```
load commqpsktxrx_sbits_100.mat; % length 174
% General simulation parameters
SimParams.M = 4; % M-PSK alphabet size
SimParams.Upsampling = 4; % Upsampling factor
SimParams.Downsampling = 2; % Downsampling factor
SimParams.Fs = 2e5; % Sample rate in Hertz
SimParams.Ts = 1/SimParams.Fs; % Sample time in sec
SimParams.FrameSize = 174; % Number of modulated symbols per frame
% LDPC matrix size, rate must be 1/2
% Warning: encoding - decoding can be very long for large LDPC
matrix!
SimParams.LdpcRow = 148;
SimParams.LdpcColumn = 296;
SimParams.Ldpcmethod = 1;% Method for creating LDPC matrix (0 =
Evencol; 1 = Evenboth)
SimParams.LdpcnoCycle = 1;% Eliminate length-4 cycle
SimParams.LdpcOnePerCol = 3;% Number of 1s per column for LDPC matrix
SimParams.LdpcStrategy = 1;% LDPC matrix reorder strategy (0 = First;
1 = Mincol; 2 = Minprod)
SimParams.LdpcIteration = 1;% Number of iteration;
 SimParams.LdpcH = makeLdpc(SimParams.LdpcRow, ...
                      SimParams.LdpcColumn, ...
                      SimParams.Ldpcmethod, ...
                      SimParams.LdpcnoCycle, ...
                      SimParams.LdpcOnePerCol);
 [SimParams.LdpcNewH,SimParams.LdpcU,SimParams.LdpcL] =
 makeParityChk(SimParams.LdpcH, SimParams.LdpcStrategy);
 % Tx parameters
 SimParams.BarkerLength = 26; % Number of Barker code symbols
 SimParams.DataLength = (SimParams.FrameSize -
 SimParams.BarkerLength)*2; % Number of data payload bits per frame
 SimParams.ScramblerBase = 2;
 SimParams.ScramblerPolynomial = [1 1 1 0 1];
 SimParams.ScramblerInitialConditions = [0 0 0 0];
  SimParams.sBit = sBit; % Payload bits
 SimParams.RxBufferedFrames = 10; % Received buffer length (in frames)
```

\_\_\_\_\_\_QPSKTXRXSim.m-----

```
SimParams.RaisedCosineFilterSpan = 10; % Filter span of Raised Cosine
Tx Rx filters (in symbols)
SimParams.MessageLength = 112;
SimParams.FrameCount = 100; % Number of frames transmitted
% Channel parameters
SimParams.PhaseOffset = 0; % in degrees
SimParams.EbNo = 20; % in dB
SimParams.FrequencyOffset = 0; % Frequency offset introduced by
channel impairments in Hertz
SimParams.DelayType = 'Triangle'; % select the type of delay for
channel distortion
% Rx parameters
SimParams.CoarseCompFrequencyResolution = 25; % Frequency resolution
for coarse frequency compensation
% Look into model for details for details of PLL parameter choice.
Refer equation 7.30 of "Digital Communications - A Discrete-Time
Approach" by Michael Rice.
K = 1;
A = 1/sqrt(2);
SimParams.PhaseRecoveryLoopBandwidth = 0.01; % Normalized loop
bandwidth for fine frequency compensation
{\tt SimParams.PhaseRecoveryDampingFactor} = 1; \ {\tt \% \ Damping \ Factor} \ {\tt for \ fine}
frequency compensation
SimParams.TimingRecoveryLoopBandwidth = 0.01; % Normalized loop
bandwidth for timing recovery
SimParams.TimingRecoveryDampingFactor = 1; % Damping Factor for
timing recovery
SimParams.TimingErrorDetectorGain = 2.7*2*K*A^2+2.7*2*K*A^2; % K p
for Timing Recovery PLL, determined by 2KA^2*2.7 (for binary PAM),
QPSK could be treated as two individual binary PAM, 2.7 is for raised
cosine filter with roll-off factor 0.5
%QPSK modulated Barker code header
BarkerCode = [+1; +1; +1; +1; +1; -1; -1; +1; +1; -1; +1; -1; +1; +1;
+1; +1; +1; +1; -1; -1; +1; -1; +1; -1; +1]; % Bipolar Barker
SimParams.ModulatedHeader = sqrt(2)/2 * (-1-1i) * BarkerCode;
% Generate square root raised cosine filter coefficients (required
only for MATLAB example)
SimParams.Rolloff = 0.5;
```

```
% Square root raised cosine transmit filter
SimParams.TransmitterFilterCoefficients = ...
 rcosdesign(SimParams.Rolloff, SimParams.RaisedCosineFilterSpan, ...
 SimParams.Upsampling);
% Square root raised cosine receive filter
SimParams.ReceiverFilterCoefficients = ...
 rcosdesign(SimParams.Rolloff, SimParams.RaisedCosineFilterSpan, ...
 SimParams.Upsampling);
prmQPSKTxRx = SimParams; % QPSK system parameters
printData = true; %true if the received data is to be printed
useScopes = true; % true if scopes are to be used
% Initialize the components
    % Create and configure the transmitter System object
    hTx = QPSKTransmitterR(...
       'UpsamplingFactor', prmQPSKTxRx.Upsampling, ...
       'MessageLength', prmQPSKTxRx.MessageLength, ...
 'TransmitterFilterCoefficients',prmQPSKTxRx.TransmitterFilterCoeffici
       'DataLength', prmQPSKTxRx.DataLength, ...
       'ScramblerBase', prmQPSKTxRx.ScramblerBase, ...
       'ScramblerPolynomial', prmQPSKTxRx.ScramblerPolynomial, ...
       'ScramblerInitialConditions',
 prmQPSKTxRx.ScramblerInitialConditions,...
        'LdpcNewH',prmQPSKTxRx.LdpcNewH,...
        'LdpcU',prmQPSKTxRx.LdpcU,...
        'LdpcL',prmQPSKTxRx.LdpcL);
    % Create and configure the AWGN channel System object
    hChan = QPSKChannelR('DelayType', prmQPSKTxRx.DelayType, ...
        'RaisedCosineFilterSpan',
 prmQPSKTxRx.RaisedCosineFilterSpan, ...
        'PhaseOffset', prmQPSKTxRx.PhaseOffset, ...
        'SignalPower', 1/prmQPSKTxRx.Upsampling, ...
        'FrameSize', prmQPSKTxRx.FrameSize, ...
```

```
'UpsamplingFactor', prmQPSKTxRx.Upsampling, ...
       'EbNo', prmQPSKTxRx.EbNo, ...
       'BitsPerSymbol',
prmQPSKTxRx.Upsampling/prmQPSKTxRx.Downsampling, ...
       'FrequencyOffset', prmQPSKTxRx.FrequencyOffset, ...
       'SampleRate', prmQPSKTxRx.Fs);
   % Create and configure the receiver System object
   hRx = QPSKReceiverR('DesiredAmplitude',
1/sqrt(prmQPSKTxRx.Upsampling), ...
       'ModulationOrder', prmQPSKTxRx.M, ...
       'DownsamplingFactor', prmQPSKTxRx.Downsampling, ...
       'CoarseCompFrequencyResolution',
prmQPSKTxRx.CoarseCompFrequencyResolution, ...
       'PhaseRecoveryDampingFactor',
prmQPSKTxRx.PhaseRecoveryDampingFactor, ...
       'PhaseRecoveryLoopBandwidth',
prmQPSKTxRx.PhaseRecoveryLoopBandwidth, ...
       'TimingRecoveryDampingFactor',
prmQPSKTxRx.TimingRecoveryDampingFactor, ...
       'TimingRecoveryLoopBandwidth',
prmQPSKTxRx.TimingRecoveryLoopBandwidth, ...
       'TimingErrorDetectorGain',
prmQPSKTxRx.TimingErrorDetectorGain, ...
       'PostFilterOversampling',
prmQPSKTxRx.Upsampling/prmQPSKTxRx.Downsampling, ...
       'FrameSize', prmQPSKTxRx.FrameSize, ...
       'BarkerLength', prmQPSKTxRx.BarkerLength, ...
       'MessageLength', prmQPSKTxRx.MessageLength, ...
       'SampleRate', prmQPSKTxRx.Fs, ...
       'DataLength', prmQPSKTxRx.DataLength, ...
       'ReceiverFilterCoefficients',
prmQPSKTxRx.ReceiverFilterCoefficients, ...
       'DescramblerBase', prmQPSKTxRx.ScramblerBase, ...
       'DescramblerPolynomial', prmQPSKTxRx.ScramblerPolynomial, ...
       'DescramblerInitialConditions',
prmQPSKTxRx.ScramblerInitialConditions,...
       'LdpcNewH',prmQPSKTxRx.LdpcNewH,...
      'LdpcIteration',prmQPSKTxRx.LdpcIteration,...
      'LdpcU',prmQPSKTxRx.LdpcU,...
      'LdpcL',prmQPSKTxRx.LdpcL,...
       'PrintOption', printData);
```

if useScopes

```
% Create the System object for plotting all the scopes
      hScopes = QPSKScopes;
   end
hRx.PrintOption = printData;
for count = 1:prmQPSKTxRx.FrameCount
   [transmittedSignal] = step(hTx); % Transmitter
   corruptSignal = step(hChan, transmittedSignal, 0); % AWGN Channel
   [RCRxSignal,coarseCompBuffer, timingRecBuffer,BER] =
step(hRx,corruptSignal); % Receiver
end
if useScopes
    releaseQPSKScopes(hScopes);
end
fprintf('Error rate = %f.\n',BER(1));
fprintf('Number of detected errors = %d.\n', BER(2));
fprintf('Total number of compared samples = %d.\n',BER(3));
```

'BernoulliLength', obj.DataLength-2\*obj.MessageLength, ...

'ScramblerInitialConditions', obj.ScramblerInitialConditions,...

'ScramblerPolynomial', obj.ScramblerPolynomial, ...

'ScramblerBase', obj.ScramblerBase, ...

'LdpcNewH',obj.LdpcNewH,...

'LdpcU',obj.LdpcU,...
'LdpcL',obj.LdpcL);

```
obj.pQPSKModulator = comm.QPSKModulator('BitInput',true, ...
             'PhaseOffset', pi/4);
         obj.pTransmitterFilter = dsp.FIRInterpolator(obj.UpsamplingFactor, ...
             obj.TransmitterFilterCoefficients);
    end
    function [transmittedSignal,transmittedData,modulatedData] = stepImpl(obj)
          % Generates the data to be transmitted
         [transmittedData, ~] = step(obj.pBitGenerator);
         % Modulates the bits into QPSK symbols
         modulatedData = step(obj.pQPSKModulator, transmittedData);
         % Square root Raised Cosine Transmit Filter
         transmittedSignal = step(obj.pTransmitterFilter, modulatedData);
    end
    function resetImpl(obj)
         reset(obj.pBitGenerator);
         reset(obj.pQPSKModulator );
         reset(obj.pTransmitterFilter);
    end
    function releaseImpl(obj)
         release(obj.pBitGenerator);
         release(obj.pQPSKModulator );
         release(obj.pTransmitterFilter);
    end
    function N = getNumInputsImpl(~)
         N = 0:
    end
end
```

end

```
-----QPSKBitsGeneratorR.m-----
classdef QPSKBitsGeneratorR < matlab.System</pre>
%#codegen
% Generates the bits for each frame
   Copyright 2012 The MathWorks, Inc.
   properties (Nontunable)
      MessageLength = 105;
      BernoulliLength = 69;
      ScramblerBase = 2;
      ScramblerPolynomial = [1 1 1 0 1];
      ScramblerInitialConditions = [0 0 0 0];
      LdpcNewH=zeros(148,296);
      LdpcU=zeros(148,148);
      LdpcL=zeros(148,148);
   end
   properties (Access=private)
      pHeader
      pScrambler
      pMsgStrSet
      pCount
   end
   methods
      function obj = QPSKBitsGeneratorR(vararqin)
         setProperties(obj,nargin,varargin(:));
      end
   end
   methods (Access=protected)
      function setupImpl(obj, ~)
         +1 -1 -1 +1 +1 -1 +1 -1 +1]; % Bipolar Barker Code
         ubc = ((bbc + 1) / 2)'; % Unipolar Barker Code
         temp = (repmat(ubc, 1, 2))';
         obj.pHeader = temp(:);
         obj.pCount = 0;
         obj.pScrambler = comm.Scrambler(obj.ScramblerBase, ...
            obj.ScramblerPolynomial,
obj.ScramblerInitialConditions);
         obj.pMsgStrSet = ['Hello world 1000';...
```

```
'Hello world 1001';...
'Hello world 1002';...
'Hello world 1003';...
'Hello world 1004';...
'Hello world 1005';...
'Hello world 1006';...
'Hello world 1007';...
'Hello world 1008';...
'Hello world 1009';...
'Hello world 1010';...
'Hello world 1011';...
'Hello world 1012';...
'Hello world 1013';...
'Hello world 1014';...
'Hello world 1015';...
'Hello world 1016';...
'Hello world 1017';...
'Hello world 1018';...
'Hello world 1019';...
'Hello world 1020';...
'Hello world 1021';...
'Hello world 1022';...
'Hello world 1023';...
'Hello world 1024';...
'Hello world 1025';...
'Hello world 1026';...
'Hello world 1027';...
'Hello world 1028';...
'Hello world 1029';...
'Hello world 1030';...
'Hello world 1031';...
'Hello world 1032';...
'Hello world 1033';...
'Hello world 1034';...
'Hello world 1035';...
'Hello world 1036';...
'Hello world 1037';...
'Hello world 1038';...
'Hello world 1039';...
'Hello world 1040';...
'Hello world 1041';...
'Hello world 1042';...
'Hello world 1043';...
'Hello world 1044';...
```

```
'Hello world 1045';...
'Hello world 1046';...
'Hello world 1047';...
'Hello world 1048';...
'Hello world 1049';...
'Hello world 1050';...
'Hello world 1051';...
'Hello world 1052';...
'Hello world 1053';...
'Hello world 1054';...
'Hello world 1055';...
'Hello world 1056';...
'Hello world 1057';...
'Hello world 1058';...
'Hello world 1059';...
'Hello world 1060';...
'Hello world 1061';...
'Hello world 1062';...
'Hello world 1063';...
'Hello world 1064';...
'Hello world 1065';...
'Hello world 1066';...
'Hello world 1067';...
'Hello world 1068';...
'Hello world 1069';...
'Hello world 1070';...
'Hello world 1071';...
'Hello world 1072';...
'Hello world 1073';...
'Hello world 1074';...
'Hello world 1075';...
'Hello world 1076';...
'Hello world 1077';...
'Hello world 1078';...
'Hello world 1079';...
'Hello world 1080';...
'Hello world 1081';...
'Hello world 1082';...
'Hello world 1083';...
'Hello world 1084';...
'Hello world 1085';...
'Hello world 1086';...
'Hello world 1087';...
'Hello world 1088';...
```

```
'Hello world 1089';...
           'Hello world 1090';...
           'Hello world 1091';...
           'Hello world 1092';...
           'Hello world 1093';...
           'Hello world 1094';...
           'Hello world 1095';...
           'Hello world 1096';...
           'Hello world 1097';...
           'Hello world 1098';...
           'Hello world 1099'];
      end
      function [y,msg] = stepImpl(obj)
          % Converts the message string to bit format
          cycle = mod(obj.pCount,100);
         msgStr = obj.pMsgStrSet(cycle+1,:);
          msgBin = de2bi(int8(msgStr),7,'left-msb');
          msg = reshape(double(msgBin).',obj.MessageLength,1);
          data = [msg ; randi([0 1], obj.BernoulliLength/2, 1)];
          % Scramble the data
          scrambledData = step(obj.pScrambler, data);
          ldpccode = mod(obj.LdpcU\(obj.LdpcL\mod(obj.LdpcNewH(:, 148
+ 1:end)*scrambledData, 2)), 2);
          CodescrambledData = [ldpccode; scrambledData];
          % Append the scrambled bit sequence to the header
          y = [obj.pHeader ; CodescrambledData];
          obj.pCount = obj.pCount+1;
       end
       function resetImpl(obj)
          obj.pCount = 0;
          reset(obj.pScrambler);
       end
       function releaseImpl(obj)
          release(obj.pScrambler);
       end
```

```
function N = getNumInputsImpl(~)
    N = 0;
end

function N = getNumOutputsImpl(~)
    N = 2;
end
end
end
```

```
classdef QPSKReceiverR < matlab.System
```

% Copyright 2012-2015 The MathWorks, Inc.

```
properties (Nontunable)
    DesiredAmplitude = 1/sqrt(2);
    ModulationOrder = 4;
    DownsamplingFactor = 2;
    CoarseCompFrequencyResolution = 50;
    PhaseRecoveryLoopBandwidth = 0.01;
    PhaseRecoveryDampingFactor = 1;
    TimingRecoveryDampingFactor = 1;
    TimingRecoveryLoopBandwidth = 0.01;
    TimingErrorDetectorGain = 5.4;
    PostFilterOversampling = 2;
    FrameSize = 100;
    BarkerLength = 26;
    MessageLength = 105;
    SampleRate = 200000;
    DataLength = 148;
    ReceiverFilterCoefficients = 1;
    DescramblerBase = 2;
    DescramblerPolynomial = [1 1 1 0 1];
    DescramblerInitialConditions = [0 0 0 0];
    LdpcNewH=zeros(148,296);
    LdpcU=zeros(148,148);
    LdpcL=zeros(148,148);
    Ldpclteration=1;
    PrintOption = false;
end
properties (Access = private)
    pAGC
    pRxFilter
     pCoarseFreqEstimator
     pCoarseFreqCompensator
     pFineFreqCompensator
     pTimingRec
     pFrameSync
     pDataDecod
     pBER
 end
```

```
properties (Access = private, Constant)
        pUpdatePeriod = 4 % Defines the size of vector that will be processed in AGC system
object
        1; -1; +1; +1; -1; +1; -1; +1]; % Bipolar Barker Code
        pModulatedHeader = sqrt(2)/2 * (-1-1i) * QPSKReceiverR.pBarkerCode;
    end
    methods
        function obj = QPSKReceiverR(varargin)
            setProperties(obj,nargin,varargin{:});
        end
    end
   methods (Access = protected)
        function setupImpl(obj, ~)
            obj.pAGC = comm.AGC;
            obj.pRxFilter = dsp.FIRDecimator( ...
                'Numerator', obj.ReceiverFilterCoefficients, ...
                'DecimationFactor', obj.DownsamplingFactor);
            obj.pCoarseFreqEstimator = comm.PSKCoarseFrequencyEstimator( ...
                'ModulationOrder',
                                       obj.ModulationOrder, ...
                'Algorithm',
                                       'FFT-based', ...
                'FrequencyResolution', obj.CoarseCompFrequencyResolution, ...
                'SampleRate',
                                       obj.SampleRate);
            obj.pCoarseFreqCompensator = comm.PhaseFrequencyOffset( ...
                'PhaseOffset'.
                                        0, ...
                'FrequencyOffsetSource', 'Input port', ...
                'SampleRate',
                                         obj.SampleRate);
            obj.pFineFreqCompensator = comm.CarrierSynchronizer( ...
                'Modulation'.
                                           'QPSK', ...
                'ModulationPhaseOffset',
                                          'Auto', ...
                'SamplesPerSymbol',
                                            obj.PostFilterOversampling, ...
                'DampingFactor',
                                           obj.PhaseRecoveryDampingFactor, ...
                'NormalizedLoopBandwidth', obj.PhaseRecoveryLoopBandwidth);
           obj.pTimingRec = comm.SymbolSynchronizer( ....
                'TimingErrorDetector',
                                         'Zero-Crossing (decision-directed)', ...
                'SamplesPerSymbol',
                                           obj.PostFilterOversampling, ...
```

```
'NormalizedLoopBandwidth', obj.TimingRecoveryLoopBandwidth, ...
                                              obj.TimingErrorDetectorGain);
                  'DetectorGain',
             obj.pFrameSync = FrameFormation( ...
                                              obj.FrameSize, ...
                  'OutputFrameLength',
                  'PerformSynchronization', true, ...
                                              obj.pModulatedHeader);
                  'FrameHeader',
             obj.pDataDecod = QPSKDataDecoderR('FrameSize', obj.FrameSize, ...
                  'BarkerLength', obj.BarkerLength, ...
                  'ModulationOrder', obj.ModulationOrder, ...
                  'DataLength', obj.DataLength, ...
                  'MessageLength', obj.MessageLength, ...
                  'DescramblerBase', obj.DescramblerBase, ...
                  'DescramblerPolynomial', obj.DescramblerPolynomial, ...
                  'DescramblerInitialConditions', obj.DescramblerInitialConditions, ...
                  'LdpcNewH',obj.LdpcNewH,...
                  'Ldpclteration',obj.Ldpclteration,...
                  'LdpcU',obj.LdpcU,...
                  'LdpcL',obj.LdpcL,...
                  'PrintOption', obj.PrintOption);
         end
         function [RCRxSignal, fineCompSignal, timingRecBuffer,BER] = stepImpl(obj, bufferSignal)
             % AGC control
             AGCSignal = obj.DesiredAmplitude*step(obj.pAGC, bufferSignal);
             % Pass the signal through Square-Root Raised Cosine Received Filter
             RCRxSignal = step(obj.pRxFilter, AGCSignal);
             % Coarse frequency offset estimation
             freqOffsetEst = step(obj.pCoarseFreqEstimator, RCRxSignal);
             % Coarse frequency compensation
                                        step(obj.pCoarseFreqCompensator,
                                                                               RCRxSignal,
             coarseCompSignal
freqOffsetEst);
             % Fine frequency compensation
             fineCompSignal = step(obj.pFineFreqCompensator, coarseCompSignal);
             % Symbol timing recovery
             [timingRecSignal, timingRecBuffer] = step(obj.pTimingRec, fineCompSignal);
```

'DampingFactor',

obj.TimingRecoveryDampingFactor, ...

```
% Frame synchronization
         [symFrame, isFrameValid] = step(obj.pFrameSync, timingRecSignal);
         if isFrameValid % Decode frame of symbols
             obj.pBER = step(obj.pDataDecod, symFrame);
         end
         BER = obj.pBER;
    end
    function resetImpl(obj)
         obj.pBER = zeros(3, 1);
         reset(obj.pAGC);
         reset(obj.pRxFilter);
         reset(obj.pCoarseFreqEstimator);
         reset(obj.pCoarseFreqCompensator);
         reset(obj.pFineFreqCompensator);
         reset(obj.pTimingRec);
         reset(obj.pFrameSync);
         reset(obj.pDataDecod);
    end
    function releaseImpl(obj)
         release(obj.pAGC);
         release(obj.pRxFilter);
         release(obj.pCoarseFreqEstimator);
         release(obj.pCoarseFreqCompensator);
         release(obj.pFineFreqCompensator);
         release(obj.pTimingRec);
         release(obj.pFrameSync);
         release(obj.pDataDecod);
    end
    function N = getNumOutputsImpl(~)
         N = 4;
    end
end
```

end

```
----- QPSKDataDecoderR------
classdef QPSKDataDecoderR < matlab.System
% Copyright 2012-2014 The MathWorks, Inc.
    properties (Nontunable)
        FrameSize = 100;
        BarkerLength = 13;
        ModulationOrder = 4;
        DataLength = 174;
       MessageLength = 105;
       DescramblerBase = 2;
       DescramblerPolynomial = [1 1 1 0 1];
       DescramblerInitialConditions = [0 0 0 0];
       LdpcNewH=zeros(148,296);
       LdpcU=zeros(148,148);
       LdpcL=zeros(148,148);
       Ldpclteration=1;
       PrintOption = false;
   end
   properties (Access = private)
       pCorrelator
       pQPSKDemodulator
       pDescrambler
       pBitGenerator
       pErrorRateCalc
   end
   properties (Constant, Access = private)
       1; -1; +1; +1; -1; +1; -1; +1]; % Bipolar Barker Code
       pModulatedHeader = sqrt(2)/2 * (-1-1i) * QPSKDataDecoderR.pBarkerCode;
   end
   methods
       function obj = QPSKDataDecoderR(varargin)
           setProperties(obj,nargin,varargin{:});
       end
   end
   methods (Access = protected)
```

```
obj.pCorrelator = dsp.Crosscorrelator;
            obj.pQPSKDemodulator = comm.QPSKDemodulator('PhaseOffset',pi/4, ...
                 'BitOutput', true);
             obj.pDescrambler = comm.Descrambler(obj.DescramblerBase, ...
                  obj.DescramblerPolynomial, obj.DescramblerInitialConditions);
             obj.pBitGenerator = QPSKBitsGeneratorR('MessageLength', obj.MessageLength, ...
                  'BernoulliLength', obj.DataLength-2*obj.MessageLength, ...
                  'ScramblerBase', obj.DescramblerBase, ...
                  'ScramblerPolynomial', obj.DescramblerPolynomial, ...
                  'ScramblerInitialConditions', obj.DescramblerInitialConditions,...
                  'LdpcNewH',obj.LdpcNewH,...
                  'LdpcU',obj.LdpcU,...
                  'LdpcL',obj.LdpcL);
             obj.pErrorRateCalc = comm.ErrorRate;
        end
        function BER = stepImpl(obj, data)
             % Phase offset estimation
             phaseEst
                                      round(angle(mean(conj(obj.pModulatedHeader)
data(1:obj.BarkerLength)))*2/pi)/2*pi;
             % Compensating for the phase offset
             phShiftedData = data .* exp(-1i*phaseEst);
             % Demodulating the phase recovered data
             demodOut = step(obj.pQPSKDemodulator, phShiftedData);
             demodOutMsg=demodOut( ...
                  obj.BarkerLength*log2(obj.ModulationOrder)+1:...
                  obj.FrameSize*log2(obj.ModulationOrder));
             vhat = decodeBitFlip(demodOutMsg',obj.LdpcNewH,obj.LdpcIteration);
             deScrDataMsg=vhat(149:end)';
             % Performs descrambling
             deScrData = step(obj.pDescrambler, deScrDataMsg);
             % Recovering the message from the data
             Received = deScrData(1:obj.MessageLength);
```

function setupImpl(obj, ~)

```
bits2ASCII(obj, Received);
        [\sim, transmittedMessage] = step(obj.pBitGenerator);
        BER = step(obj.pErrorRateCalc, transmittedMessage, Received);
    end
    function resetImpl(obj)
        reset(obj.pCorrelator);
        reset(obj.pQPSKDemodulator);
        reset(obj.pDescrambler);
        reset(obj.pBitGenerator);
         reset(obj.pErrorRateCalc);
    end
    function releaseImpl(obj)
         release(obj.pCorrelator);
         release(obj.pQPSKDemodulator);
         release(obj.pDescrambler);
         release(obj.pBitGenerator);
         release(obj.pErrorRateCalc);
    end
end
methods (Access=private)
    function bits2ASCII(obj,u)
         coder.extrinsic('disp')
         % Convert binary-valued column vector to 7-bit decimal values.
         w = [64 32 16 8 4 2 1]; % binary digit weighting
         Nbits = numel(u);
         Ny = Nbits/7;
         y = zeros(1,Ny);
          for i = 0:Ny-1
              y(i+1) = w*u(7*i+(1:7));
          end
          % Display ASCII message to command window
          if(obj.PrintOption)
               disp(char(y));
          end
     end
end
```

end

```
function H = makeLdpc(M, N, method, noCycle, onePerCol)
% Create R = 1/2 low density parity check matrix
              : Number of row
% N
              : Number of column
% method : Method for distributing non-zero element
%
               {0} Evencol: For each column, place 1s uniformly at random
               {1} Evenboth: For each column and row, place 1s uniformly at random
%
%
  noCyle : Length-4 cycle
%
                {0} Ignore (do nothing)
%
                {1} Eliminate
%
  onePerCol: Number of ones per column
%
             : Low density parity check matrix
% H
%
%
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% http://bsnugroho.googlepages.com
% Number of ones per row (N/M ratio must be 2)
if N/M \sim= 2
   fprintf('Code rate must be 1/2\n');
onePerRow = (N/M)*onePerCol;
fprintf('Creating LDPC matrix...\n');
switch method
   % Evencol
   case {0}
      % Distribute 1s uniformly at random within column
      for i = 1:N
         onesInCol(:, i) = randperm(M)';
      end
      % Create non zero elements (1s) index
      r = reshape(onesInCol(1:onePerCol, :), N*onePerCol, 1);
      tmp = repmat([1:N], onePerCol, 1);
      c = reshape(tmp, N*onePerCol, 1);
```

% Create sparse matrix H

----- makeLdpc-----

```
H = full(sparse(r, c, 1, M, N));
   % Evenboth
   case {1}
      % Distribute 1s uniformly at random within column
      for i = 1:N
          onesInCol(:, i) = randperm(M)';
      end
      % Create non zero elements (1s) index
      r = reshape(onesInCol(1:onePerCol, :), N*onePerCol, 1);
      tmp = repmat([1:N], onePerCol, 1);
      c = reshape(tmp, N*onePerCol, 1);
      % Make the number of 1s between rows as uniform as possible
      % Order row index
      [r, ix] = sort(r);
      % Order column index based on row index
      for i = 1:N*onePerCol
          cSort(i, :) = c(ix(i));
      end
      % Create new row index with uniform weight
      tmp = repmat([1:M], onePerRow, 1);
      r = reshape(tmp, N*onePerCol, 1);
      % Create sparse matrix H
      % Remove any duplicate non zero elements index using logical AND
      S = and(sparse(r, cSort, 1, M, N), ones(M, N));
      H = full(S);
end % switch
% Check rows that have no 1 or only have one 1
for i = 1:M
   n = randperm(N);
   % Add two 1s if row has no 1
   if length(find(r == i)) == 0
       H(i, n(1)) = 1;
       H(i, n(2)) = 1;
   % Add one 1 if row has only one 1
```

```
elseif length(find(r == i)) == 1
       H(i, n(1)) = 1;
    end
end % for i
% If desired, eliminate any length-4 cycle
if noCycle == 1
    for i = 1:M
       % Look for pair of row - column
       for j = (i + 1):M
           w = and(H(i, :), H(j, :));
           c1 = find(w);
           lc = length(c1);
           if tc > 1
               % If found, flip one 1 to 0 in the row with less number of 1s
               if \ length(find(H(i,:))) < length(find(H(j,:))) \\
                   % Repeat the process until only one column left
                   for cc = 1:|c - 1|
                      H(j, c1(cc)) = 0;
                   end
               else
                   for cc = 1:|c - 1|
                      H(i, c1(cc)) = 0;
                   end
               end % if
           end % if
       end % for j
   end % for i
end % if
fprintf('LDPC matrix is created.\n');
```

```
function [newH,U,L] = makeParityChk( H, strategy)
% Generate parity check vector bases on LDPC matrix H using sparse LU decomposition
  dSource: Binary source (0/1)
            : LDPC matrix
% H
   strategy: Strategy for finding the next non-zero diagonal elements
               {0} First : First non-zero found by column search
%
               {1} Mincol: Minimum number of non-zeros in later columns
%
               {2} Minprod: Minimum product of:
                              - Number of non-zeros its column minus 1
%
                              - Number of non-zeros its row minus 1
%
            : Check bits
% C
%
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% Get the matric dimension
[M, N] = size(H);
% Set a new matrix F for LU decomposition
F = H:
% LU matrices
L = zeros(M, N - M);
U = zeros(M, N - M);
% Re-order the M x (N - M) submatrix
for i = 1:M
    % strategy {0 = First; 1 = Mincol; 2 = Minprod}
    switch strategy
       % Create diagonally structured matrix using 'First' strategy
       case {0}
           % Find non-zero elements (1s) for the diagonal
           [r, c] = find(F(:, i:end));
           % Find non-zero diagonal element candidates
           rowIndex = find(r == i);
           % Find the first non-zero column
```

\_\_\_\_\_ makeParityChk------

```
chosenCol = c(rowIndex(1)) + (i - 1);
 % Create diagonally structured matrix using 'Mincol' strategy
 case {1}
     % Find non-zero elements (1s) for the diagonal
     [r, c] = find(F(:, i:end));
     colWeight = sum(F(:, i:end), 1);
     % Find non-zero diagonal element candidates
     rowIndex = find(r == i);
     % Find the minimum column weight
     [x, ix] = min(colWeight(c(rowIndex)));
     % Add offset to the chosen row index to match the dimension of the...
     % original matrix F
     chosenCol = c(rowIndex(ix)) + (i - 1);
  % Create diagonally structured matrix using 'Minprod' strategy
  case {2}
      % Find non-zero elements (1s) for the diagonal
      [r, c] = find(F(:, i:end));
      colWeight = sum(F(:, i:end), 1) - 1;
      rowWeight = sum(F(i, :), 2) - 1;
      % Find non-zero diagonal element candidates
      rowIndex = find(r == i);
      % Find the minimum product
      [x, ix] = min(colWeight(c(rowIndex))*rowWeight);
      % Add offset to the chosen row index to match the dimension of the...
      % original matrix F
      chosenCol = c(rowIndex(ix)) + (i - 1);
      fprintf('Please select columns re-ordering strategy!\n');
end % switch
% Re-ordering columns of both H and F
tmp1 = F(:, i);
tmp2 = H(:, i);
F(:, i) = F(:, chosenCol);
```

```
H(:, i) = H(:, chosenCol);
  F(:, chosenCol) = tmp1;
  H(:, chosenCol) = tmp2;
  % Fill the LU matrices column by column
  L(i:end, i) = F(i:end, i);
   U(1:i, i) = F(1:i, i);
  % There will be no rows operation at the last row
   if i < M
      % Find the later rows with non-zero elements in column i
       [r2, c2] = find(F((i + 1):end, i));
      \% Add current row to the later rows which have a 1 in column i
      F((i + r2), :) = mod(F((i + r2), :) + repmat(F(i, :), length(r2), 1), 2);
   end % if
end % for i
% Find B.dsource
%z = mod(H(:, (N - M) + 1:end)*dSource, 2);
% Parity check vector found by solving sparse LU
%c = mod(U\setminus(L\setminus z), 2);
% Return the rearrange H
newH = H;
%fprintf('Message encoded.\n');
```

```
function vHat = decodeBitFlip(ci, H, iteration)
% Hard-decision/bit flipping sum product algorithm LDPC decoder
              : Received signal vector (column vector)
               : LDPC matrix
  Н
%
   iteration : Number of iteration
               : Decoded vector (0/1)
%
   vHat
%
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[M,N] = size(H);
% Prior hard-decision
\%ci = 0.5*(sign(rx') + 1);
% Initialization
rji = zeros(M, N);
% Associate the ci matrix with non-zero elements of H
qij = H.*repmat(ci, M, 1);
% Iteration
for n = 1:iteration
    %fprintf('Iteration: %d\n', n);
    % ----- Horizontal step -----
    for i = 1:M
       % Find non-zeros in the column
       c1 = find(H(i, :));
       % Get the summation of qij\c1(k)
       for k = 1:length(c1)
           rji(i, c1(k)) = mod(sum(qij(i, c1)) + qij(i, c1(k)), 2);
       end % for k
```

----- decodeBitFlip-----

```
end % for i
   % ----- Vertical step -----
   for j = 1:N
      % Find non-zero in the row
      r1 = find(H(:, j));
      % Number of 1s in a row
      numOfOnes = length(find(rji(r1, j)));
      for k = 1:length(r1)
          % Update qij, set '1' for majority of 1s else '0', excluding r1(k)
          if numOfOnes + ci(j) >= length(r1) - numOfOnes + rji(r1(k), j)
              qij(r1(k),j)=1;
          else
              qij(r1(k),j)=0;
          end
       end % for k
       % Bit decoding
       if numOfOnes + ci(j) >= length(r1) - numOfOnes
          vHat(j) = 1;
       else
          vHat(j) = 0;
       end
   end % for j
end % for n
```

posterior (