

Swing-up and Balance Controller for an Inverted Pendulum on a Cart

(Energy Shaping + LQR hybrid with automatic fallback)

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Abstract

This document summarizes the dynamical model, controller design, and the hybrid control architecture used in the provided implementation. It contains the mathematical derivation of the energy-shaping swing-up law, the linearization used for LQR balance design, the finite-state machine (FSM) for mode switching, implementation notes mapping variables to code, pseudocode for the algorithms, and practical tuning and safety guidelines.

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1 System description and notation

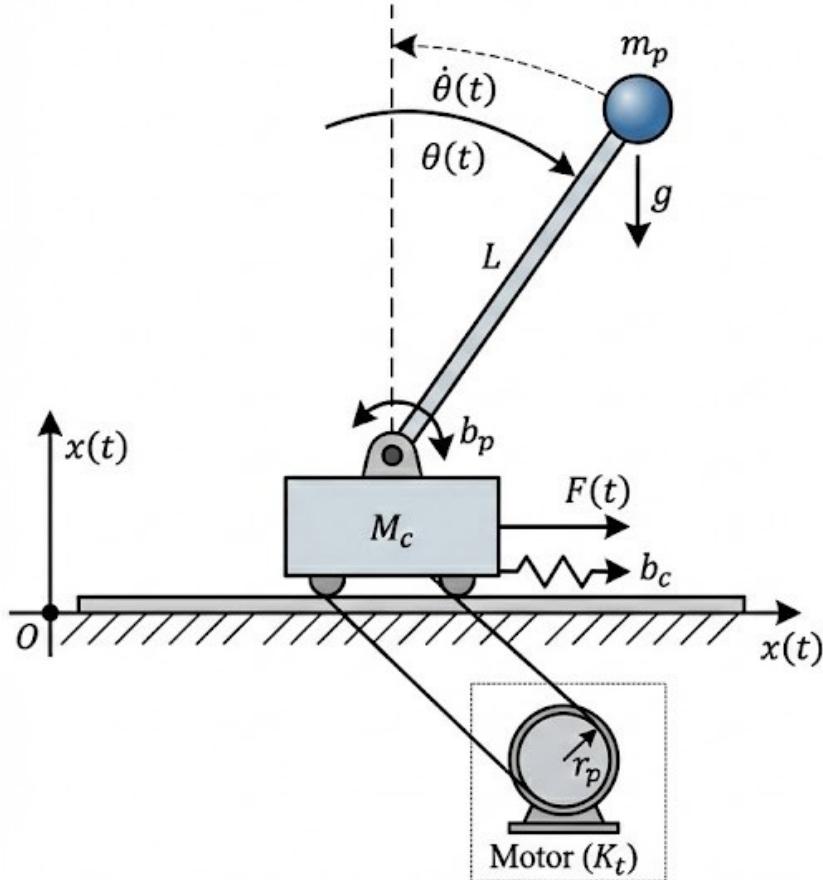


Figure 1: Schematic of the pendulum-on-cart (pulley) system and variable mapping.

The following variables are used:

- $x(t)$: cart horizontal position (m).
- $\dot{x}(t)$: cart velocity (m/s).
- $\theta(t)$: pendulum angle in radians, with $\theta = 0$ when upright. (Note: in code angle is often in degrees; mapping is described below.)
- $\dot{\theta}(t)$: pendulum angular velocity (rad/s).
- L : distance from pivot to pendulum centre-of-mass (m).
- m (or m_p): pendulum mass (kg).
- g : gravitational acceleration (m/s^2).
- $F(t)$: horizontal force applied to the cart (N). The actuator provides torque τ ; through the pulley radius r_p this maps to $F = \tau/r_p$.
- Motor torque constant: $\tau = K_t I$. The code uses $I = (Fr_p)/K_t$.

2 Dynamical model (simplified)

The standard underactuated pendulum-on-cart model is used. Taking the pendulum as a point mass m at distance L from pivot (small-angle linearization later), the nonlinear equations (including cart mass M_c for generality) are:

$$(M_c + m)\ddot{x} + mL\ddot{\theta} \cos \theta - mL\dot{\theta}^2 \sin \theta = F(t) - b_c \dot{x}, \quad (1)$$

$$mL^2\ddot{\theta} + mL\ddot{x} \cos \theta + mgL \sin \theta = -b_p \dot{\theta}, \quad (2)$$

where b_c and b_p represent viscous friction terms (cart friction and pendulum damping). For the energy-shaping design the pendulum's energy is the primary focus.

2.1 Pendulum energy

Define potential and kinetic energy:

$$PE(\theta) = mgL \cos \theta, \quad (3)$$

$$KE(\dot{\theta}) = \frac{1}{2}mL^2\dot{\theta}^2. \quad (4)$$

Total pendulum energy:

$$E(\theta, \dot{\theta}) = mgL \cos \theta + \frac{1}{2}mL^2\dot{\theta}^2.$$

Upright target energy:

$$E_{\text{target}} = mgL.$$

Energy error:

$$E_{\text{err}} = E_{\text{target}} - E.$$

3 Energy shaping swing-up controller

3.1 Control law

A continuous energy-shaping input is used to modify pendulum energy toward E_{target} . The energy injection term is:

$$u_{\text{energy}} = K_{\text{energy}} E_{\text{err}} \dot{\theta} \cos \theta.$$

Damping and cart terms are combined with this to form the actuator command (in units of force before conversion to current):

$$u_{\text{pump}} = u_{\text{energy}} - K_{\text{damp}} \dot{\theta} - K_{\text{center}} x - K_{x,\text{dot}} \dot{x}.$$

A virtual wall force $F_{\text{wall}}(x, \dot{x})$ (spring-damper) is computed when the cart approaches limits. The final motor current command is:

$$I_{\text{cmd}} = \text{clip}\left(u_{\text{pump}} + \frac{F_{\text{wall}} r_p}{K_t}, -I_{\max}, I_{\max}\right),$$

where r_p is the pulley radius and K_t is the motor torque constant.

3.2 Rationale

- The factor $\dot{\theta}$ aligns the energy injection with pendulum motion sign — energy is provided in-phase with motion.
- The $\cos \theta$ factor enforces correct phase dependence around top and bottom.
- Multiplication by E_{err} scales injection so it diminishes near the upright energy.
- The damping term $-K_{\text{damp}}\dot{\theta}$ limits overshoot and reduces kinetic energy near capture.

4 Virtual wall (safety) model

To protect against hitting physical limits, a virtual wall at $x = \pm X_{\text{wall}}$ is used with spring-damper action:

$$F_{\text{wall}}(x, \dot{x}) = \begin{cases} -K_{\text{wall}}(x - X_{\text{wall}}) - D_{\text{wall}}\dot{x}, & x > X_{\text{wall}}, \\ -K_{\text{wall}}(x + X_{\text{wall}}) - D_{\text{wall}}\dot{x}, & x < -X_{\text{wall}}, \\ 0, & \text{otherwise.} \end{cases}$$

Converted to current:

$$I_{\text{wall}} = \frac{F_{\text{wall}} r_p}{K_t}.$$

Outward pumping is suppressed while the wall is active.

5 Linearization and LQR design

5.1 Linearization about the upright

Linearizing the full equations about $\theta = 0$ results in a linear time-invariant model in the state vector $z = [x, \dot{x}, \theta, \dot{\theta}]^\top$:

$$\dot{z} = Az + Bu,$$

with A, B dependent on the physical parameters M_c, m, L, b_c, b_p, g . For controller design the conceptual LTI model is used; numerical values require M_c and friction coefficients.

5.2 LQR control law

For the linearized system $\dot{z} = Az + Bu$, select weighting matrices $Q \succeq 0$ and $R \succ 0$ and solve the continuous algebraic Riccati equation for P . The optimal full-state feedback gain is:

$$K_{\text{LQR}} = R^{-1}B^\top P,$$

and the control law (force) is:

$$F(t) = -K_{\text{LQR}} z(t).$$

Mapping the computed force to motor current:

$$I(t) = \frac{F(t) r_p}{K_t}.$$

In the implementation the hard-coded LQR gain vector is:

$$K = [K_1 \ K_2 \ K_3 \ K_4] = [-10.0 \ -11.7749 \ 50.7317 \ 6.3],$$

so the discrete implementation uses:

$$I(t) = \text{clip}\left(\frac{-(Kz(t)) r_p}{K_t}, -I_{\max}, I_{\max}\right).$$

Interpretation of the gain vector:

- K_3 acts as angle stiffness around the upright,
- K_4 damps angular rate and is critical for capturing high-speed arrival at the top,
- K_1, K_2 regulate cart position and velocity.

6 Hybrid control logic (FSM)

A finite-state machine with two states is used:

$$\text{SWING_UP} \xrightarrow{\text{if } |\theta| < \theta_{\text{enter}}, |\dot{\theta}| < \omega_{\text{enter}}} \text{LQR}$$

$$\text{LQR} \xrightarrow{\text{if } |\theta| > \theta_{\text{fall}} \text{ for } N \text{ frames}} \text{SWING_UP}$$

Values used in the implementation:

- Enter LQR: $\theta_{\text{enter}} = 15^\circ$, $\omega_{\text{enter}} = 1.0$ (rad/s).
- Fall back to swing-up: $\theta_{\text{fall}} = 35\text{--}45^\circ$, debounced for $N = 5$ frames at 200 Hz.

7 Pseudocode

Algorithm 1 MAIN_LOOP (FSM)

```

1: Initialize hardware: drv, enc
2: while true do
3:   start_rot  $\leftarrow$  RUN_SWINGUP()
4:   if start_rot == None then
5:     break                                      $\triangleright$  sensor failure or exit
6:   end if
7:   RUN_LQR(start_rot)
8: end while
9: drv.setDeviceToIdle()

```

Algorithm 2 RUN_SWINGUP

```
1: Read start_rot from motor (timeout on failure)
2: Wait small delay (e.g. 1.5 s)
3: Initialize filters and previous states
4: loop
5:   Read sensors:  $(s_1, \text{raw\_angle}) \leftarrow \text{enc.getAbsoluteAngle}()$ 
6:    $(s_2, \text{raw\_rot}) \leftarrow \text{drv.getMotorPosition}()$ 
7:   if  $s_1 \neq 0$  or  $s_2 \neq 0$  then
8:      $\text{drv.setTorqueControl}(0.0, 0)$                                  $\triangleright$  fail-safe zero current
9:     continue
10:    end if
11:     $\theta_{\text{deg}} \leftarrow \text{MAP\_ANGLE}(\text{raw\_angle})$ 
12:     $\theta \leftarrow \text{rad}(\theta_{\text{deg}})$ 
13:     $x \leftarrow (\text{raw\_rot} - \text{start\_rot}) \cdot 2\pi r_p$ 
14:    if prev is None then
15:      store prev values and continue
16:    end if
17:    Compute  $\dot{\theta}$  (with unwrap) and filter; compute  $\dot{x}$ 
18:     $PE \leftarrow mgL \cos \theta, KE \leftarrow \frac{1}{2}mL^2\dot{\theta}^2$ 
19:     $E \leftarrow PE + KE, E_{\text{err}} \leftarrow E_{\text{target}} - E$ 
20:     $u_{\text{energy}} \leftarrow K_{\text{ENERGY}} E_{\text{err}} \dot{\theta} \cos \theta$ 
21:     $u_{\text{damp}} \leftarrow -K_{\text{DAMP}} \dot{\theta}$ 
22:     $u_{\text{center}} \leftarrow -K_{\text{CENTER}} x - K_{\text{XDOT}} \dot{x}$ 
23:     $u_{\text{pump}} \leftarrow u_{\text{energy}} + u_{\text{damp}} + u_{\text{center}}$ 
24:    Compute wall force  $F_{\text{wall}}(x, \dot{x})$  and suppress outward pump if active
25:     $I \leftarrow \text{clip}(u_{\text{pump}} + F_{\text{wall}}r_p/K_t, -I_{\text{max}}, I_{\text{max}})$ 
26:     $\text{drv.setTorqueControl}(I, 3000)$ 
27:    if  $|\theta_{\text{deg}}| < \theta_{\text{enter}}$  and  $|\dot{\theta}| < \omega_{\text{enter}}$  then
28:       $\text{drv.setTorqueControl}(0.0, 0); \text{return start\_rot}$ 
29:    end if
30: end loop
```

Algorithm 3 RUN_LQR(*start_rot*)

```
1: Initialize state  $z = [0, 0, 0, 0]$ , velocity filters, fall_count  $\leftarrow 0$ 
2: loop
3:   Read sensors:  $(s_1, \text{raw\_angle}), (s_2, \text{raw\_rot})$ 
4:   if  $s_1 \neq 0$  or  $s_2 \neq 0$  then
5:     continue
6:   end if
7:    $\theta_{\text{deg}} \leftarrow \text{raw\_angle} - \text{UPRIGHT\_OFFSET}$  (wrap to  $[-180, 180]$ )
8:    $\theta \leftarrow \text{rad}(\theta_{\text{deg}})$ 
9:    $x \leftarrow (\text{raw\_rot} - \text{start\_rot}) \cdot 2\pi r_p$ 
10:  Compute raw derivatives and EMA-filtered  $\dot{x}, \dot{\theta}$ 
11:   $z \leftarrow [x, \dot{x}_{\text{filt}}, \theta, \dot{\theta}_{\text{filt}}]$ 
12:  if  $|\theta_{\text{deg}}| > \text{FALL\_ANGLE}$  then
13:    fall_count  $+ = 1$ 
14:  else
15:    fall_count  $\leftarrow 0$ 
16:  end if
17:  if fall_count  $\geq \text{FALL\_COUNT\_MAX}$  then
18:    drv.setTorqueControl(0.0,0); return                                 $\triangleright$  switch to swing-up
19:  end if
20:   $F \leftarrow -K \cdot z$                                                $\triangleright$  LQR force
21:   $I \leftarrow \text{clip}(F r_p / K_t, -I_{\text{max}}, I_{\text{max}})$ 
22:  drv.setTorqueControl(I,3000)
23: end loop
```

Algorithm 4 WATCHDOG_THREAD

```
1: Shared variables: desired_current, last_update_time
2: loop(run in separate thread at high rate)
3:   if time.now() - last_update_time  $> \text{timeout\_s}$  then
4:     drv.setTorqueControl(0.0,0)                                          $\triangleright$  fail-safe
5:   else
6:     drv.setTorqueControl(desired_current,3000)
7:   end if
8:   sleep(small_interval)
9: end loop
```

8 Parameter table (mapping to python variables)

9 Implementation notes and safety

- **Actuator mapping:** The code computes current commands. To convert a desired linear force F to current, use $I = (Fr_p)/K_t$.
- **Watchdog:** Because some motor drivers hold the last command if software stalls, a hardware or software watchdog that sets current to zero if the control loop stalls is strongly recommended.
- **Hysteresis:** Use asymmetric thresholds (enter at small angle, exit at larger fall angle) to avoid chattering.
- **Tuning suggestions:**

Symbol	Meaning	Python variable / unit
M	pendulum mass	M (kg)
L	pendulum length to CoM	L (m)
g	gravity	G (m/s ²)
r_p	pulley radius	PULLEY_RADIUS (m)
K_t	motor torque constant	KT (Nm/A)
I_{\max}	current limit	MAX_CURRENT (A)
K_{energy}	energy gain	K_ENERGY
K_{damp}	pendulum damping gain	K_DAMP
K_{center}	cart centering	K_CENTER
K_{x_dot}	cart damping	K_XDOT
$K_{\text{wall}}, D_{\text{wall}}$	virtual wall gains	K_WALL, D_WALL

Table 1: Key parameters and variable mapping.

- If LQR cannot catch because $\dot{\theta}$ is too large: increase angular speed feedback gain $K[3]$.
- If pendulum overshoots the upright violently: increase K_{damp} .
- If cart drifts: increase K_{x_dot} or introduce a small K_{center} during swing-up.