

IIR Filter Design

AIM

- To design a Low Pass Butterworth IIR filter and to plot its frequency response.
- To design a High Pass Butterworth IIR and to plot its frequency response.
- To Demonstrate the designed filters by filtering a composite frequency signal and plotting the filtered output

THEORY

A low-pass Butterworth filter is an analog all-pole filter with a squared magnitude response given by,

$$|H_C(j\Omega)|^2 = \frac{1}{1 + \left(\frac{j\Omega}{j\Omega_C}\right)^{2N}} = \frac{1}{1 + \epsilon^2 \left(\frac{j\Omega}{j\Omega_P}\right)^{2N}}$$

Where N is the order of the filter (number of poles in the system function), Ω_C is the 3-dB cut-off frequency and Ω_P is the passband edge frequency

$$\epsilon = \left(\frac{\Omega_P}{\Omega_C}\right)^N$$

As the order (N) increases, the transition band becomes narrower. i.e, the filter characteristics becomes sharper. The frequency response of the Butterworth filter decreases monotonically with increasing Ω . $|H_C(j\Omega)|^2$ is monotonic in both the passband and stopband.

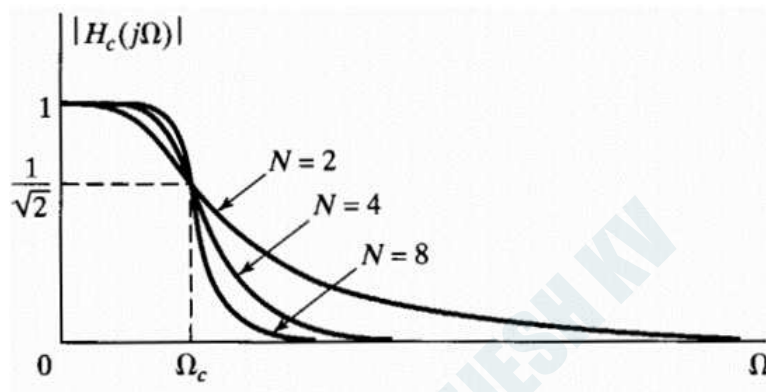
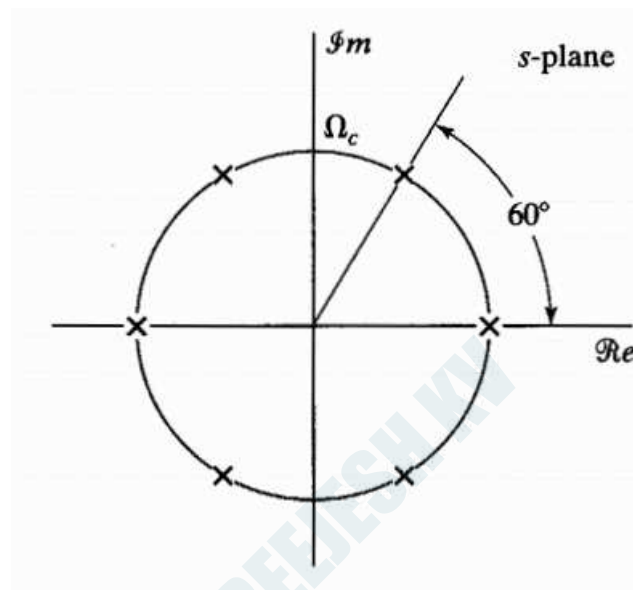


Figure 7.1: Magnitude Response of an analog Butterworth filter

The poles of the Butterworth filter are symmetrically located with respect to the imaginary axis. A pole never falls on the imaginary axis, and one occurs on the real axis for odd N , but not for even N .

The poles of the magnitude-squared function always occur in pairs; i.e., if there is a pole at $s = s_k$, then a pole also occurs at $s = -s_k$. Therefore, to construct $H_C(s)$ from the magnitude-squared function, we would choose one pole from each such pair: To obtain a stable and causal filter, we should choose the poles on the left-half-plane part of the s -plane. The angular spacing between the poles on the circle is π/N radians. For example, for $N = 3$, the poles are spaced by $\pi/3$ radians as shown below.

Figure 7.2: Pole locations of $H_C(s)H_C(-s)$ for a third order analog Butterworth filter

Design of IIR digital filters from analog filters

The traditional approach for designing discrete-time IIR filters is through transformation of a continuous-time filter into a discrete-time filter meeting prescribed specifications.

In designing a discrete-time filter by transforming a prototype continuous-time filter, the specifications for the continuous-time filter are obtained by a transformation of the specifications for the desired discrete-time filter.

Then the system function $H_C(s)$ or the impulse response $h_C(t)$ of the continuous-time filter is obtained through one of the established approximation methods.

Next, the system function $H(z)$ or impulse response $h[n]$ for the discrete-time filter is obtained by applying a transformation to $H_C(s)$ or $h_C(t)$.

In such transformations, the following essential properties of the continuous-time frequency response should be preserved in the frequency response of the resulting discrete-time filter.

- The imaginary axis of the s -plane should map onto the unit circle of the z -plane
- A stable continuous-time filter should be transformed to a stable discrete-time filter. i.e., if the continuous-time system has poles only in the left half of the s -plane, then the discrete-time filter must have poles only inside the unit circle in the z -plane

Bilinear transformation

This transformation avoids aliasing by using an algebraic transformation between the variables s and z that maps the entire $j\Omega$ -axis in the s -plane to one revolution of the unit circle in the z -plane.

The transformation is non-linear, since $-\infty \leq \Omega \leq \infty$ maps to $-\pi \leq \omega \leq \pi$. Therefore, Bilinear transformation is used only when warping of frequency axis is acceptable. The transformation is done by replacing s with,

$$s = \frac{2}{T_d} \left(\frac{1 - z^{-1}}{1 + z^{-1}} \right)$$

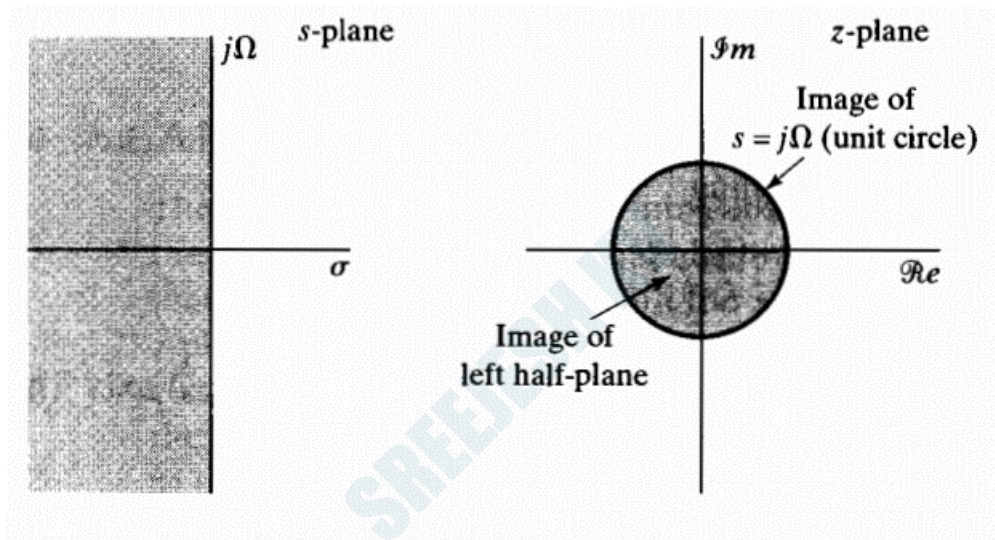


Figure 7.3: Bilinear Transformation

Thus, BLT avoids aliasing by mapping the entire imaginary axis of s plane to one complete revolution of the unit circle of z -plane. The price paid for this is the non-linear compression of frequency axis.

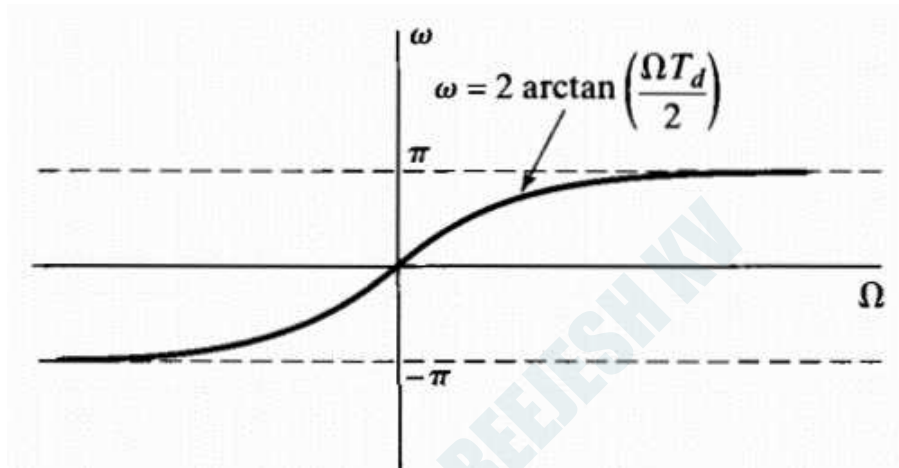


Figure 7.4: Mapping of the continuous-time frequency axis onto the discrete-time frequency axis by bilinear transformation

The relation between continuous-time frequency(Ω) and discrete-time frequency (ω) for bilinear transformation is given by,

$$\omega = 2 \arctan\left(\frac{\Omega T_d}{2}\right)$$

Prewarping in Bilinear Transformation

The relationship between the analog frequency Ω and the digital frequency ω is almost linear for small values of ω , but becomes nonlinear for large values of ω leading to a distortion (or warping) of the digital frequency response.

This effect is normally compensated for by **prewarping** the analog filter before applying the bilinear transformation. i.e., we prewarp one or more critical frequencies before applying the Bilinear Transformation. For example, for a lowpass filter we often prewarp the cutoff frequency as follows:

$$\Omega = \frac{2}{T_d} \tan(\omega/2)$$

MATLAB FUNCTIONS USED

buttord

Butterworth filter order and cutoff frequency

`[n,Wn] = buttord(Wp,Ws,Rp,Rs)` returns the lowest order, n , of the digital Butterworth filter with no more than R_p dB of passband ripple and at least R_s dB of attenuation in the stopband.

W_p and W_s are respectively the passband and stopband edge frequencies of the filter, normalized from 0 to 1, where 1 corresponds to π rad/sample. The scalar (or vector) of corresponding cutoff frequencies, W_n , is also returned.

To design a Butterworth filter, use the output arguments `n` and `Wn` as inputs to **butter**

butter

Butterworth filter design

`[b,a] = butter(n,Wn)` returns the transfer function coefficients of an n th-order lowpass digital Butterworth filter with normalized cutoff frequency `Wn`.
`[b,a] = butter(n,Wn,ftype)` designs a lowpass, highpass, bandpass, or bandstop Butterworth filter, depending on the value of `ftype` and the number of elements of `Wn`. The resulting bandpass and bandstop designs are of order $2n$.

freqz

Frequency response of digital filter

`[h,w] = freqz(b,a,n)` returns the n -point frequency response vector `h` and the corresponding angular frequency vector `w` for the digital filter with transfer function coefficients stored in `b` and `a`.
`freqz(__)` with no output arguments plots the frequency response of the filter.
`[h,f]=freqz(__,n,fs)` returns the frequency response vector `h` and the corresponding physical frequency vector `f` for a digital filter designed to filter signals sampled at a rate `fs`.

Note:

For digital filter design, **butter** uses **bilinear** to convert the analog filter into a digital filter through a bilinear transformation with frequency prewarping

ALGORITHM

- Step 1. Start
- Step 2. Input the pass band ripple `rp`, stop band ripple `rs`, pass band frequency `fp` and stop band frequency `fs` and the sampling frequency `fsamp`
- Step 3. Convert the frequencies to digital and normalize them.
- Step 4. Find the cut off frequency and lowest order `N` of the filter satisfying the given specifications using MATLAB's **buttord** function
- Step 5. Using the above values find the coefficients of the low pass and high pass filters using **butter** function

- Step 6. Find the frequency response of the filters and plot them
- Step 7. Create a composite signal as the sum of two sinusoids: one having a frequency less than f_p and the other having a frequency higher than f_s
- Step 8. Filter the composite signal using the designed lowpass and highpass filters and observe the output in time domain.
- Step 9. Stop

PROGRAM

```

1 %Title: Program to
2 % 1) Design a Low Pass Butterworth IIR filter and to plot its ...
   frequency response.
3 % 2) Design a High Pass Butterworth IIR and to plot its frequency ...
   response.
4 % 3) Demonstrate the designed filters by filtering a composite
5 % frequency signal and plotting the filtered output
6
7 %Author: Sreejesh K V, Dept. of ECE, GCEK
8 %Date: 05/11/2022
9
10 clc;
11 clear;
12 close all;
13
14 fp=200;%the pass band edge frequency in Hz
15 fs=400;%the stop band edge frequency in Hz
16 rp=1;% peak-to-peak passband ripple in dB
17 rs=40;%the minimum stopband attenuation in dB
18 fsamp=3000;% Sampling frequency in Hz
19
20 wp=2*fp/fsamp;%normalized digital passband edge frequency(=dig ...
   frequency/pi)
21 ws=2*fs/fsamp;%normalized digital stopband edge frequency
22
23 % --- Designing the Low Pass Butterworth Filter --- %
24 [N,wn]=buttord(wp,ws,rp,rs);% returns the lowest order N of a digital
25 %Butterworth filter satisfying the given PB & SB ripple conditions.
26 %Wp and Ws are the PB & SB edge frequencies.
27 % Wn is the Butterworth natural(3dB) frequency
28
29
30 [b,a]=butter(N,wn,'low');% designs an Nth order lowpass digital
31 %Butterworth filter and returns the filter coefficients in length
32 % N+1 vectors B (numerator) and A (denominator)
33 figure,
34 freqz(b,a,500,fsamp);%plotting the frequency response of the LPF

```

```

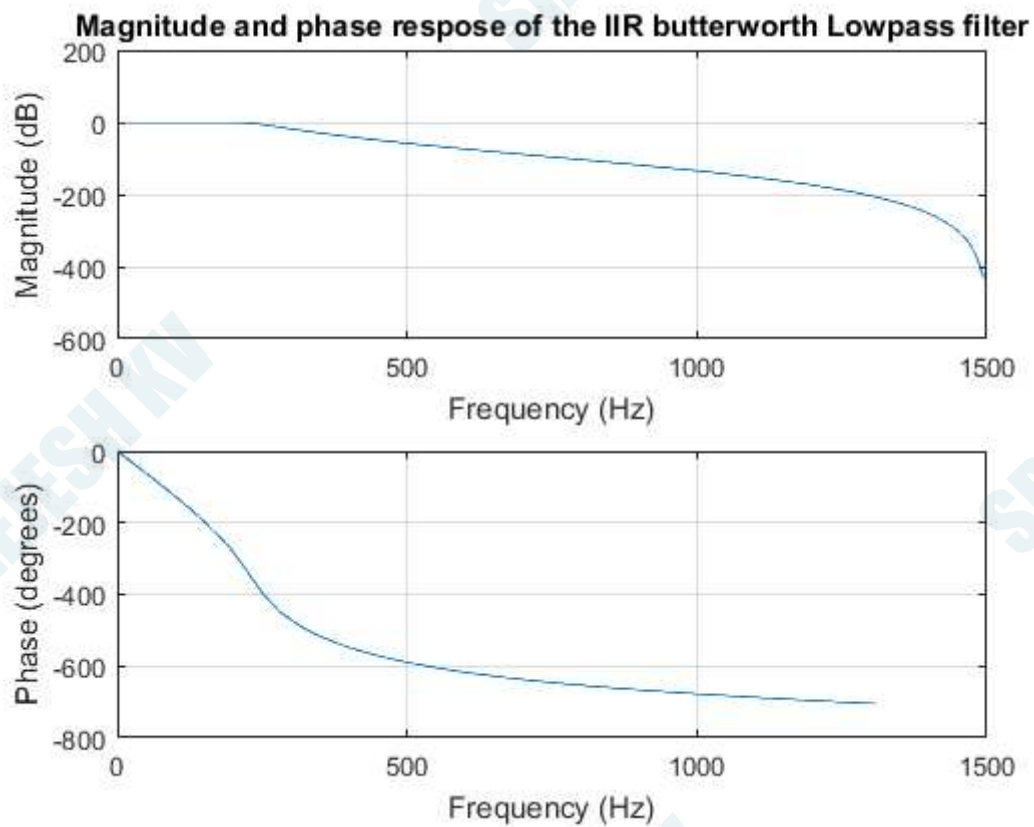
35 title('Magnitude and phase response of the IIR butterworth Lowpass ...
    filter');
36
37 %--- Designing the High Pass Butterworth Filter --- %
38 [bh,ah]=butter(N,wn,'high');% designs an Nth order lowpass digital
39 %Butterworth filter and returns the filter coefficients in length
40 % N+1 vectors B (numerator) and A (denominator)
41
42 figure,
43 freqz(bh,ah,500,fsamp);%plotting the frequency response of the HPF
44 title('Magnitude and phase response of the IIR butterworth ...
    HighPass filter');
45
46 %--- Demonstration of the Filters --%
47 f1=100;%100Hz : in the passband of LPF and in the SB of HPF
48 f2=500;%500Hz: in the stopband of LPF and in the PB of HPF
49 T=1/f1;
50 t=0:1/fsamp:3*T;
51 s1=sin(2*pi*f1*t);%100Hz signal(approximation using only 3 periods)
52 s2=sin(2*pi*f2*t);%500Hz signal(approximation)
53 s=s1+s2;%the composite signal containing 100Hz and 500Hz ...
    components(mainly)
54
55 %---Filtering Process--%
56 y=filter(b,a,s);%low pass filtering
57 y1=filter(bh,ah,s);%highpass filtering
58
59 %---Plotting the signals--%
60 figure;
61 subplot(2,2,1);
62 plot(t,s1,'r','LineWidth',2);
63 hold on
64 plot(t,s2);
65 legend('s1','s2');
66 title(['Signals s1 ( f = ' num2str(f1) ' Hz) & s2 ( f = ' ...
    num2str(f2) ' Hz) ']);
67 xlabel('time');
68 ylabel('Amplitude');
69
70 subplot(2,2,2);
71 plot(t,s);
72 title('Composite signal=s1+s2');
73 xlabel('time');
74 ylabel('Amplitude');
75
76 subplot(2,2,3);
77 plot(t,y);
78 title('LowPass Filtered Output');
79 xlabel('time');
80 ylabel('Amplitude');
81

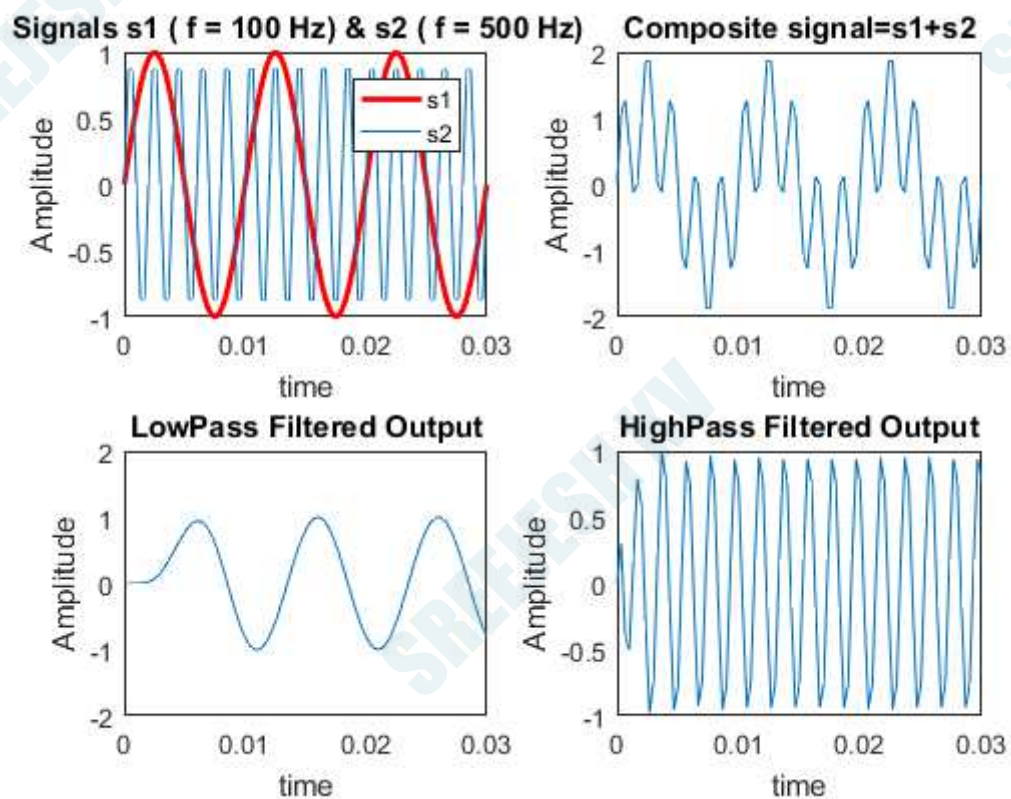
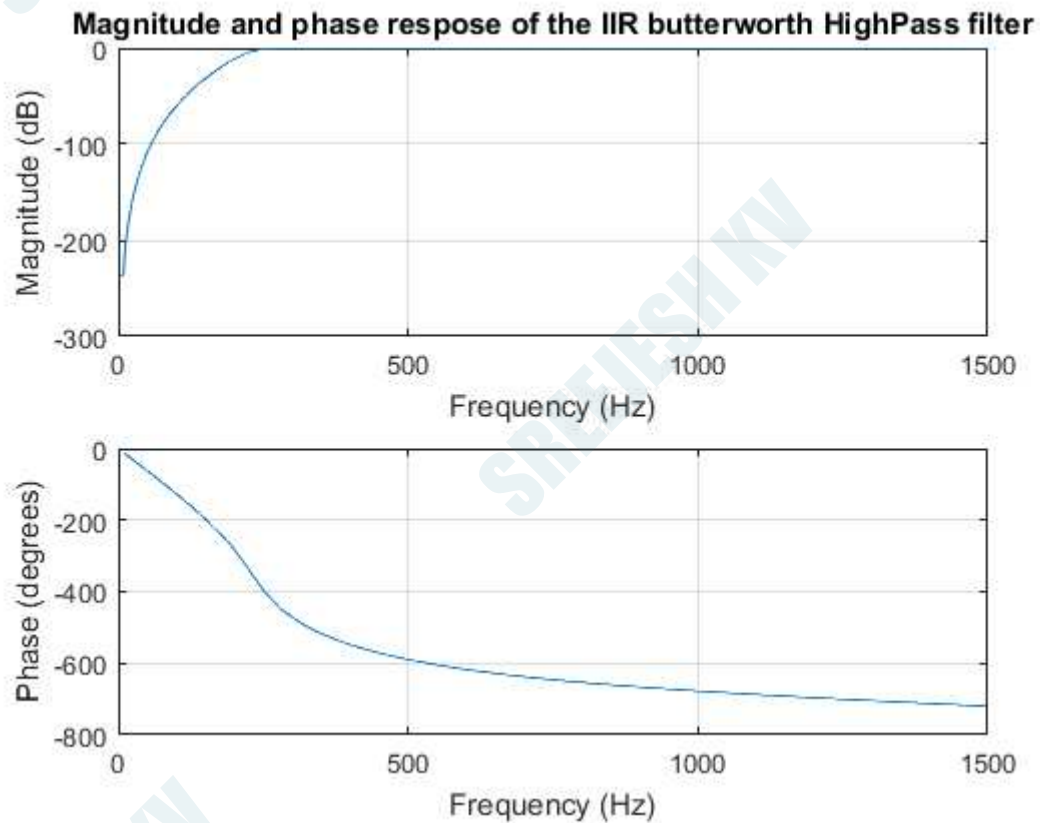
```

```
82 subplot(2,2,4);  
83 plot(t,y1);  
84 title('HighPass Filtered Output');  
85 xlabel('time');  
86 ylabel('Amplitude');
```

OUTPUT & OBSERVATIONS

Figure Window Outputs:





RESULTS

- (a) A Low Pass Butterworth IIR filter was designed using MATLAB and its frequency response was plotted

- (b) A High Pass Butterworth IIR filter was designed using MATLAB and its frequency response was plotted
- (c) The designed filters were demonstrated by filtering a composite frequency signal and plotting the filtered output