

Conversion of the Continuous AWGN Channel into a Vector Channel

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- Continuous signal to vector representation

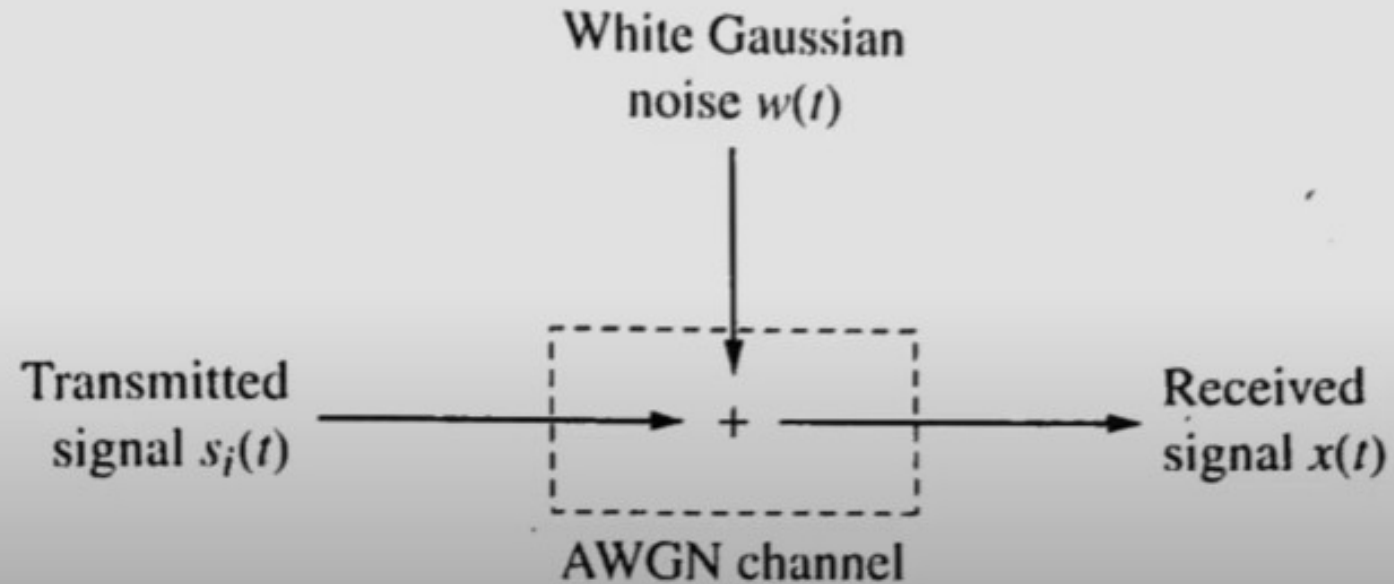
$$S_i(t) \text{ ----- } S_i =$$

Similarly we convert continuous AWGN channel to its vector model.

AWGN – additive white Gaussian noise

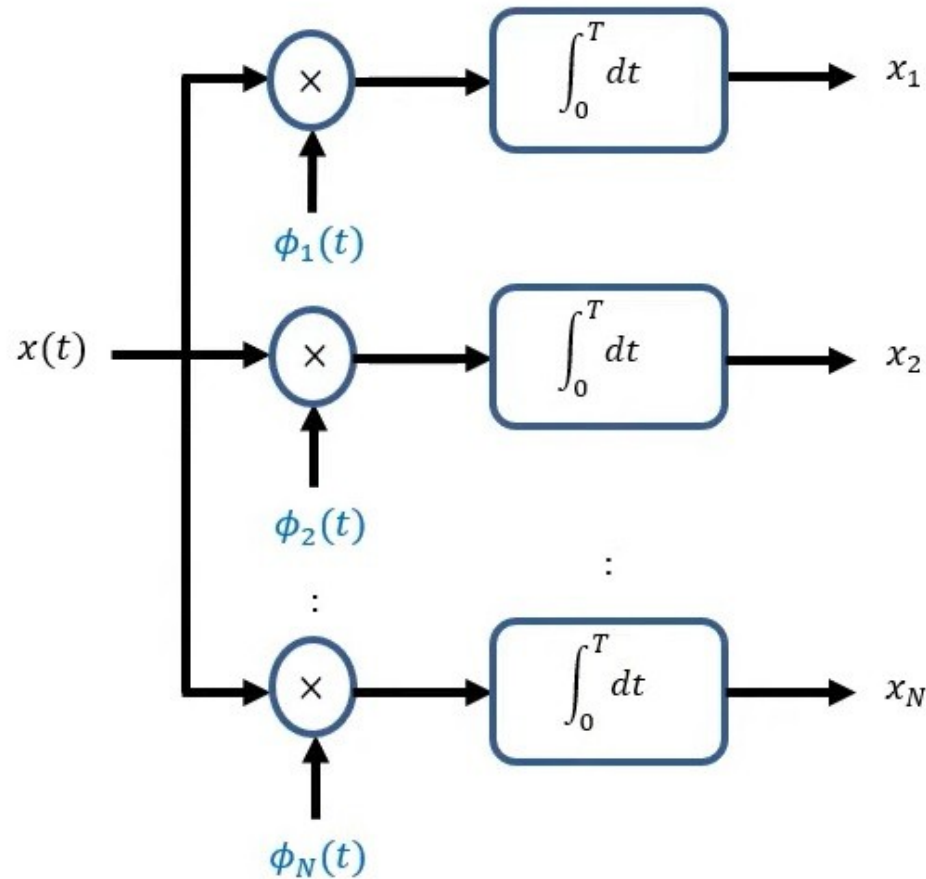
- **Additive** – As its name suggests, noise is added to a signal. Noise is generated randomly.
- **White** –noise has the same power distribution at every frequency. Therefore, white noise has a constant Power Spectral Density (the measure of a signal's power compared to frequency) across all frequencies.
- **Gaussian** – Due to noise source's random nature, a mathematical model is used to calculate the probability of events. Noise pdf is a Gaussian distribution

Model of AWGN channel



----- (1)

- Consider a bank of N pdt integrators : correlator



Observation vector \mathbf{x}

- Consider j^{th} correlator from the set of N correlators. o/p of j^{th} correlator

$$= \int s_j(t) dt$$

$$= \int s_j(t) dt$$

$$= \int s_j(t) dt + \int w_j(t) dt$$

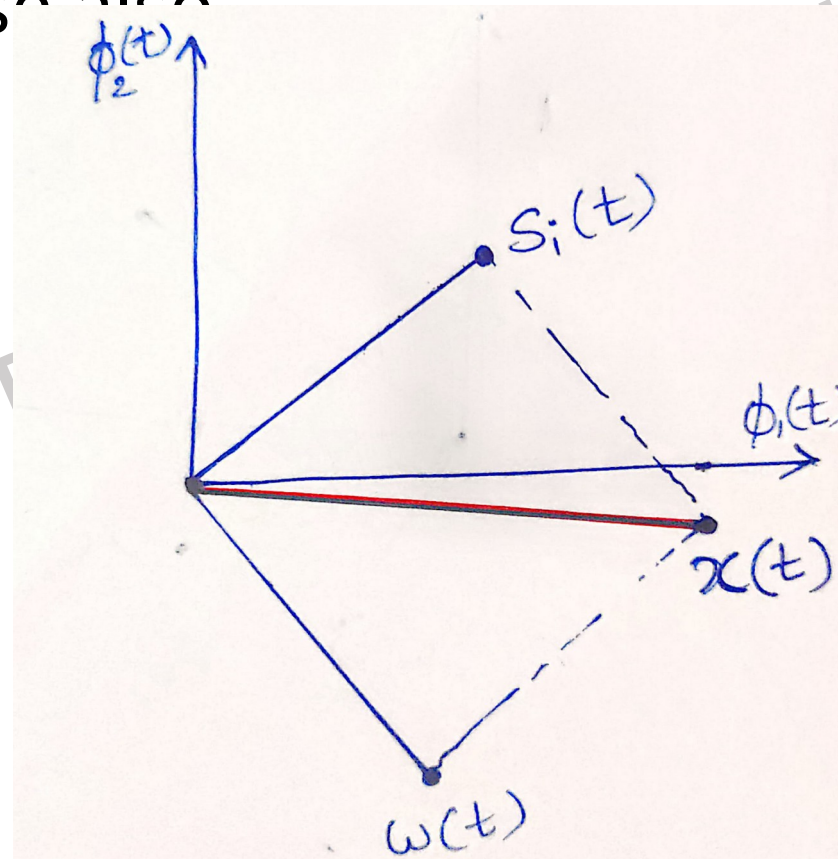
$$= s_{ij} + w_j \quad \text{----(2)} \quad \text{where } w_j \text{ is the noise component in the } j^{\text{th}} \text{ correlator o/p}$$

-deterministic s/g

$w(t)$ - random s/g

- Let $X(t)$ be a random process
- So $x(t)$ can be considered as the sample function of $X(t)$
- Let $W(t)$ be a random process defining noise
- So $w(t)$ can be considered as the sample function of $W(t)$

- We can represent a signal vector $s_i(t)$ using orthonormal basis functions.
- But when we consider rx'd s/g $x(t)$ it becomes difficult to represent it with the same orthonormal basis functions as it includes noise also.



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- Using synthesizer, we can represent s/g $s_i(t)$ from its vector elements.

If we have , then $s_i(t) =$

But if we have , then $x(t)$

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- Now Let's define $X'(t)$ be a random process
- So $x'(t)$ can be considered as the sample function of $X'(t)$

$$x'(t) = x(t) - \text{-----}(3)$$

Substitute eqn (1) & (2) in (3)

$$\begin{aligned} x'(t) &= - \\ &= - - \\ &= - \\ &= w'(t) \end{aligned}$$

$$x'(t) = w'(t)$$

- From eqn (3) , $x(t) = x'(t) +$

$$x(t) = + w'(t)$$

This is the vector representation of channel.

$w'(t)$ – remainder term.

Statistical characterization of vector channel

1. Mean
2. Variance
3. Co-variance

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1. Mean

- Consider a random vector $X =$
- X_1, X_2, \dots, X_N random variables.
- x_1, x_2, \dots, x_N are sample values of random variables X_1, X_2, \dots, X_N .

Mean of $X_j =$

$$= E[X_j]$$

From eqn (2) , $= s_{ij} + w_j$

$$X = S_{ij} + W_j$$

$$= E[S_{ij} + W_j]$$

$$= E[S_{ij}] + E[W_j]$$

There is no randomness in **deterministic s/g**.

So $E[S_{ij}] = S_{ij}$

Noise is **AWGN** . That is zero mean .

$$E[W_j] = 0$$

$$\text{Mean} = S_{ij}$$

2. Variance

• We know that variance = $E[(x-\mu)^2]$

$$= E[(X_j - \mu)^2]$$

$$= E[(X_j - S_{ij})^2]$$

$$= E[W_j^2]$$

and

$$W_j = \int_0^t \sigma_j(u) du$$

$$= E\left[\int_0^t \sigma_j(u) du \int_0^t \sigma_j(u) du\right]$$

$$= \int_0^t \int_0^t \sigma_j(u) \sigma_j(u) du dt E[1]$$

$E[] = R_w(t, u)$ is the autocorrelation function of noise $w(t)$

- But for an AWG noise, autocorr fn =

$$= \int_{-\infty}^{\infty} \delta(u) dt du$$

$$= \delta(t) \delta(u)$$

We know that by time shifting property

$$dt = x(t_0)$$

$$=) du$$

$$= du$$

- du = energy of orthonormal basis fn =1

$$\text{Variance} =$$

3. Co-variance

$$\begin{aligned}\text{Cov}[X_j, X_k] &= E[(X_j - \bar{X}_j)(X_k - \bar{X}_k)] \\ &= E[(X_j - S_{ij})(X_k - S_{ik})] \\ &= E[(W_j)(W_k)]\end{aligned}$$

Where $W_j = \int_0^t \sigma_j(t) dt$

$$W_k = \int_0^t \sigma_k(u) du$$

$$\text{Cov}[X_j, X_k] = E\left[\int_0^t \sigma_j(t) dt \int_0^t \sigma_k(u) du\right]$$

- $\text{Cov} [X_j X_k] = \int_k(u) E[w(t)w(u)] dt du$

- $E[] = R_w(t,u) =$

- $\text{Cov} [X_j X_k] = \int_k(u) dt du$
 $= \int_j(t) dt du$

We know that by time shifting property

$$dt = x(t_0)$$

$$\text{Cov} [X_j X_k] = \int_j(u) du$$

$\int du = 0$ property of orthonormal basis fn.

Therefore

$$\text{Cov} [X_j X_k] = 0$$

ie X_j 's are uncorrelated.

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