

TRANSISTOR EQUIVALENT CIRCUIT MODEL

- **TOTAL RESPONSE OF AN AMPLIFIER:**
 - DC RESPONSE + AC RESPONSE
- **AC ANALYSIS**
 - SMALL SIGNAL ANALYSIS
 - LARGE SIGNAL ANALYSIS
- **What is an Equivalent Model??**
- An equivalent model is the combination of circuit elements properly chosen to represent the actual behavior of the device under specific operating point
- **Why Equivalent model?**
- Cannot apply network theorems directly on practical devices. So need equivalent model to use these theorems to find out different network parameters.
- **TRANSISTOR MODELS**
 - Hybrid Model
 - re Model
 - Hybrid Pi Model

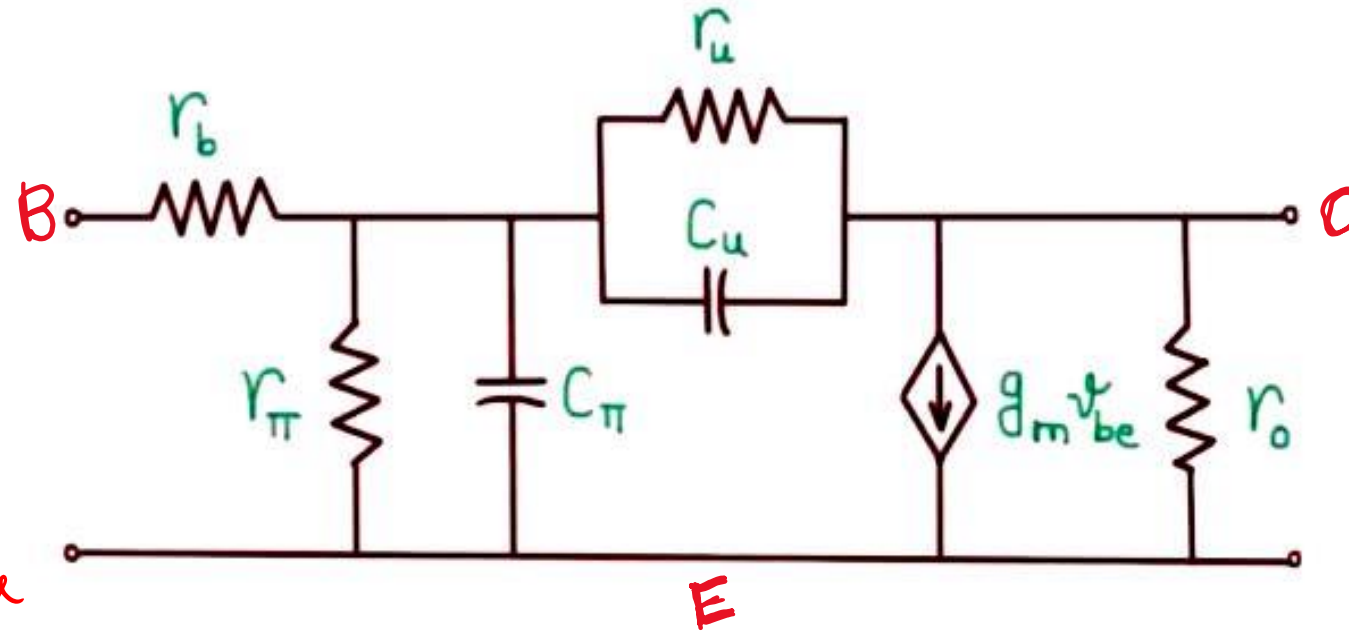
HYBRID π MODEL

- C_{μ} - Transition capacitance between base and collector. Early effect representation
- C_{π} - Diffusion Capacitance. Minority carriers stored in Base region
- r_{π} - small input resistance between base and emitter seen from base
- r_{μ} - collector base reverse resistance (large value)
- r_b - base terminal resistance (small value)
- r_o - output resistance seen from output, collector emitter resistance
- $g_m v_{be}$ - current source

$$i_c = g_m v_{be}$$

$$i_c = \beta i_b$$

$g_m \rightarrow$ Trans conductance



VOLTAGE CONTROLLED CURRENT SOURCE
CURRENT CONTROLLED CURRENT SOURCE

HYBRID π MODEL

DC conditions of a transistor

$$I_C = I_0 e^{(V_{BE}/V_T)}$$

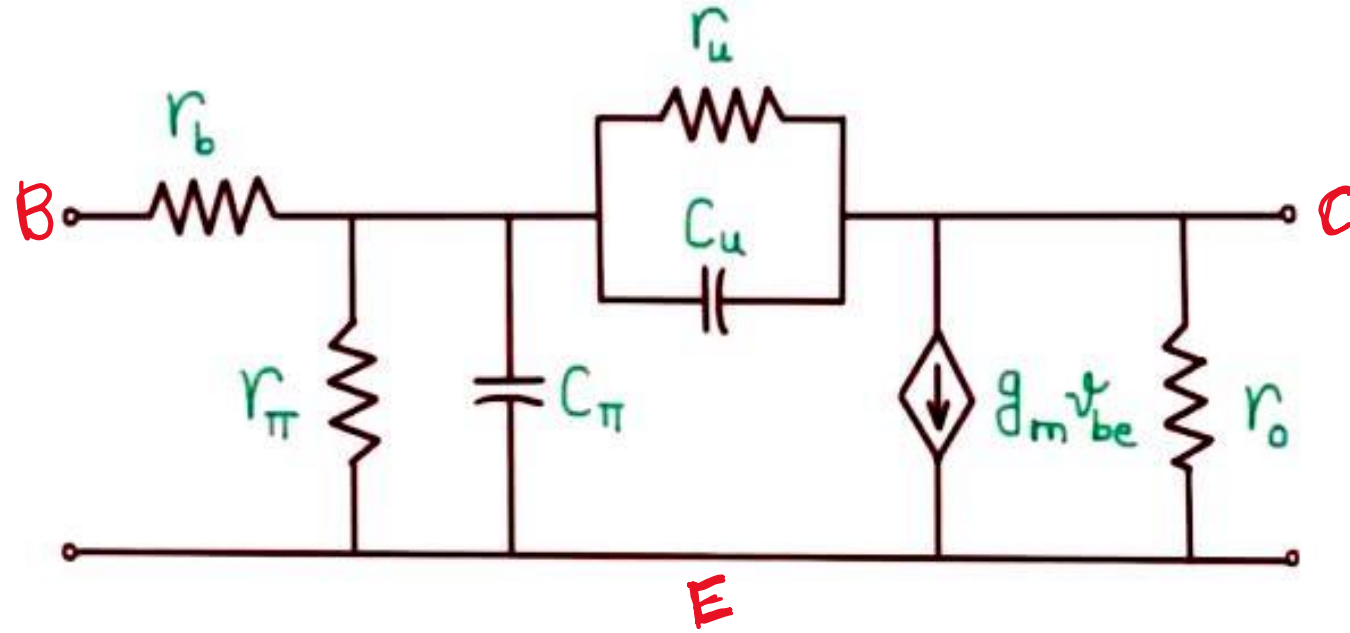
If an input signal v_{be} also applied to the base

$$\begin{aligned} i_C &= I_0 e^{(V_{BE} + v_{be})/V_T} \\ &= I_0 e^{V_{BE}/V_T} \cdot e^{v_{be}/V_T} \end{aligned}$$

$$i_C = I_C \cdot e^{v_{be}/V_T} = I_C \left(1 + \frac{v_{be}}{V_T} \right)$$

$$i_C = I_C + \underbrace{I_C \cdot \frac{v_{be}}{V_T}}_{\text{Signal Component}}$$

\nwarrow DC component



$$e^x = 1 + \frac{x}{1!} + \frac{x^2}{2!} + \dots$$

$$e^{v_{be}/V_T} = 1 + \frac{v_{be}}{V_T} + \dots$$

$v_{be} \ll V_T$

HYBRID π MODEL

$$i_c = I_c \cdot \frac{V_{be}}{V_T}$$

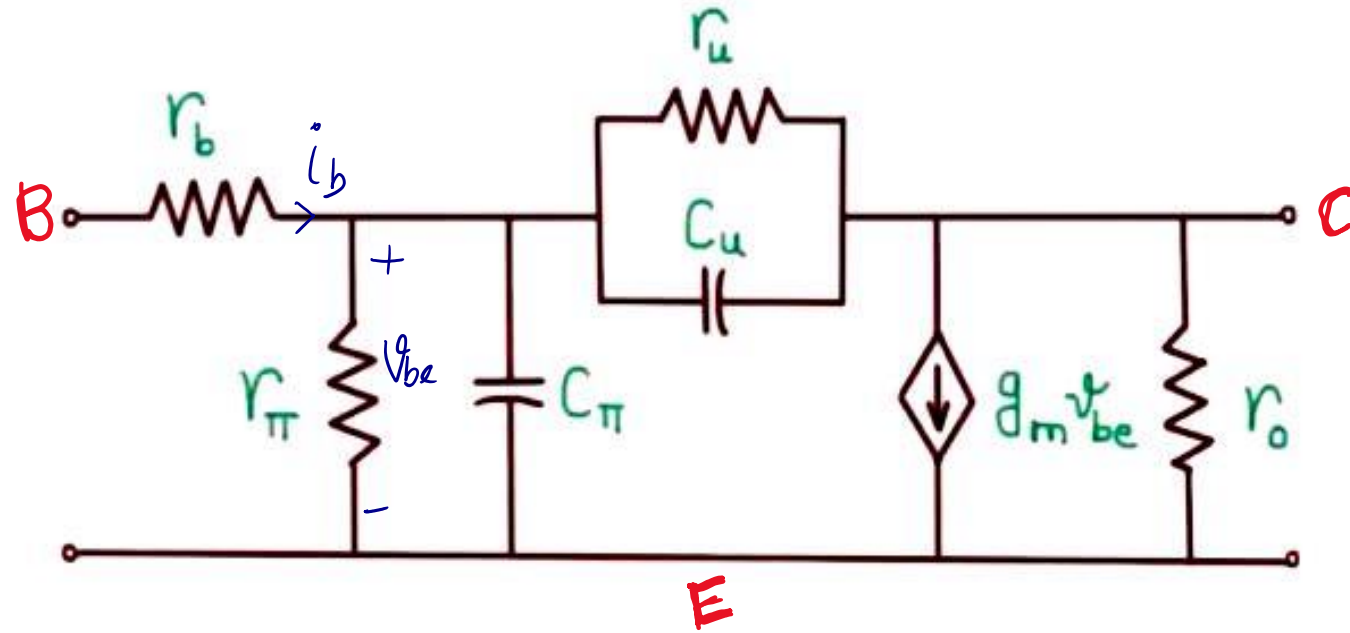
TRANSCONDUCTANCE, $g_m = \frac{I_c}{V_T}$

$$i_c = g_m V_{be}$$

$$i_b = \frac{i_c}{\beta} \Rightarrow i_b = \frac{g_m V_{be}}{\beta}$$

$$\alpha_{\pi} = \frac{V_{be}}{i_b} = \frac{\beta}{g_m}$$

$$g_m = \frac{I_c}{V_T} \Rightarrow \alpha_{\pi} = \frac{\beta V_T}{I_c} = \frac{V_T}{I_B}$$



$$i_c = g_m V_{be}$$

$$i_b = \frac{g_m V_{be}}{\beta}$$

$$\alpha_{\pi} = \frac{V_{be}}{i_b} = \frac{\beta}{g_m} = \frac{V_T}{I_B}$$

HYBRID π MODEL

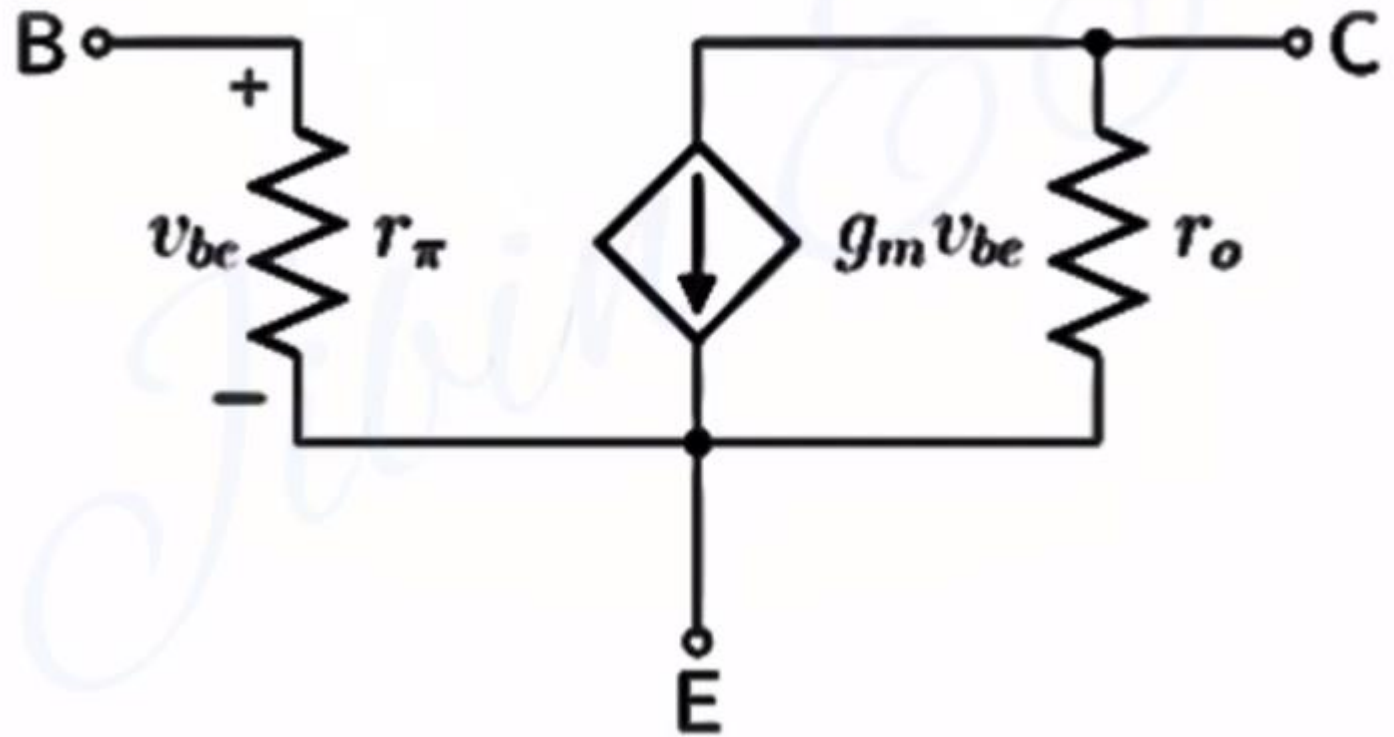
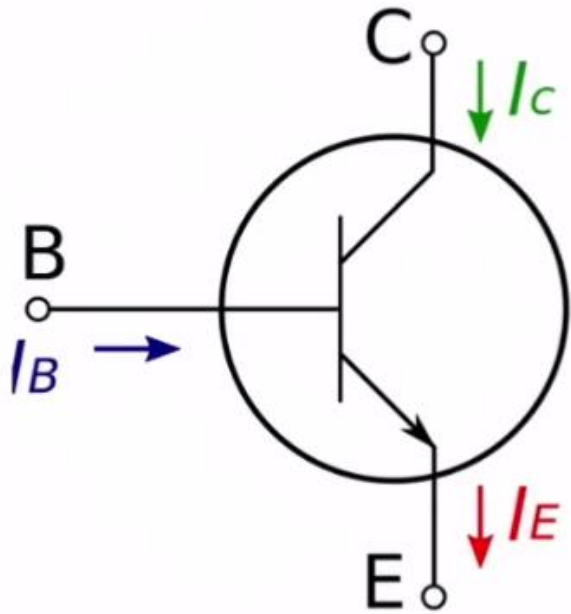


r_b - very small
 r_m - very large

$$X_c = \frac{1}{2\pi f C}$$

$f \downarrow \rightarrow X_c \uparrow$ (OPEN CIRCUIT)

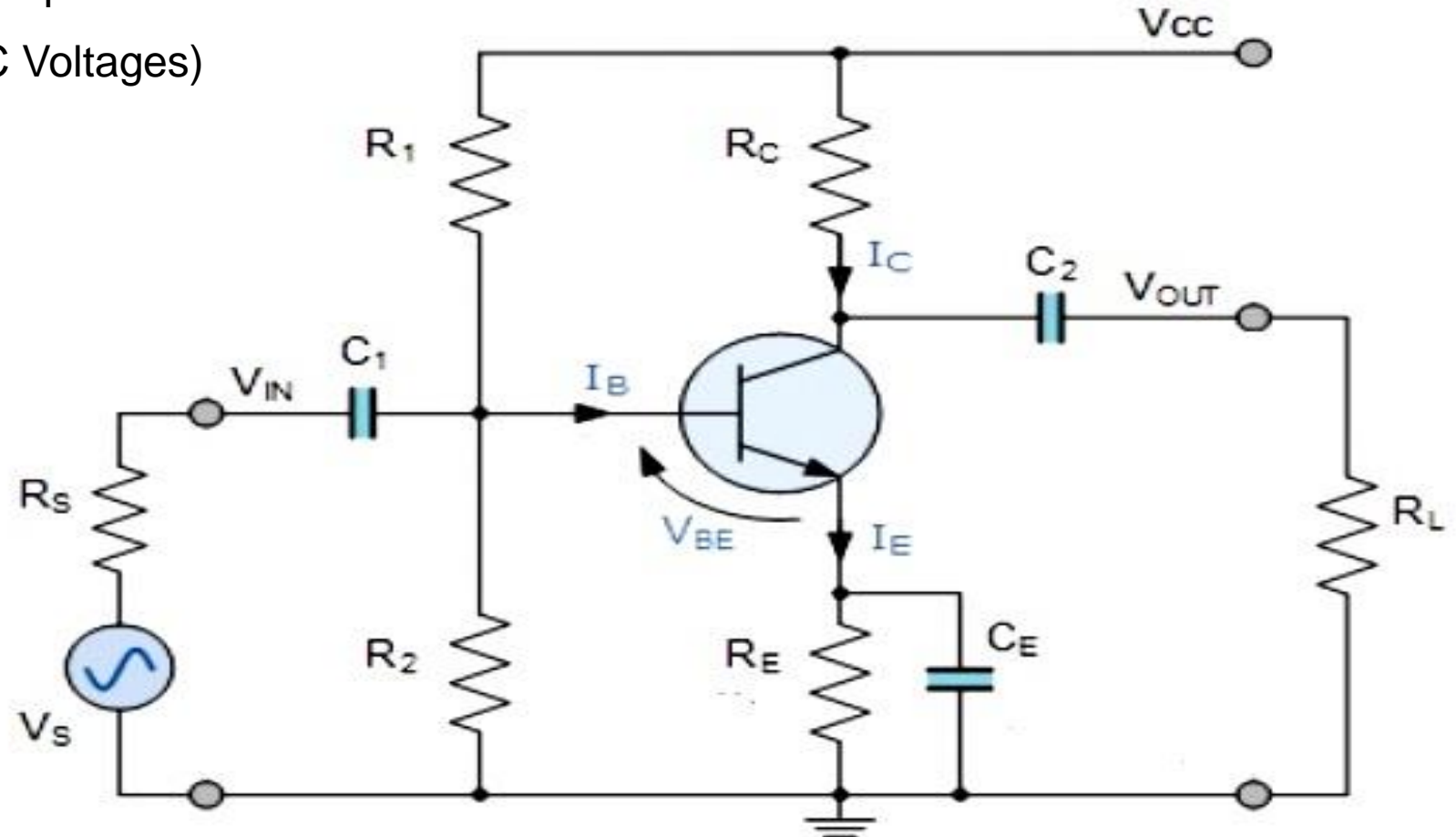
HYBRID π MODEL



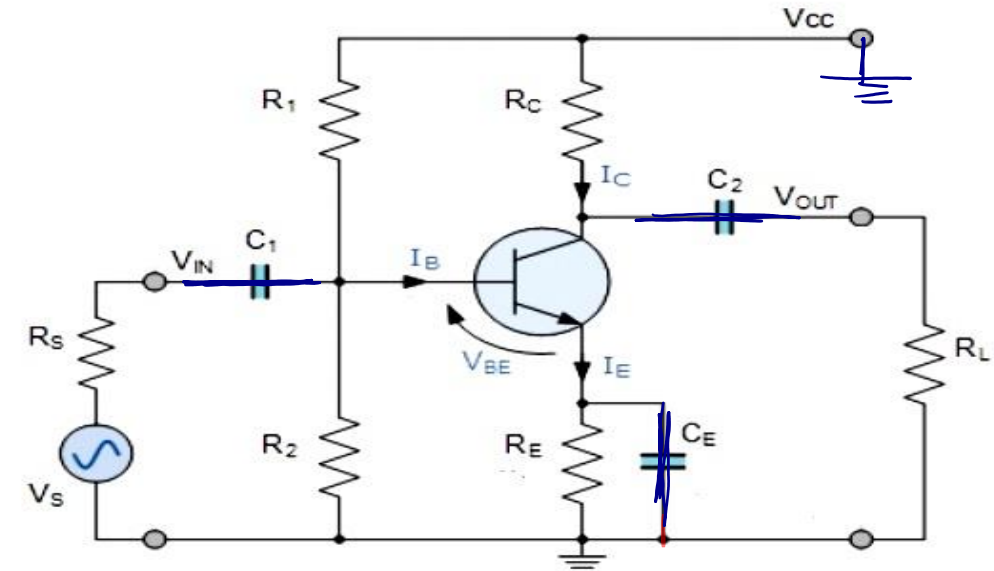
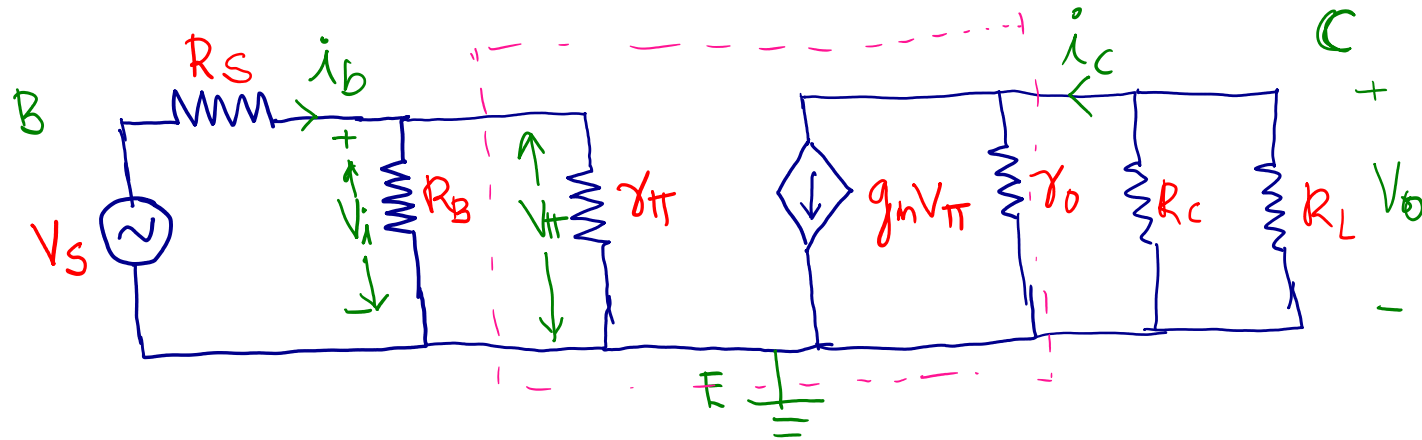
CE AMPLIFIER WITH BYPASS CAPACITOR

STEPS TO CONVERT TO HYBRID π MODEL

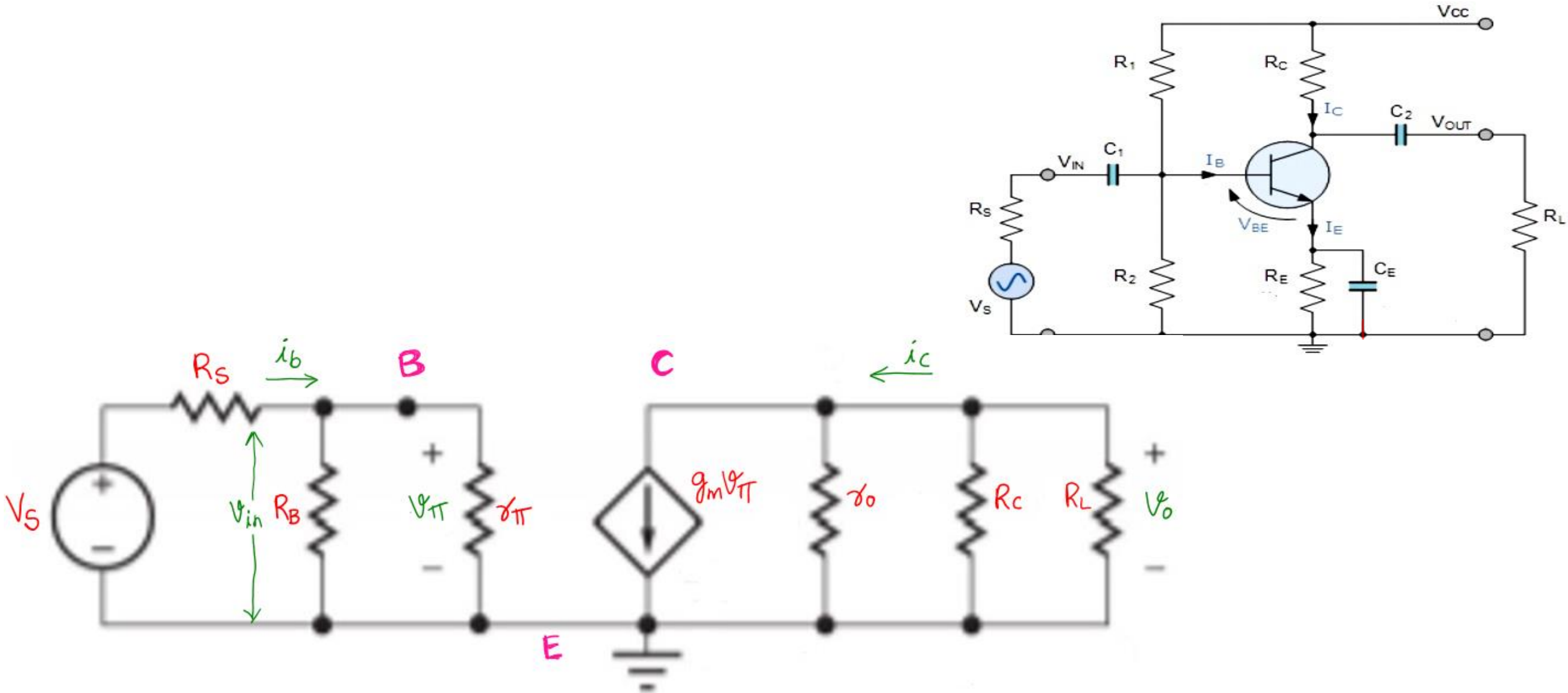
- Short circuit all Bypass and Coupling Capacitors
- Remove all external power supply (DC Voltages) and connect that node to ground



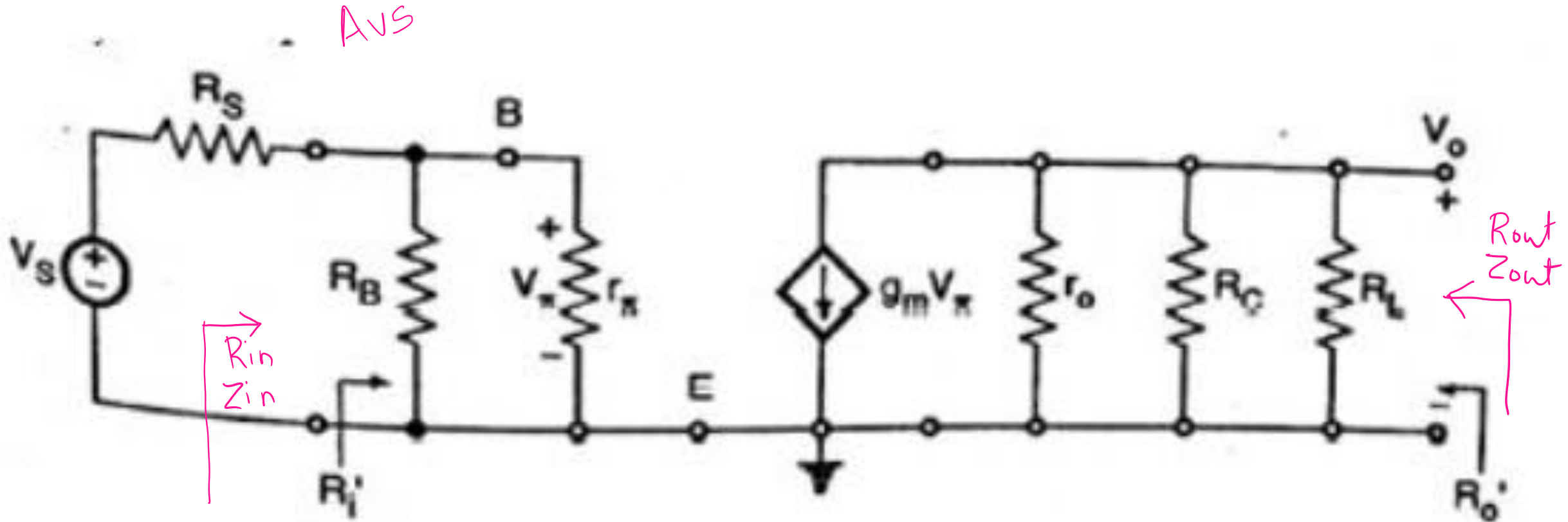
EQUIVALENT CIRCUIT (CE amplifier with bypass capacitor)



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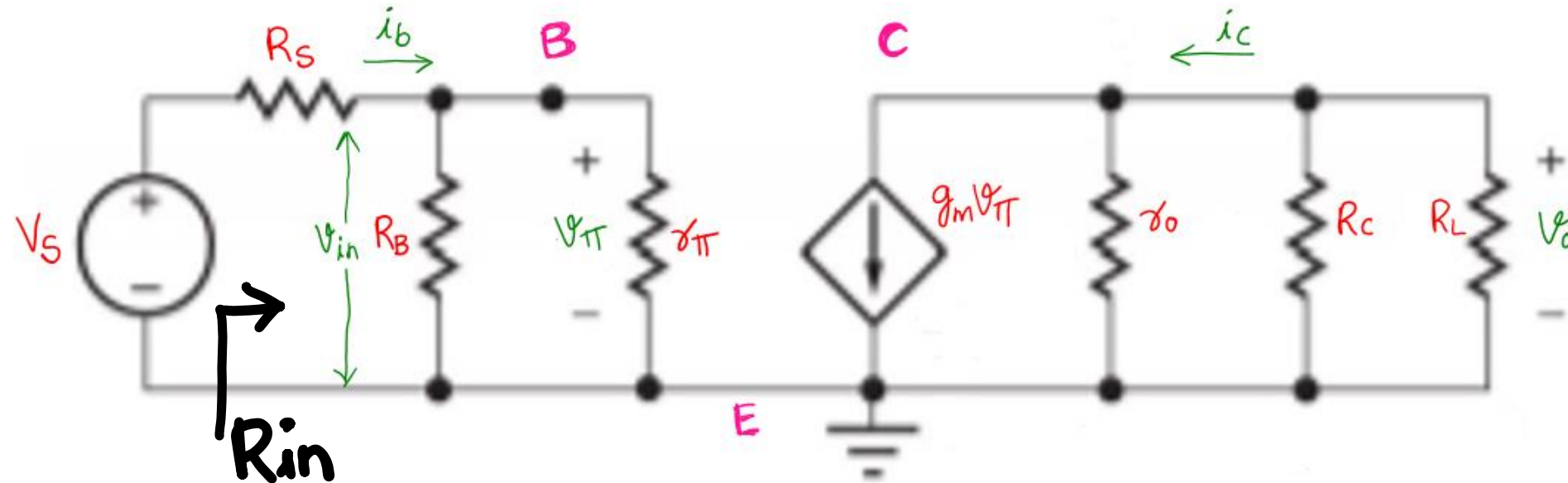
EQUIVALENT CIRCUIT (CE amplifier with bypass capacitor)

PARAMETERS TO DETERMINE

- INPUT RESISTANCE (R_{in})
- OUTPUT RESISTANCE (R_{OUT})
- VOLTAGE GAIN WITHOUT V_S (A_V)
- VOLTAGE GAIN WITH V_S (A_{VS})
- CURRENT GAIN (A_i)



Input Resistance (R_{in}) or Input Impedance (Z_{in})



$$R_{in} = Z_{in} = R_B \parallel r_{\pi}$$

Voltage Gain (without V_s) (A_v)

$$A_v = \frac{V_{out}}{V_{in}} = \frac{i_{out} \times R_{out}}{V_{in}}$$

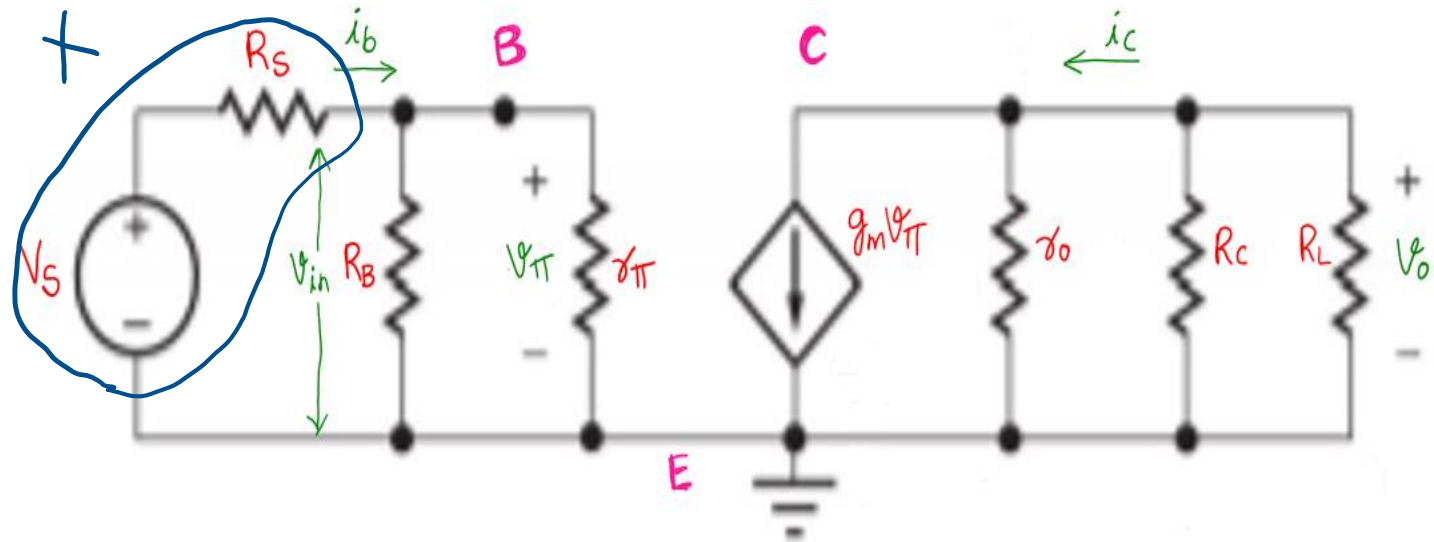
$$i_{out} = -g_m V_{\pi}$$

$$R_{out} = (R_L \parallel R_C \parallel r_o)$$

$$V_{\pi} \cong V_{in}$$

$$i_{out} = -g_m V_{in}$$

$$A_v = \frac{-g_m V_{in} \times (R_L \parallel R_C \parallel r_o)}{V_{in}}$$



$$A_v = -g_m (R_L \parallel R_C \parallel r_o)$$

Voltage Gain (with V_s) (A_{V_s})

$$A_{V_s} = \frac{V_{out}}{V_s}$$

$$R_B \parallel r_{\pi} \cong R_{in}$$

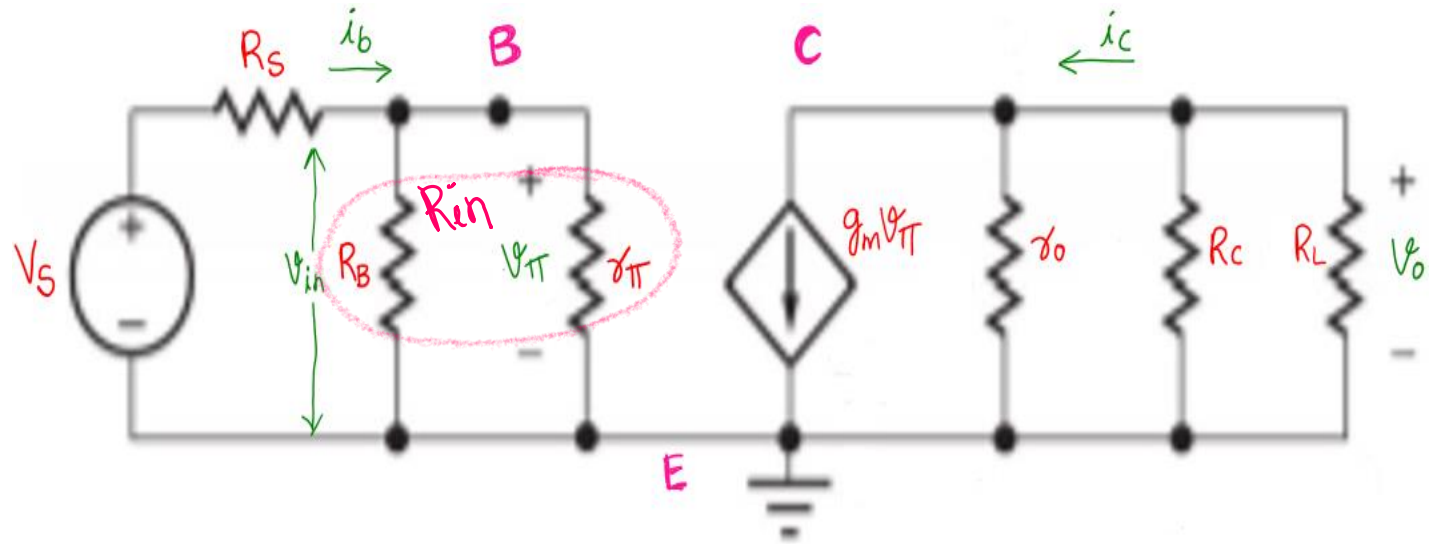
APPLYING VOLTAGE DIVIDER RULE AT INPUT SIDE

$$V_{in} = \frac{V_s \cdot R_{in}}{R_{in} + R_s}$$

$$\frac{V_{in}}{V_s} = \frac{R_{in}}{R_{in} + R_s}$$

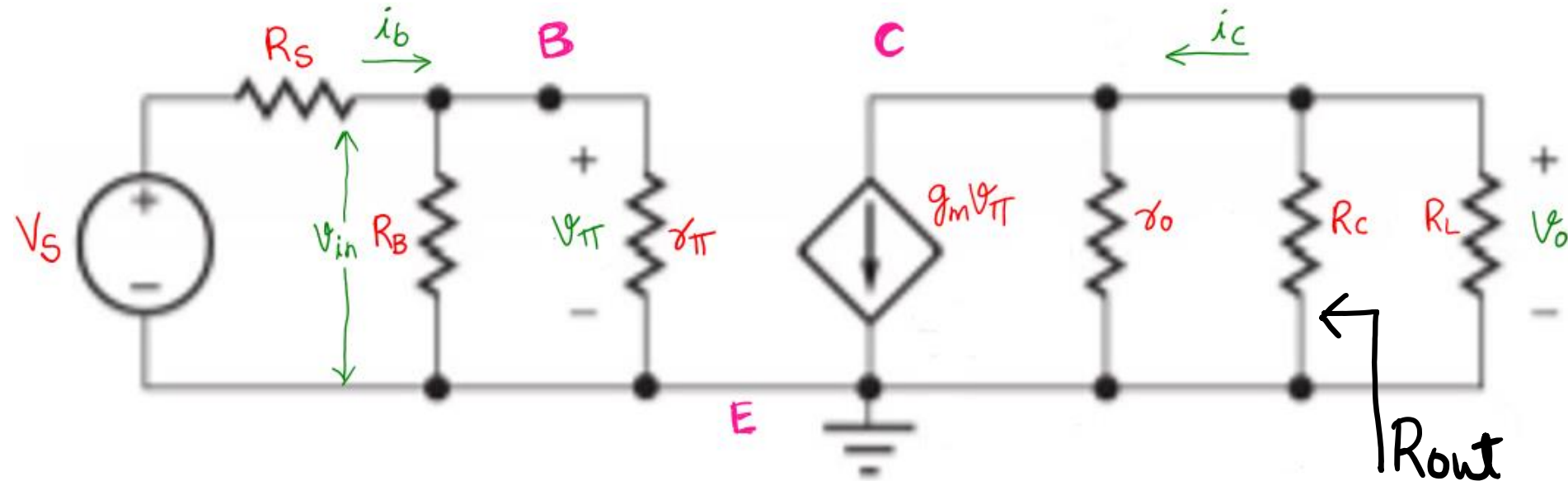
$$A_{V_s} = \frac{V_{out}}{V_s} = \frac{V_{out}}{V_s} \times \frac{V_{in}}{V_{in}} = \frac{V_{out}}{V_{in}} \times \frac{V_{in}}{V_s}$$

$$A_{V_s} = A_v \left(\frac{R_{in}}{R_{in} + R_s} \right)$$



$$A_{V_s} = -g_m (R_C \parallel R_L \parallel r_o) \left(\frac{R_{in}}{R_{in} + R_s} \right)$$

Output Resistance (R_{out}) or Output Impedance (Z_{out})



$$R_{out} = Z_{out} = (R_C \parallel r_o)$$



Current Gain (A_i)

$$A_i = \frac{i_{out}}{i_{in}}$$

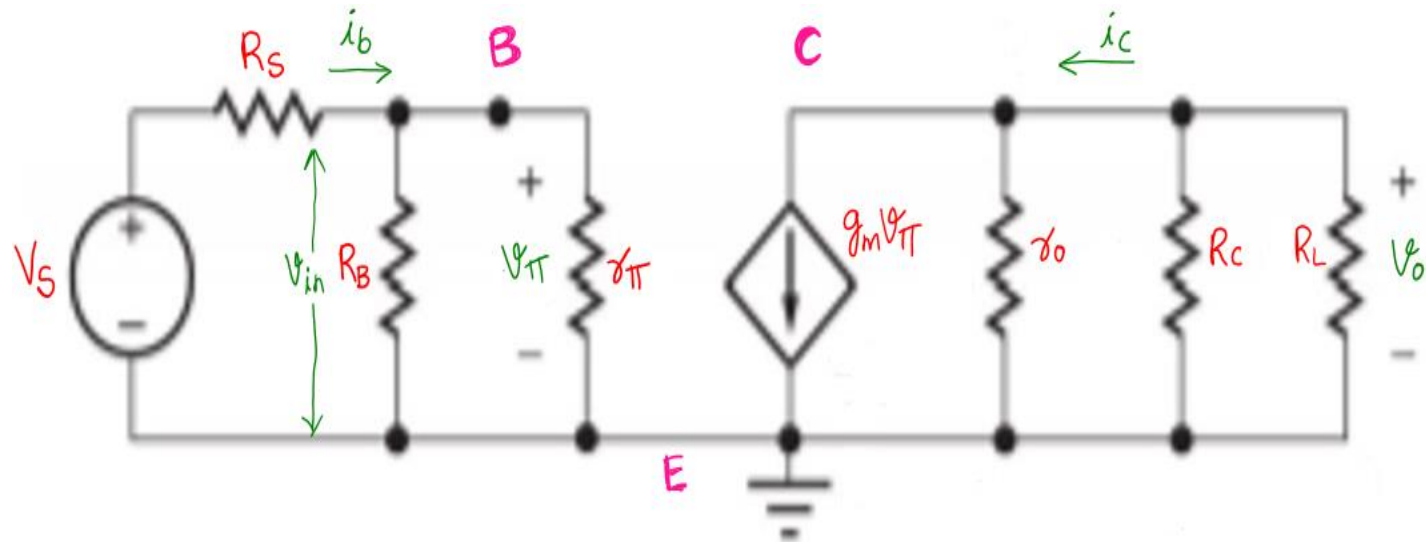
$$i_{out} = V_{out}/R_L$$

$$R_{in} \cong R_B \parallel \beta \pi$$

$$i_{in} = V_{in}/R_{in}$$

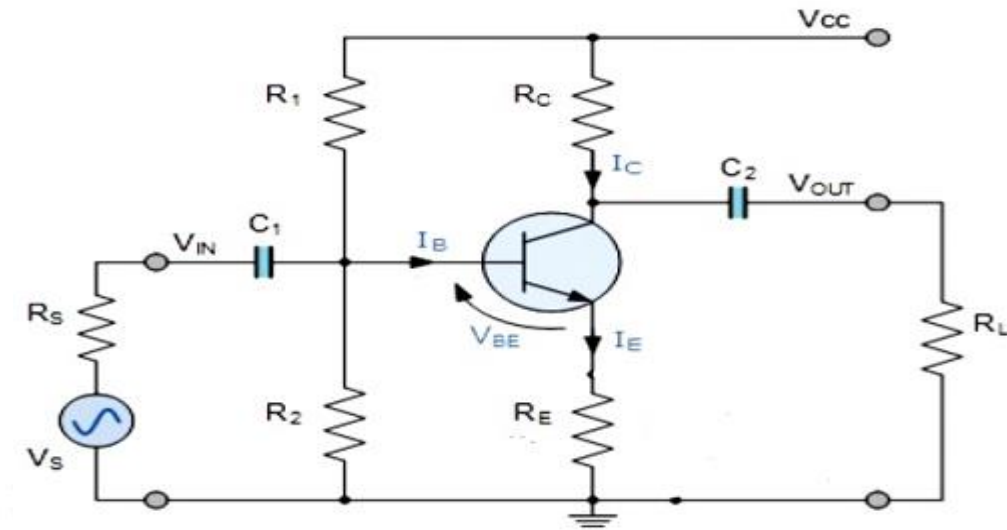
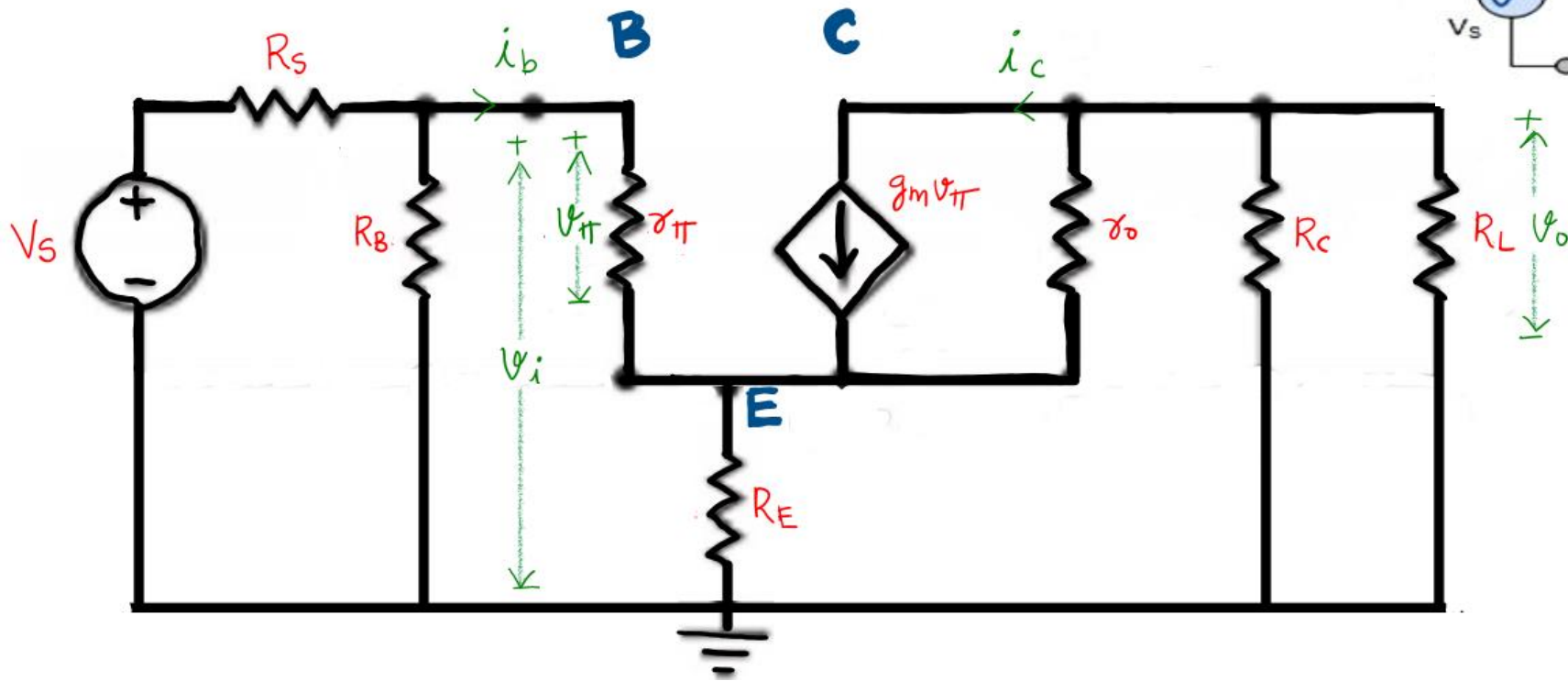
$$A_i = \frac{i_{out}}{i_{in}} = \frac{V_{out}/R_L}{V_{in}/R_{in}} = \frac{V_{out}}{V_{in}} \times \frac{R_{in}}{R_L}$$

$$A_i = A_v \cdot \frac{R_{in}}{R_L}$$



$$A_i = -g_m (R_C \parallel R_L \parallel r_o) \cdot \frac{(R_B \parallel \beta \pi)}{R_L}$$

CE amplifier without bypass capacitor



CE amplifier without bypass capacitor

PARAMETERS TO DETERMINE

- INPUT RESISTANCE (R_{in})
- VOLTAGE GAIN WITHOUT V_S (A_V)
- VOLTAGE GAIN WITH V_S (A_{VS})
- CURRENT GAIN (A_i)



Input Resistance (R_{in}) or Input Impedance (Z_{in})

$$R_{in} = R_B \parallel R_i' \quad R_i' = \frac{V_{in}}{i_b}$$

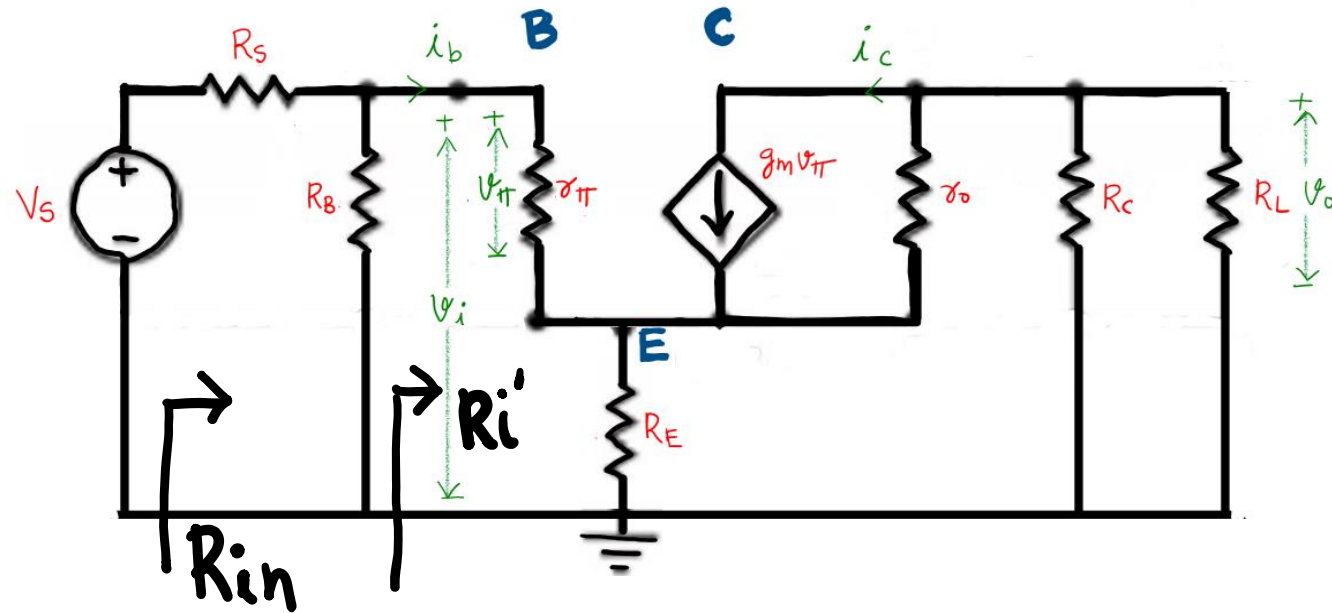
$$V_{in} = i_b r_{\pi} + (i_b + g_m V_{\pi}) R_E$$

$$V_{in} = i_b \left(r_{\pi} + \left(1 + g_m r_{\pi} \right) R_E \right)$$

$$V_{in} = i_b \left(r_{\pi} + (1 + g_m r_{\pi}) R_E \right)$$

$$R_i' = \frac{V_{in}}{i_b} = \frac{i_b (r_{\pi} + (1 + g_m r_{\pi}) R_E)}{i_b}$$

$$R_{in} = R_B \parallel (r_{\pi} + (1 + g_m r_{\pi}) R_E)$$



$$g_m r_{\pi} = \beta$$

$$R_i' = R_E (1 + \beta) + r_{\pi}$$

$$R_{in} = R_B \parallel (r_{\pi} + (1 + \beta) R_E)$$

Voltage Gain without V_s (A_v)

$$A_v = \frac{V_{out}}{V_{in}} \quad V_{out} = i_{out} \times R_{out}$$

NEGLECTING r_o term

$$V_{out} = (-g_m v_{\pi}) \cdot (R_c \parallel R_L)$$

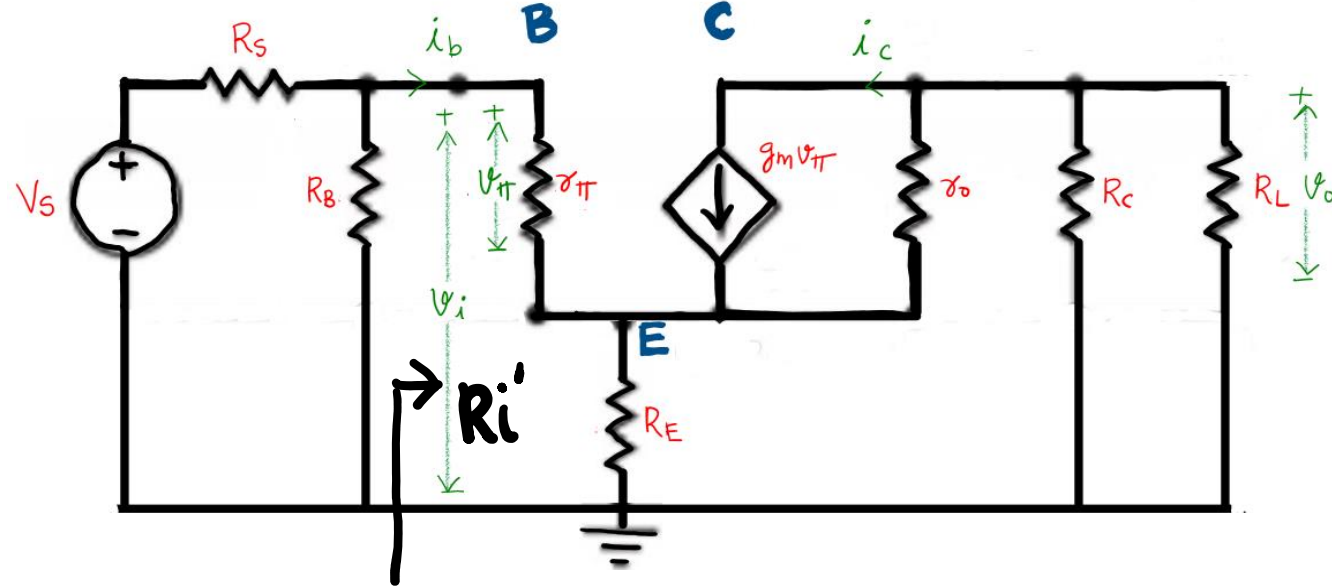
$$v_{\pi} = i_b r_{\pi} \quad V_{in} = R_{i'} \cdot i_b$$

$$A_v = \frac{(-g_m i_b r_{\pi}) \cdot (R_c \parallel R_L)}{R_{i'} \cdot i_b}$$

$$R_{i'} = R_E(1 + \beta) + r_{\pi} = R_E + \beta R_E + r_{\pi}$$

$$R_{i'} \simeq \beta R_E \quad \beta = g_m r_{\pi}$$

$$R_{i'} = g_m r_{\pi} R_E$$



$$A_v = \frac{-g_m r_{\pi} (R_c \parallel R_L)}{g_m r_{\pi} R_E}$$

$$A_v = -\frac{(R_c \parallel R_L)}{R_E}$$

Voltage Gain with Vs (Avs)

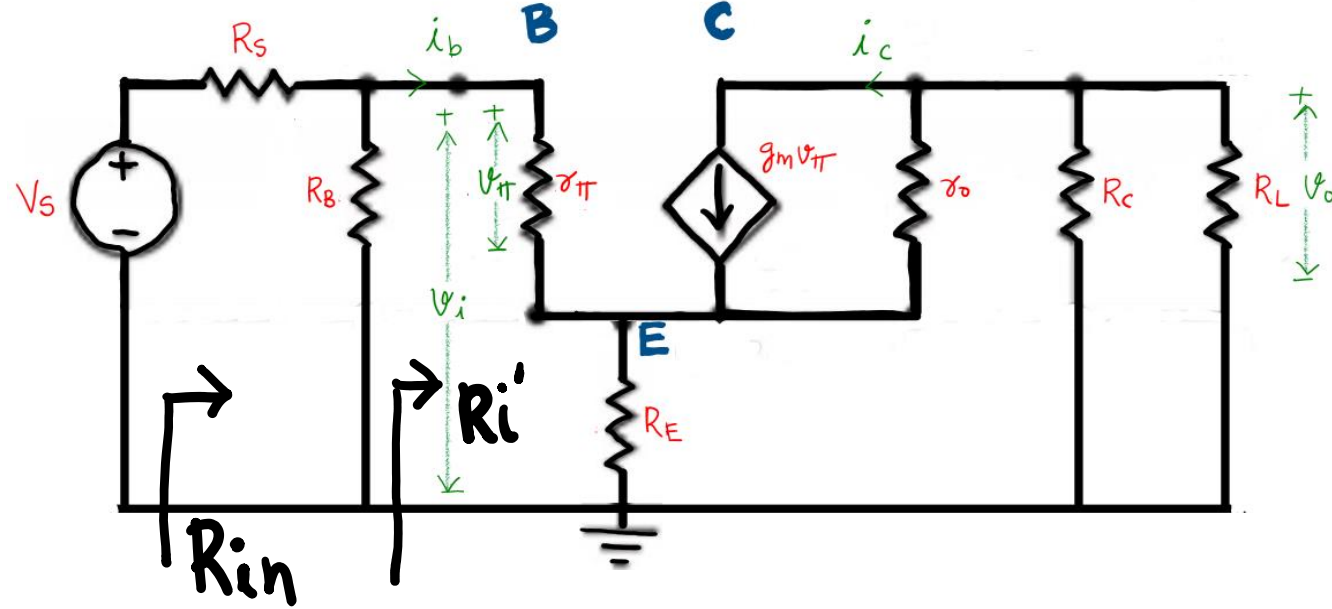
$$A_{V_s} = \frac{V_{out}}{V_s}$$

APPLYING VOLTAGE DIVIDER RULE AT INPUT SIDE

$$V_{in} = \frac{V_s \cdot R_{in}}{R_s + R_{in}} \quad \frac{V_{in}}{V_s} = \frac{R_{in}}{R_s + R_{in}}$$

$$A_{V_s} = \frac{V_{out}}{V_s} = \frac{V_{out}}{V_s} \times \frac{V_{in}}{V_{in}} = \frac{V_{out}}{V_{in}} \times \frac{V_{in}}{V_s}$$

$$A_{V_s} = A_v \times \left(\frac{R_{in}}{R_{in} + R_s} \right)$$



$$A_{V_s} = \frac{-(R_C \parallel R_L)}{R_E} \left(\frac{R_{in}}{R_{in} + R_s} \right)$$

Current Gain (A_i)

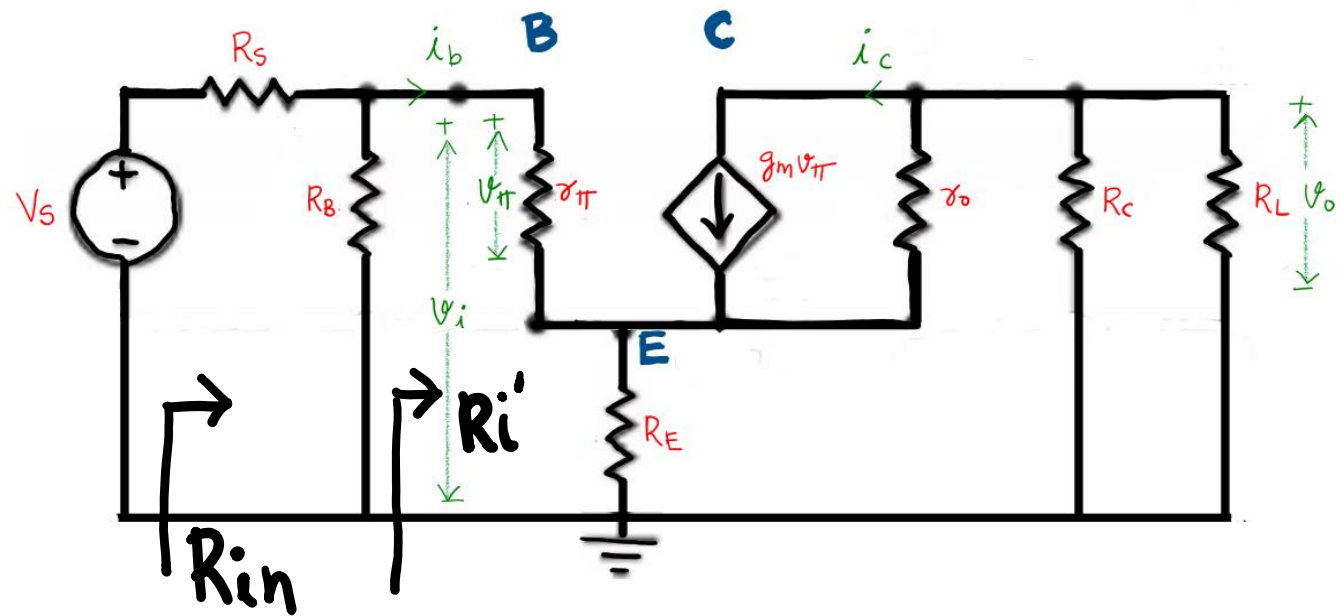
$$A_i = \frac{i_{out}}{i_{in}}$$

$$i_{out} = V_{out}/R_L$$

$$i_{in} = V_{in}/R_{in}$$

$$A_i = \frac{i_{out}}{i_{in}} = \frac{V_{out}/R_L}{V_{in}/R_{in}} = \frac{V_{out}}{V_{in}} \times \left(\frac{R_{in}}{R_L} \right)$$

$$A_i = A_v \left(\frac{R_{in}}{R_L} \right)$$



$$A_i = - \frac{(R_C \parallel R_L)}{R_E} \left(\frac{R_{in}}{R_L} \right)$$

