

Poles of a Transfer function $H(s)$

Consider the transfer function

$$H(s) = \frac{s-a}{(s-b)(s-c)}.$$

Poles of the transfer function $H(s)$, is the values of 's' for which the function $H(s)$ approaches infinity [or denominator of $H(s)$ becomes zero].

$$\text{at poles: } (s-b)(s-c) = 0$$

$$\text{ie either } (s-b) = 0 \text{ or } (s-c) = 0$$

$$\text{so the poles are } \underline{s=b}, \underline{s=c}.$$

$$\text{at zeros: The function } H(s) \text{ becomes zero.}$$

$$\text{ie } H(s) = 0$$

$$\text{for that numerator of } H(s) = 0$$

$$\text{ie } (s-a) = 0$$

$$\underline{s=a}.$$

For example.

$$H(s) = \frac{s-1}{(s+1)(s+2)}.$$

$$\text{Zeros: } (s-1) = 0 \Rightarrow s=1. \text{ zeros denoted '0'}$$

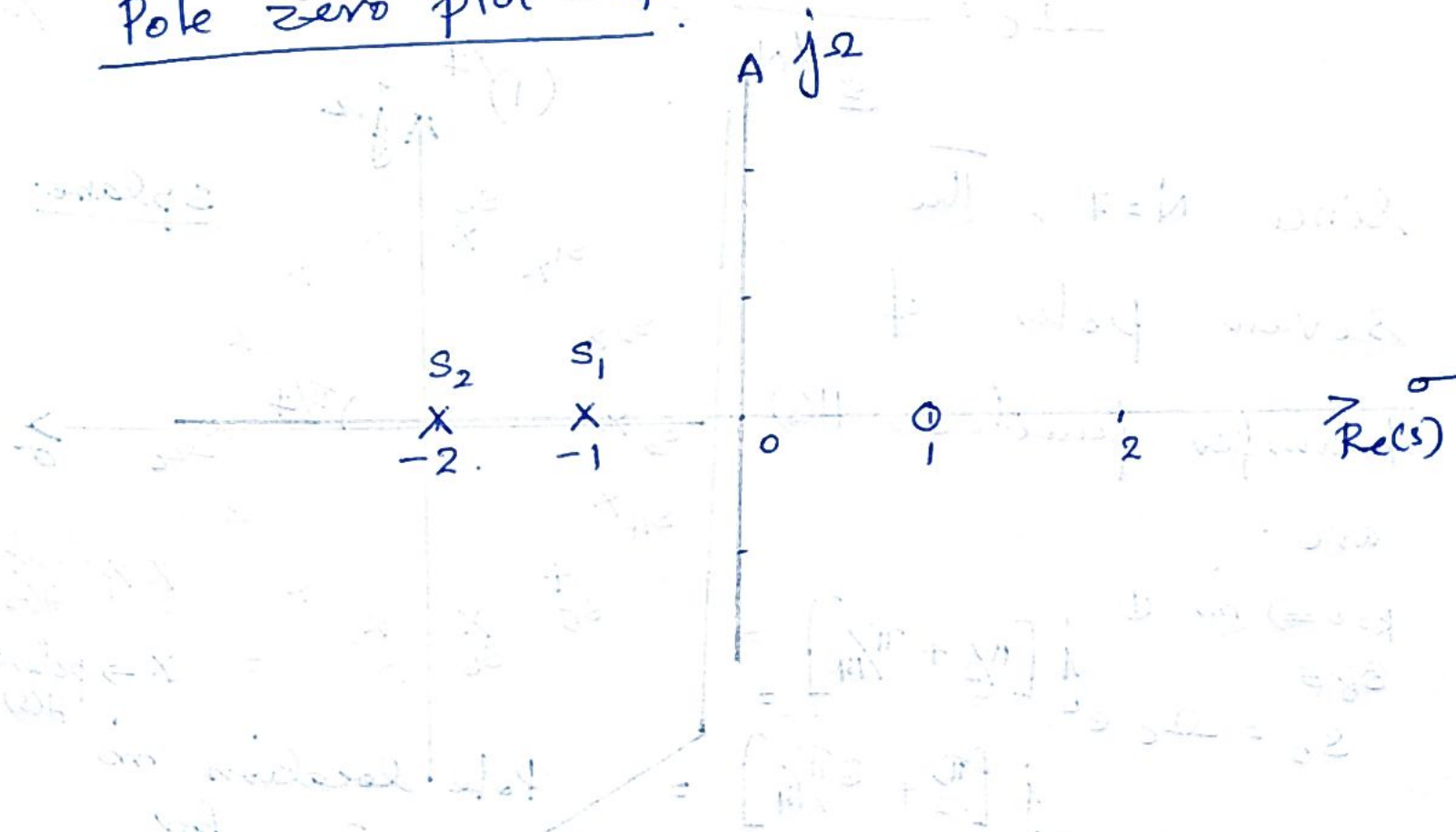
poles : $(s+1) = 0 \Rightarrow s = -1$

$(s+2) = 0 \Rightarrow s = -2.$

$\therefore H(s)$ has two poles at $s = -1$ and $s = -2.$

Poles are denoted as 'x'

Pole zero plot of $H(s)$



2) Determine the order and the poles of lowpass Butterworth filter that has a 3 dB attenuation at 500 Hz and an attenuation of 40 dB at 1000 Hz.

Ans.

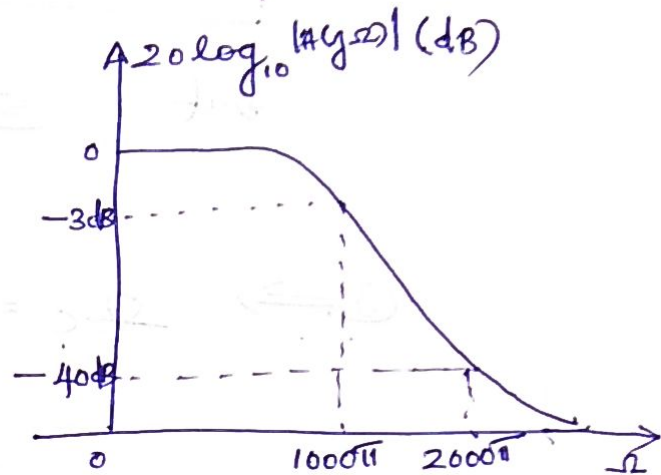
Given $\alpha_p = 3 \text{ dB}$.

$f_p = 500 \text{ Hz}$.

$$\therefore \omega_p = 2\pi f_p$$

$$= 2\pi \times 500$$

$$= 1000\pi \text{ rad/sec}$$



and $\alpha_s = 40 \text{ dB}$

$f_s = 1000 \text{ Hz}$.

$$\therefore \omega_s = 2\pi f_s$$

$$= 2\pi \times 1000 = 2000\pi \text{ rad/sec}$$

$$\boxed{\omega = 2\pi f}$$

$$\lambda = \sqrt{10^{0.1\alpha_s} - 1} =$$

$$\varepsilon = \sqrt{10^{0.1\alpha_p} - 1} =$$

$$N \geq \frac{\log_{10} (\lambda/\varepsilon)}{\log_{10} (\omega_s/\omega_p)} = 6.6$$

Rounding 'N' to nearest higher value we get $N = 7$

Poles of Butter worth function is given by

$$s_k = -\Omega_c e^{j \left[\frac{\pi}{2} + (2k+1) \frac{\pi}{2N} \right]}$$

①

$k = 0, 1, \dots, N-1$

Cut off frequency

$$\Omega_c = \frac{\Omega_p}{\varepsilon^{1/N}} = \frac{1000\pi}{(1)^{1/4}} = 1000\pi \text{ rad/sec.}$$

Since $N=7$, The seven poles of transfer function $H(s)$ are.

$k=0 \Rightarrow m$ ①

$$s_0 = -\Omega_c e^{j \left[\frac{\pi}{2} + \frac{\pi}{14} \right]} =$$

$$s_1 = -\Omega_c e^{j \left[\frac{\pi}{2} + 3\frac{\pi}{14} \right]} =$$

$$s_2 = -\Omega_c e^{j \left[\frac{\pi}{2} + 5\frac{\pi}{14} \right]} =$$

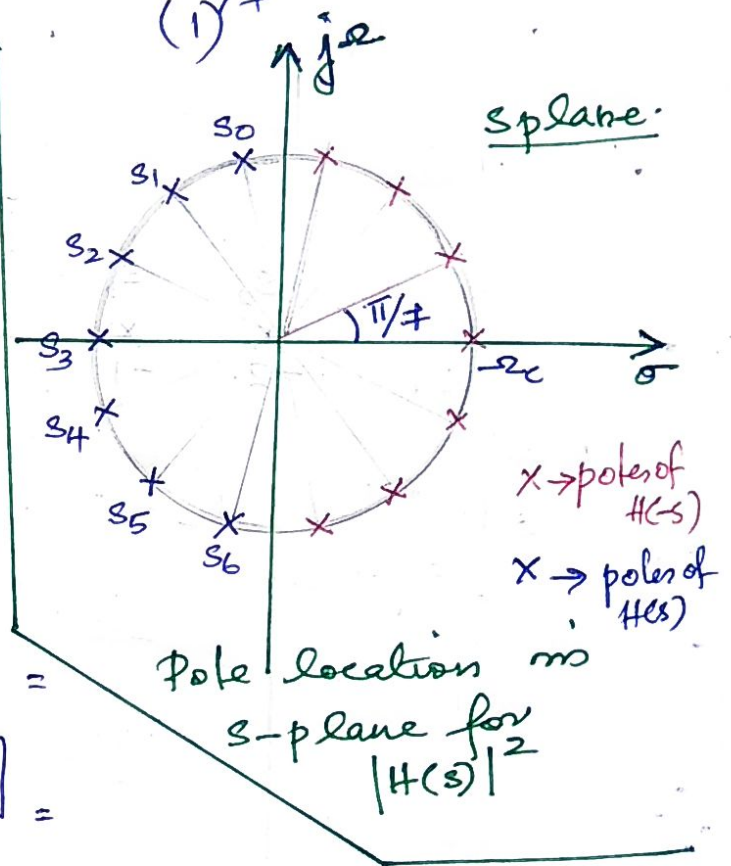
s_3, s_4, s_5, s_6 (do as Homework).

Then the transfer function

$$H(s) = \frac{1}{(s-s_0)(s-s_1)(s-s_2)(s-s_3)(s-s_4)(s-s_5)(s-s_6)}$$

[\therefore Butter worth polynomial in all pole transfer function]

[For a stable transfer function $H(s)$ poles are on Left half of s-plane]



The following table 5.1 gives Butterworth polynomials for various values of N for $\Omega_c = 1$ rad/sec.

Table 5.1 List of Butterworth Polynomials

N	Denominator of $H(s)$
1	$s + 1$
2	$s^2 + \sqrt{2}s + 1$
3	$(s + 1)(s^2 + s + 1)$
4	$(s^2 + 0.76537s + 1)(s^2 + 1.8477s + 1)$
5	$(s + 1)(s^2 + 0.61803s + 1)(s^2 + 1.61803s + 1)$
6	$(s^2 + 1.931855s + 1)(s^2 + \sqrt{2}s + 1)(s^2 + 0.51764s + 1)$
7	$(s + 1)(s^2 + 1.80194s + 1)(s^2 + 1.247s + 1)(s^2 + 0.445s + 1)$

The Eq. (5.12) gives us the pole locations of Butterworth filter for $\Omega_c = 1$ rad/sec and are known as normalized poles. In general, the unnormalized poles are given by

$$s_k' = \Omega_c s_k \quad (5.14)$$

The transfer function of such type of Butterworth filter can be obtained by substituting $s \rightarrow s/\Omega_c$ in the transfer function of Butterworth filter.

5.6 Steps to design an analog Butterworth lowpass filter

1. From the given specifications find the order of the filter N .
2. Round off it to the next higher integer.
3. Find the transfer function $H(s)$ for $\Omega_c = 1$ rad/sec for the value of N .
4. Calculate the value of cutoff frequency Ω_c .
5. Find the transfer function $H_a(s)$ for the above value of Ω_c by substituting $s \rightarrow \frac{s}{\Omega_c}$ in $H(s)$.

Example 5.4 Design an analog Butterworth filter that has a -2 dB passband attenuation at a frequency of 20 rad/sec and at least -10 dB stopband attenuation at 30 rad/sec.

Solution

Given $\alpha_p = 2$ dB; $\Omega_p = 20$ rad/sec
 $\alpha_s = 10$ dB; $\Omega_s = 30$ rad/sec

$$N \geq \frac{\log \sqrt{\frac{10^{0.1\alpha_s} - 1}{10^{0.1\alpha_p} - 1}}}{\log \frac{\Omega_s}{\Omega_p}}$$

$$\begin{aligned} &\geq \frac{\log \sqrt{\frac{10-1}{10^{0.2}-1}}}{\log \frac{30}{20}} \\ &\geq 3.37 \end{aligned}$$

Rounding off N to the next highest integer we get

$$N = 4$$

The normalized lowpass Butterworth filter for $N = 4$ can be found from table 5.1 as

$$H(s) = \frac{1}{(s^2 + 0.76537s + 1)(s^2 + 1.8477s + 1)}$$

From Eq. (5.31) we have

$$\Omega_c = \frac{\Omega_p}{(10^{0.1\alpha_p} - 1)^{1/2N}} = \frac{20}{(10^{0.2} - 1)^{1/8}} = 21.3868$$

Note:

To find the cutoff frequency Ω_c either (5.31) or (5.32a) can be used. The Eq. (5.31) satisfies passband specification at Ω_p , while the stopband specification at Ω_s is exceeded. The Eq. (5.32a) satisfies the stopband specification Ω_s , while the passband specification at Ω_p is exceeded. All the examples in this chapter are solved using Eq. (5.31). Students are advised to solve the exercise problems using Eq. (5.32a).

The transfer function for $\Omega_c = 21.3868$ can be obtained by substituting

$$s \rightarrow \frac{s}{21.3868} \text{ in } H(s)$$

$$\begin{aligned} \text{i.e., } H(s) &= \frac{1}{\left(\frac{s}{21.3868}\right)^2 + 0.76537 \left(\frac{s}{21.3868}\right) + 1} \\ &\quad \frac{1}{\left(\frac{s}{21.3868}\right)^2 + 1.8477 \left(\frac{s}{21.3868}\right) + 1} \\ &= \frac{0.20921 \times 10^6}{(s^2 + 16.3686s + 457.394)(s^2 + 39.5176s + 457.394)} \end{aligned}$$