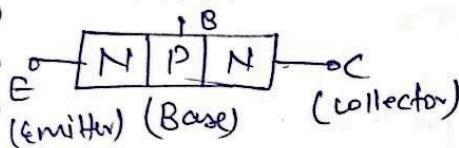


## Bipolar Junction Transistor (BJT)

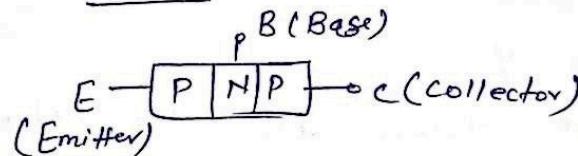
BJT is a three terminal Device which is mainly used as an Amplifier but also used as switch

### Types

→ NPN



→ PNP



### Construction of BJT :-

Emitter (E) :- Highly Doped

Size → Moderate

Collector (C) :- Doping → Moderate

Size → Large

Base (B) = Doping → lightly doped

Size → thin

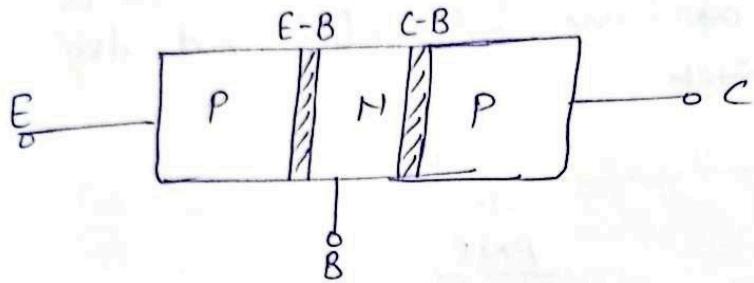
\* Doping

$$E > C > B$$

\* size

$$C > E > B$$

## Working of BJT

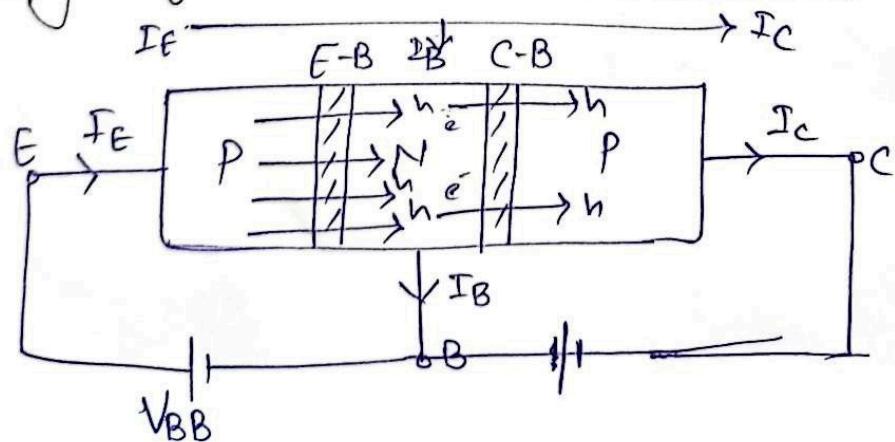


### Modes of operation

since there are 2 Junctions in BJT. Therefore we have to connect 2 Batteries one at each terminal and according to the connection of Batteries across the Junction the BJT will operate in total 4 modes, as follows:-

Mode	E-B Junction	C-B Junction	Application
Active	Forward	Reverse	Amplifier
Saturation	Forward	Forward	]
Cut off	Reverse	Reverse	ON switch OFF switch
Inverse-Active	Reverse	Forward	PreAmplification

## Working of BJT as an Amplifier



① E-B Junction should be Forward Biased.

E-B  $\rightarrow$  depletion layer gets reduced

- The holes will move from emitter to Base and the no. of holes moving from emitter will constitute the emitter current ( $I_E$ )
- As the holes enter the base they will recombine with the e<sup>-</sup>'s but since base is lightly doped, therefore very few holes will recombine in the base and the no. of holes recombining in the base will constitute the base current ( $I_B$ ).
- The remaining holes in the base will act as minority charge carriers and since these minority charge carriers have to move to the collector, therefore the Base Junction should always be reverse biased and hence the no. of holes will move from base to collector and constitute the collector current ( $I_C$ ).

$$I_E = I_B + I_C$$

$$I_C = I_E - I_B$$

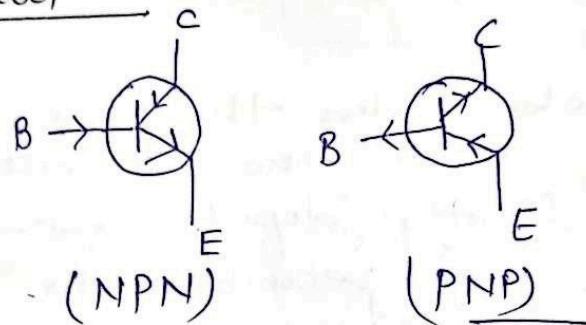
↓                  ↓  
Majority      Minority

That's why Bipolar called.

(3)

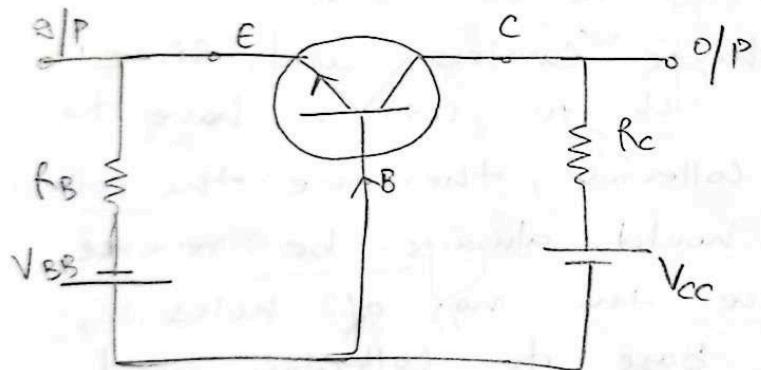
For Amplification Action since the signal is transferring from low resistance (E.B.) to High Resistance (C.B.) therefore it is named as transistor. (Transfer + Resistor)

### Transistor symbol

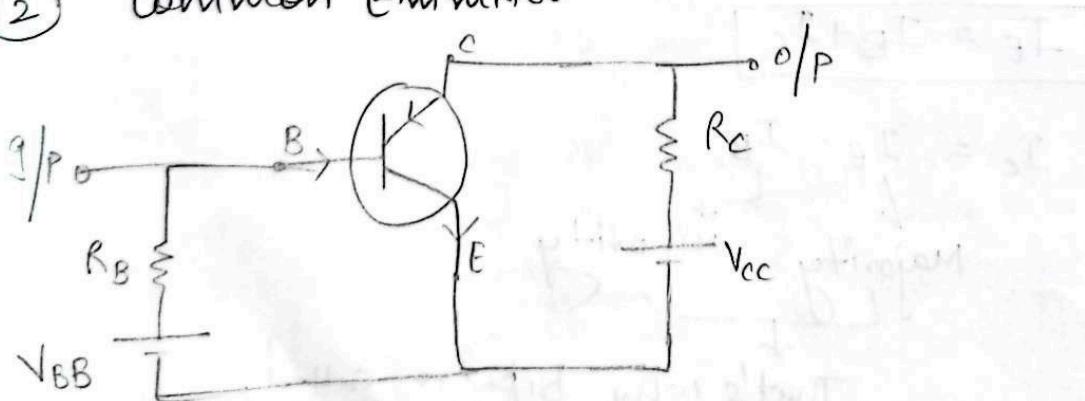


### Transistor Configuration :-

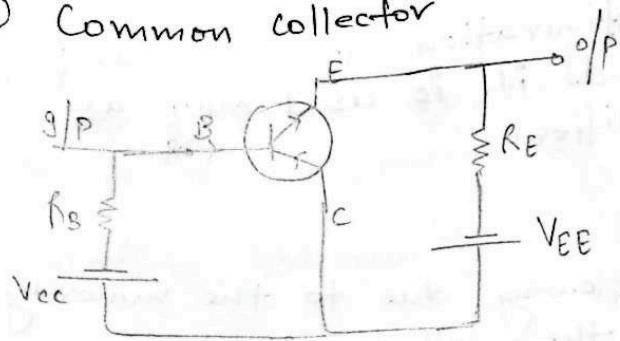
#### ① common Base



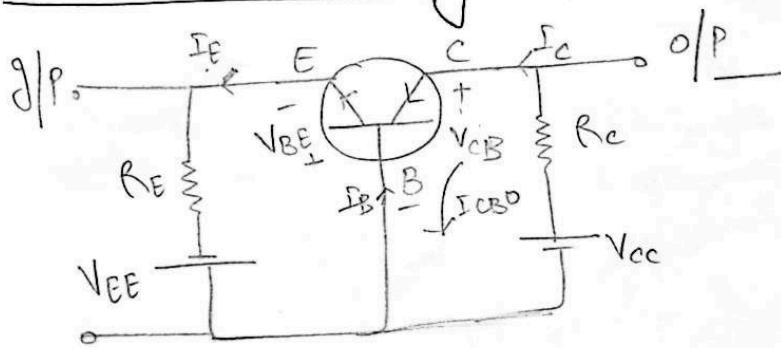
#### ② common Emitter



③ Common collector



Common base configuration



$I_{CBO}$   
↳ collector to  
Base with  
emitter terminal  
open.

I/P current :-  $I_E$

I/P Voltage =  $V_{BE}$        $\{ V_{EB} = -V_{BE} \}$

O/P current =  $I_C$

O/P Voltage =  $V_{CB}$

current gain :-  $\frac{\text{O/P current}}{\text{I/P current}}$

$$\frac{I_C}{I_E} = \alpha$$

Amplification factor  
or current gain  
for common base config.

$$\alpha \leq 1$$

$$[I_C = \alpha I_E + I_{CBO}]$$

Injected current

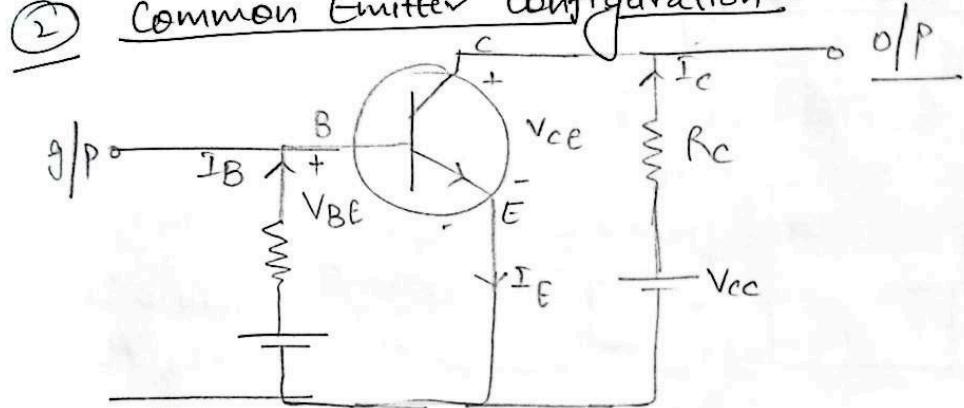
Reverse saturation current / leakage current

Since ( $\alpha$ ) is less than or equal to (1) therefore common base configuration does not amplify the current and it is used only as a voltage Amplifier.

### \* $I_{CBO}$

It is the current flowing due to the minority charge carriers of the collector.

### ② Common Emitter configuration



$$\text{Input current} = I_B$$

$$\text{Input Voltage} = V_{BE}$$

$$\text{Output Voltage} = V_{CE}$$

$$\text{Output current} = I_C$$

$$\text{current gain} = \frac{\text{O/P current}}{\text{I/P current}} = \frac{I_C}{I_B} = \underline{\underline{\beta}}$$

$\beta$  = It is the amplification factor or current gain for the common emitter configuration.

$$\boxed{\beta \gg 1}$$

before  
slify

→ since  $I_c$  is very much greater than  $I_B$   
therefore  $(\beta)$  will be very high and hence  
common emitter circuit is used as  
a power Amplifier

### Relation between $\alpha$ and $\beta$

$$I_E = I_B + I_C$$

Divide whole eq<sup>n</sup> with  $I_E$

$$\frac{I_E}{I_C} = \frac{I_B}{I_C} + 1$$

$$\frac{1}{\alpha} = \frac{1}{\beta} + 1$$

$$\boxed{\alpha = \frac{\beta}{\beta + 1}}$$

$$\text{or } \boxed{\beta = \frac{\alpha}{1 - \alpha}}$$

### Equation for $I_C$ :-

$$I_C = \alpha I_E + I_{CBO}$$

$$I_C = \alpha (I_B + I_C) + I_{CBO}$$

$$(1 - \alpha) I_C = \alpha I_B + I_{CBO}$$

$$I_C = \frac{\alpha}{1 - \alpha} I_B + \left(\frac{1}{1 - \alpha}\right) I_{CBO}$$

$$I_C = \beta I_B + \left(\frac{1}{1 - \alpha}\right) I_{CBO}$$

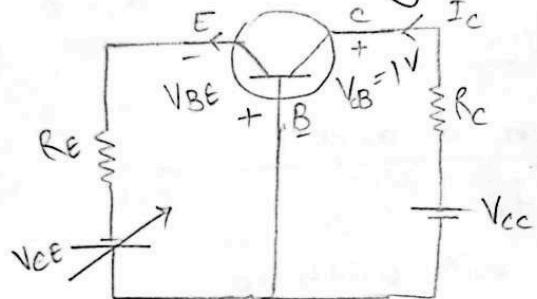
$$\boxed{I_C = \beta I_B + (\beta + 1) I_{CBO}}$$

↓  
leakage current more → disadvantage

$$\boxed{I_C = \beta I_B + I_{CEO}}$$

where  $I_{CEO} = (\beta + 1) I_{CBO}$

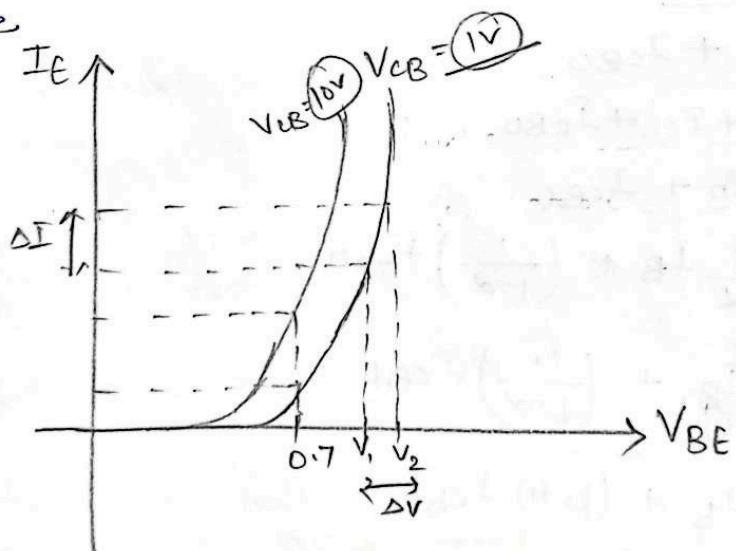
Input and output voltage characteristics  
(imp) of Common Base configuration.



### Input characteristics

Input current  $V/s$  Input Voltage Keeping output Voltage constant  $I_E \text{ } V/s \text{ } V_{BE}$  keeping  $V_{CB}$  constant

\* since Emitter Base Junction is forward Biased therefore the Input characteristics of the circuit will be similar to the forward Biased diode

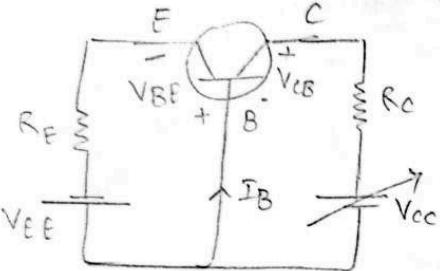


→ On Increasing the output Voltage ( $V_{CB}$ ) the input current ( $I_E$ ) will increase

$$R_{in} = \frac{\Delta V}{\Delta I}$$

$R_{in}$  is very low ( $\Delta V \ll \Delta I$ )

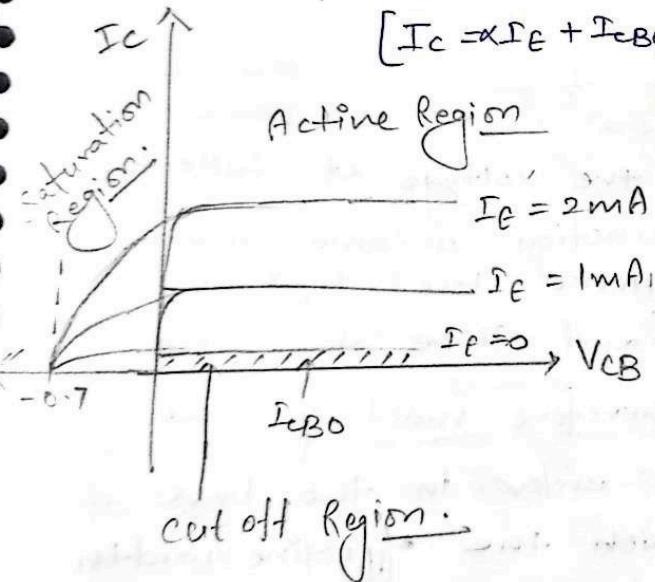
### Output characteristics



o/p current v/s o/p Voltage  
keeping g/p current as constant  
I\_C v/s V\_{CB} keeping I\_E const.

→ Since the collector Base junction is Reverse Biased therefore, the output characteristics will be similar to the Reverse Bias diode and hence the diode current will remain const with respect to the output voltage

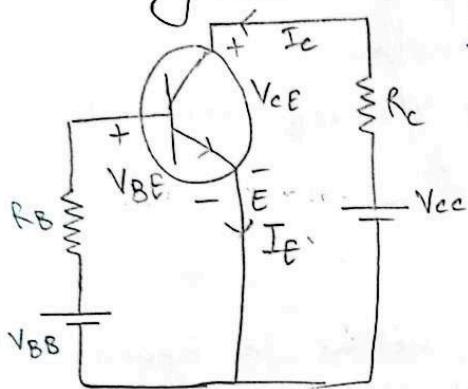
$$[I_C = \alpha I_E + I_{CBO}]$$



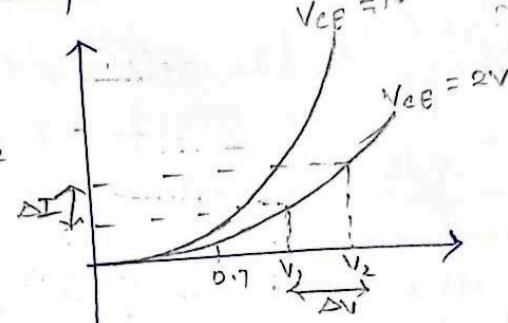
$$R_{out} = \frac{\Delta V_{CB}}{\Delta I_C}$$

$R_{out} \rightarrow$  very high

Input and output characteristics of common emitter configuration.



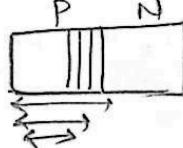
Input characteristic



$$\frac{out}{I_C \cdot V_I}$$

B  
+  
V<sub>BE</sub>

$I_B$  vs  $V_{BE}$  keeping  $V_{CE}$  constant  
Early effect or Base width modulation



on increasing the voltage at collector the collector base junction become more reverse bias and hence the depletion layer gets increased and since the base is lightly doped therefore most of the depletion layer will enter in the base region due to which the effective width of the base get decreased and the base current also reduced. on increasing the output voltage, the base current gets decreased.

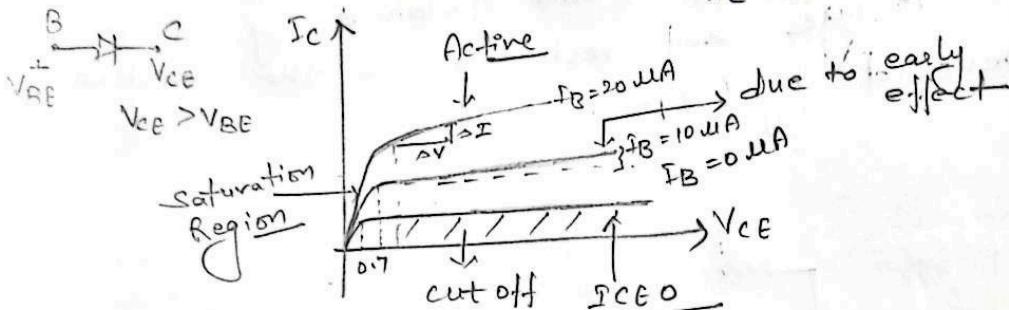
why

- i)  A
- ii)  S
- iii)  T
- iv)  L

emitter

Output characteristics  
 $I_C$  V/s  $V_{CE}$  keeping  $I_B$  constant

$$I_C = \beta I_B + I_{CEO}$$
$$I_C = \beta I_B + (\beta + 1) I_{CBQ}$$



$$R_{in} = \frac{\Delta V_{BE}}{\Delta I_B}$$

but

th

or

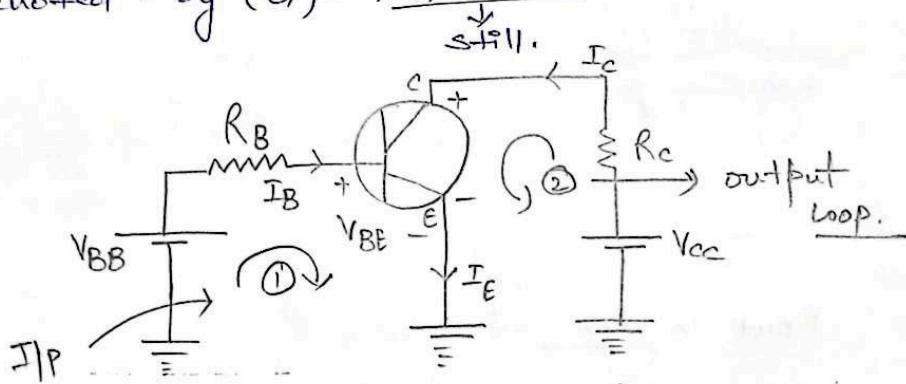
$R_{out}$  is high

why do we prefer common emitter over common base.

- i) common Emitter Amplifies the voltage as well as the current while common base Amplifies the voltage.
- ii) Since the difference b/w input and output resistance of common emitter is comparatively lower in common emitter circuit as compare to the common base therefore signal transfer will be higher in common emitter over the common base.

## Operating point

Operating point is defined as the output values of the circuit obtained by applying DC voltages and resistance to the transistor. Denoted by  $(Q)$  → Quiescent  
still.



① Apply KVL in Input loop

$$V_{BB} - I_B R_B - V_{BE} = 0$$

Generally  $V_{BEQ} = 0.7V$

$$I_B = I_{BQ} = \frac{V_{BB} - 0.7}{R_B} \quad \text{--- ①}$$

$$\text{② } [I_c = I_{cQ} = \beta I_{BQ}] \quad \text{--- ②}$$

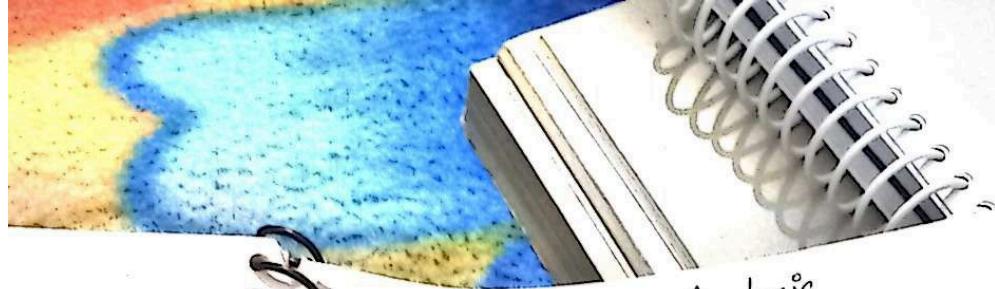
③ Apply KVL in O/P loop

$$V_{cc} - I_c R_C - V_{CE} = 0$$

Put value of  $I_c$  from eq ②

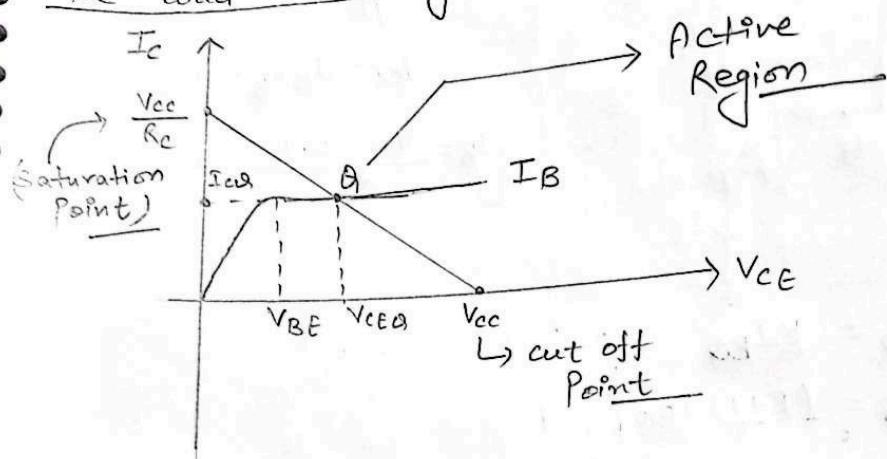
$$V_{CEQ} = V_{cc} - I_{cQ} R_C \rightarrow (\text{DC load line})$$

$$I_{EQ} = I_{BQ} + I_{cQ} \\ = (\beta + 1) I_{BQ}$$



Input  
Applying  
transistor.

### DC Load Line Analysis

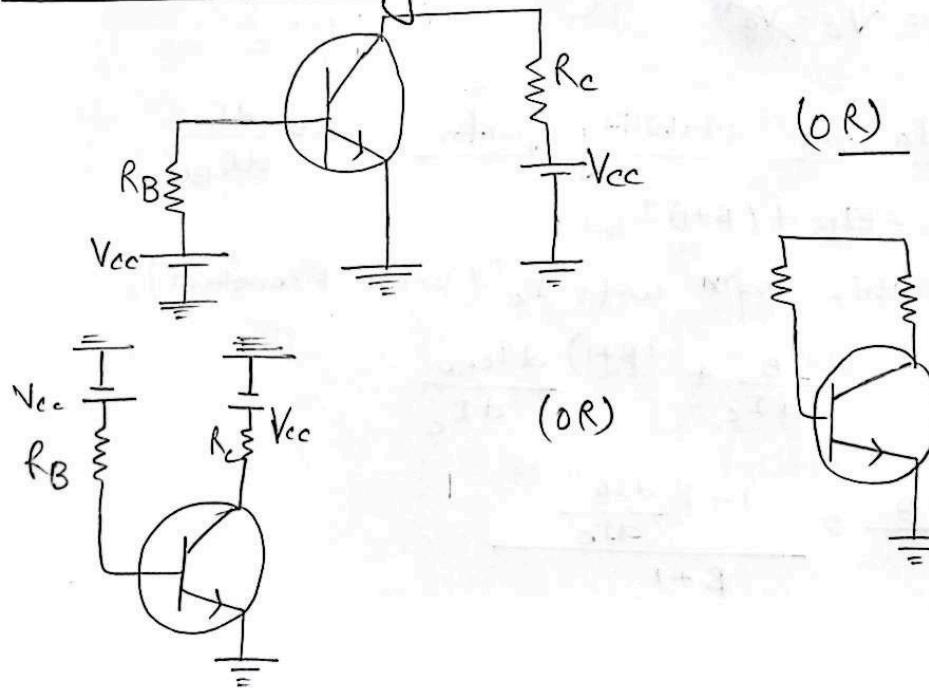


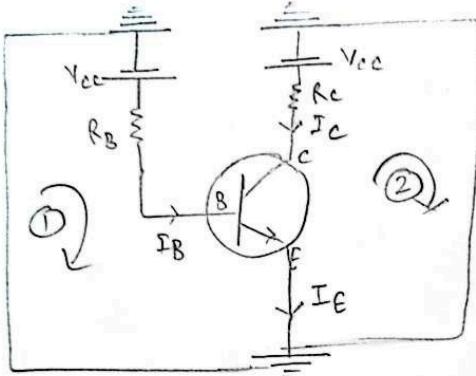
$$y = mx + c$$

$$I_c = -\frac{V_{ce}}{R_C} + \frac{V_{cc}}{R_C} \quad (\text{From DC load line eqn})$$

$$m = -\frac{1}{R_C} \rightarrow \underline{\text{load}}$$

### Transistor Biasing





① I/P Loop (KVL)

$$V_{cc} - I_B R_B - V_{BE} = 0$$

$$\left[ I_B = \frac{V_{cc} - 0.7}{R_B} = I_{BQ} \right]$$

②  $\left[ \begin{array}{l} I_{CO} = \beta I_{BQ} \\ I_{EQ} = (\beta + 1) I_{BQ} \end{array} \right]$

③ O/P Loop (KVL)

$$V_{cc} - I_C R_C - V_{CE} = 0$$

$$\left[ V_{CEQ} = V_{cc} - I_{CQ} R_C \right]$$

$$\left\{ \begin{array}{l} V_{BE} = V_B - V_E \\ V_{CE} = V_C - V_E \end{array} \right. \quad \left\{ \begin{array}{l} V_{BC} = V_B - V_C \\ = V_{BE} - V_{CE} \end{array} \right.$$

Formula for stability factor  $S = \frac{dI_C}{dI_{CBO}}$

$$I_C = \beta I_B + (\beta + 1) I_{BQ}$$

Dif. this eqn wrt.  $I_C$  (keep  $\beta$  constant)

$$1/S = \frac{\beta dI_B}{dI_C} + (\beta + 1) \frac{dI_{CBO}}{dI_C}$$

$$\frac{dI_{CBO}}{dI_C} = \frac{1 - \beta \frac{dI_B}{dI_C}}{\beta + 1}$$

$$\begin{array}{l} \text{KVL} \\ \frac{V_{BE}}{0.7} = 0 \end{array}$$

$\beta_0$

$$S = \frac{dI_C}{dI_{CBO}} = \frac{\beta + 1}{1 - \beta \frac{dI_B}{dI_C}}$$

For Fixed Bias CKT :-

$$V_{CC} - I_B R_B - V_{BE} = 0$$

Differentiate w.r.t.  $I_C$

$$0 - \frac{dI_B}{dI_C} \times R_B - 0 = 0$$

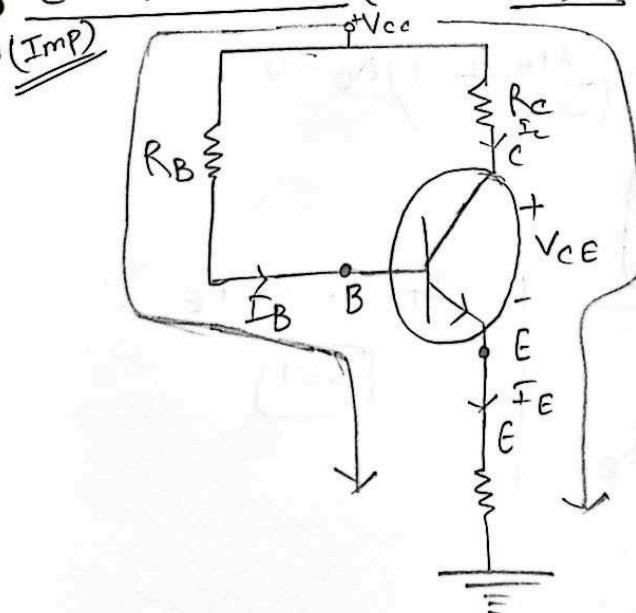
$$\frac{dI_B}{dI_C} = 0$$

So,

$$S = \beta + 1$$

Since  $\beta$  is very high therefore the stability factor of fixed bias CKT is also very high and hence this CKT is not a stable CKT.

Emitter Bias (Self Bias) :-



① F/P Loop (KVL)

$$V_{CC} - I_B R_B - V_{BE} - I_E R_E = 0$$

$$\text{Put } I_E = (\beta + 1) I_B$$

$$V_{CC} - I_B R_B - 0.7 - (\beta + 1) I_B R_E = 0$$

$$I_{BQ} = \frac{V_{CC} - 0.7}{R_B + (\beta + 1) R_E}$$

$$\textcircled{2} \quad \left. \begin{aligned} I_{CQ} &= \beta I_{BQ} \\ I_{EQ} &= (\beta + 1) I_{BQ} \end{aligned} \right] \quad | \quad \boxed{I_C \approx I_E}$$

\textcircled{3} O/P loop (KVL)

$$V_{CC} - I_e R_C - V_{CE} - I_E R_E = 0$$

$$\left[ V_{CEQ} = V_{CC} - I_{EQ} R_C - I_{EQ} R_E \right]$$

$$\left. \begin{aligned} V_E - I_E R_E &= 0 \\ V_E &= I_E R_E \end{aligned} \right| \quad \begin{aligned} V_B &= V_{BE} + V_E \\ V_C &= V_{CE} + V_E \end{aligned}$$

stability factor

$$V_{CC} - I_B R_B - V_{BE} - I_E R_E = 0$$

$$V_{CC} - I_B R_B - V_{BE} - (I_B + I_C) R_E = 0$$

Dif. w.r.t.  $I_C$

$$0 - \frac{dI_B}{dI_C} R_B - 0 - \left( \frac{dI_B}{dI_C} + 1 \right) R_E = 0$$

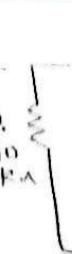
$$\left[ \frac{dI_B}{dI_C} = \frac{-R_E}{R_B + R_E} \right]$$

$$S = \frac{\beta + 1}{1 + \frac{\beta R_E}{R_B + R_E}}$$

$$S < \beta + 1$$

If  $R_B \ll R_E$

$$\boxed{S \approx 1}$$



De.  
TPu