesized Mux Module from VHDL